TOSHIBA

TOSHIBA Original CMOS 4-Bit Microcontroller

TLCS-47 Series

TMP47C243NG
TMP47C443NG
TMP47C243MG
TMP47C443MG
TMP47C243DMG
TMP47C443DMG

TOSHIBA CORPORATION

Semiconductor Company

Document Change Notification

The purpose of this notification is to inform customers about the launch of the Pb-free version of the device. The introduction of a Pb-free replacement affects the datasheet. Please understand that this notification is intended as a temporary substitute for a revision of the datasheet.

Changes to the datasheet may include the following, though not all of them may apply to this particular device.

1. Part number

   Example: TMPxxxxxxF → TMPxxxxxxFG

   All references to the previous part number were left unchanged in body text. The new part number is indicated on the prelims pages (cover page and this notification).

2. Package code and package dimensions

   Example: LQFP100-P-1414-0.50C → LQFP100-P-1414-0.50F

   All references to the previous package code and package dimensions were left unchanged in body text. The new ones are indicated on the prelims pages.

3. Addition of notes on lead solderability

   Now that the device is Pb-free, notes on lead solderability have been added.

4. RESTRICTIONS ON PRODUCT USE

   The previous (obsolete) provision might be left unchanged on page 1 of body text. A new replacement is included on the next page.

5. Publication date of the datasheet

   The publication date at the lower right corner of the prelims pages applies to the new device.

1. Part number

2. Package code and dimensions

Previous Part Number (in Body Text)	Previous Package Code (in Body Text)	New Part Number	New Package Code	OTP
TMP47C243N	P-SDIP28-400-1.78	TMP47C243NG	SDIP28-P-400-1.78	TMP47P443VNG
TMP47C243M	P-SOP28-450-1.27	TMP47C243MG	SOP28-P-450-1.27B	TMP47P443VMG
TMP47C243DM	P-SSOP30-56-0.65	TMP47C243DMG	SSOP30-P-56-0.65	TMP47P443VDMG
TMP47C443N	P-SDIP28-400-1.78	TMP47C443NG	SDIP28-P-400-1.78	TMP47P443VNG
TMP47C443M	P-SOP28-450-1.27	TMP47C443MG	SOP28-P-450-1.27B	TMP47P443VMG
TMP47C443DM	P-SSOP30-56-0.65	TMP47C443DMG	SSOP30-P-56-0.65	TMP47P443VDMG

*: For the dimensions of the new package, see the attached Package Dimensions diagram.

3. Addition of notes on lead solderability

The following solderability test is conducted on the new device.

Lead solderability of Pb-free devices (with the G suffix)

Test	Test Conditions	Remark
Solderability	(1) Use of Lead (Pb) ·solder bath temperature = 230°C ·dipping time = 5 seconds ·the number of times = once ·use of R-type flux (2) Use of Lead (Pb)-Free ·solder bath temperature = 245°C ·dipping time = 5 seconds ·the number of times = once ·use of R-type flux	Leads with over 95% solder coverage till lead forming are acceptable.

4. RESTRICTIONS ON PRODUCT USE

The following replaces the "RESTRICTIONS ON PRODUCT USE" on page 1 of body text.

RESTRICTIONS ON PRODUCT USE 20070701-EN

- The information contained herein is subject to change without notice.
- TOSHIBA is continually working to improve the quality and reliability of its products. Nevertheless, semiconductor devices in general can malfunction or fail due to their inherent electrical sensitivity and vulnerability to physical stress. It is the responsibility of the buyer, when utilizing TOSHIBA products, to comply with the standards of safety in making a safe design for the entire system, and to avoid situations in which a malfunction or failure of such TOSHIBA products could cause loss of human life, bodily injury or damage to property.
  In developing your designs, please ensure that TOSHIBA products are used within specified operating ranges as set forth in the most recent TOSHIBA products specifications. Also, please keep in mind the precautions and conditions set forth in the "Handling Guide for Semiconductor Devices," or "TOSHIBA Semiconductor Reliability Handbook" etc.
- The TOSHIBA products listed in this document are intended for usage in general electronics applications (computer, personal equipment, office equipment, measuring equipment, industrial robotics, domestic appliances, etc.).These TOSHIBA products are neither intended nor warranted for usage in equipment that requires extraordinarily high quality and/or reliability or a malfunction or failure of which may cause loss of human life or bodily injury ("Unintended Usage"). Unintended Usage include atomic energy control instruments, airplane or spaceship instruments, transportation instruments, traffic signal instruments, combustion control instruments, medical instruments, all types of safety devices, etc.. Unintended Usage of TOSHIBA products listed in his document shall be made at the customer's own risk.
- The products described in this document shall not be used or embedded to any downstream products of which manufacture, use and/or sale are prohibited under any applicable laws and regulations.
- The information contained herein is presented only as a guide for the applications of our products. No responsibility is assumed by TOSHIBA for any infringements of patents or other rights of the third parties which may result from its use. No license is granted by implication or otherwise under any patents or other rights of TOSHIBA or the third parties.
- Please contact your sales representative for product-by-product details in this document regarding RoHS compatibility. Please use these products in this document in compliance with all applicable laws and regulations that regulate the inclusion or use of controlled substances. Toshiba assumes no liability for damage or losses occurring as a result of noncompliance with applicable laws and regulations.
- For a discussion of how the reliability of microcontrollers can be predicted, please refer to Section 1.3 of the chapter entitled Quality and Reliability Assurance/Handling Precautions.

5. Publication date of the datasheet

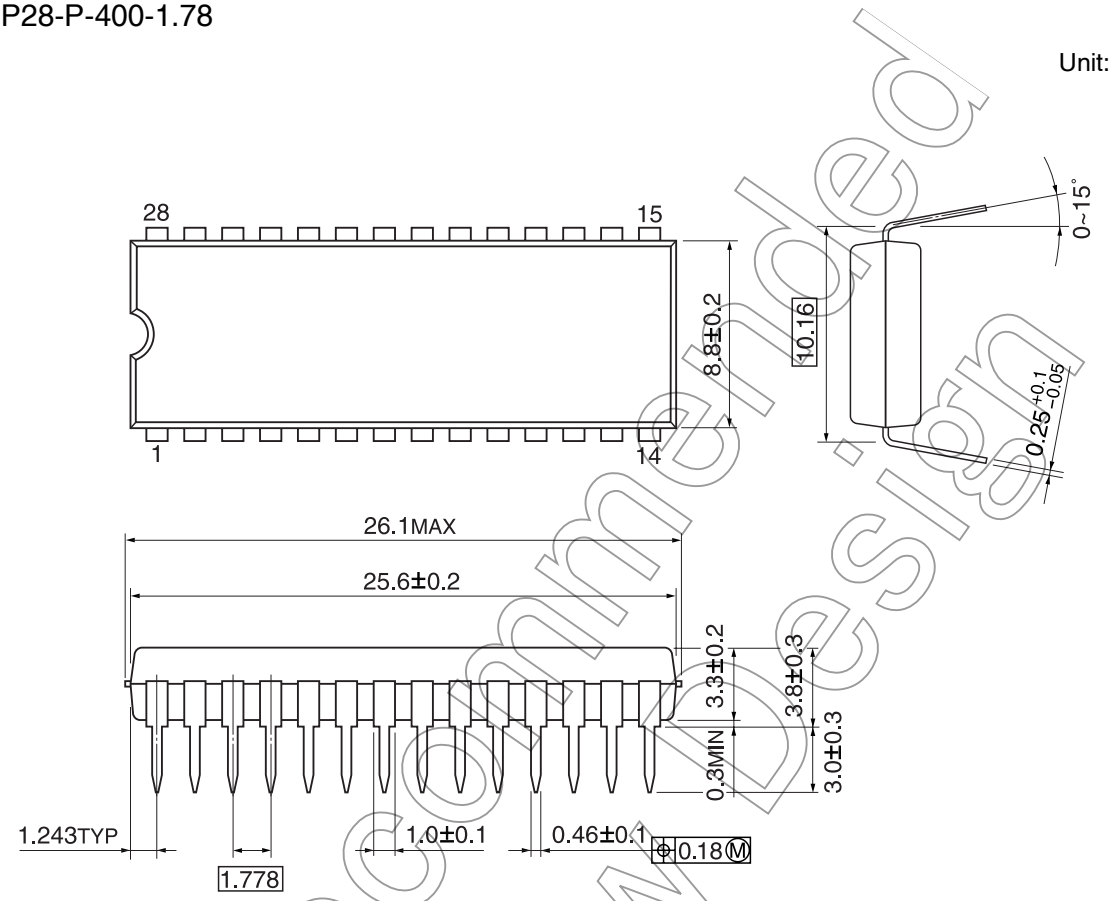
The publication date of this datasheet is printed at the lower right corner of this notification.

(Annex)

