# INTEGRATED CIRCUITS

# DATA SHEET

# 89C535/89C536/89C538 CMOS single-chip 8-bit microcontrollers with FLASH program memory

Preliminary specification

1997 June 05

IC20 Data Handbook





# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### **DESCRIPTION**

The 89C535/89C536/89C538 are Single-Chip 8-Bit Microcontrollers manufactured in advanced CMOS process and are derivatives of the 80C51 microcontroller family. All the devices have the same instruction set as the 80C51.

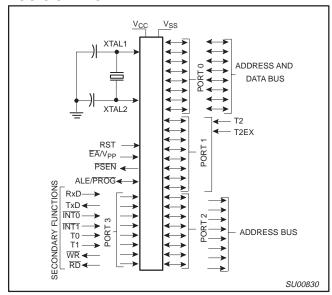
The devices also have four 8-bit I/O ports, three 16-bit timer/event counters, a multi-source, two-priority-level, nested interrupt structure, UART and on-chip oscillator and timing circuits. For systems that require extra data memory capability up to 64k bytes, each can be expanded using standard TTL-compatible memories and logic.

The 89C535/89C536/89C538 contain a non-volatile FLASH EPROM program memory (8K bytes in 89C535, 16k bytes in the 89C536, and 64k bytes in the 89C538). The devices have 512 bytes of RAM data memory.

### **FEATURES**

- 80C51 Central Processing Unit
- 8k x 8 (89C535) 16k × 8 (89C536) or 64k × 8 (89C538), FLASH EPROM Program Memory
- 512 × 8 RAM, externally expandable to 64k × 8 Data Memory
- Three 16-bit counter/timers
- Up to 3 external interrupt request inputs
- 6 interrupt sources with 2 priority levels
- Four 8-bit I/O ports
- Full-duplex UART
- Power control modes
  - Idle mode
  - Power down mode, with wakeup from power down using external interrupt
- 44-pin PLCC and QFP packages

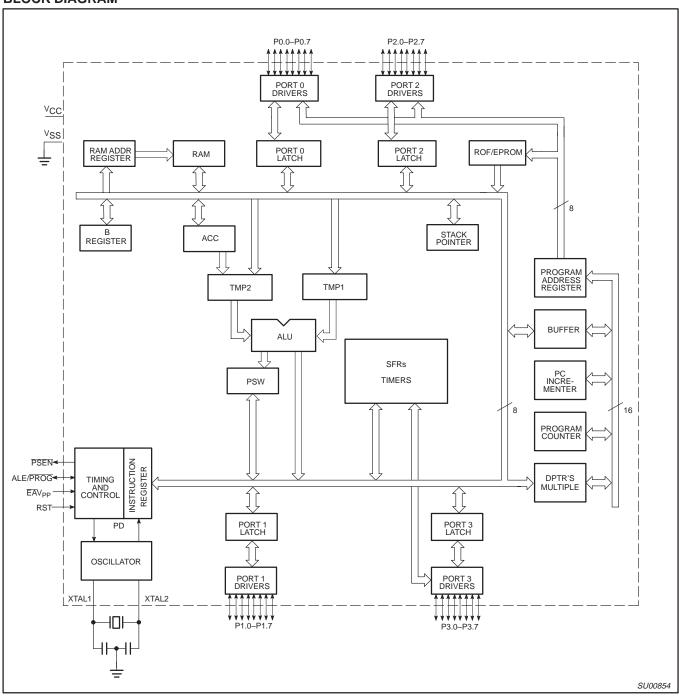
#### LOGIC SYMBOL



### **ORDERING INFORMATION**

PART NUMBER	MEMORY SIZE	TEMPERATURE RANGE (°C) AND PACKAGE	FREQ. (MHz)	DRAWING NUMBER
P89C535NBA A	8k bytes	0 to +70, 44-pin Plastic Leaded Chip Carrier	33	SOT187-2
P89C536NBA A	16k bytes	0 to +70, 44-pin Plastic Leaded Chip Carrier	33	SOT187-2
P89C536NBB B	16k bytes	0 to +70, 44-pin Plastic Quad Flat Package	33	SOT307-2
P89C538NBA A	64k bytes	0 to +70, 44-pin Plastic Leaded Chip Carrier	33	SOT187-2
P89C538NBB B	64k bytes	0 to +70, 44-pin Plastic Quad Flat Package	33	SOT307-2

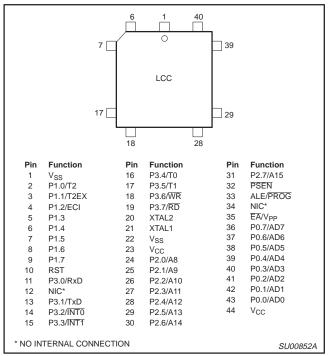
### **BLOCK DIAGRAM**



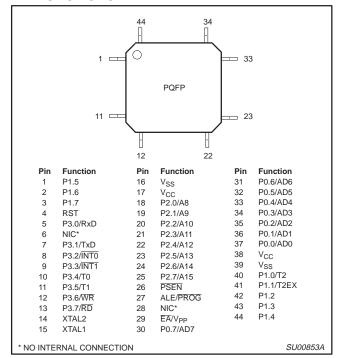
# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

# CERAMIC AND PLASTIC LEADED CHIP CARRIER PIN FUNCTIONS



# PLASTIC QUAD FLAT PACK PIN FUNCTIONS



# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

### **PIN DESCRIPTIONS**

	PIN NUMBER			
MNEMONIC	LCC	QFP	TYPE	NAME AND FUNCTION
V <sub>SS</sub>	1, 22	16, 39	ı	Ground: 0V reference.
V <sub>CC</sub>	23, 44	17, 38	ı	<b>Power Supply:</b> This is the power supply voltage for normal, idle, and power-down operation.
P0.0-0.7	43–36	37–30	I/O	<b>Port 0:</b> Port 0 is an open-drain, bidirectional I/O port. Port 0 pins that have 1s written to them float and can be used as high-impedance inputs. Port 0 is also the multiplexed low-order address and data bus during accesses to external program and data memory. In this application, it uses strong internal pull-ups when emitting 1s. Port 0 also outputs the code bytes during program verification and received code bytes during EEPROM programming. External pull-ups are required during program verification.
P1.0-P1.7	2–9	40–44, 1–3	I/O	<b>Port 1:</b> Port 1 is an 8-bit bidirectional I/O port with internal pull-ups. Port 1 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, port 1 pins that are externally pulled low will source current because of the internal pull-ups. (See DC Electrical Characteristics: I <sub>IL</sub> ). Port 1 also receives the low-order address byte during program memory verification.
				Alternate functions for Port 1 include:
	2	40	I/O	T2 (P1.0): Timer/Counter 2 external count input
	3	41	- 1	T2EX (P1.1): Timer/Counter 2 Reload/Capture
P2.0-P2.7	24–31	18–25	I/O	Port 2: Port 2 is an 8-bit bidirectional I/O port with internal pull-ups. Port 2 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, port 2 pins that are externally being pulled low will source current because of the internal pull-ups. (See DC Electrical Characteristics: I <sub>IL</sub> ). Port 2 emits the high-order address byte during fetches from external program memory and during accesses to external data memory that use 16-bit addresses (MOVX @DPTR). In this application, it uses strong internal pull-ups when emitting 1s. Some Port 2 pins receive the high order address bits during EEPROM programming and verification.
P3.0-P3.7	11, 13–19	5, 7–13	I/O	<b>Port 3:</b> Port 3 is an 8-bit bidirectional I/O port with internal pull-ups. Port 3 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, port 3 pins that are externally being pulled low will source current because of the pull-ups. (See DC Electrical Characteristics: I <sub>IL</sub> ). Port 3 also serves the special features of the 80C51 family, as listed below:
	11	5	- 1	RxD (P3.0): Serial input port
	13	7	0	TxD (P3.1): Serial output port
	14	8		INTO (P3.2): External interrupt
	15 16	9 10	l ¦	INT1 (P3.3): External interrupt T0 (P3.4): Timer 0 external input
	17	11	l i	T1 (P3.5): Timer 1 external input
	18	12	0	WR (P3.6): External data memory write strobe
	19	13	0	RD (P3.7): External data memory read strobe
RST	10	4	I	<b>Reset:</b> A high on this pin for two machine cycles while the oscillator is running, resets the device. An internal diffused resistor to V <sub>SS</sub> permits a power-on reset using only an external capacitor to V <sub>CC</sub> .
ALE/PROG	33	27	0	Address Latch Enable/Program Pulse: Output pulse for latching the low byte of the address during an access to external memory. In normal operation, ALE is emitted at a constant rate of 1/6 the oscillator frequency, and can be used for external timing or clocking. Note that one ALE pulse is skipped during each access to external data memory. This pin is also the program pulse input (PROG) during EEPROM programming.
PSEN	32	26	0	<b>Program Store Enable:</b> The read strobe to external program memory. When the processor is executing code from the external program memory, <u>PSEN</u> is activated twice each machine cycle, except that two <u>PSEN</u> activations are skipped during each access to external data memory. <u>PSEN</u> is not activated during fetches from internal program memory.
EA/V <sub>PP</sub>	35	29	I	External Access Enable/Programming Supply Voltage: $\overline{EA}$ must be externally held low to enable the device to fetch code from external program memory. If $\overline{EA}$ is held high, the device executes from internal program memory. This pin also receives the 12V programming supply voltage (V <sub>PP</sub> ) during EPROM programming. $\overline{EA}$ is internally latched on Reset.
XTAL1	21	15	1	Crystal 1: Input to the inverting oscillator amplifier and input to the internal clock generator circuits.
XTAL2	20	14	0	Crystal 2: Output from the inverting oscillator amplifier.
/ \	20			

### NOTE:

To avoid "latch-up" effect at power-on, the voltage on any pin at any time must not be higher than  $V_{CC}$  + 0.5V or  $V_{SS}$  – 0.5V, respectively.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

