

## **AUTO-ZEROED OPERATIONAL AMPLIFIERS**

#### **FEATURES**

First Monolithic Chopper-Stabilize	ed Amplifier
With On-Chip Nulling Capacitors	
Offset Voltage	<b>5</b> μ <b>V</b>
Offset Voltage Drift	0.05μV/°C
Low Supply Current	350μA
High Common-Mode Rejection	116dB
Single Supply Operation	4.5V to 16V
High Slew Rate	<b>2.5V/</b> μs
Wide Bandwidth	1.5MHz
High Open-Loop Voltage Gain	
$(R_L = 10 kΩ)$	120dB
Low Input Voltage Noise	
(0.1 Hz to 1 Hz)	0.65μV <sub>P-P</sub>
Pin Compatible With ICL7650	,
Lower System Parts Count	

#### ORDERING INFORMATION

Part No.	Package	Temperature Range	Maximum Offset Voltage
TC911ACOA	8-Pin SOIC	0°C to +70°C	15μV
TC911ACPA	8-Pin Plastic DIP	0°C to +70°C	15μV
TC911BCOA	8-Pin SOIC	0°C to +70°C	30μV
TC911BCPA	8-Pin Plastic DIP	0°C to +70°C	30μV

### **GENERAL DESCRIPTION**

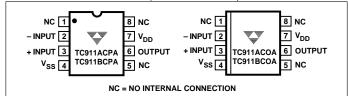
The TC911 CMOS auto-zeroed operational amplifier is the first complete monolithic chopper-stabilized amplifier. Chopper operational amplifiers like the ICL7650/7652 and LTC1052 require user-supplied, external offset compensation storage capacitors. **External capacitors are not required with the TC911.** Just as easy to use as the conventional OP07 type amplifier, the TC911 significantly reduces offset voltage errors. Pinout matches the OP07/741/7650 8-pin mini-DIP configuration.

Several system benefits arise by eliminating the external chopper capacitors: lower system parts count, reduced assembly time and cost, greater system reliability, reduced PC board layout effort and greater board area utilization. Space savings can be significant in multiple-amplifier designs.

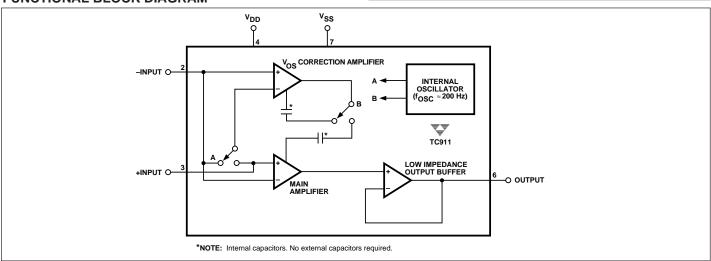
Electrical specifications include  $15\mu V$  maximum offset voltage,  $0.15\mu V/^{\circ}C$  maximum offset voltage temperature coefficient. Offset voltage error is five times lower than the premium OP07E bipolar device. The TC911 improves offset drift performance by eight times.

The TC911 operates from dual or single power supplies. Supply current is typically  $350\mu A$ . Single 4.5V to 16V supply operation is possible, making single 9V battery operation possible. The TC911 is available in 2 package types: 8-pin plastic DIP and SOIC.

#### PIN CONFIGURATION (SOIC and DIP)



#### **FUNCTIONAL BLOCK DIAGRAM**



# AUTO-ZEROED MONOLITHIC OPERATIONAL AMPLIFIERS

# TC911A TC911B

#### **ABSOLUTE MAXIMUM RATINGS\***

Total Supply Voltage (V <sub>DD</sub> to V <sub>SS</sub> )	+18V
Input Voltage (V <sub>DD</sub> + 0.3	3V) to $(V_{SS} - 0.3V)$
Current into Any Pin	10mA
While Operating	100μΑ
Storage Temperature Range	65°C to +150°C
Lead Temperature (Soldering, 10 sec)	+300°C
Operating Temperature Range	
C. Device	$0^{\circ}$ C to $+70^{\circ}$ C