Package Dimensions

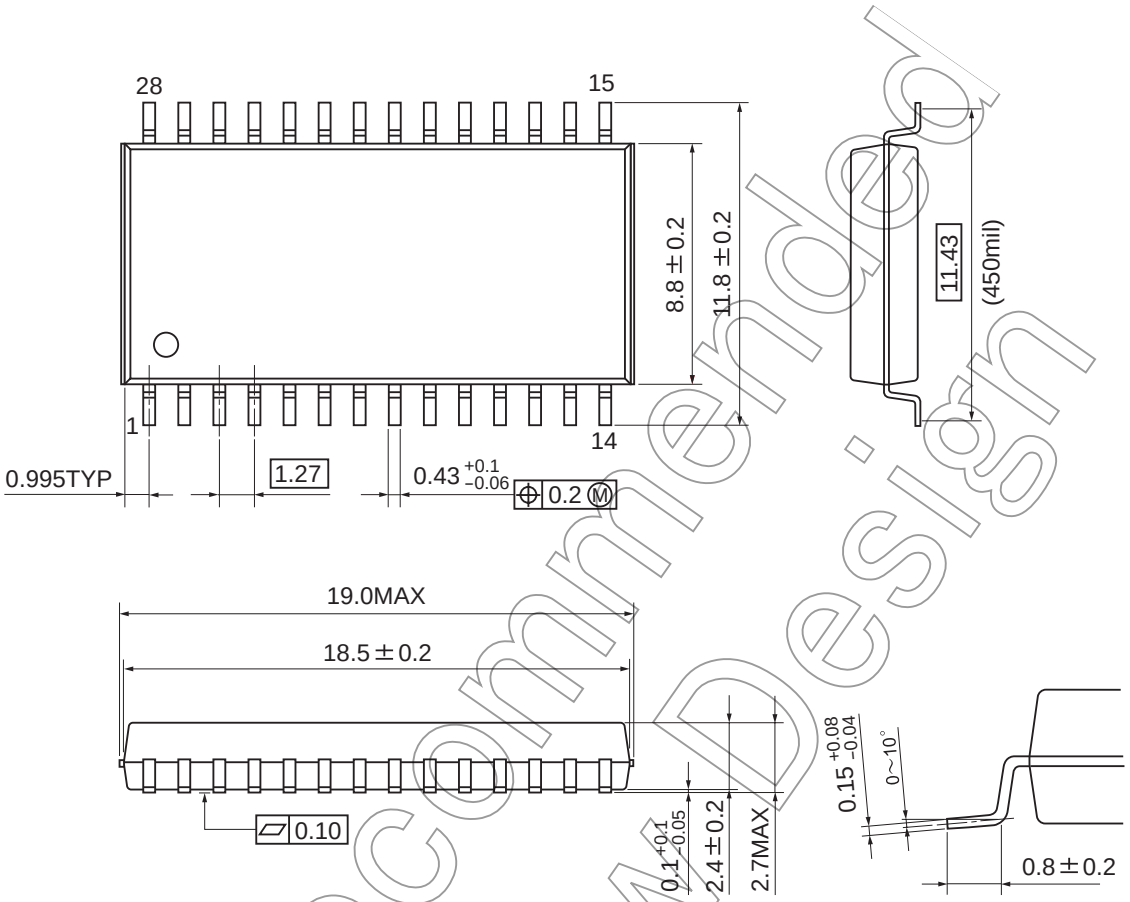
SDIP28-P-400-1.78

Unit: mm



SOP28-P-450-1.27B

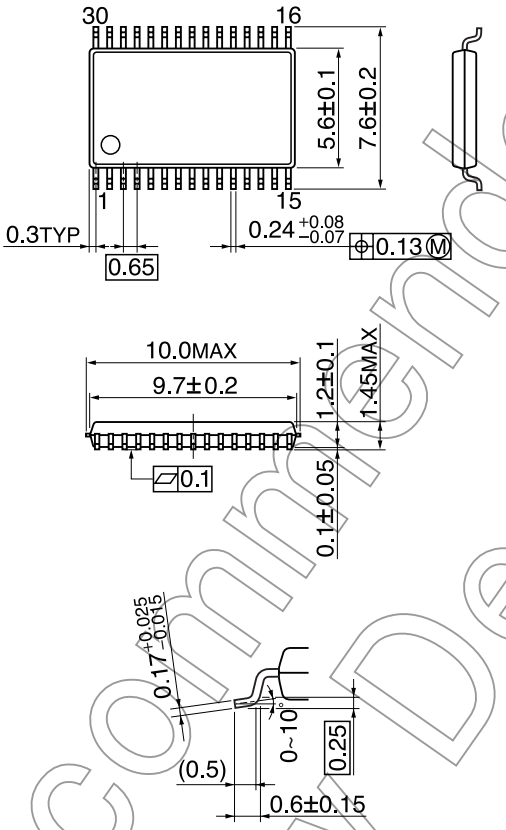
Unit: mm



Note: Palladium plated

SSOP30-P-56-0.65

Unit: mm



CMOS 4-Bit Microcontroller

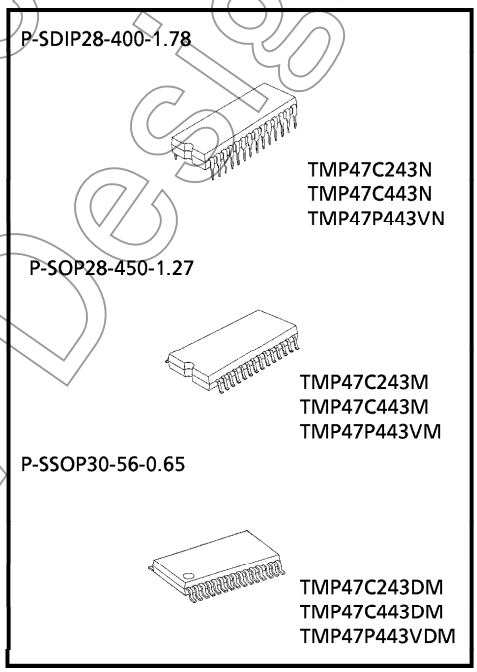
TMP47C243N, TMP47C443N
TMP47C243M, TMP47C443M
TMP47C243DM, TMP47C443DM

The TMP47C243/443 are the high speed and high performance 4-bit single chip microcomputers, with 8-bit AD converter, 8-bit serial interface, pulse output and zero-cross detector based on the TLCS-470 series.

Part No.	ROM	RAM	Rackage	OTP
TMP47C243N	2048 × 8-bit	128 × 4-bit	P-SDIP28-400-1.78	TMP47P443VN
TMP47C243M			P-SOP28-450-1.27	TMP47P443VM
TMP47C243DM			P-SSOP30-56-0.65	TMP47P443VDM
TMP47C443N	4096 × 8-bit	256 × 4-bit	P-SDIP28-400-1.78	TMP47P443VN
TMP47C443M			P-SOP28-450-1.27	TMP47P443VM
TMP47C443DM			P-SSOP30-56-0.65	TMP47P443VDM

Features

- ◆4-bit single chip microcomputer
- ◆Instruction execution time: 1.0 μs (at 8 MHz)
- ◆Low voltage operation: 2.2 V (at 4.2 MHz)
- ◆92 basic instructions
  - Table look-up instructions
  - 5-bit to 8-bit data conversion instruction
- ◆Subroutine nesting: 15 levels max
- ◆6 interrupt sources (External: 2, Internal: 4)
  All sources have independent latches each, and multiple interrupt control is available.
- ◆I/O port (23 pins)
- ◆Interval Timer
- ◆Two 12-bit Timer/Counters
  Timer, event counter,and pulse width measurement mode
- ◆Serial Interface with 8-bit buffer
  - Simultaneous transmission and reception capability
  - 8/4-bit transfer, external/internal clock, and leading/trailing edge shift mode



000707EBA1

◆8-bit successive approximate type AD converter
- With sample and hold
- 8 analog inputs
- Conversion time: 24 μs (at 8 MHz)

◆Pulse output
Buzzer drive/Remocon carrier

◆Zero-cross detector

◆High current outputs
LED direct drive capability (typ. 20 mA × 8 bits)

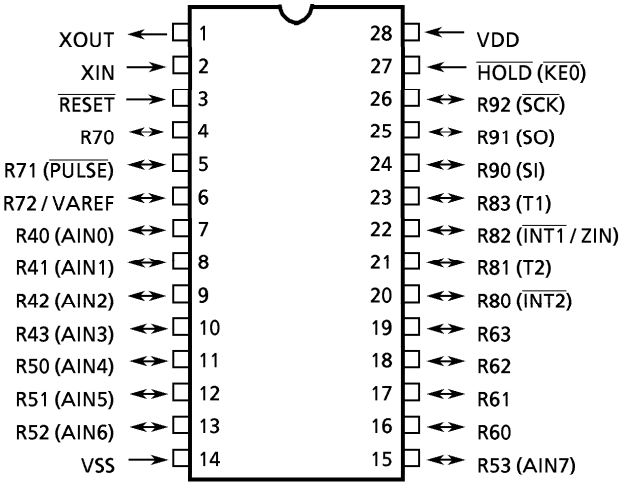
◆RESET function
- Watchdog Timer

◆Hold function
Battery/Capacitor back-up

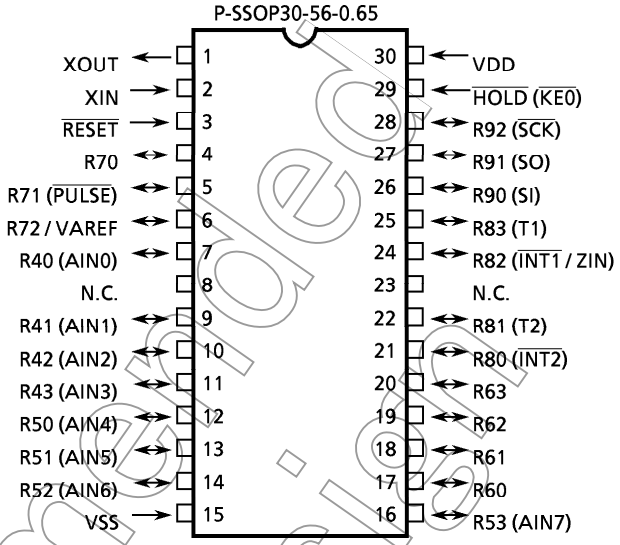
◆Emulation pod: BM47C443

Pin Assignments (Top View)

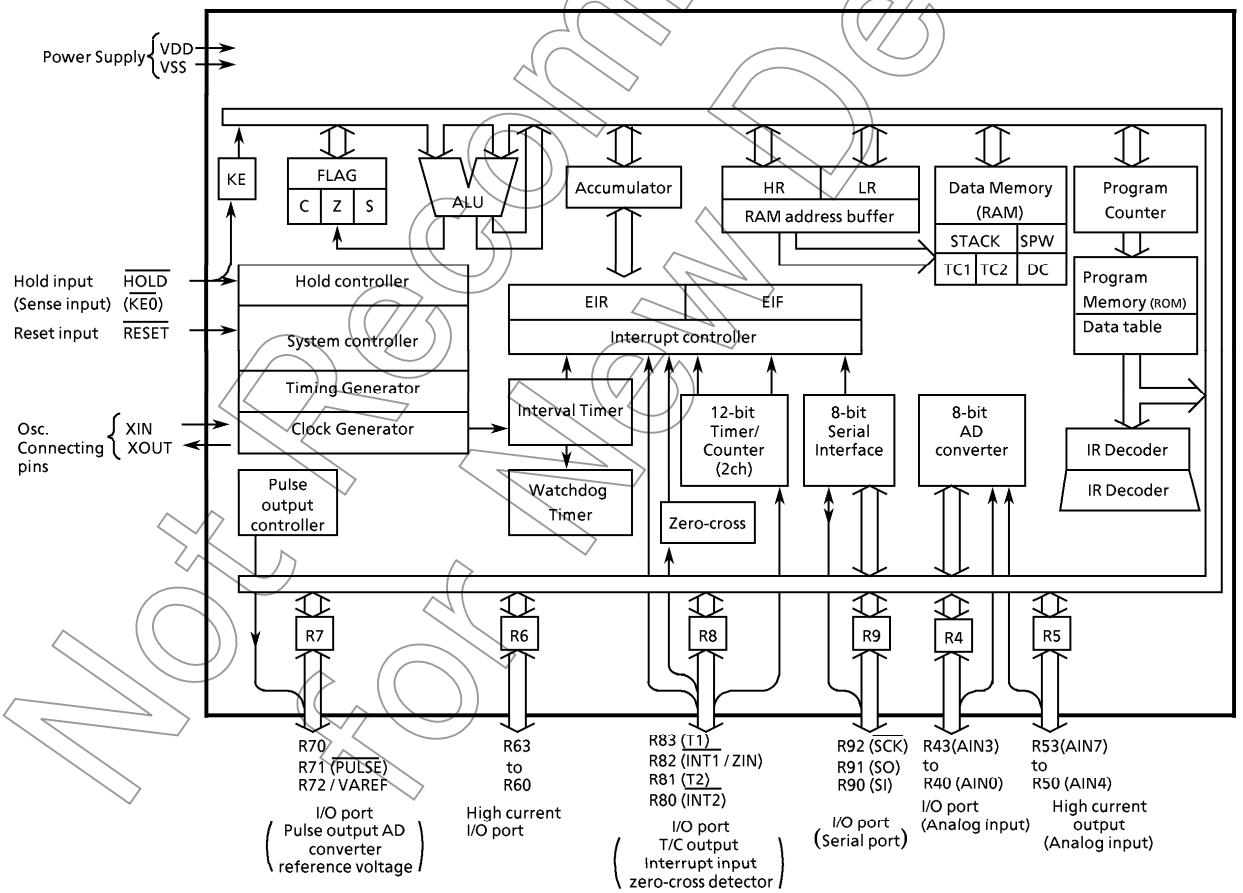
P-SDIP28-400-1.78 / P-SOP28-450-1.27



P-SSOP30-56-0.65



Block Diagram



Pin Function

Pin Name	Input / Output	Functions	
R43 to R40	I/O (input)	4-bit I/O port with latch (R7 port has 3-bit). Every bit data is possible to be set, cleared and tested by the bit manipulation instruction of the L-register indirect addressing. In the R5 and R6 port, 8-bit data are output by the 5-bit to 8-bit data conversion instruction [OUTB @HL] .	AD converter analog input
R53 to R50			
R63 to R60	I/O		
R72 /VAREF	I/O		AD converter reference voltage
R71 ($\overline{\text{PULSE}}$)	I/O (Output)		Pulse output
R70	I/O		
R83 (T1)	I/O (Input)	4-bit I/O port with latch. When used as input port, external interrupt input pin, or timer/counter external input pin, the latch must be set to "1".	Timer/Counter 1 external input
R82 ($\overline{\text{INT1}}$/ZIN)			External interrupt 1 and zero-cross input
R81 (T2)			Timer/Counter 2 external input
R80 ($\overline{\text{INT2}}$)			External interrupt 1 input
R92 ($\overline{\text{SCK}}$)	I/O (I/O)	3-bit I/O port with latch. When used as input port or serial port, the latch must be set to "1".	Serial clock I/O
R91 (SO)	I/O (Output)		Serial data output
R90 (SI)	I/O (Input)		Serial data input
XIN	Input	Resonator connecting pins. For inputting external clock, XIN is used and XOUT is opened.	
XOUT	Output		
$\overline{\text{RESET}}$	Input	Reset signal input	
$\overline{\text{HOLD}}$ ($\overline{\text{KE0}}$)	Input (Input)	Hold request/release signal input	Sense input
VDD	Power Supply	+ 5 V	
VSS		0 V (GND)	

Operational Description

Concerning the TMP47C243/443 the configuration and functions of hardwares are described. The basic instruction of configuration in the TMP47C243/443 is the same as those of TLCS-470 serise.