**Table 1. Special Function Registers** 

SYMBOL	DESCRIPTION	DIRECT ADDRESS	BIT A	DDRESS	, SYMBOI	L, OR ALT	ΓERNATIV	E PORT	FUNCTIO	N LSB	RESET VALUE
ACC*	Accumulator	E0H	E7	E6	E5	E4	E3	E2	E1	E0	00H
B*	B register	F0H	F7	F6	F5	F4	F3	F2	F1	F0	00H
DPTR:	Data Pointer (2 bytes)										
DPH	Data Pointer High	83H									00H
DPL	Data Pointer Low	82H									00H
			AF	AE	AD	AC	AB	AA	A9	A8	
IE*	Interrupt Enable	A8H	EA	-	ET2	ES	ET1	EX1	ET0	EX0	00H
			BF	BE	BD	ВС	BB	BA	B9	B8	
IP*	Interrupt Priority	В8Н	_	_	PT2	PS	PT1	PX1	PT0	PX0	x0000000B
			87	86	85	84	83	82	81	80	
P0*	Port 0	80H	AD7	AD6	AD5	AD4	AD3	AD2	AD1	AD0	FFH
			97	96	95	94	93	92	91	90	1
P1*	Port 1	90H	_	_	_	_	_	_	T2EX	T2	FFH
			A7	A6	A5	A4	А3	A2	A1	A0	1
P2*	Port 2	A0H	AD15	AD14	AD13	AD12	AD11	AD10	AD9	AD8	FFH
			B7	B6	B5	B4	B3	B2	B1	В0	1
P3*	Port 3	вон	RD	WR	T1	T0	ĪNT1	ĪNT0	TxD	RxD	FFH
PCON#	Power Control	87H	SMOD	_	_		GF1	GF0	PD	IDL	0xxxx000B
1 001111	1 ower control	0711	D7	D6	D5	<u>L</u> D4	D3	D2	D1	D0	OXXXXOOOB
PSW*	Program Status Word	D0H	CY	AC	F0	RS1	RS0	OV	_	P	00H
RACAP2H#	Timer 2 Capture High	CBH	<u> </u>	7.0			1.00	<u> </u>			00H
RACAP2L#	Timer 2 Capture Low	CAH									00H
SBUF	Serial Data Buffer	99H									xxxxxxxxB
			9F	9E	9D	9C	9B	9A	99	98	
SCON*	Serial Control	98H	SM0	SM1	SM2	REN	TB8	RB8	TI	RI	00H
SP	Stack Pointer	81H									07H
			8F	8E	8D	8C	8B	8A	89	88	
TCON*	Timer Control	88H	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	00H
			CF	CE	CD	СС	СВ	CA	C9	C8	1
T2CON*	Timer 2 Control	C8H	TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2	CP/RL2	00H
TH0	Timer High 0	8CH									00H
TH1	Timer High 1	8DH									00H
TH2#	Timer High 2	CDH									00H
TL0	Timer Low 0	8AH									00H
TL1	Timer Low 1	8BH CCH									00H
TL2#	Timer Low 2		CATE	C/ <del>T</del>	144	MO	CATE	C/ <del>T</del>	N/4	MO	00H
TMOD	Timer Mode	89H	GATE	C/T	M1	M0	GATE	C/T	M1	M0	00H

<sup>\*</sup> SFRs are bit addressable.

<sup>#</sup> SFRs are modified from or added to the 80C51 SFRs.

Reserved bits.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### OSCILLATOR CHARACTERISTICS

XTAL1 and XTAL2 are the input and output, respectively, of an inverting amplifier. The pins can be configured for use as an on-chip oscillator.

To drive the device from an external clock source, XTAL1 should be driven while XTAL2 is left unconnected. There are no requirements on the duty cycle of the external clock signal, because the input to the internal clock circuitry is through a divide-by-two flip-flop. However, minimum and maximum high and low times specified in the data sheet must be observed.

#### **RESET**

A reset is accomplished by holding the RST pin high for at least two machine cycles (24 oscillator periods), while the oscillator is running. To insure a good power-on reset, the RST pin must be high long enough to allow the oscillator time to start up (normally a few milliseconds) plus two machine cycles. At power-on, the voltage on  $V_{CC}$  and RST must come up at the same time for a proper start-up. Ports 1, 2, and 3 will asynchronously be driven to their reset condition when a voltage above  $V_{IH1}$  (min.) is applied to RESET.

### **LOW POWER MODES**

#### Idle Mode

In the idle mode (see Table 2), the CPU puts itself to sleep while all of the on-chip peripherals stay active. The instruction to invoke the idle mode is the last instruction executed in the normal operating mode before the idle mode is activated. The CPU contents, the on-chip RAM, and all of the special function registers remain intact during this mode. The idle mode can be terminated either by any enabled interrupt (at which time the process is picked up at the interrupt service routine and continued), or by a hardware reset which starts the processor in the same manner as a power-on reset.

### **Power-Down Mode**

To save even more power, a Power Down mode (see Table 2) can be invoked by software. In this mode, the oscillator is stopped and the instruction that invoked Power Down is the last instruction executed. The on-chip RAM and Special Function Registers retain their values down to 2.0V and care must be taken to return  $V_{CC}$  to the minimum specified operating voltages before the Power Down Mode is terminated.

Either a hardware reset or external interrupt can be used to exit from Power Down. Reset redefines all the SFRs but does not change the on-chip RAM. An external interrupt allows both the SFRs and the on-chip RAM to retain their values.

To properly terminate Power Down the reset or external interrupt should not be executed before  $V_{\rm CC}$  is restored to its normal operating level and must be held active long enough for the oscillator to restart and stabilize (normally less than 10ms).

With an external interrupt, INT0 and INT1 must be enabled and configured as level-sensitive. Holding the pin low restarts the oscillator but bringing the pin back high completes the exit. Once the interrupt is serviced, the next instruction to be executed after RETI will be the one following the instruction that put the device into Power Down.

### **Design Consideration**

 To eliminate the possibility of an unexpected write when Idle is terminated by reset, the instruction following the one that invokes Idle should not be one that writes to a port pin or to memory.

#### ONCE™ Mode

The ONCE ("On-Circuit Emulation") Mode facilitates testing and debugging of systems without the device having to be removed from the circuit. The ONCE Mode is invoked by:

- 1. Pull ALE low while the device is in reset and PSEN is high;
- 2. Hold ALE low as RST is deactivated.

While the device is in ONCE Mode, the Port 0 pins go into a float state, and the other port pins and ALE and PSEN are weakly pulled high. The oscillator circuit remains active. While the 8XC51FA/FB is in this mode, an emulator or test CPU can be used to drive the circuit. Normal operation is restored when a normal reset is applied.

Table 2. External Pin Status During Idle and Power-Down Mode

MODE	PROGRAM MEMORY	ALE	PSEN	PORT 0	PORT 1	PORT 2	PORT 3
Idle	Internal	1	1	Data	Data	Data	Data
Idle	External	1	1	Float	Data	Address	Data
Power-down	Internal	0	0	Data	Data	Data	Data
Power-down	External	0	0	Float	Data	Data	Data

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### **TIMER 2 OPERATION**

#### Timer 2

Timer 2 is a 16-bit Timer/Counter which can operate as either an event timer or an event counter, as selected by C/T2\* in the special function register T2CON (see Figure 1). Timer 2 has three operating modes:Capture, Auto-reload, and Baud Rate Generator, which are selected by bits in the T2CON as shown in Table 3.

### **Capture Mode**

In the capture mode there are two options which are selected by bit EXEN2 in T2CON. If EXEN2=0, then timer 2 is a 16-bit timer or counter (as selected by C/T2\* in T2CON) which, upon overflowing sets bit TF2, the timer 2 overflow bit. This bit can be used to generate an interrupt (by enabling the Timer 2 interrupt bit in the IE register/SFR table). If EXEN2= 1, Timer 2 operates as described above, but with the added feature that a 1- to -0 transition at external input T2EX causes the current value in the Timer 2 registers, TL2 and TH2, to be captured into registers RCAP2L and RCAP2H, respectively. In addition, the transition at T2EX causes bit EXF2 in T2CON to be set, and EXF2 like TF2 can generate an interrupt (which vectors to the same location as Timer 2 overflow interrupt. The Timer 2 interrupt service routine can interrupate TF2 and EXF2 to determine which event caused the interrupt). The capture mode is

illustrated in Figure 2 (There is no reload value for TL2 and TH2 in this mode. Even when a capture event occurs from T2EX, the counter keeps on counting T2EX pin transitions or osc/12 pulses.).

#### **Auto-Reload Mode**

In the 16-bit auto-reload mode, Timer 2 can be configured as either a timer or counter (C/T2\* in T2CON).

Figure 3 shows the auto—reload mode of Timer 2. In this mode there are two options selected by bit EXEN2 in T2CON register. If EXEN2=0, then Timer 2 counts up to 0FFFFH and sets the TF2 (Overflow Flag) bit upon overflow. This causes the Timer 2 registers to be reloaded with the 16-bit value in RCAP2L and RCAP2H.

The values in RCAP2L and RCAP2H are preset by software. If EXEN2=1, then a 16-bit reload can be triggered either by an overflow or by a 1-to-0 transition at input T2EX. This transition also sets the EXF2 bit. The Timer 2 interrupt, if enabled, can be generated when either TF2 or EXF2 are 1.

The external flag EXF2 toggles when Timer 2 underflows or overflows. This EXF2 bit can be used as a 17th bit of resolution if needed. The EXF2 flag does not generate an interrupt in this mode of operation.