Package Power Dissipation ( $T_A = \le 70^{\circ}C$ )	
Plastic DIP	730mW
Plastic SOIC	470mW

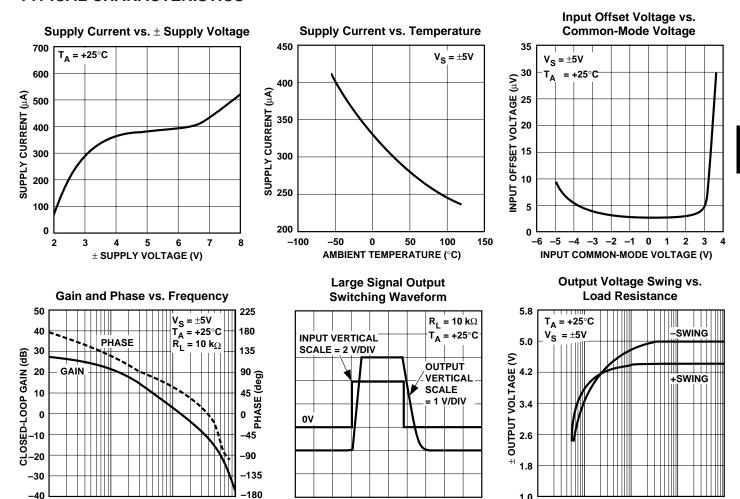
<sup>\*</sup>Static-sensitive device. Unused devices should be stored in conductive material. Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied.

## **ELECTRICAL CHARACTERISTICS:** $V_S = \pm 5V$ , $T_A = +25$ °C, unless otherwise indicated.

Symbol	Parameter	Test Conditions	TC911A			TC911B			
			Min	Тур	Max	Min	Тур	Max	Unit
Vos	Input Offset Voltage	T <sub>A</sub> = +25°C	_	5	15	_	15	30	μV
TCV <sub>OS</sub>	Average Temperature Coefficient of Input Offset Voltage	$0^{\circ}C \le T_{A} \le +70^{\circ}C$ -25°C \le T_{A} \le +85°C (Note 1)	_	0.05 0.05	0.15 0.15	_	0.1 0.1	0.25 0.25	μV/°C μV/°C
I <sub>B</sub>	Average Input Bias Current	$T_A = +25^{\circ}C$ $0^{\circ}C \le T_A \le +70^{\circ}C$ $-25^{\circ}C \le T_A \le +85^{\circ}C$	_ _ _	_ _ _	70 3 4		_ _ _	120 4 6	pA nA nA
I <sub>OS</sub>	Average Input Offset Current	$T_A = +25^{\circ}C$ $T_A = +85^{\circ}C$	_	5 —	20 1	_	10 —	40 1	pA nA
e <sub>N</sub>	Input Voltage Noise	0.1 to 1 Hz, $R_S \le 100\Omega$ 0.1 to 10 Hz, $R_S \le 100\Omega$		0.65 11	_	_	0.65 11	_	μV <sub>P-P</sub> μV <sub>P-P</sub>
CMRR	Common-Mode Rejection Ratio	$V_{SS} \le V_{CM} \le V_{DD} - 2.2$	110	116	_	105	110	_	dB
CMVR	Common-Mode Voltage Range		V <sub>SS</sub>	_	V <sub>DD</sub> – 2	V <sub>SS</sub>	_	V <sub>DD</sub> – 2	V
A <sub>OL</sub>	Open-Loop Voltage Gain	$R_L = 10 \text{ k}\Omega, V_{OUT} = \pm 4V$	115	120	_	110	120	_	dB
V <sub>OUT</sub>	Output Voltage Swing	$R_L = 10 \text{ k}\Omega$	$V_{SS} + 0.3$	_	$V_{DD} - 0.9$	V <sub>SS</sub> + 0.3	_	$V_{DD} - 0.9$	V
BW	Closed Loop Bandwidth	Closed Loop Gain = +1	_	1.5	_	_	1.5	_	MHz
SR	Slew Rate	$R_L = 10 \text{ k}\Omega, C_L = 50 \text{ pF}$	_	2.5	_	_	2.5	_	V/μs
PSRR	Power Supply Rejection Ratio	±3.3V to ±5.5V	112	_	_	105	_	_	dB
Vs	Operating Supply Voltage Range	Split Supply Single Supply	±3.3 6.5	_	±8 16	±3.3 6.5	_	±8 16	V
Is	Quiescent Supply Current	$V_S = \pm 5V$	_	350	600	_	_	800	μА

NOTES: 1. Characterized; not 100% tested.