1. System Configuration

◆Internal CPU Function

- 2.1 Program Counter (PC)
- 2.2 Program Memory (ROM)
- 2.3 H Register, L Register
- 2.4 Data Memory (RAM)
  - a. Stack, b. Stack Pointer Word (SPW), c. Data Counter (DC)
- 2.5 ALU, Accumulator
- 2.6 Flags
- 2.7 Clock Generator and Timing Generator
- 2.8 Interrupt Function
- 2.9 Reset Function
  - · Watchdog Timer Reset

◆Peripheral Hardware Function Watchdog Timer Reset

- 3.1 I/O Ports
- 3.2 Interval Timer
- 3.3 Timer/Counters (TC1, TC2)
- 3.4 AD Converter
- 3.5 Pulse output
- 3.6 Zero-cross detector
- 3.7 Serial Interface

2. Internal CPU Function

2.1 Program Counter (PC)

The program counter is a 12-bit binary counter which indicates the address of the program memory storing the next instruction to be executed. Normally, the PC is incremented by the number of bytes of the instruction every time it is fetched. When a branch instruction or a subroutine instruction has been executed or an interrupt has been accepted, the specified values listed in Table 2-1 are set to the PC. The PC is initialized to "0" during reset.

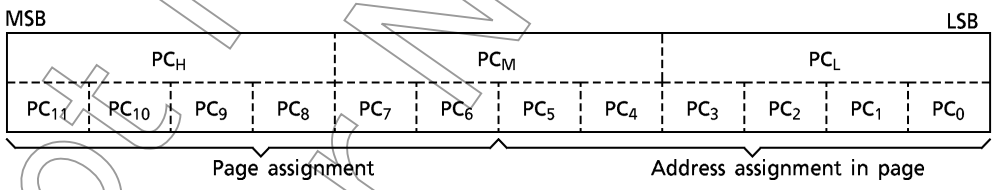


Figure 2-1. Configuration of Program Counter

The PC can directly address a 4096-byte address space. However, with the short branch and subroutine call instructions, the following points must be considered. In the TMP47C243/443, the long branch instruction [BSL a] is invalid.

(1) Short branch instruction [BSS a]

In [BSS a] instruction execution, when the branch condition is satisfied, the value specified in the instruction is set to the lower 6 bits of the PC. That is, [BSS a] becomes the in-page branch instruction. When [BSS a] is stored at the last address of the page, the upper 6 bits of the PC point the next page, so that branch is made to the next page.

(2) Subroutine call instruction [CALL a]

In [CALL a] instruction execution, the contents of the PC are saved to the stack then the value specified in the instruction is set to the PC. The address which can be specified by the instruction consists of 11 bits and the most significant bit of the PC is always "0". Therefore, the entry address of the subroutine should be within an addrress range of 000_H through $7FF_H$.

Table 2-1. Status Change of Program Counter

Instruction or Operation		Condition		Program Counter (PC)											
				PC_{11}	PC_{10}	PC_9	PC_8	PC_7	PC_6	PC_5	PC_4	PC_3	PC_2	PC_1	PC_0
Execution of Instruction	BS a	SF = 1 (Branch condition is satisfied)		Immediate data specified by the instruction											
		SF = 0 (Branch condition is not satisfied)		+ 2											
	BSS a	SF = 1	Lower 6-bit address ≠ 111111	Hold						Immediate data specified by the instruction					
			Lower 6-bit address = 111111 (last address in page)	+ 1						Immediate data specified by the instruction					
		SF = 0		+ 1											
	CALL a			0	Immediate data specified by the instruction										
	CALLS a			0	0	0	0	The data generated by the immediate data specified by the instruction					1	1	0
	RET			The return address restored from stack											
	RETI			The return address restored from stack											
	Others			Incremented by the number of bytes in the instruction											
Interrupt acceptance				0	0	0	0	0	0	0	0	Interrupt vector			0
Reset				0	0	0	0	0	0	0	0	0	0	0	0

2.2 Program Memory (ROM)

Programs and fixed data are stored in the program memory. The instruction to be executed next is read from the address indicated by the contents of the PC.

The fixed data can be read by using the table look-up instructions or 5-bit to 8-bit data conversion instruction.

(1) Table look-up instructions

[LDL A, @DC], [LDH A, @DC+]

The table look-up instructions read the lower and upper 4 bits of the fixed data stored at the address specified in the data counter (DC) to place them into the accumulator. [LDL A, @DC] instruction reads the lower 4 bits of fixed data, and [LDH A, @DC+] instruction reads the upper 4 bits.

The DC is a 12-bit register, allowing it to address the entire program memory space.

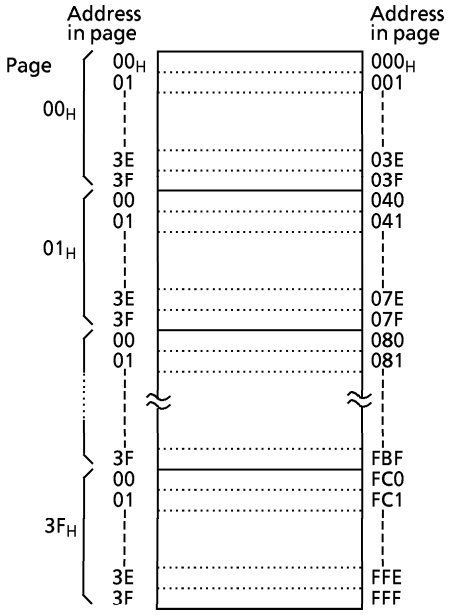


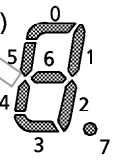
Figure 2-2. Configuration of Program Memory

(2) 5-bit to 8-bit data conversion instruction [OUTB @HL]

The 5-bit to 8-bit data conversion instruction reads the fixed data (8 bits) from the data conversion table in the program memory to output the upper 4 bits to port R6 and the lower 4 bits to port R5. The table is located in the last 32-byte space (addresses,$7E0_H$ through $7FF_H$ for the TMP47C243, $FE0_H$ through FFF_H for the TMP47C443) in the program memory with the lower address consisting of the 5 bits obtained by concatenating the contents of the data memory specified by the HL register pair and the content of the carry flag. This instruction is usable for such applications as converting BCD data into an output code to the 7-segment display elements.

Example: The following shows that the BCD data at address $2F_H$ in the data memory is converted into the 7-segment code (e.g., anode common LED) to be output to ports R6 and R5.

```
LD      HL, #2FH ;  HL←2FH (Data memory address is set)
TEST    CF       ;  CF←0 (The table is specified at addresses FE0H to FEFH)
OUTB    @HL      ;  Ports R6, R5←fixed data
  ⋮
ORG     0FE0H    ;  Data conversion table
DATA    0C0H, 0F9H, 0A4H, 0B0H, 99H, 92H, 82H, 0D8H, 80H, 98H
```

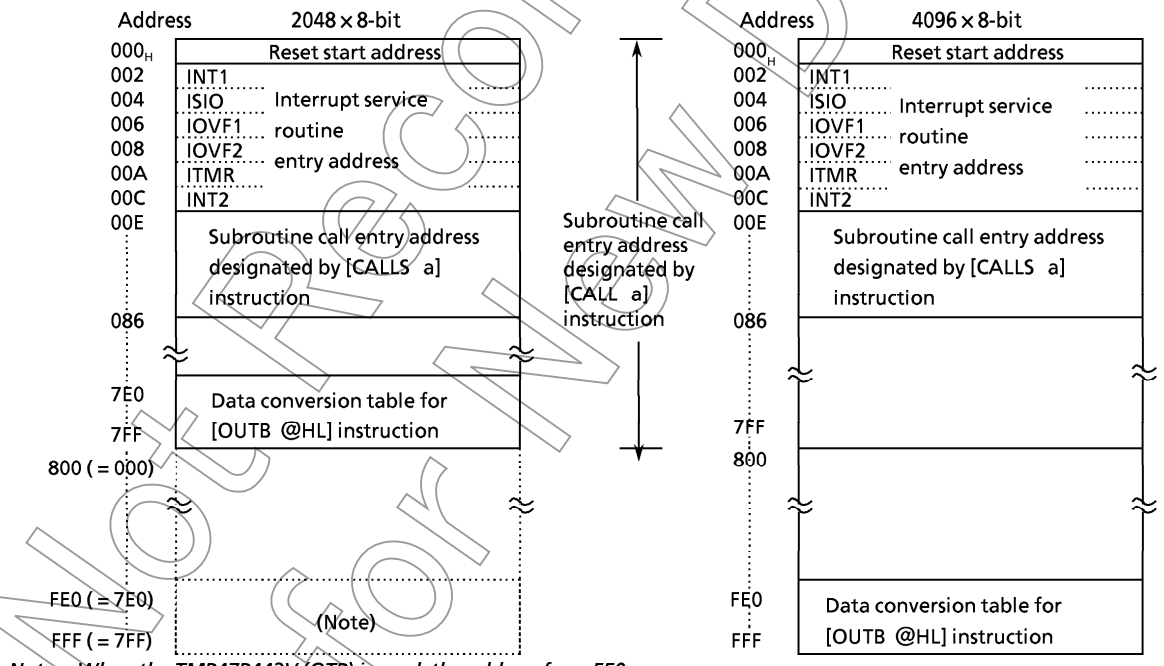


2.2.1 Program Memory Map

Figure 2-3 shows the program memory map. Address 000_H to 086_H and $FE0_H$ to FFF_H (000_H to 086_H and $7E0_H$ to $7FF_H$ for the TMP47C243) of the program memory are also used for special purposes.

2.2.2 Program Memory Capacity

The TMP47C243 has 2048 × 8 bits (addresses 000_H through $7FF_H$) of program memory (mask ROM), the TMP47C443 has 4096 × 8 bits (addresses 000_H through FFF_H).



Note: When the TMP47P443V (OTP) is used, the address from $FE0_H$ to FFF_H are used as the data conversion table area.

(a) TMP47C243 (b) TMP47C443

Figure 2-3. Program Memory Map

On the TMP47C243, no physical program memory exists in the address range 800_H through FFF_H. However, if this space is accessed by program, the most significant bit of each address is always regarded as "0" and the contents of the program memory corresponding to the address 000_H through $7FF_H$ are read.

2.3 H Register and L Register

The H register and the L register are 4-bit general registers. They are also used as a register pair (HL) for the data memory (RAM) addressing pointer. The RAM consists of pages, each page being 16 words long (1 word = 4 bits). The H register specifies a page and the L register specifies an address in the page.

The L register has the auto-post-increment/decrement capability, implementing the execution of composite instructions. For example, [ST A, @HL+] instruction automatically increments the contents of the L register after data transfer.