	(	(MSB)							(LSB)	
		TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2	CP/RL2	
Symbol	Positi	on Nar	ne and Sigi	nificance						
TF2	T2CO		er 2 overflov en either RC			overflow and	d must be cl	eared by so	oftware. TF2	will not be set
EXF2	T2CO	EXI		en Timer 2	interrupt is	enabled, EX	F2 = 1  will  0	,	0	ition on T2EX and r to the Timer 2
RCLK	T2CO								ow pulses fo eceive clock	r its receive clock
TCLK	T2CO								ow pulses fo	or its transmit clock k.
EXEN2	T2CO	tran		EX if Timer						of a negative ses Timer 2 to
TR2	T2COI	N.2 Sta	rt/stop contr	ol for Timer	2. A logic 1	starts the ti	mer.			
C/T2	T2CO	N.1 Tim	Timer or counter select. (Timer 2)  0 = Internal timer (OSC/12)  1 = External event counter (falling edge triggered).							
CP/RL2	T2COI	clea 2 = 1. Whe	Capture/Reload flag. When set, captures will occur on negative transitions at T2EX if EXEN2 = 1. V cleared, auto-reloads will occur either with Timer 2 overflows or negative transitions at T2EX when Vhen either RCLK = 1 or TCLK = 1, this bit is ignored and the timer is forced to auto-reload on Timer 2 overflow.							T2EX when

Figure 1. Timer/Counter 2 (T2CON) Control Register

Table 3. Timer 2 Operating Modes

RCLK + TCLK	CP/RL2	TR2	MODE
0	0	1	16-bit Auto-reload
0	1	1	16-bit Capture
1	X	1	Baud rate generator
X	Х	0	(off)

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

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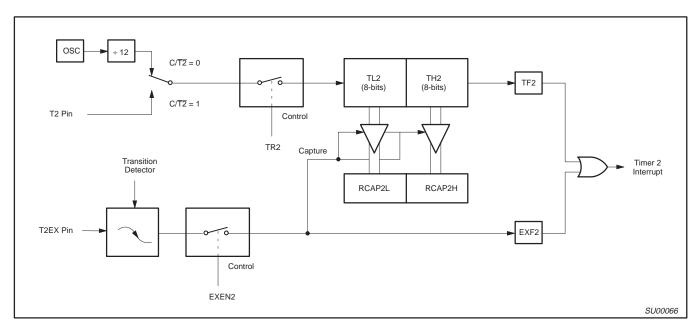


Figure 2. Timer 2 in Capture Mode

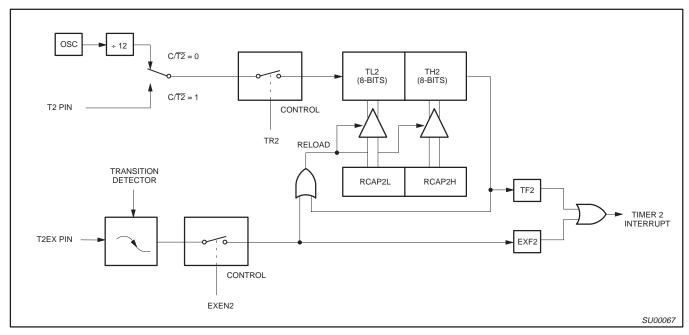


Figure 3. Timer 2 in Auto-Reload Mode

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# 89C535/89C536/89C538

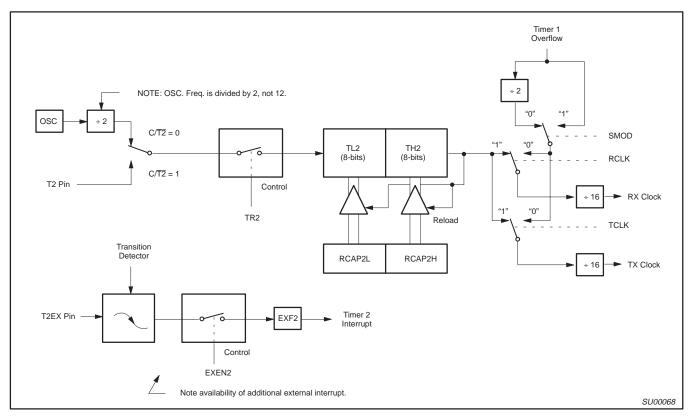


Figure 4. Timer 2 in Baud Rate Generator Mode

Table 4. Timer 2 Generated Commonly Used Baud Rates

Baud Rate	Osc Freq	Time	er 2
Baud Nate	OSC Freq	RCAP2H	RCAP2L
375K	12MHz	FF	FF
9.6K	12MHz	FF	D9
2.8K	12MHz	FF	B2
2.4K	12MHz	FF	64
1.2K	12MHz	FE	C8
300	12MHz	FB	1E
110	12MHz	F2	AF
300	6MHz	FD	8F
110	6MHz	F9	57

### **Baud Rate Generator Mode**

Bits TCLK and/or RCLK in T2CON (Table 3) allow the serial port transmit and receive baud rates to be derived from either Timer 1 or Timer 2. When TCLK= 0, Timer 1 is used as the serial port transmit baud rate generator. When TCLK= 1, Timer 2 is used as the serial port transmit baud rate generator. RCLK has the same effect for the serial port receive baud rate. With these two bits, the serial port can have different receive and transmit baud rates – one generated by Timer 1, the other by Timer 2.

Figure 4 shows the Timer 2 in baud rate generation mode. The baud rate generation mode is like the auto-reload mode,in that a rollover in TH2 causes the Timer 2 registers to be reloaded with the 16-bit value in registers RCAP2H and RCAP2L, which are preset by software.

The baud rates in modes 1 and 3 are determined by Timer 2's overflow rate given below:

Modes 1 and 3 Baud Rates = 
$$\frac{\text{Timer 2 Overflow Rate}}{16}$$

The timer can be configured for either "timer" or "counter" operation. In many applications, it is configured for "timer" operation (C/T2\*=0). Timer operation is different for Timer 2 when it is being used as a baud rate generator.

Usually, as a timer it would increment every machine cycle (i.e., 1/12 the oscillator frequency). As a baud rate generator, it increments every state time (i.e., 1/2 the oscillator frequency). Thus the baud rate formula is as follows:

Modes 1 and 3 Baud Rates =

 $\frac{\text{Oscillator Frequency}}{[32 \times [65536 - (RCAP2H, RCAP2L)]]}$ 

Where: (RCAP2H, RCAP2L)= The content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.

The Timer 2 as a baud rate generator mode shown in Figure 4, is valid only if RCLK and/or TCLK = 1 in T2CON register. Note that a rollover in TH2 does not set TF2, and will not generate an interrupt. Thus, the Timer 2 interrupt does not have to be disabled when Timer 2 is in the baud rate generator mode. Also if the EXEN2 (T2 external enable flag) is set, a 1-to-0 transition in T2EX (Timer/counter 2 trigger input) will set EXF2 (T2 external flag) but will not cause a reload from (RCAP2H, RCAP2L) to (TH2,TL2). Therefore when Timer 2 is in use as a baud rate generator, T2EX can be used as an additional external interrupt, if needed.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

When Timer 2 is in the baud rate generator mode, one should not try to read or write TH2 and TL2. As a baud rate generator, Timer 2 is incremented every state time (osc/2) or asynchronously from pin T2; under these conditions, a read or write of TH2 or TL2 may not be accurate. The RCAP2 registers may be read, but should not be written to, because a write might overlap a reload and cause write and/or reload errors. The timer should be turned off (clear TR2) before accessing the Timer 2 or RCAP2 registers.

Table 4 shows commonly used baud rates and how they can be obtained from Timer 2.

### **Summary Of Baud Rate Equations**

Timer 2 is in baud rate generating mode. If Timer 2 is being clocked through pin T2(P1.0) the baud rate is:

Baud Rate = 
$$\frac{\text{Timer 2 Overflow Rate}}{16}$$

If Timer 2 is being clocked internally, the baud rate is:

Baud Rate = 
$$\frac{f_{OSC}}{[32 \times [65536 - (RCAP2H, RCAP2L)]]}$$

Where f<sub>OSC</sub>= Oscillator Frequency

To obtain the reload value for RCAP2H and RCAP2L, the above equation can be rewritten as:

RCAP2H, RCAP2L = 
$$65536 - \left(\frac{f_{OSC}}{32 \times Baud \ Rate}\right)$$

### **Timer/Counter 2 Set-up**

Except for the baud rate generator mode, the values given for T2CON do not include the setting of the TR2 bit. Therefore, bit TR2 must be set, separately, to turn the timer on. see Table 5 for set-up of Timer 2 as a timer. Also see Table 6 for set-up of Timer 2 as a counter.

Table 5. Timer 2 as a Timer

	T2CON				
MODE	INTERNAL CONTROL (Note 1)	EXTERNAL CONTROL (Note 2)			
16-bit Auto-Reload	00H	08H			
16-bit Capture	01H	09H			
Baud rate generator receive and transmit same baud rate	34H	36H			
Receive only	24H	26H			
Transmit only	14H	16H			

### Table 6. Timer 2 as a Counter

	ТМ	OD	
MODE	INTERNAL CONTROL (Note 1)	EXTERNAL CONTROL (Note 2)	
16-bit	02H	0AH	
Auto-Reload	03H	0BH	

#### NOTES:

- 1. Capture/reload occurs only on timer/counter overflow.
- 2. Capture/reload occurs on timer/counter overflow and a 1-to-0 transition on T2EX (P1.1) pin except when Timer 2 is used in the baud rate generator mode.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### Serial Interface

The 89C538/536 has a standard 80C51 serial port. This serial port can operate in 4 modes:

Mode 0: Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted/received (LSB first). The baud rate is fixed at 1/12 the oscillator frequency.

Mode 1: 10 bits are transmitted (through TxD) or received (through RxD): a start bit (0), 8 data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in Special Function Register SCON. The baud rate is variable.

Mode 2: 11 bits are transmitted (throughTxD) or received (through RxD): start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). On Transmit, the 9th data bit (TB8 in SCON) can be assigned the value of 0 or 1, Or, for example, the parity bit (P, in the PSW) could be moved into TB8. On receive, the 9th data bit goes into RB8 in Special Function Register SCON, while the stop bit is ignored. The baud rate is programmable to either 1/32 or 1/64 the oscillator frequency.

Mode 3: 11 bits are transmitted (through TxD) or received (through jRxD): a start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). In fact, Mode 3 is the same as Mode 2 in all respects except baud rate. The baud rate in Mode 3 is variable.

In all four modes, transmission is initiated by any instruction that uses SBUF as a destination register. Reception is initiated in Mode 0 by the condition RI = 0 and REN = 1. Reception is initiated in the other modes by the incoming start bit if REN = 1.

### **Serial Port Control Register**

The serial port control and status register is the Special FunctionRegister SCON, shown in Figure 5. This register contains not only the mode selection bits, but also the 9th data bit for transmit and receive (TB8 and RB8), and the serial port interrupt bits (TI and RI).