#### TYPICAL CHARACTERISTICS



HORIZONTAL SCALE =  $2 \mu s/DIV$ 

100

10k

LOAD RESISTANCE ( $\Omega$ )

100k

1M

10k

100k

FREQUENCY (Hz)

10M

### Pin Compatibility

The CMOS TC911 is pin compatible with the industry standard ICL7650 chopper-stabilized amplifier. The ICL7650 must use external 0.1  $\mu\text{F}$  capacitors connected at pins 1 and 8. With the TC911, external offset voltage error canceling capacitors are not required. On the TC911 pins 1, 8 and 5 are not connected internally. The ICL7650 uses pin 5 as an optional output clamp connection. External chopper capacitors and clamp connections are not necessary with the TC911. External circuits connected to pins 1, 8 and 5 will have no effect. The TC911 can be quickly evaluated in existing ICL7650 designs. Since external capacitors are not required, system part count, assembly time, and total system cost are reduced. Reliability is increased and PC board layout eased by having the error storage capacitors integrated on the TC911 chip.

The TC911 pinout matches many existing op amps: 741, LM101, LM108, OP05–OP08, OP-20, OP-21, ICL7650 and ICL7652. In many applications operating from +5V supplies the TC911 offers superior electrical performance and can be a functional pin-compatible replacement. Offset voltage correction potentiometers, compensation capacitors, and chopper-stabilization capacitors can be removed when retrofitting existing equipment designs.

#### Thermocouple Errors

Heating one joint of a loop made from two different metallic wires causes current flow. This is known as the Seebeck effect. By breaking the loop, an open circuit voltage

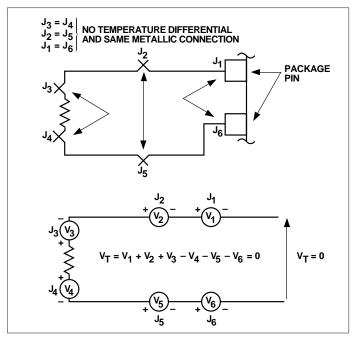


Figure 1. Unwanted Thermocouple Errors Eliminated by Reducing Thermal Gradients and Balancing Junctions

(Seebeck voltage) can be measured. Junction temperature and metal type determine the magnitude. Typical values are  $0.1\mu V/^{\circ}C$  to  $10\mu V/^{\circ}C$ . Thermal-induced voltages can be many times larger than the TC911 offset voltage drift. Unless unwanted thermocouple potentials can be controlled, system performance will be less than optimum.

Unwanted thermocouple junctions are created when leads are soldered or sockets/connectors are used. Low thermo-electric coefficient solder can reduce errors. A 60% Sn/36% Pb solder has 1/10 the thermal voltage of common 64% Sn/36% Pb solder at a copper junction.

The number and type of dissimilar metallic junctions in the input circuit loop should be balanced. If the junctions are kept at the same temperature, their summation will add to zero-canceling errors (Figure 1).

Shielding precision analog circuits from air currents — especially those caused by power dissipating components and fans — will minimize temperature gradients and thermocouple-induced errors.

## **Avoiding Latch-Up**

Junction-isolated CMOS circuits inherently contain a parasitic p-n-p-n transistor circuit. Voltages exceeding the supplies by 0.3V should not be applied to the device pins. Larger voltages can turn the p-n-p-n device on, causing excessive device power supply current and excessive power dissipation. TC911 power supplies should be established at the same time or before input signals are applied. If this is not possible input current should be limited to 0.1mA to avoid triggering the p-n-p-n structure.

#### **Overload Recovery**

The TC911 recovers quickly from the output saturation. Typical recovery time from positive output saturation is 20msec. Negative output saturation recovery time is typically 5msec.

#### **TYPICAL APPLICATIONS**