During the execution of [SET @L], [CLR @L], or [TEST @L] instructions, the L register is also used to specify the bits corresponding to I/O port pins R72 through R40 (the indirect addressing of port bits by the L register). Note that the TMP47C243/443 has not Data Memory Bank (DMB), so that it can not use the data memory bank control instruction

Example 1: To write immediate values "5" and "F" to data memory addresses 10_H and 11_H.

```
LD      HL,#10H          ;  HL←10H
ST      #5,@HL+          ;  RAM [10H] ←5H, LR←LR + 1
ST      #0FH,@HL+        ;  RAM [11H] ←FH, LR←LR + 1
```

Example 2: The output latch of R71 pin set "1" by the L register indirest addressing bit manipulation instruction.

```
LD      L,#1101B         ;  Sets R71 pin address to L register.
SET     @L               ;  R71←1
```

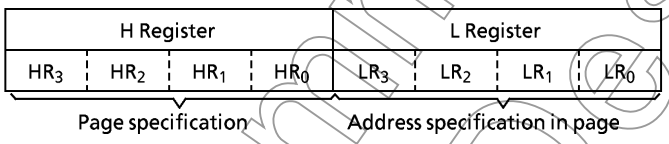


Figure 2-4. Configuration of H and L Registers

2.4 Data Memory (RAM)

The data memory stores user-processed data. One page of this memory is 16 words long (1 word = 4 bits). It has 8 pages.

The data memory is addressed in one of three ways (addressing modes):

The RAM is addressed in one of the three ways (addressing modes):

(1) Register-indirect addressing mode

In this mode, a page is specified by the H register and an address in the page by the L register.

Example: LD A, @HL ; Acc←RAM [HL]

(2) Direct addressing mode

In this mode, an address is directly specified by the 8 bits of the second byte (operand) in the instruction field.

Example: LD A, 2CH ; Acc←RAM [$2C_H$]

(3) Zero-page addressing mode

In this mode, an address in zero-page (addresses 00_H through $0F_H$) is specified by the lower 4 bits of the second byte (operand) in the instruction field.

Example: ST #3, 05H ; RAM [05_H] ←3

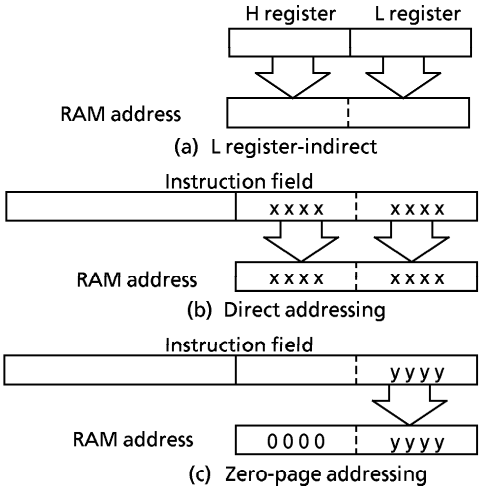
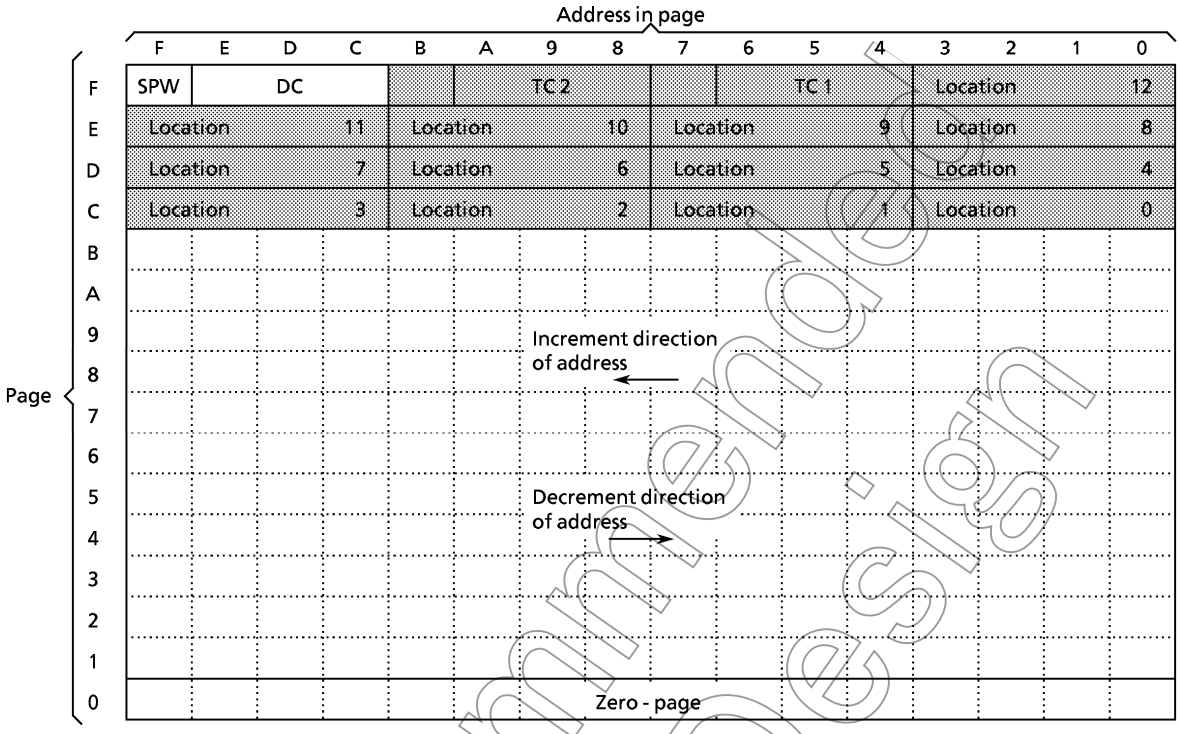


Figure 2-5. Addressing mode

2.4.1 Data Memory Map

Figure 2-6 shows the data memory map. The data memory is also used for the following special purpose.

① Stack & Stack Pointer Word (SPW)
② Data Counter (DC)
③ Count registers of the timer/counters (TC1, TC2)
④ Zero-page



Note 1: ▨ *denotes the stack area.*
Note 2: *The TC1 and TC2 areas are shared by the locations 13 and 14.*

Figure 2-6. Data Memory Map

(1) Stack

The stack provides the area in which the return address is saved before a jump is performed to the processing routine at the execution of a subroutine call instruction or the acceptance of an interrupt. When a subroutine call instruction is executed, the contents (the return address) of the program counter are saved; when an interrupt is accepted, the contents of the program counter and flags are saved.

When returning from the processing routine, executing the subroutine return instruction [RET] restores the contents of the program counter from the stack; executing the interrupt return instruction [RETI] restores the contents of the program counter and flags.

The stack consists of up to 15 levels (locations 0 through 14) which are provided in the data memory (addresses $C0_H$ through FB_H). Each location consists of 4-word data memory. Locations 13 and 14 are shared with the count registers of the timer/counters (TC1, TC2) to be described later.

The save/restore locations in the stack are determined by the stack pointer word (SPW). The SPW is automatically decremented after save, and incremented before restore. That is, the value of the SPW indicates the stack location number for the next save.

(2) Stack Pointer Word (SPW)

Address FF_H in the data memory is called the stack pointer word, which identifies the location in the stack to be accessed (save or restore).

Generally, location number 0 to 12 can be set to the SPW, providing up to 13 levels of stack nesting. Locations 13 and 14 are shared with the timer/counters to be described later; therefore, when the timer/counters are not used, the stack area of up to 15 levels is available. Address FF_H is assigned to the SPW, so that the contents of the SPW cannot be set "15" in any case.

The SPW is automatically updated when a subroutine call is executed or an interrupt is accepted. However, if it is used in excess of the stack area permitted by the data memory allocating configuration, the user-processed data may be lost. (For example, when the user-processed data area is in an address range 00_H through CF_H, up to location 4 of the stacks are usable. If an interrupt is accepted with location 4 already used, the user-processed data stored in addresses CC_H through CF_H corresponding to the location 3 area is lost.)
The SPW is not initialized by hardware, requiring to write the initial value (the location with which the use of the stack starts) by using the initialization routine. Normally, the initial value of "12" is used.

Example: To initialize the SPW (when the stack is used from location 12)

```
LD      A,#12     ;  SPW←12
ST      A,0FFH
```

(3) Data Counter (DC)

The data counter is a 12-bit register to specify the address of the data table to be referenced in the program memory (ROM). Data table reference is performed by the table look-up instructions [LDL A, @DC] and [LDH A, @DC+]. The data table may be located anywhere within the program memory address space.
The DC is assigned with a RAM address in unit of 4 bits. Therefore, the RAM manipulation instruction is used to set the initial value or read the contents of the DC.

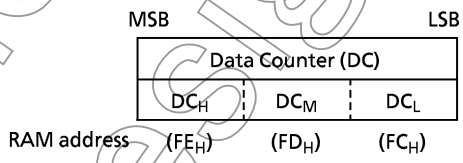


Figure 2-7. Data Counter

Example: To set the DC to 780_H.

```
LD      HL,#0FCH    ;  Sets RAM address of DCL to HL register pair.
ST      #0H,@HL+    ;  DC ← 780H
ST      #8H,@HL+
ST      #7H,@HL+
```

(4) Count registers of the timer/counters (TC1, TC2)

The TMP47C243/443 has two channels of 12-bit timer/counters. The count register of the timer/counter is assigned with a RAM addresses in unit of 4 bits, so that the initial value is set and the contents are read by using the RAM manipulation instruction.
The count registers are shared with the stack area (locations 13 and 14) described earlier, so that the stack is usable from location 13 when the timer/counter 1 is not used. When none of timer/counter 1 and timer/counter 2 are used, the stack is usable from location 14.
When both timer/counter 1 and timer/counter 2 are used, the data memory locations at addresses $F7_H$ and FB_H can be used to store the user-processed data.

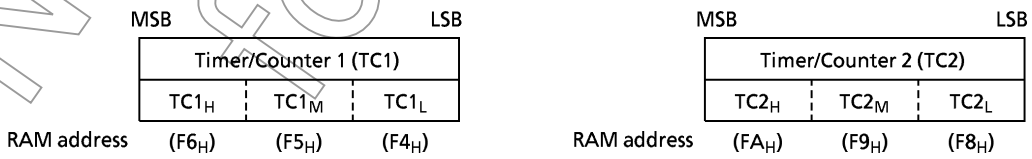
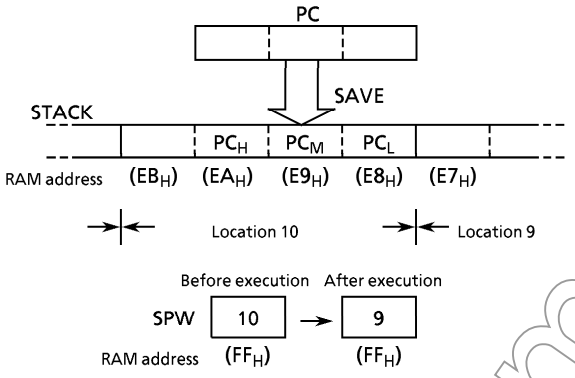


Figure 2-8. Count Registers of the Timer/Counters (TC1, TC2)

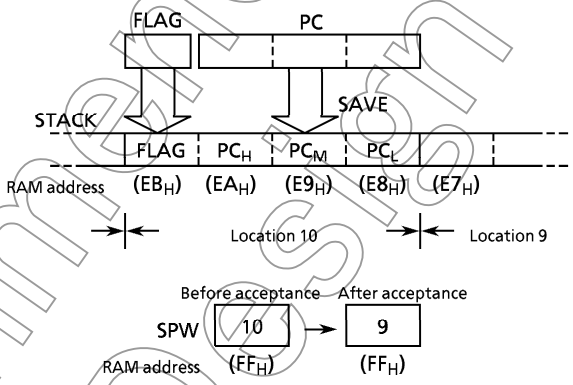
(5) Zero-page

The 16 words (at addresses 00_H through $0F_H$) of the zero page of the data memory can be used as the user flags or pointers by using zero-page addressing mode instructions (comparison, addition, transfer, and bit manipulation), providing enhanced efficiency in programming.