Additional details of serial port operation may be found in the 80C51 Family Hardware Description found in the *Philips 80C51–Based 8–Bit Microcontroller Data Handbook, IC20.* 

	S	CON Add	lress = 98H						1	Reset Value = 0000 0000B
	Bit Add	ressable								
		SM0	SM1	SM2	REN	TB8	RB8	TI	RI	
	Bit:	7	6	5	4	3	2	1	0	
Symbol	Func	tion								
SM0	Serial	Port Mode	e Bit 0							
SM1	Serial	Port Mode	e Bit 1							
	SM0	SM1	Mode	Descr	iption	Baud Rate	**			
	0	0	0	shift re	egister	f <sub>OSC</sub> /12				
	0	1	1	8-bit L	JART	variable				
	1	0	2	9-bit L		f <sub>OSC</sub> /64 or	f <sub>OSC</sub> /32			
	1	1	3	9-bit L	JART	variable				
SM2	receiv In Mo	red 9th dat de 1, if SM	a bit (RB8) is	1, indica I will not b	ting an ad be activate	dress, and the dunless a va	e received	byte is a Gi	ven or Bro	ot be set unless the badcast Address. e received byte is a
REN	Enabl	es serial re	eception. Set	by softwa	are to enal	ble reception.	Clear by s	oftware to d	lisable red	eption.
TB8	The 9	th data bit	that will be tr	ansmitted	d in Modes	2 and 3. Set	or clear by	software a	s desired.	
RB8			3, the 9th dais not used.	ta bit that	was recei	ved. In Mode	1, if SM2 =	0, RB8 is t	he stop bi	t that was received.
TI						nd of the 8th le cleared by s		Node 0, or a	t the begi	nning of the stop bit in the
RI						nd of the 8th b see SM2). M				ough the stop bit time in
TE:										

Figure 5. SCON: Serial Port Control Register

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

### **Interrupt Priority Structure**

The 89C535/536/538 has a 6-source two-level interrupt structure (see Table 7). There are 2 SFRs associated with the interrupts on the 89C535/536/538. They are the IE and IP. (See Figures 6 and 7.)

The function of the IPH SFR is simple and when combined with the IP SFR determines the priority of each interrupt. The priority of each interrupt is determined as shown in the following table:

PRIORITY BITS	INTERRUPT PRIORITY I EVE			
IP.x	INTERRUPT PRIORITY LEVEL			
0	Level 0 (lowest priority)			
1	Level 1 (highest priority)			

An interrupt will be serviced as long as an interrupt of equal or higher priority is not already being serviced. If an interrupt of equal or higher level priority is being serviced, the new interrupt will wait until it is finished before being serviced. If a lower priority level interrupt is being serviced, it will be stopped and the new interrupt serviced. When the new interrupt is finished, the lower priority level interrupt that was stopped will be completed.

Table 7. Interrupt Table

SOURCE	POLLING PRIORITY	REQUEST BITS	HARDWARE CLEAR?	VECTOR ADDRESS
X0	1	IE0	N (L) <sup>1</sup> Y (T) <sup>2</sup>	03H
T0	2	TP0	Y	0BH
X1	3	IE1	N (L) Y (T)	13H
T1	4	TF1	Υ	1BH
SP	5	R1, TI	N	23H
T2	6	TF2, EXF2	N	2BH

### NOTES:

- 1. L = Level activated
- 2. T = Transition activated

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

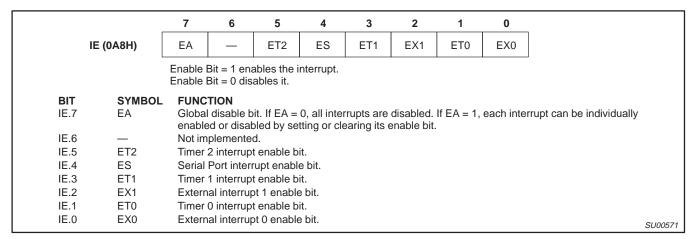


Figure 6. IE Registers

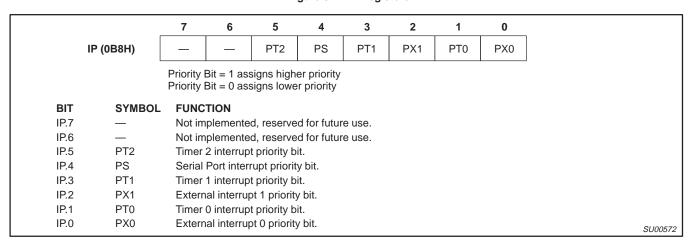


Figure 7. IP Registers

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

### **Expanded Data RAM Addressing**

The 89C535/536/538 has internal data memory that is mapped into four separate segments: the lower 128 bytes of RAM, upper 128 bytes of RAM, 128 bytes Special Function Register (SFR), and 256 bytes expanded RAM (ERAM).

The four segments are:

- The Lower 128 bytes of RAM (addresses 00H to 7FH) are directly and indirectly addressable.
- The Upper 128 bytes of RAM (addresses 80H to FFH) are indirectly addressable only.
- The Special Function Registers, SFRs, (addresses 80H to FFH) are directly addressable only.
- The 256-bytes expanded RAM (ERAM, 00H FFH) are indirectly accessed by move external instruction, MOVX.

The Lower 128 bytes can be accessed by either direct or indirect addressing. The Upper 128 bytes can be accessed by indirect addressing only. The Upper 128 bytes occupy the same address space as the SFRs. That means they have the same address, but are physically separate from SFR space.

When an instruction accesses an internal location above address 7FH, the CPU knows whether the access is to the upper 128 bytes of data RAM or to SFR space by the addressing mode used in the instruction. Instructions that use direct addressing access SFR space. For example:

### MOV 0A0H,#data

accesses the SFR at location 0A0H (which is P2). Instructions that use indirect addressing access the Upper 128 bytes of data RAM.

For example:

#### MOV @R0,#data

where R0 contains 0A0H, accesses the data byte at address 0A0H, rather than P2 (whose address is 0A0H).

The ERAM can be accessed by indirect addressing and MOVX instructions. This part of memory is physically located on-chip, logically occupies the first 256-bytes of external data memory.

The ERAM is indirectly addressed, using the MOVX instruction in combination with any of the registers R0, R1 of the selected bank or DPTR. An access to ERAM will not affect ports P0, P3.6 (WR#) and P3.7 (RD#). P2 SFR is output during external addressing. For example,

#### MOVX @R0,#data

where R0 contains 0A0H, accesses the ERAM at address 0A0H rather than external memory. An access to external data memory locations higher than FFH (i.e., 0100H to FFFFH) will be performed with the MOVX DPTR instructions in the same way as in the standard 80C51, so with P0 and P2 as data/address bus, and P3.6 and P3.7 as write and read timing signals. Refer to Figure 8.

External data memory cannot be accessed using the MOVX with R0 or R1. This will always access the ERAM.

The stack pointer (SP) may be located anywhere in the 256 bytes RAM (lower and upper RAM) internal data memory. The stack may not be located in the ERAM.

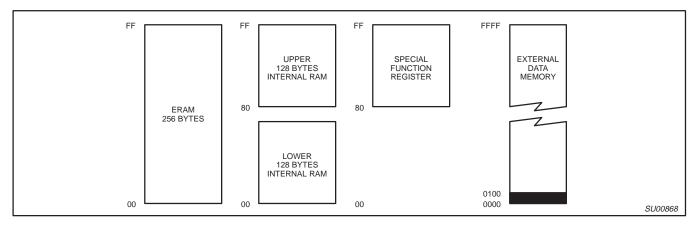


Figure 8. Internal and External Data Memory Address Space

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

# **ABSOLUTE MAXIMUM RATINGS**1, 2, 3

PARAMETER	RATING	UNIT
Operating temperature under bias	0 to +70	°C
Storage temperature range	-65 to +150	°C
Voltage on EA/V <sub>PP</sub> pin to V <sub>SS</sub>	0 to +13.0	V
Voltage on any other pin to V <sub>SS</sub>	-0.5 to +6.5	V
Maximum I <sub>OL</sub> per I/O pin	15	mA
Power dissipation (based on package heat transfer limitations, not device power consumption)	1	W

#### NOTES:

1997 Jun 05 16

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Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any conditions other than those described in the AC and DC Electrical Characteristics section of this specification is not implied.

This product includes circuitry specifically designed for the protection of its internal devices from the damaging effects of excessive static charge. Nonetheless, it is suggested that conventional precautions be taken to avoid applying greater than the rated maxima.
 Parameters are valid over operating temperature range unless otherwise specified. All voltages are with respect to V<sub>SS</sub> unless otherwise

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### DC ELECTRICAL CHARACTERISTICS

 $T_{amb} = 0^{\circ}C \text{ to } +70^{\circ}C; 5V \pm 10\%; V_{SS} = 0V$ 

OVMDOL	DADAMETED	TEST	LIM	IITS	
SYMBOL	PARAMETER	CONDITIONS	MIN	MAX	UNIT
V <sub>IL</sub>	Input low voltage	4.5V < V <sub>CC</sub> < 5.5V	-0.5	0.2V <sub>CC</sub> -0.1	V
V <sub>IH</sub>	Input high voltage (ports 0, 1, 2, 3, EA)		0.2V <sub>CC</sub> +0.9	V <sub>CC</sub> +0.5	V
V <sub>IH1</sub>	Input high voltage, XTAL1, RST		0.7V <sub>CC</sub>	V <sub>CC</sub> +0.5	V
V <sub>OL</sub>	Output low voltage, ports 1, 2, 3 <sup>6</sup>	$V_{CC} = 4.5V$ $I_{OL} = 1.6 \text{mA}^1$		0.4	V
V <sub>OL1</sub>	Output low voltage, port 0, ALE, PSEN 5, 6	$V_{CC} = 4.5V$ $I_{OL} = 3.2 \text{mA}^{1}$		0.4	V
V <sub>OH</sub>	Output high voltage, ports 1, 2, 3 <sup>2</sup>	$V_{CC} = 4.5V$ $I_{OH} = -30\mu A$	V <sub>CC</sub> - 0.7		V
V <sub>OH1</sub>	Output high voltage (port 0 in external bus mode), ALE <sup>7</sup> , PSEN <sup>2</sup>	$V_{CC} = 4.5V$ $I_{OH} = -800\mu A$	V <sub>CC</sub> - 0.7		V
I <sub>IL</sub>	Logical 0 input current, ports 1, 2, 3	$V_{IN} = 0.4V$	-1	-50	μΑ
I <sub>TL</sub>	Logical 1-to-0 transition current, ports 1, 2, 3	V <sub>IN</sub> = 2.0V See note 3		-650	μΑ
ILI	Input leakage current, port 0	$0.45 < V_{IN} < V_{CC} - 0.3$		±10	μΑ
Icc	Power supply current (see Figure 16): Active mode Idle mode Power-down mode or clock stopped (see Figure 20 for conditions)	See note 4 $V_{CC} = 5.5V$ FREQ = 24  MHz $T_{amb} = 0^{\circ}\text{C to } 70^{\circ}\text{C}$		60 25 100	mA mA μA
R <sub>RST</sub>	Internal reset pull-down resistor		40	225	kΩ

### NOTES:

- Capacitive loading on ports 0 and 2 may cause spurious noise to be superimposed on the V<sub>OL</sub>s of ALE and ports 1 and 3. The noise is due to external bus capacitance discharging into the port 0 and port 2 pins when these pins make 1-to-0 transitions during bus operations. In the worst cases (capacitive loading > 100pF), the noise pulse on the ALE pin may exceed 0.8V. In such cases, it may be desirable to qualify ALE with a Schmitt Trigger, or use an address latch with a Schmitt Trigger STROBE input. IOL can exceed these conditions provided that no single output sinks more than 5mA and no more than two outputs exceed the test conditions.
- 2. Capacitive loading on ports 0 and 2 may cause the V<sub>OH</sub> on ALE and PSEN to momentarily fall below the V<sub>CC</sub>-0.7 specification when the address bits are stabilizing.
- Pins of ports 1, 2 and 3 source a transition current when they are being externally driven from 1 to 0. The transition current reaches its maximum value when V<sub>IN</sub> is approximately 2V.