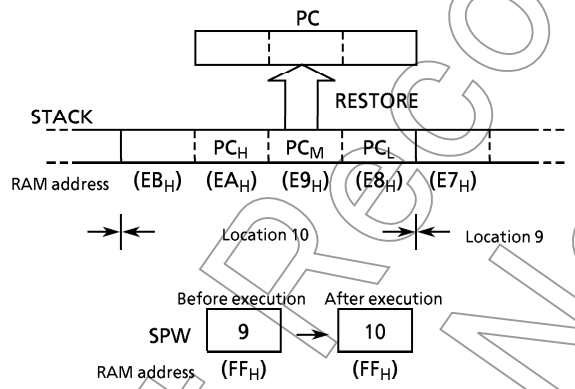
```
Example:  To write immediate data "8" to address 09H if bit 2 at address 04H in the RAM is "1".
          TEST      04H,2      ;  Skips if bit 2 at address 04H in the RAM is "0".
          B         SKIP
          ST        #8,09H     ;  Writes "8" to address 09H in the RAM
        SKIP:
```



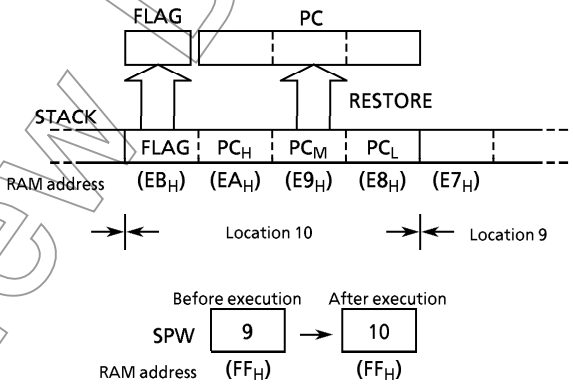
(a) At execution of the subroutine call instruction



(b) At acceptance of an interrupt



(c) At execution of the subroutine return instruction



(d) At execution of the interrupt return instruction

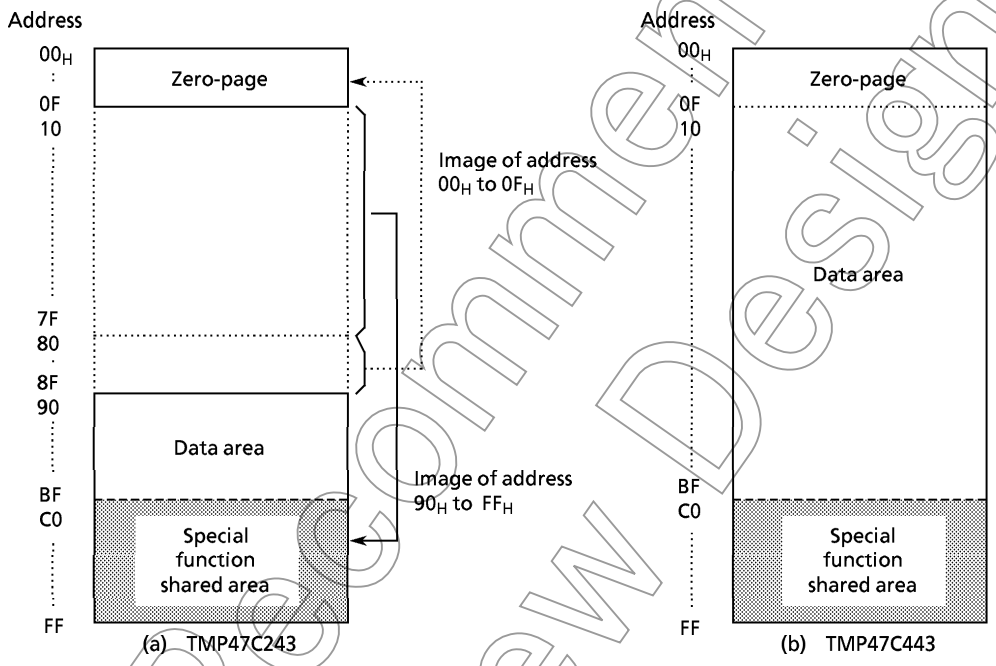
Figure 2-9. Accessing Stack (Save/Restore)

2.4.2 Data Memory Capacity

The TMP47C243 has 128 × 4 bits (addresses 00_H to $7F_H$) of the data memory (RAM), and the TMP47C443 has 256 × 4 bits (addresses 00_H to FF_H). When power-on performed, the contents of the RAM become unpredicable, so that they must be initialized by the initialization routine.

Example: To clear RAM (use common to the TMP47C243 and TMP47C443)

```
        LD      HL, #00H        ; HL←00H
SCLRRAM: ST     #0, @HL+        ; RAM [HL] ←0, LR←LR + 1
        B       SCLRRAM
        ADD     H, #1           ; HR←HR + 1
        B       SCLRRAM
```



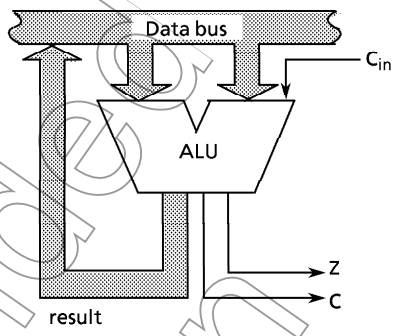
Note: With the TMP47C243, the most significant bit of the RAM address is always regarded as "0", so that addresses 90_H to FF_H may be accessed as addrresses 10_H to $7F_H$. However, programming should be performed assuming that the RAM is assigned to addresses 00_H to $0F_H$ and 90_H to FF_H as shown in Figure 2-10 (a) by considering the application software evaluation with a TMP47P443V (OTP) or development tools. Address 10 to $8F_H$ should not be accessed.

Figure 2-10. Data Memory Capacity and Address Assignment

2.5 ALU and Accumulator

2.5.1 Arithmetic / Logic Unit (ALU)

The ALU performs the arithmetic and logic operations specified by instructions on 4-bit binary data and outputs the result of the operation, the carry information (C), and the zero detect information (Z).



Note: C_{in} indicates the carry input specified by instruction

Figure 2-11. ALU

(1) Carry information (C)

The carry information indicates a carry-out from the most significant bit in an addition. A subtraction is performed as addition of two's complement, so that, with a subtraction, the carry information indicates that there is no borrow to the most significant bit. With a rotate instruction, the information indicates the data to be shifted out from the accumulator.

(2) Zero detect information (Z)

This information is "1" when the operation result or the data to be transferred to the accumulator/data memory is "0000_B".

Example: The carry information (C) and zero detect information (Z) for 4-bit additions and subtractions.

Operation	Result	C	Z
4 + 2 =	6	0	0
7 + 9 =	0	1	1
8 − 1 =	7	1	0
2 − 2 =	0	1	1
5 − 8 =	−3 (1101_B)	0	0

2.5.2 Accumulator (Acc)

The accumulator is a 4-bit register used to hold source data or results of the operations and data manipulations.

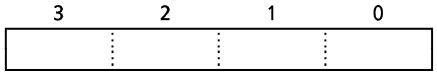


Figure 2-12. Accumulator

2.6 Flags

There are a carry flag (CF), a zero flag (ZF) and a status flag (SF), each consisting of 1 bit. These flags are set or cleared according to the condition specified by an instruction. When an interrupt is accepted, the flags are saved on the stack along with the program counter. When the [RETI] instruction is executed, the flags are restored from the stack to the states set before interrupt acceptance.

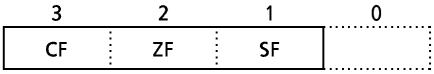


Figure 2-13. Flags

(1) Carry flag (CF)

The carry flag holds the carry information received from the ALU at the execution of an addition/subtraction with carry instruction, a compare instruction, or a rotate instruction. With a carry flag test instruction, the CF holds the value specified by it.

① Addition/subtraction with carry instructions [ADDC A, @HL], [SUBRC A, @HL]
The CF becomes the input (C_{in}) to the ALU to hold the carry information.

② Compare instructions [CMPR A, @HL], [CMPR A, #k]
The CF holds the carry information (non-borrow).

③ Rotate instructions [ROLC A], [RORC A]
The CF is shifted into the accumulator to hold the carry information (the data shifted out from the accumulator).

④ Carry flag test instructions [TESTP CF], [TEST CF]
With [TESTP CF] instruction, the content of the CF is transferred to the SF then the CF is set to "1".
With [TEST CF] instruction, the value obtained by inverting the content of the CF is transferred to the SF then the CF is cleared to "0".

(2) Zero flag (ZF)

The zero flag holds the zero detect information (Z) received from the ALU at the execution of an operational instruction, a rotate instruction, an input instrcution, or a transfer-to-accumulator instruction.

(3) Status flag (SF)

The status flag provides the branch condition for a branch instruction. Branch is performed when this flag is set to "1". Normally the SF is set to "1", so that any branch instruction can be regarded as an unconditional branch instruction. When a branch instruction is executed upon set or clear of the SF according to the condition specified by an instruction, this instruction becomes a conditional branch instruction. During reset, the SF is initialized to "1", other flags are not affected.

Example: When the following instructions are executed with the accumulator, H register, L register, data memory (address 07_H), and carry flag being set to "C_H", "0", "7", "5", and "1" respectively, the contents of the accumulator and flags become as follows:

Instruction	Acc after execution	Flag after execution		
		CF	ZF	SF
ADDC A, @HL	2_H	1	0	0
SUBRC A, @HL	9_H	0	0	0
CMPR A, @HL	C_H	0	0	1
AND A, @HL	4_H	1	0	1
LD A, @HL	5_H	1	0	1

Instruction	Acc after execution	Flag after execution		
		CF	ZF	SF
LD A, #0	0_H	1	1	1
ADD A, #4	0_H	1	1	0
DEC A	B_H	1	0	1
ROLC A	9_H	1	0	0
RORC A	E_H	0	0	1

2.7 Clock Generator and Timing Generator

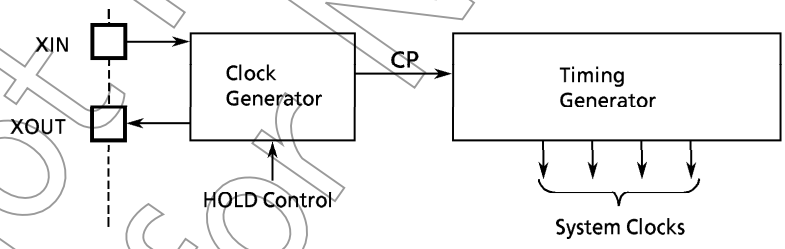
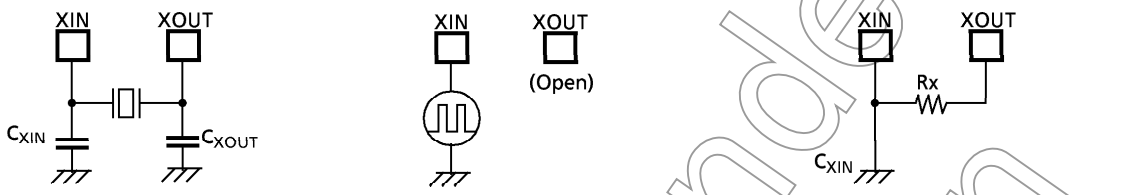


Figure 2-14. Clock Generator and Timing Generator

2.7.1 Clock Generator

The clock generator provides the basic clock pulse (CP) by which the system clock to be supplied to the CPU and the peripheral hardware is produced. The CP can be easily obtained by connecting the resonator to the XIN and XOUT pins (RC oscillation is also possible, depending on the mask option). The clock from the external oscillator is also available. In the hold operating mode, the clock generator stops oscillating.