- See Figures 17 through 20 for I<sub>CC</sub> test conditions.
   Active mode: I<sub>CC(MAX)</sub> = 0.9 × FREQ. + 1.1mA
   Idle mode: I<sub>CC(MAX)</sub> = 0.18 × FREQ. + 1.0mA; See Figure 16.
   Load capacitance for port 0, ALE, and PSEN = 100pF, load capacitance for all other outputs = 80pF.
- 6. Under steady state (non-transient) conditions, I<sub>OL</sub> must be externally limited as follows:

Maximum IOL per port pin: 15mA Maximum I<sub>OL</sub> per 8-bit port: 26mA Maximum total I<sub>OL</sub> for all outputs: 71mA

- If IoI exceeds the test condition, Vol may exceed the related specification. Pins are not guaranteed to sink current greater than the listed
- 7. ALE is tested to V<sub>OH1</sub>, except when ALE is off then V<sub>OH</sub> is the voltage specification.
- 8. Pin capacitance is characterized but not tested. Pin capacitance is less than 25pF. Pin capacitance of ceramic package is less than 15pF (except EA is 25pF).

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

# AC ELECTRICAL CHARACTERISTICS Tamb = 0°C to +70°C, $V_{CC}$ = 5V ±10%, $V_{SS}$ = 0V<sup>1, 2, 3</sup>

			VARIABL	E CLOCK	33MHz		
SYMBOL	FIGURE	PARAMETER	MIN	MAX	MIN	MAX	דואט
1/t <sub>CLCL</sub>	9	Oscillator frequency	3.5	33			
		Speed versions: N (33MHz)			3.5	33	MHz
<b>t.</b>	9	ALE pulse width	2t <sub>CLCL</sub> -40		21	- 55	ns
t <sub>LHLL</sub>	9	Address valid to ALE low			5	-	
t <sub>AVLL</sub>	-	Address hold after ALE low	t <sub>CLCL</sub> -25		5		ns
t <sub>LLAX</sub>	9		t <sub>CLCL</sub> -25	44 05	5		ns
t <sub>LLIV</sub>	9	ALE low to valid instruction in		4t <sub>CLCL</sub> -65		55	ns
t <sub>LLPL</sub>	9	ALE low to PSEN low	t <sub>CLCL</sub> -25		5	-	ns
t <sub>PLPH</sub>	9	PSEN pulse width	3t <sub>CLCL</sub> -45	0. 00	45	00	ns
t <sub>PLIV</sub>	9	PSEN low to valid instruction in		3t <sub>CLCL</sub> -60		30	ns
t <sub>PXIX</sub>	9	Input instruction hold after PSEN	0		0		ns
t <sub>PXIZ</sub>	9	Input instruction float after PSEN		t <sub>CLCL</sub> -25		5	ns
t <sub>AVIV</sub>	9	Address to valid instruction in		5t <sub>CLCL</sub> -80		70	ns
t <sub>PLAZ</sub>	9	PSEN low to address float		10		10	ns
Data Mem	ory			_			
t <sub>RLRH</sub>	10, 11	RD pulse width	6t <sub>CLCL</sub> -100		82		ns
$t_{WLWH}$	10, 11	WR pulse width	6t <sub>CLCL</sub> -100		82		ns
t <sub>RLDV</sub>	10, 11	RD low to valid data in		5t <sub>CLCL</sub> -90		60	ns
t <sub>RHDX</sub>	10, 11	Data hold after RD	0		0		ns
t <sub>RHDZ</sub>	10, 11	Data float after RD		2t <sub>CLCL</sub> -28		32	ns
t <sub>LLDV</sub>	10, 11	ALE low to valid data in		8t <sub>CLCL</sub> -150		90	ns
t <sub>AVDV</sub>	10, 11	Address to valid data in		9t <sub>CLCL</sub> -165		105	ns
t <sub>LLWL</sub>	10, 11	ALE low to RD or WR low	3t <sub>CLCL</sub> -50	3t <sub>CLCL</sub> +50	40	140	ns
t <sub>AVWL</sub>	10, 11	Address valid to WR low or RD low	4t <sub>CLCL</sub> -75		45		ns
t <sub>QVWX</sub>	10, 11	Data valid to WR transition	t <sub>CLCL</sub> -30		0		ns
t <sub>WHQX</sub>	10, 11	Data hold after WR	t <sub>CLCL</sub> -25		5		ns
t <sub>QVWH</sub>	11	Data valid to WR high	7t <sub>CLCL</sub> -130		80		ns
t <sub>RLAZ</sub>	10, 11	RD low to address float	GLGE	0		0	ns
t <sub>WHLH</sub>	10, 11	RD or WR high to ALE high	t <sub>CLCL</sub> -25	t <sub>CLCL</sub> +25	5	55	ns
External C		3 4 3	0202	OLOL			
t <sub>CHCX</sub>	13	High time	17	t <sub>CLCL</sub> -t <sub>CLCX</sub>			ns
t <sub>CLCX</sub>	13	Low time	17	t <sub>CLCL</sub> -t <sub>CHCX</sub>			ns
	13	Rise time		5	<del>                                     </del>	<del>                                     </del>	ns
t <sub>CLCH</sub>	13	Fall time		5			ns
t <sub>CHCL</sub> Shift Regi	<u> </u>	i dii diilo	<u> </u>				113
	12	Serial port clock cycle time	12+		360		ns
t	12	Output data setup to clock rising edge	12t <sub>CLCL</sub>		167		_
t <sub>QVXH</sub>		1 0 0	10t <sub>CLCL</sub> -133				ns
t <sub>XHQX</sub>	12	Output data hold after clock rising edge	2t <sub>CLCL</sub> -80		50	-	ns
t <sub>XHDX</sub>	12	Input data hold after clock rising edge	0	10: 155	0		ns
t <sub>XHDV</sub>	12	Clock rising edge to input data valid		10t <sub>CLCL</sub> -133		167	ns

### NOTES:

Parameters are valid over operating temperature range unless otherwise specified.
 Load capacitance for port 0, ALE, and PSEN = 100pF, load capacitance for all other outputs = 80pF.

<sup>3.</sup> Interfacing the microcontroller to devices with float times up to 45ns is permitted. This limited bus contention will not cause damage to Port 0 drivers.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### **EXPLANATION OF THE AC SYMBOLS**

Each timing symbol has five characters. The first character is always 't' (= time). The other characters, depending on their positions, indicate the name of a signal or the logical status of that signal. The designations are:

A - Address

C - Clock

D - Input data

H - Logic level high

I – Instruction (program memory contents)

L - Logic level low, or ALE

P - PSEN

Q - Output data

 $R - \overline{RD}$  signal

t - Time

V - Valid

W- WR signal

X - No longer a valid logic level

Z - Float

**Examples:** t<sub>AVLL</sub> = Time for address valid to ALE low.

 $t_{LLPL}$  =Time for ALE low to  $\overline{PSEN}$  low.

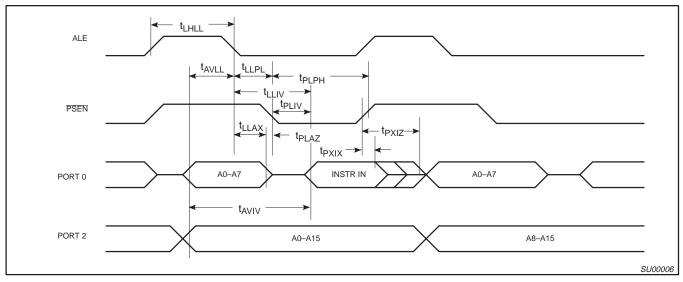


Figure 9. External Program Memory Read Cycle

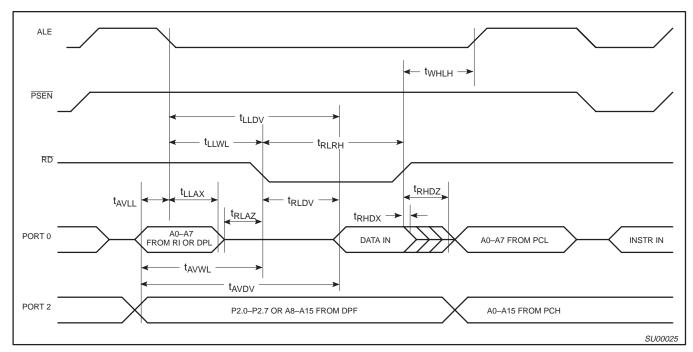


Figure 10. External Data Memory Read Cycle

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

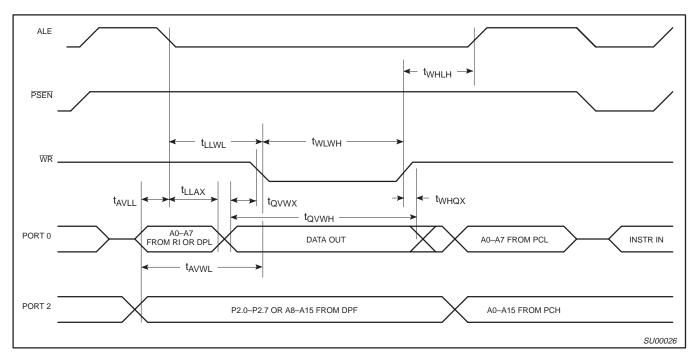


Figure 11. External Data Memory Write Cycle

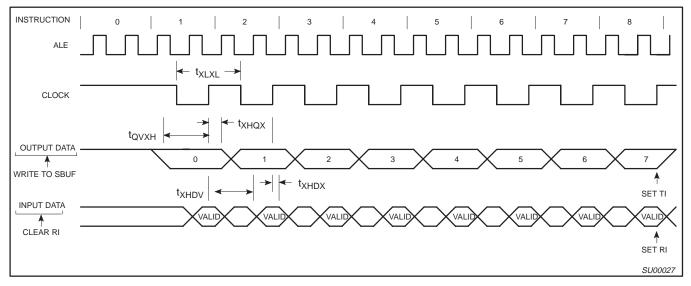


Figure 12. Shift Register Mode Timing

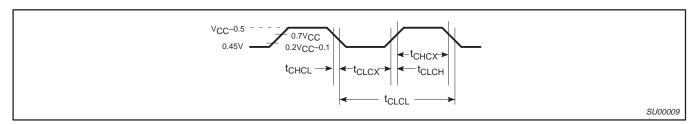
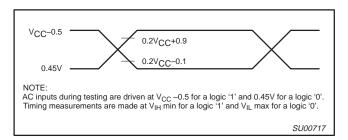


Figure 13. External Clock Drive

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538



VLOAD VLOAD+0.1V REFERENCE POINTS VOH-0.1V NOTE: For timing purposes, a port is no longer floating when a 100mV change from load voltage occurs, and begins to float when a 100mV change from the loaded VOH/VoL level occurs.  $I_{OH}/I_{OL} \ge \pm 20$ mA.