(a) Crystal or ceramic resonator (b) External oscillator (c) RC oscillation (mask option)

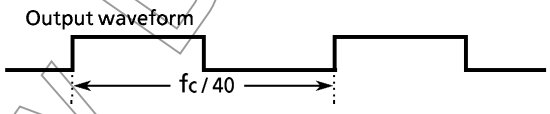
Figure 2-15. Examples of Oscillator Connection

Note: Accurate adjustment of the oscillation frequency

Although the hardware to externally and directly monitor the CP is not provided, the oscillation frequency can be adjusted by making the program to output the pulse with a fixed frequency to the port with the all interrupts disabled and timer/counters stopped and monitoring this pulse. With a system requiring the oscillation frequency adjustment, the adjusting program must be created beforehand.

Example: To output the oscillation frequency adjusting monitor pulse to port R70.

```
SFCCHK:  SET   %OP07,0
         CLR   %OP07,0
         BSS   SFCCHK
```



2.7.2 Timing Generator

The timing generator produces the system clocks from basic clock pulse which are supplied to the CPU and the peripheral hardware.

The timing generater consists of a 19-stage binary counter with a divide-by-3 prescaler. The basic clock (frequency: fc) provides the prescaler. Therefor, the output frequency at the last stage is $fc/2^{22}$ [Hz].

During reset, the binary counter is cleared to "0", however, the prescaler is not cleared.

The timing generater provides the following functions:

① Internal pulse for interval timer
② Internal pulse for timer/counters
③ Source clock for a pulse output
④ Internal serial clock for a serial interface
⑤ Warm-up time at release of the hold operation
⑥ Source clock for a watchdog timer

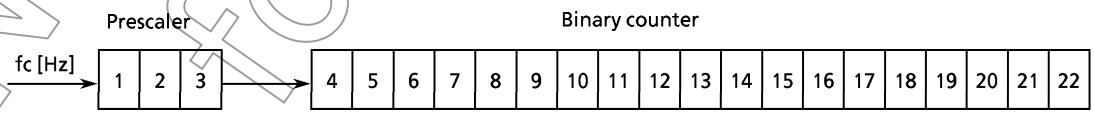


Figure 2-16. Configuration of Timing Generater

2.7.3 Instruction Cycle

The instruction execution and the on-chip peripheral hardware operations are performed in synchronization with the basic clock pulse (CP: fc [Hz]). The smallest unit of instruction execution is called an instruction cycle. The instruction set of the TLCS-47 series consists of 1-cycle instructions and 2-cycle instructions. The former requires 1 cycle for their execution; the latter, 2 cycles. Each instruction cycle consists of 4 states (S1 through S4). Each state consists of 2 basic clock pulses.
In the TMP47C243/443, [BSL a] instruction can not be used.

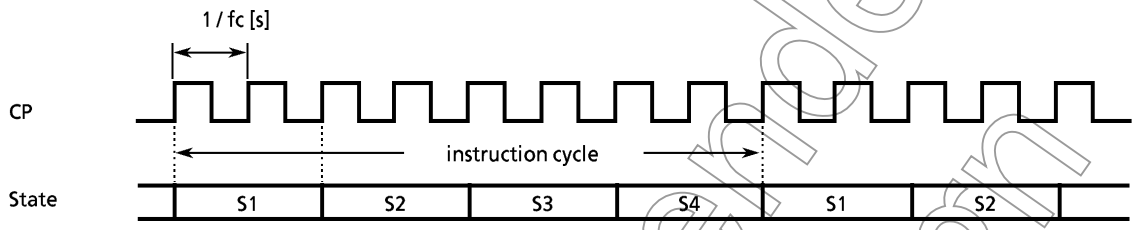


Figure 2-17. Instruction Cycle

2.7.4 Hold Operating Mode

The hold feature stops the system and holds the the system's internal states active before stop with a low power. The hold operation is controlled by the command register (OP10) and the $\overline{\text{HOLD}}$ pin input. The $\overline{\text{HOLD}}$ pin input state can be known by the status register (IP0E). The $\overline{\text{HOLD}}$ pin is shared with the $\overline{\text{KE0}}$ pin.

(1) Starts HOLD Operating Mode

The HOLD operating mode consister of the level-sensitive release mode and the edge-sensitive release mode.

The HOLD operation is started when the command is set to the command register and holds the following states during the hold operation:

① The oscillator stops and the system's internal operations are all held up.
② The interval timer is cleared to "0".
③ The states of the data memory, registers, and latches valid immediately before the system is put in the hold state are all held.
④ The program counter holds the address of the instruction to be executed after the instruction which starts the hold operating mode.

Hold operating mode command register

(Port address: OP10 Initial value: ****)

3	2	1	0
HLDMS		HWUT	

HLDMS	Sets mode/starts hold operation

01: Starts hold operation in edge-release mode

11: Starts hold operation in level-release mode

*0: Unused

HWUT	Sets the warm-up time at release of the hold operating mode

Example: At fc = 4 MHz

00: 2^{18} / fc [s] ..... 65.5 [ms]

01: 2^{14} / fc 4.1

10: Unused

11: 2^{6} / fc 0.016

Hold operating mode status register (Port address: IP0E)

3	2	1	0
(SIOF)	(SEF)		HOLD (KE0)

HOLD	$\overline{\text{HOLD}}$ pin input state

0: $\overline{\text{HOLD}}$ pin is high

1: $\overline{\text{HOLD}}$ pin is low (HOLD operation request)

*Note 1: *; Don't care*

Note 2: fc; Basic clock frequency [Hz]

Note 3: Do not access the OP10 when HOLD mode is not used.

Note 4: The status KE0 is usable in single clock mode only.

Figure 2-18. Hold Operating Mode Command Register/Status Register

a. Level-sensitive release mode

In this mode, the hold operation is released by setting the $\overline{\text{HOLD}}$ pin to the high level. This mode is used for the capacitor backup with power off or for the battery backup for long hours.

If the instruction to start the hold operation is executed with the $\overline{\text{HOLD}}$ pin input being high, the hold operation does not start but the release sequence (warm-up) starts immediately. Therefore, to start the hold operation in the level-sensitive release mode, that the $\overline{\text{HOLD}}$ pin input being low (the hold operation request) must be recognized in program. This recognition is performed in one of the two ways below:

① Testing HOLD (bit 0 of the status register)

② Applying the $\overline{\text{HOLD}}$ pin input also to the $\overline{\text{INT1}}$ pin to generate the external interrupt 1 request.

Example: To test HOLD to start the hold operation in the level-sensitive release mode (the warm-up time = 2^{14}/fc).

```
SHOLDH: TEST    %IP0E, 0      ;  Waits until HOLD pin input goes low.
        B       SHOLDH
        LD      A, #1101B     ;  OP10←1101B
        OUT     A, %OP10
```

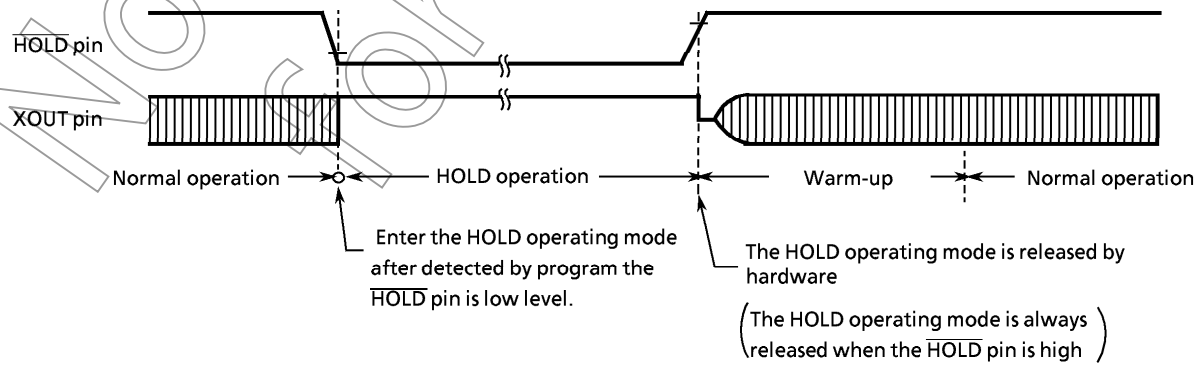


Figure 2-19. Level-sensitive release mode

(2) Edge-sensitive release mode

In this mode, the hold operation is released at the rising edge of the $\overline{HOLD}$ pin input. This mode is used for applications in which a relatively short-time program processing is repeated at a certain cycle. This cyclic signal (for example, the clock supplied from the low power dissipation oscillator). In the edge-sensitive mode, even if the $\overline{HOLD}$ pin input is high, the hold operation is performed.

Example: To start the hold operation in the edge-sensitive release mode (the warm-up time = 2^{14}/fc).

```
LD      A, #0101B     ;  OP10←0101B
OUT     A, %OP10
```

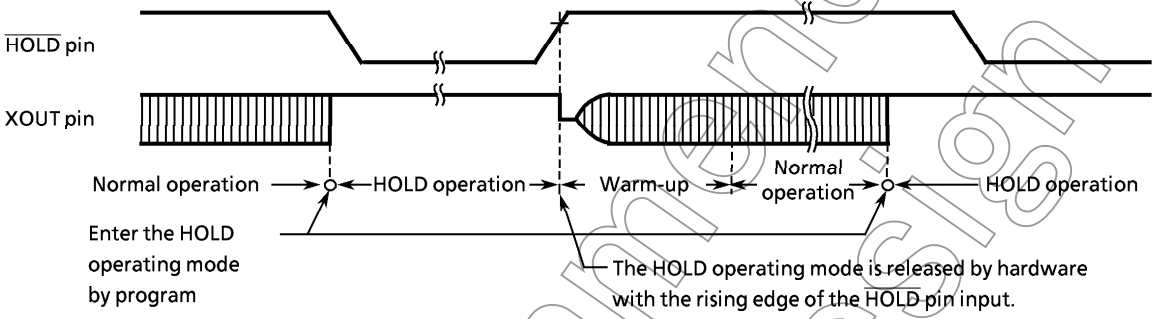


Figure 2-20. Edge-sensitive release mode

Note: In the hold operation, the dissipation of the power associated with the oscillator and the internal hardware is lowered; however, the power dissipation associated with the pin interface (depending on the external circuitry and program) is not directly determined by the hardware operation of the hold feature. This point should be considered in the system design and the interface circuit design.

In the CMOS circuitry, a current does not flow when the input level is stable at the power voltage level (V_{DD}/V_{SS}); however, when the input level gets higher than the power voltage level (by approximately 0.3 to 0.5 V), a current begins to flow. Therefore, if cutting off the output transistor at an I/O port (the open drain output pin with an input transistor connected) puts the pin signal into the high-impedance state, a current flows across the ports input transistor, requiring to fix the level by pull-up or other means.

(3) Releases Hold Operating Mode

The hold operating mode is released in the following sequence:

① The oscillator starts

② Warm-up is performed to acquire the time for stabilizing oscillation. During the warm-up , the internal operations are all stopped. One of three warm-up times can be selected by program depending on the characteristics of the oscillator used.

③ When the warm-up time has passed, an ordinary operation restarts from the instruction next to the instruction which starts the hold operation. At this time, the interval timer starts from the reset state "0".

* The warm-up time is obtained by dividing the basic clock by the interval timer, so that, if the frequency at releasing the hold operation is unstable, the warm-up time shown in Figure 5-1 includes an error. Therefore, the warm-up time must be handled as an approximate value. The hold operation is also released by setting the $\overline{RESET}$ pin to the low level. In this case, the normal reset operation follows immediately.

2.8 Interrupt Function

(1) Interrupt Controller

There are 6 interrupt sources (2 external and 4 internal). The prioritized multiple interrupt capability is supported. The interrupt latches (IL_5 through IL_0) to hold interrupt requests are provided for the interrupt sources. Each interrupt latch is set to "1" when an interrupt request is made, asking the CPU to accept the interrupt. The acceptance of interrupt can be permitted or prohibited by program through the interrupt enable master flip-flop (EIF) and interrupt enable register (EIR). When two or more interrupts occur simultaneously, the one with the highest priority determined by hardware is serviced first.

Table 2-2. Interrupt Sources

Interrupt Source			Priority	Interrupt Latch	Enable conditions	Entry address
External	Extenal Interrupt 1	(INT1)	(highest) 1	IL_5	EIF = 1	002_H
Internal	Serial Interface Interrupt	(ISIO)	2	IL_4	EIF = 1, EIR_3 = 1	004_H
	TC1 overflow Interrupt	(IOVF1)	3	IL_3	EIF = 1, EIR_2 = 1	006_H
	TC2 overflow Interrupt	(IOVF2)	4	IL_2	EIF = 1, EIR_1 = 1	008_H
	Interval Timer Interrupt	(ITMR)	5	IL_1		$00A_H$
External	External Interrupt 2	(INT2)	(lowest) 6	IL_0	EIF = 1, EIR_0 = 1	$00C_H$

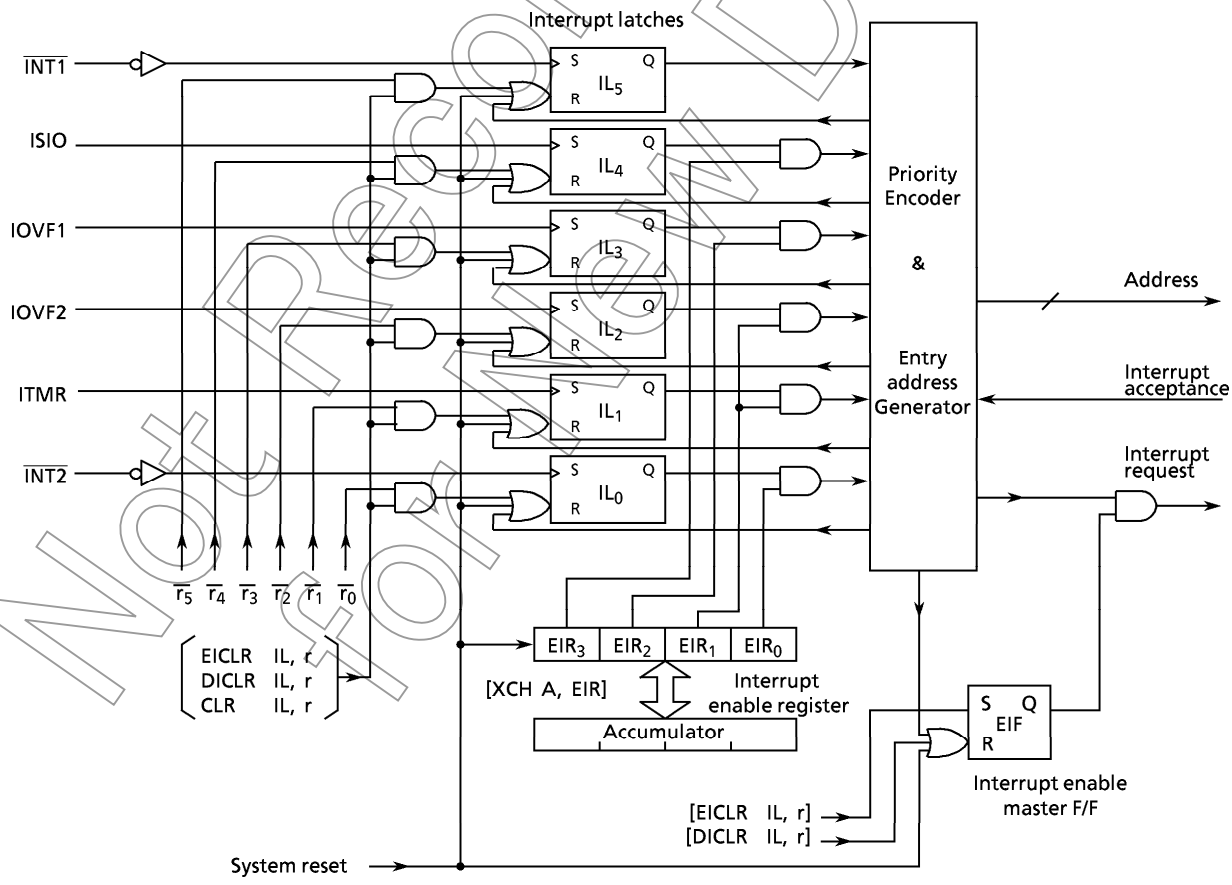


Figure 2-21. Interrupt Controller Block Diagram

a. Interrupt enable master flip-flop (EIF)

The EIF controls the enable/disable of all interrupts. When this flip-flop is cleared to "0", all interrupts are disabled; when it is set to "1", the interrupts are enabled.

When an interrupt is accepted, the EIF is cleared to "0", temporarily disabling the acceptance of subsequent interrupts. When the interrupt service program has been executed, the EIF is set to "1" by the execution of the interrupt return instruction [RETI], being put in the enabled state again.

Set or clear of the EIF in program is performed by instructions [EICLR IL, r] and [DICLR IL, r], respectively. The EIF is initialized to "0" during reset.

b. Interrupt enable register (EIR)

The EIR is a 4-bit register specifies the enable or disable of each interrupt except INT1. An interrupt is enabled when the corresponding bit of the EIR is "1", and an interrupt is disabled when the corresponding bit of the EIR is "0". Bit 1 of the EIR (EIR_1) is shared by both IOVF2 and ITMR interrupts.

Read/write on the EIR is performed by executing [XCH A, EIR] instruction. The EIR is initialized to "0" during reset.

c. Interrupt latch (IL_5 through IL_0)

An interrupt latch is provided for each interrupt source. The IL is set to "1" when an interrupt request is made to ask the CPU for accepting the interrupt. Each IL is cleared to "0" upon acceptance of the interrupt. It is initialized to "0" during reset.

The ILs can be cleared independently by interrupt latch operation instructions ([EICLR IL, r], [DICLR IL, r], and [CLR IL, r]) to make them cancel interrupt requests or initialize by program. When the value of instruction field (r) is "0", the interrupt latch is cleared; when the value is "1", the IL is held. Note that the ILs cannot be set by instruction.

Example 1: To enable IOVF1, INT1, and INT2 interrupts.

```
LD      A,#0101B    ;  EIR←0101B
XCH     A,EIR
EICLR   IL,111111B  ;  EIF←1
```

Example 2: To set the EIF to "1", and to clear the interrupt latches except ISIO to "0".

```
EICLR   IL,010000B  ;  EIF←1, IL5←0, IL3 – IL0←0
```

(2) Interrupt Processing

An interrupt request is held until the interrupt is accepted or the IL is cleared by the reset or the interrupt latch operation instruction. The interrupt acknowledge processing is performed in 2 instruction cycles after the end of the current instruction execution (or after the timer/counter processing if any). The interrupt service program terminates upon execution of the interrupt return instruction [RETI].

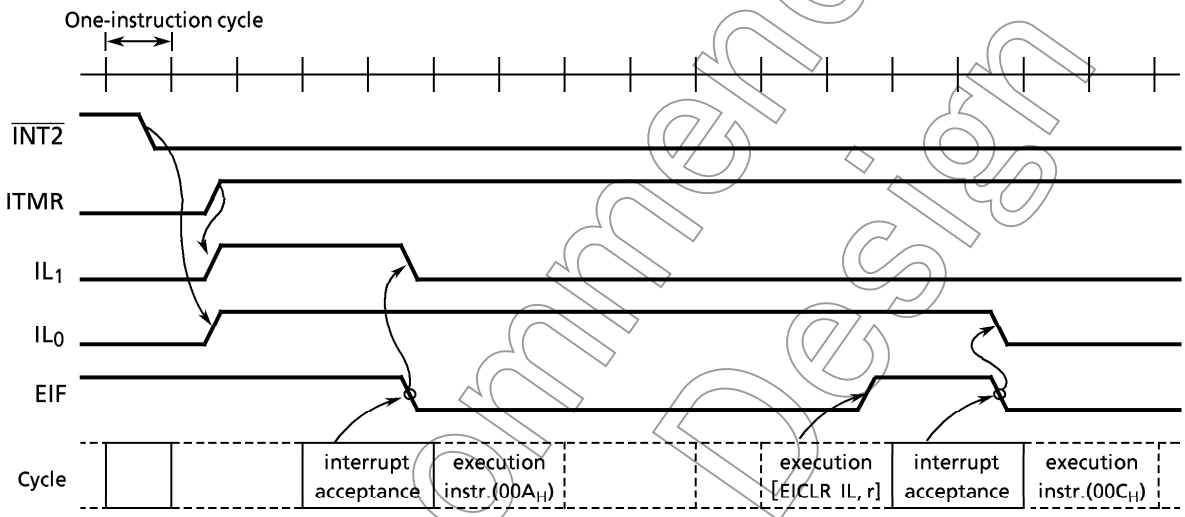
The interrupt acknowledge processing consists of the following sequence:

① The contents of the program counter and the flags are saved on the stack.
② The interrupt entry address corresponding to the interrupt source is set to the program counter.
③ The status flag is set to "1".
④ The EIF is cleared to "0", temporarily disabling the acceptance of subsequent interrupts.
⑤ The interrupt latch for the accepted interrupt source is cleared to "0".
⑥ The instruction stored at the interrupt entry address is executed. (Generally, in the program memory space at the interrupt entry address, the branch instruction to each interrupt processing program is stored.)

To perform the multi-interrupt, the EIF is set to "1" in the interrupt service program, and the acceptable interrupt source is selected by the EIR. However, for the INT1 interrupt, the interrupt service is disabled under software control because it is not disabled by the EIR.

Example: The INT1 interrupt service is disabled under software control (Bit 0 of RAM [05_H] are assigned to the disabling switch of interrupt service).

```
PINT1:  TEST    05H,0    ;  Skips if RAM [05H] 0 is "1"
        B       SINT1
        RETI
SINT1:   ⋮
```



Note 1: It is assumed that there is no other interrupt request and EIR = 0011_B.
Note 2: The value r in the [EICLR IL, r] instruction is assumed as 111111_B.
Note 3: [ ] denotes the execution of an instruction.

Figure 2-22. Interrupt Timing chart (Example)

The interrupt return instruction [RETI] performs the following operations:

① Restores the contents of the program counter and the flags from the stack.
② Sets the EIF to "1" to provide the interrupt enable state again.

Note: When the time required for the interrupt service is longer than that foe the interrupt request, only the interrupt service program is executed without executing the main program.

In the interrupt processing, the program counter and flags are automatically saved or restored but the accumulator and other registers are not. If it is necessary to save or restore them, it must be performed by program as shown in the following example. To perform the multi-interrupt, the saving RAM area never be overlapped.

Example: To save and restore the accumulator and HL register pair.

```
XCH     HL, GSAV1      ;  RAM [GSAV1] ↔ HL
XCH     A, GSAV1 + 2   ;  RAM [GSAV1 + 2] ↔ Acc
```

Note: The lower 2 bits of GSAV1 should be "0's".

(3) External Interrupt

When an external interrupt (INT1 or INT2) occurs, the interrupt latch is set at the falling edge of the corresponding pin input ($\overline{INT1}$ or $\overline{INT2}$).

The INT1 interrupt cannot be disabled by the EIR, so that it is always accepted in the interrupt enable state (EIF = "1"). Therefore, INT1 is used for an interrupt with high priority such as an emergency interrupt. When R82 ($\overline{INT1}$) pin is used for the I/O port, the INT1 interrupt occurs at the falling edge of the pin input, so that the interrupt return [RETI] instruction must be stored at the interrupt entry address to perform dummy interrupt processing.