Figure 14. AC Testing Input/Output

Figure 15. Float Waveform

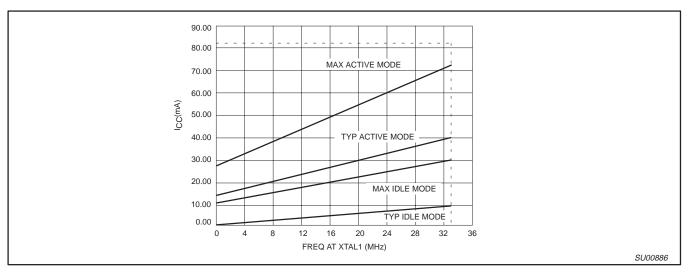


Figure 16. I<sub>CC</sub> vs. FREQ Valid only within frequency specifications of the device under test

1997 Jun 05 21

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# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

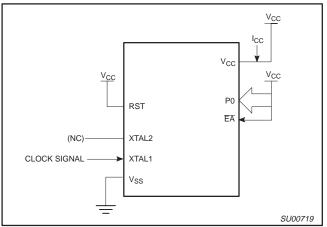


Figure 17. I<sub>CC</sub> Test Condition, Active Mode
All other pins are disconnected

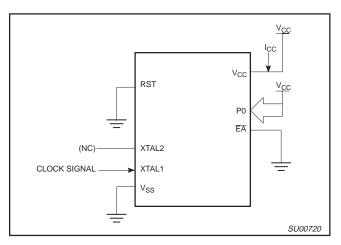


Figure 18. I<sub>CC</sub> Test Condition, Idle Mode All other pins are disconnected

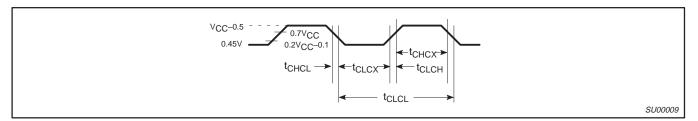


Figure 19. Clock Signal Waveform for  $I_{CC}$  Tests in Active and Idle Modes  $t_{CLCH} = t_{CHCL} = 5 ns$ 

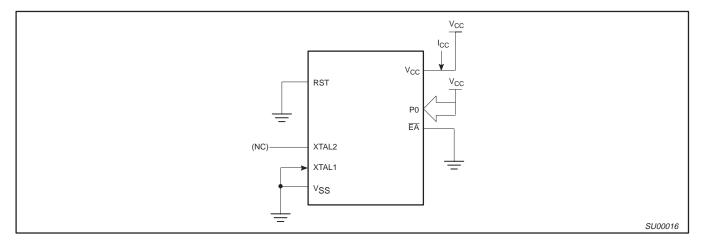


Figure 20.  $I_{CC}$  Test Condition, Power Down Mode All other pins are disconnected.  $V_{CC}$  = 2V to 5.5V

1997 Jun 05 22

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# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### FLASH EPROM PROGRAM MEMORY

#### **FEATURES**

- 8K (89C535), 16K (89C536), 64K (89C538) or electrically erasable internal program.
- Up to 64 Kilobyte external program memory if the internal program memory is switched off (EA = 0)..
- Programming and erasing voltage 12V ±5%
- Command register architecture
  - Byte Programming (10 us typical)
  - Auto chip erase 5 seconds typical (including preprogramming time)
- Auto Erase and auto program
  - DATA polling
  - Toggle bit
- 100 minimum erase/program cycles
- Advanced CMOS FLASH EPROM memory technology

#### **GENERAL DESCRIPTION**

The 89C535/536/538 FLASH EPROM memory augments EPROM functionality with In–circuit electrical erasure and programming. The 89C535/536/538 uses a command register to manage this functionality.

The FLASH EPROM reliably stores memory contents even after 100 erase and program cycles. The cell is designed to optimize the erase and programming mechanisms. In addition, the combination of advanced tunnel oxide processing and low internal electric fields for erase and programming operations produces reliable cycling. The 89C535/536/538 uses a 12.0V  $\pm 5\% V_{PP}$  supply to perform the Auto Program/Erase algorithms.

### **Automatic Programming**

The 89C535/536/538 is byte programmable using the Automatic Programming algorithm. The Automatic Programming algorithm

does not require the system to time out or verify the data programmed. The typical room temperature chip programming time of the 89C535/536/538 is less than 5 seconds.

### **Automatic Chip Erase**

The device may be erased using the automatic Erase algorithm. The automatic Erase algorithm automatically programs the entire array prior to electrical erase. The timing and verification of electrical erase are controlled internal to the device.

### **Automatic Programming Algorithm**

The 89C535/536/538 automatic Programming algorithm requires the user to only write a program set—up command and a program command (program data and address). The device automatically times the programming pulse width, provides the program verify, and counts the number of sequences. A status bit similar to  $\overline{\text{DATA}}$  polling and a status bit toggling between consecutive read cycles, provide feedback to the user as to the status of the programming operation.

#### **AUTOMATIC ERASE ALGORITHM**

The 89C535/536/538 Automatic Erase algorithm requires the user to only write an erase set–up command and erase command. The device will automatically pre–program and verify the entire array. Then the device automatically times the erase pulse width, provides the erase verify, and counts the number of sequences. A status bit similar to DATA polling and a status bit toggling between consecutive read cycles, provide feedback to the user as to the status of the erase operation.

Commands are written to the command register. Register contents serve as inputs to an internal state—machine which controls the erase and programming circuitry. During write cycles, the command register internally latches address and data needed for the programming and erase operations. For system design simplification, the 89C535/536/538 is designed to support either  $\overline{\text{WE}}$  or  $\overline{\text{CE}}$  controlled writes. During a system write cycle, addresses are latched on the falling edge of  $\overline{\text{WE}}$  or  $\overline{\text{CE}}$ , whichever occurs last. Data is latched on the rising edge of  $\overline{\text{WE}}$  or  $\overline{\text{CE}}$ , whichever occurs first. To simplify the following discussion, the  $\overline{\text{WE}}$  pin is used as the write cycle control pin through the rest of this text. All setup and hold times are with respect to the  $\overline{\text{WE}}$  signal.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

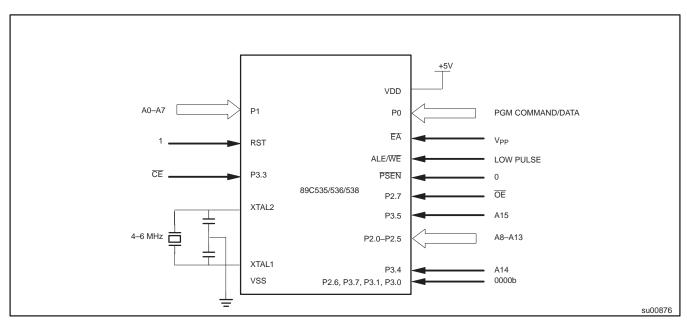


Figure 21. Erase/Programming/Verification

## Table 8. Pin Description

PIN NAME	SYMBOL	FUNCTION
P1.0–P1.7	A0-A7	Input Low Order Address Bits
P2.0-P2.5, P3.4, P3.5	A8-A13, A14-A15	Input High Order Address Bits
P0.0-P0.7	Q0-Q7	Data Input/Output
P3.3	CE	Chip Enable Input
P2.7	ŌĒ	Output Enable Input
ALE/WE	WE	Write Enable Pin
EA	V <sub>PP</sub>	Program Supply Voltage
P2.6, P3.7, P3.1, P3.0	FTEST3-FTEST0	Flash Test Mode Selection
Vcc	V <sub>CC</sub>	Power Supply Voltage (+5V)
GND	GND	Ground Pin

### **Table 9. Command Definitions**

		FIF	ST BU	S CYCLE	SEC	OND BU	S CYCLE
COMMAND	BUS CYCLES	OPERATION	ON ADE	DRESS DATA	OPERATION	ON ADDI	RESS DATA
Setup auto erase/auto erase (chip)	2	Write	Х	30H	Write	Χ	30H
Setup auto program/program	2	Write	Х	40H	Write	PA	PD
Reset	2	Write	Χ	FFH	Write	Х	FFH

#### Note:

- PA = Address of memory location to be programmed
- PD = Data to be programmed at location

### **Command Definitions**

When low voltage is applied to the  $V_{PP}$  pin, the contents of the command register default to 00H. Placing high voltage on the  $V_{PP}$  pin enables read/write operations. Device operations are selected by writing specific data patterns into the command register. Table 2 defines these 89C535/536/538 register commands. Table 3 defines the bus operations of 89C535/536/538.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### Table 10.

OPERATION		V <sub>PP</sub> (1)	CE	ŌĒ	WE	D00-D07
READ/WRITE						
	Read(2)	$V_{PPH}$	V <sub>IL</sub>	$V_{IL}$	VI <sub>H</sub>	DATA OUT(3)
	Standby(4)	$V_{PPH}$	V <sub>IH</sub>	Х	Х	Tri-State
	Write	V <sub>PPH</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>IL</sub>	Data In(5)

#### NOTES:

- 1. V<sub>PPH</sub> is the programming voltage specified for the device.
- 2. Read operation with VPP = VPPH may access array data (if write command is preceded) or silicon ID codes.
- 3. With V<sub>PP</sub> at high voltage, the standby current equals I<sub>CC</sub>+I<sub>PP</sub> (standby).
- 4. Refer to Table 38 for valid Data-In during a write operation.
- 5. X can be V<sub>IL</sub> or V<sub>IH</sub>.

### Set-Up Automatic Chip Erase/Erase Commands

The automatic chip erase does not require the device to be entirely pre–programmed prior to executing the Automatic set–up erase command and automatic chip erase command. Upon executing the Automatic chip erase command, the device automatically will program and verify the entire memory for an all–zero data pattern. When the device is automatically verified to contain an all–zero pattern, a self–timed chip erase and verify begins. The erase and verify operations are complete when the data on DQ7 is"1" at which time the device returns to the standby mode. The system is not required to provide any control or timing during these operations.

When using the Automatic Chip Erase algorithm, note that the erase automatically terminates when adequate erase margin has been achieved for the memory array (no erase verify command is required). The margin voltages are internally generated in the same manner as when the standard erase verify command is used.

The Automatic set—up erase command is a command only operation that stages the device for automatic electrical erasure of all bytes in the array. Automatic set—up erase is performed by writing 30H to the command register.

To command automatic chip erase, the command 30H must be written again to the command register. The automatic chip erase begins on the rising edge of the WE and terminates when the data on DQ7 is "1" and the data on DQ6 stops toggling for two consecutive read cycles, at which time the device returns to the standby mode.

### Set-Up Automatic Program/Program Commands

The Automatic Set—up Program is a command—only operation that stages the devices for automatic programming. Automatic Set—up Program is performed by writing 40H to the command register.

Once the Automatic Set—up Program operation is performed, the next WE pulse causes a transition to an active programming operation. Addresses are internally latched on the falling edge of the WE pulse. Data is internally latched on the rising edge of the WE pulse. The rising edge of WE also begins the programming operation. The system is not required to provide further controls or timings. The device will automatically provide an adequate internally generated program pulse and verify margin. The automatic programming operation is completed when the data read on DQ6 stops toggling for two consecutive read cycles and the data on DQ7 and DQ6 are equivalent to data written to these two bits at which time the device returns to the Read mode (no program verify command is required; but data can be read out if OE is active low).

#### **Reset Command**

A reset command is provided as a means to safely abort the erase—or program—command sequences. Following either set—up command (erase or program) with two consecutive writes of FFH will safely abort the operation. Memory contents will not be altered. Should program—fail or erase—fail happen, two consecutive writes of FFH will reset the device to abort the operation. A valid command must then be written to place the device in the desired state.

### **Write Operation Status**

### Toggle Bit-DQ6

The 89C535/536/538 features a "Toggle Bit" as a method to indicate to the host system that the Auto Program/Erase algorithms are either in progress or completed.

While the Automatic Program or Erase algorithm is in progress, successive attempts to read data from the device will result in DQ6 toggling between one and zero. Once the Automatic Program or Erase algorithm is completed, DQ6 will stop toggling and valid data will be read. The toggle bit is valid after the rising edge of the second WE pulse of the two write pulse sequences.

### Data Polling-D07

The 89C535/536/538 also features DATA Polling as a method to indicate to the host system that the Automatic Program or Erase algorithms are either in progress or completed.

While the Automatic Programming algorithm is in operation an attempt to read the device will produce the complement data of the data last written to DQ7. Upon completion of the Automatic Program algorithm an attempt to read the device will produce the true data last written to DQ7. The Data Polling feature is valid after the rising edge of the second WE pulse of the two write pulse sequences.

While the Automatic Erase algorithm is in operation, DQ7 will read "0" until the erase operation is completed. Upon completion of the erase operation, the data on DQ7 will read "1". The DATA Polling feature is valid after the rising edge of the second WE pulse of two writes pulse sequences.

The DATA Polling feature is active during Automatic Program/Erase algorithms.

### **Write Operation**

The data to be programmed into Flash should be inverted when programming. In other words to program the value '00', 'FF' should be applied to port P0.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

### System Considerations

During the switch between active and standby conditions, transient current peaks are produced on the rising and falling edges of Chip Enable. The magnitude of these transient current peaks is

dependent on the output capacitance loading of the device. At a minimum, a 0.1uF ceramic capacitor (high frequency, low inherent inductance) should be used on each device between  $V_{\mbox{CC}}$  and GND, and between V<sub>PP</sub> and GND to minimize transient effects.

SYMBOL	PARAMETER	MIN	TYP	MAX	UNIT	CONDITION
C <sub>IN</sub>	$V_{PPH}$			14	PF	$VI_N = 0V$
C <sub>OUT</sub>	$V_{PPH}$			16	pF	V <sub>OUT</sub> = 0V

Command programming/Data programming/Erase Operation

### **DC CHARACTERISTICS**

 $T_{amb}$  = 0°C to 70°C,  $V_{CC}$  = 5V  $\pm$  10%,  $V_{PP}$  = 12.0V  $\pm$  5%

SYMBOL	PARAMETER	CONDITION	MIN	TYP	MAX	UNIT
ILI	Input Leakage Current	$V_{IN}$ = GND to $V_{CC}$			10	μΑ
I <sub>LO</sub>	Output Leakage Current	$V_{OUT} = GND \text{ to } V_{CC}$			10	μΑ
I <sub>SB1</sub>	Standby V <sub>CC</sub> Current	CE = VI <sub>H</sub>			1	mA
I <sub>SB2</sub>		$\overline{\text{CE}} = V_{\text{CC}} \pm 0.3 \text{ V}$		1	100	μΑ
I <sub>CC1</sub> (Read)	Operating V <sub>CC</sub> Current	I <sub>OUT</sub> = 0 mA, f=1 MHz			30	mA
I <sub>CC2</sub>		I <sub>OUT</sub> = 0 mA, F=11MHz			50	mA
I <sub>CC3</sub> (Program)		In Programming			50	mA
I <sub>CC4</sub> (Erase)		In Erase			50	mA
I <sub>CC5</sub> (Program Verify)		In Program Verify			50	mA
I <sub>CC6</sub> (Erase Verify)		In erase Verify			50	mA
I <sub>PP1</sub> (Read)	V <sub>PP</sub> Current	V <sub>PP</sub> =12.6 V			100	μΑ
I <sub>PP2</sub> (Program)		In Programming			50	mA
I <sub>PP3</sub> (Erase)		In Erase			50	mA
I <sub>PP4</sub> (Program Verify)		In Program Verify			50	mA
I <sub>PP5</sub> (Erase Verify)		In Erase Verify			50	mA
V <sub>IL</sub>	Input Voltage		-0.5 (Note 5)		0.2V <sub>PP</sub> - 0.3	V
V <sub>IH</sub>			2.4		V <sub>CC</sub> +0.3V	V
					(Note 6)	
V <sub>OL</sub>	Output Voltage Low	I <sub>OL</sub> =2.1mA			0.45	V
V <sub>OH</sub>	Output Voltage High	I <sub>OH</sub> =400uA	2.4			V

### NOTES:

- V<sub>CC</sub> must be applied before V<sub>PP</sub> and removed after V<sub>PP</sub>.
   V<sub>PP</sub> must not exceed 14V including overshoot.
- 3. An influence may be had upon device reliability if the device is installed or removed while V<sub>PP</sub>=12V.
- 4. Do not alter  $V_{PP}$  from  $V_{IL}$  to 12V or 12V to  $V_{IL}$  when  $\overline{CE} = V_{IL}$
- $V_{IL}$  min. = -0.5V for pulse width  $\leq$  20ns.
- 6. If  $V_{IH}$  is over the specified maximum value, programming operation cannot be guaranteed.
- 7. All currents are in RMS unless otherwise noted. (Sampled, not 100% tested.).

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

#### **AC CHARACTERISTICS**

 $T_{amb}$  = 0°C to 70°C,  $V_{CC}$  = 5V  $\pm$  10%,  $V_{PP}$  = 12V  $\pm$  5%

SYMBOL	PARAMETER	CONDITION	MIN	MAX	UNIT
$\tau_{VPS}$	V <sub>PP</sub> setup time		100		ns
τ <sub>OES</sub>	OE setup time		100		ns
τ <sub>CWC</sub>	Command programming cycles		150		ns
$\tau_{\sf CEP}$	WE programming pulse width		60		ns
τ <sub>EPH1</sub>	WE programming pulse width High		20		ns
τ <sub>CEPH2</sub>	WE programming pulse width High		100		ns
$\tau_{AS}$	Address setup time		0		ns
τ <sub>AH1</sub>	Address hold time for DATA Polling		0		ns
$\tau_{DS}$	DATA setup time		50		ns
$\tau_{DH}$	DATA hold time		10		ns
τ <sub>CESP</sub>	CE setup time before DATA polling/toggle bit		100		ns
τ <sub>CES</sub>	CE setup time		0		ns
τ <sub>CESC</sub>	CE setup time before command write		100		ns
$\tau_{VPH}$	V <sub>PP</sub> hold time		100		ns
$\tau_{DF}$	Output disable time (Note 2)			35	ns
$\tau_{DPA}$	DATA polling/toggle bit access time			150	ns
τ <sub>AETC</sub>	Total erase time in auto chip erase			5(TYP)	s
τ <sub>AVT</sub>	Total programming time in auto verify		15	300	μs

### NOTES:

- 1.  $\overline{\text{CE}}$  and  $\overline{\text{OE}}$  must be fixed high during  $V_{PP}$  transition from 5V to 12V or from 12V to 5V.
- 2.  $\tau_{DF}$  is defined as the time at which the output achieves the open circuit condition and data is no longer driven.

# **Timing Waveform**

### **Automatic Programming**

One byte of data is programmed. Verifying in fast algorithm and additional programming by external control are not required because these operations are executed automatically by an internal control

circuit. Programming completion can be verified by DATA polling and toggle bit checking after automatic verify starts. Device outputs DATA during programming and DATA after programming on Q7. Q0 to Q5(Q6 is for toggle bit; see toggle bit, DATA polling, timing waveform) are in high impedance.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

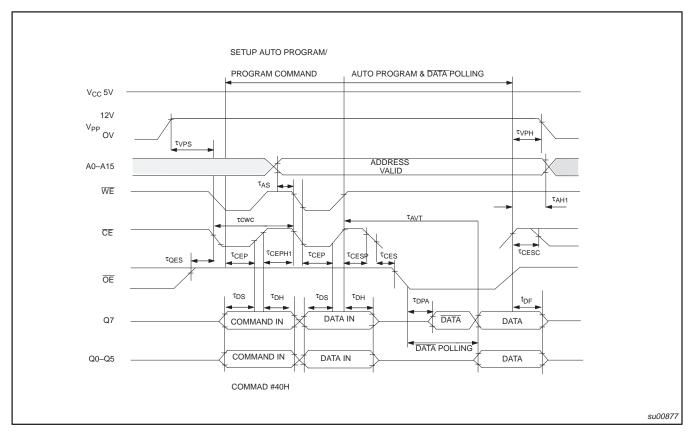


Figure 22. Automatic Programming Timing Waveform

### AUTOMATIC CHIP ERASE

All data in the FLASH memory is erased. External erase verification is not required. Erasure completion can be verified by DATA polling and toggle bit checking after automatic erase starts. Device outputs

0 during erasure and 1 after erasure on Q7, Q0 to Q5 (Q6 is for toggle bit; see toggle bit, DATA polling, timing waveform) are in high impedance.

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

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### RESET

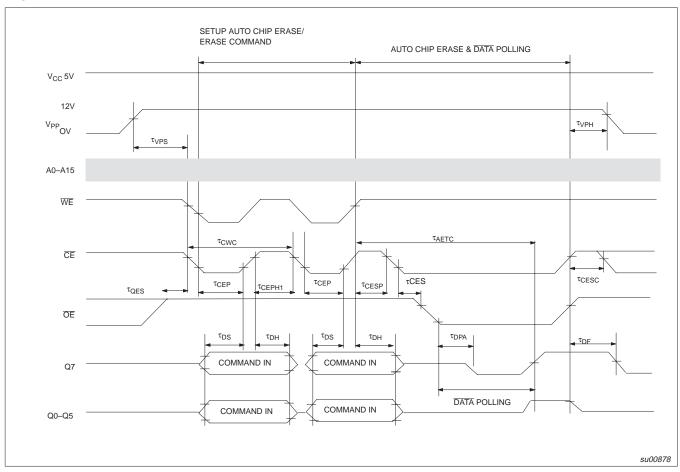


Figure 23. Automatic Chip Erase Timing Waveform

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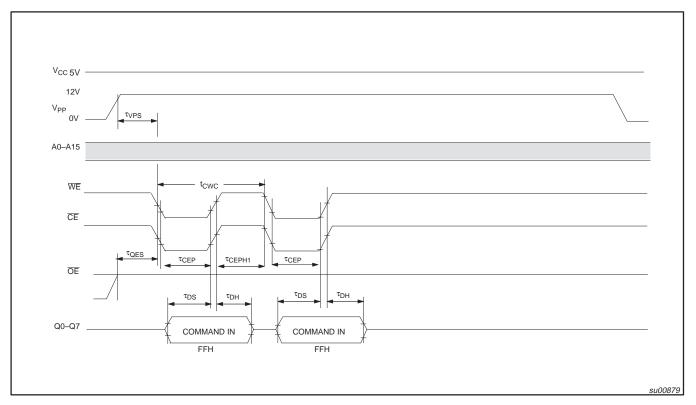


Figure 24. Reset Timing Waveform

# **Toggle Bit, Data Polling**

Toggle bit appears in Q6, when program/erase is operating. DATA polling appears in Q7 during programming or erase.

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# 89C535/89C536/89C538

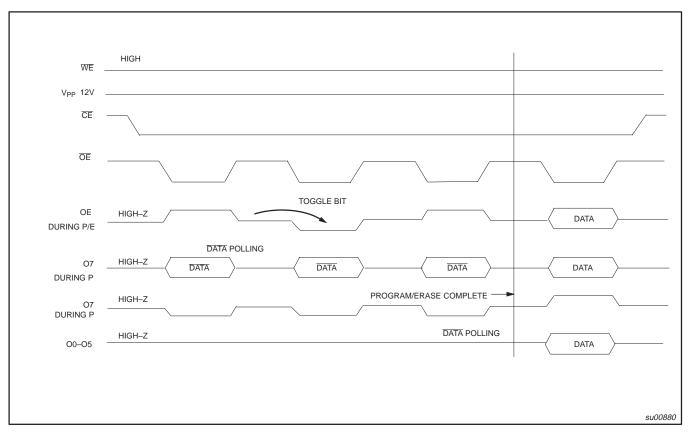


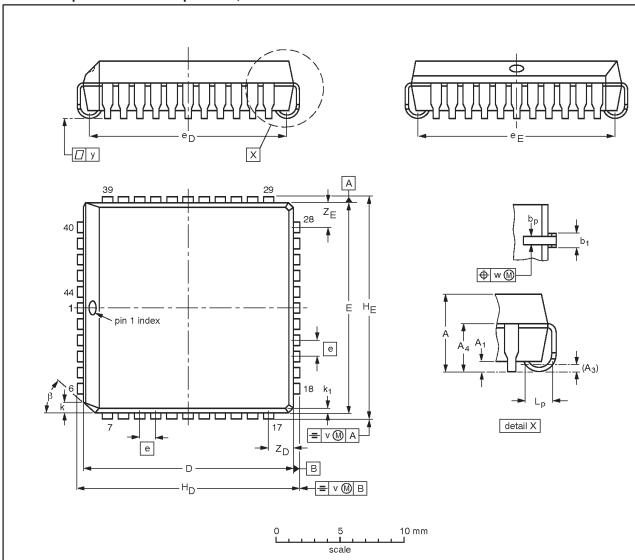
Figure 25. Toggle Bit, Data Polling Timing Waveform

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

## PLCC44: plastic leaded chip carrier; 44 leads

SOT187-2



#### DIMENSIONS (millimetre dimensions are derived from the original inch dimensions)

UNIT	Α	A <sub>1</sub> min.	A <sub>3</sub>	A <sub>4</sub> max.	bp	b <sub>1</sub>	D <sup>(1)</sup>	E <sup>(1)</sup>	е	e <sub>D</sub>	еE	H <sub>D</sub>	HE	k	k <sub>1</sub> max.	Lp	v	w	у	Z <sub>D</sub> <sup>(1)</sup> max.	Z <sub>E</sub> <sup>(1)</sup> max.	β
mm	4.57 4.19	0.51	0.25	3.05	0.53 0.33	0.81 0.66		16.66 16.51	1.27	16.00 14.99	16.00 14.99				0.51	1.44 1.02	0.18	0.18	0.10	2.16	2.16	45 <sup>0</sup>
inches	0.180 0.165	0.020	0.01					0.656 0.650	0.05	0.630 0.590	0.630 0.590	0.695 0.685	0.695 0.685	0.048 0.042	0.020	0.057 0.040	0.007	0.007	0.004	0.085	0.085	

#### Note

1. Plastic or metal protrusions of 0.01 inches maximum per side are not included.

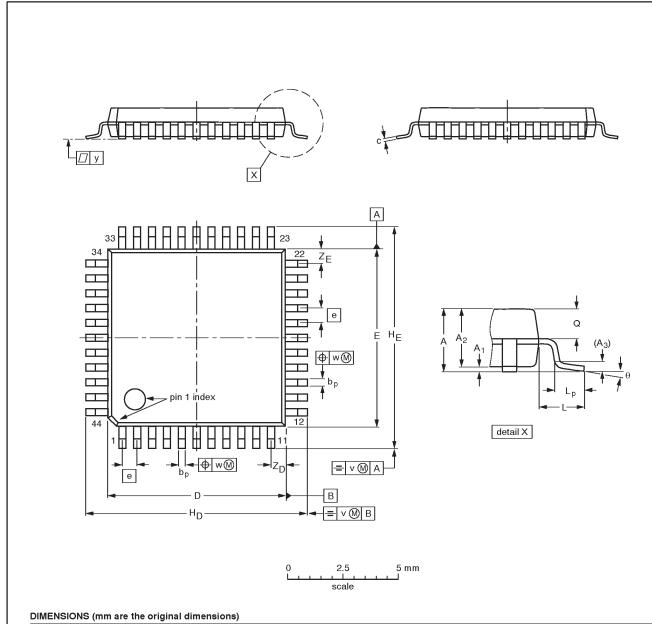
OUTLINE		REFER	EUROPEAN	ISSUE DATE	
VERSION	IEC	JEDEC	EIAJ	PROJECTION	ISSUE DATE
SOT187-2	112E10	MO-047AC			<del>92-11-17</del> 95-02-25

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

QFP44: plastic quad flat package; 44 leads (lead length 1.3 mm); body 10 x 10 x 1.75 mm

SOT307-2



UNIT	A max.	Α1	A <sub>2</sub>	<b>A</b> <sub>3</sub>	р <sub>р</sub>	O	D <sup>(1)</sup>	E <sup>(1)</sup>	е	H <sub>D</sub>	HE	L	Lp	Q	v	w	у	Z <sub>D</sub> <sup>(1)</sup>	Z <sub>E</sub> <sup>(1)</sup>	θ
mm	2.10	0.25 0.05	1.85 1.65	0.25	0.40 0.20	0.25 0.14	10.1 9.9	10.1 9.9	8.0	12.9 12.3	12.9 12.3	1.3	0.95 0.55	0.85 0.75	0.15	0.15	0.1	1.2 0.8	1.2 0.8	10° 0°

### Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE	REFERENCES				EUROPEAN	ISSUE DATE
VERSION	IEC	JEDEC	EIAJ		PROJECTION	ISSUE DATE
SOT307-2						<del>-92-11-17</del> 95-02-04

CMOS single-chip 8-bit microcontrollers with FLASH program memory

89C535/89C536/89C538

**NOTES** 

# CMOS single-chip 8-bit microcontrollers with FLASH program memory

# 89C535/89C536/89C538

DEFINITIONS					
Data Sheet Identification	Product Status	Definition			
Objective Specification	Formative or in Design	This data sheet contains the design target or goal specifications for product development. Specifications may change in any manner without notice.			
Preliminary Specification	Preproduction Product	This data sheet contains preliminary data, and supplementary data will be published at a later date. Philips Semiconductors reserves the right to make changes at any time without notice in order to improve design and supply the best possible product.			
Product Specification	Full Production	This data sheet contains Final Specifications. Philips Semiconductors reserves the right to make change at any time without notice, in order to improve design and supply the best possible product.			

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