The INT2 interrupt can be enable/disable by the EIR. When R80 ($\overline{INT2}$) pin is used as the I/O port, the INT2 interrupt ouurs at the falling edge of the pin input. However the interrupt request is not accepted by clearing bit 0 of the EIR to "0".

Because the external interrupt input is the hysteresis type, each of high and low level time requires 2 or more instruction cycles for a correct interrupt operation.

2.9 Reset Function

When the $\overline{RESET}$ pin is held to the low level for three or more instruction cycles when the power voltage is within the operating voltage range and the oscillation is stable, reset is performed to initialize the internal states.

When the $\overline{RESET}$ pin input goes high, the reset is cleared and program execution starts from address 000_H. The $\overline{RESET}$ pin is a hysteresis input with a pull-up resistor (220 kΩ typ.). Externally attaching a capacitor and a diode implement a simplified power-on-reset operation.

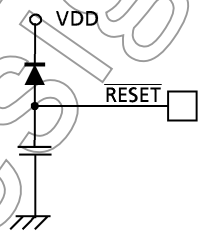


Figure 2-23. Simplified Power-On-Reset Circuit

Table 2-3. Initialization of Internal States by Reset Operation

On-chip hardware	Initial value	On-chip hardware	Initial value
Program counter (PC)	000_H	Output latch (I/O ports or Output ports)	Refer to "INPUT/OUTPUT Circuitry".
Status flag (SF)	1		
Interrupt enable master flip-flop (EIF)	0		
Interrupt enable register (EIR)	0_H	Command register	Refer to the description of each relative command register.
Interrupt latch (IL)	"0"		
Interval timer	"0"		

2.9.1 Watchdog Timer (WDT)

The watchdog timer capability is provided to quickly detect the CPU malfunction such as endless looping caused by noises or the like, and restore the CPU to the normal state. The WDT is disabled during reset.

The WDT consists of 10 binary counters, a flip-flop, and a controller. Source input clock of binary counters is $fc/2^{15}$[Hz]. The flip-flop is set to "1" during reset, and cleared to "0" on the rising edge of the binary counter output. The WDT is controlled by the command register (OP15). The command register is initialized to "1000_B" during reset.

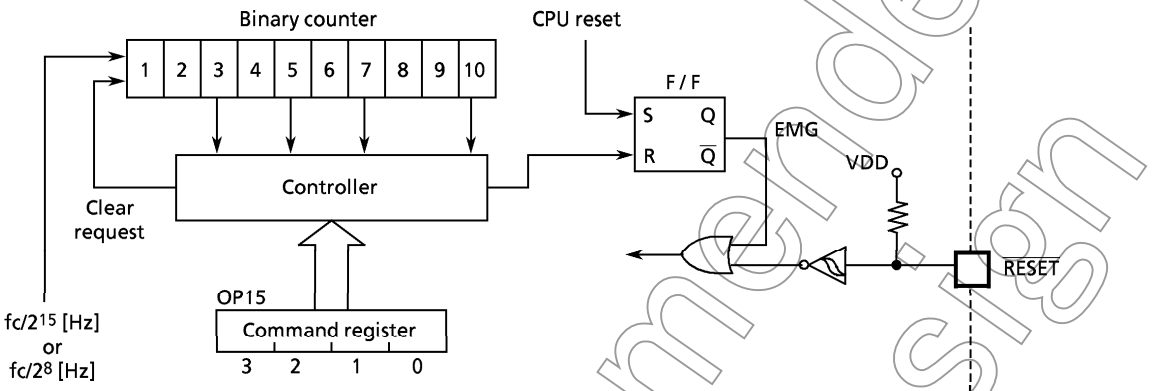


Figure 2-24. Configuration of Watchdog Timer

To detect the CPU malfunction by the WDT:

① Set the WDT detection time, and clear the binary counters.
② Enable the WDT.
③ Clear the binary counters within WDT detection time that was set in ①. If a CPU malfunction occurs, preventing the binary counters from being cleared, the flip-flop is clearred to "0" on the rising edge of the binary counter output, making the malfunction detection signal active.

Example: To enable the with detection time of $63 \times 2^{15}/fc$ [s]

```
        LD      A, #0010B       ; OP15 ← 0010B   (Set WDT detection time,
        OUT     A, %OP15                         clear binary counters)
        LD      A, #0110B       ; OP15 ← 0110B   (Enable WDT)
        OUT     A, %OP15
        :
        :
        :
        LD      A, #0110B       ; OP15 ← 0110B   (Clear binary counters)
        OUT     A, %OP15
        :
        :
```

Within WDT detection timer

Note: It is necessary to clear the binary counter prior to enabling watchdog timer.

Watchdog timer control command register
(Port address: OP15)

3	2	1	0	
RWT	EWT	TWT		(Initial value: 1000)

RWT	Clears Binary counter

0: Binary counter cleared
(after clear, it is automatically set to "1")

EWT	Watchdog timer enable/disable

0: Disable
1: Enable

TWT	Set Watchdog timer detection time

Example: At fc = 4.19 MHz

00: $3 \times 2^{15}/fc$ [s] ..... 23 [ms]
01: $15 \times 2^{15}/fc$ 117
10: $63 \times 2^{15}/fc$ 493
11: $511 \times 2^{15}/fc$ 3996

Figure 2-25. Watchdog Timer Control Command Register

3. Peripheral Hardware Function

3.1 Ports

The data transfer with the external circuit and the command/status/data transfer with the internal circuit are performed by using the I/O instructions (13 kinds). There are 4 types of ports:

① I/O port ; Data transfer with external circuit
② Command register ; Control of internal circuit
③ Status register ; Reading the status signal from internal circuit
④ Data register ; Data transfer with internal circuit

These ports are assigned with port addresses (00_H through $1F_H$). Each port is selected by specifying its port address in an I/O instruction. Table 3-2 lists the port address assignments and the I/O instructions that can access the ports.

3.1.1 I/O Timing

(1) Input timing

External data is read from an input port or an I/O port in the S3 state of the second instruction cycle during the input instruction (2-cycle instruction) execution. This timing cannot be recognized from the outside, so that the transient input such as chattering must be processed by program.

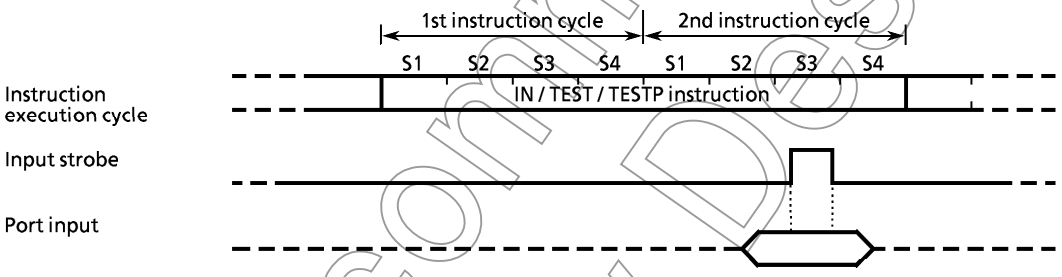


Figure 3-1. Input Timing

(2) Output timing

Data is output to an output port or an I/O port in the S4 state of the second instruction cycle during the output instruction (2-cycle instruction) execution.

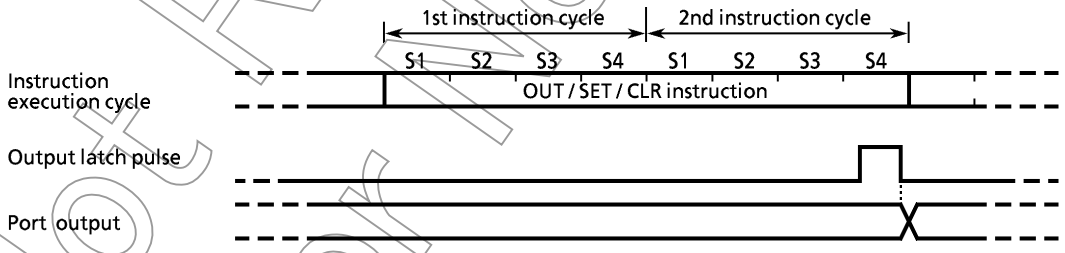


Figure 3-2. Output Timing

3.1.2 I/O Ports

The TMP47C243/443 have 7 I/O ports (23 pins) each as follows:

① R4, R5 ; 4-bit input/output (shared with AD converter analog inputs)
② R6 ; 4-bit input/output
③ R7 ; 3-bit input/output (shared with pulse output)
④ R8 ; 4-bit input/output (shared with zero-cross input, external interrupt input and timer/counter input)
⑤ R9 ; 3-bit input/output (shared with serial port)
⑥ KE ; 1-bit sense input (shared with hold request/release signal input)

Each output port contains a latch, which holds the output data. The input ports have no latch; therefore, it is desired to hold data externally until it is read or read twice or more before processing it.

(1) Ports R4, R5, R6, R7

These ports are 4-bit I/O ports with a latch (Port R7 is 3-bit). When used as an input port, the latch must be set to "1". The latch is initialized to "1" during reset.

These 4 ports (15 pins) can be set, cleared, and tested for each bit as specified by L register indirect addressing bit manipulation instructions ([SET @L], [CLR @L], and [TEST @L]). Table 3-1 lists the pins (I/O ports) that correspond to the contents of L register.

Example: To clear R43 output as specified by the L register indirect addrressing bit manipulation instruction.

```
LD      L, #0011B       ; Sets R43 pin address to L register
CLR     @L              ; R43←0
```

Table 3-1. Relationship between L register contents and I/O port bits

L register 3 2 1 0	PIN
0 0 0 0	R40
0 0 0 1	R41
0 0 1 0	R42
0 0 1 1	R43

L register 3 2 1 0	PIN
0 1 0 0	R50
0 1 0 1	R51
0 1 1 0	R52
0 1 1 1	R53

L register 3 2 1 0	PIN
1 0 0 0	R60
1 0 0 1	R61
1 0 1 0	R62
1 0 1 1	R63

L register 3 2 1 0	PIN
1 1 0 0	R70
1 1 0 1	R71
1 1 1 0	R72

(a) Ports R4 (R43 to R40) and R5 (R53 to R50)

Ports R4 and R5 are 4-bit I/O ports with latch shared by the analog inputs for AD converter. When used as an analog inputs, the latch should be set to "1". If other port is used as an output, be careful not to execute the output instruction for any port during AD conversion in order to keep accuracy of conversion. The latch is initialized to "1" and analog input is selected R40 (AIN0) pin during reset.

(b) Ports R5 (R53 to R50) and R6 (R63 to R60)

Ports R5 and R6 are 4-bit high current output ports with a latch, which can directly drive LEDs. When an input instruction is executed, the latch data is read in these ports. They can be accessed separately at port addresses OP05/IP05 and OP06/IP06. Additionally, 8-bit data can be set to these ports (the upper 4 bits to port R6, the lower 4 bits to port R5) by using the 5-bit to 8-bit data conversion instruction [OUTB @HL]. The latch is initialized to "1" during reset.

Example 1: To output immediate value "5" to port R5

```
OUT     #5,%OP05        ;  Port R5←5
```

Example 2: To read the latch data from port R6 and store it in the accumulator

```
IN      %IP06,A         ;  Acc←Port R6
```

Example 3: To read from ROM, the 8-bit data corresponding to the 5 bits obtained by linking the content (1 bit) of the carry flag with the contents (4 bits) of at address 90_H in RAM to output the 8-bit data to ports R6 and R5.

```
LD      HL,#90H         ;  HL←90H (Sets the data memory address)
OUTB    @HL             ;  Ports R6, R5←ROM data
```

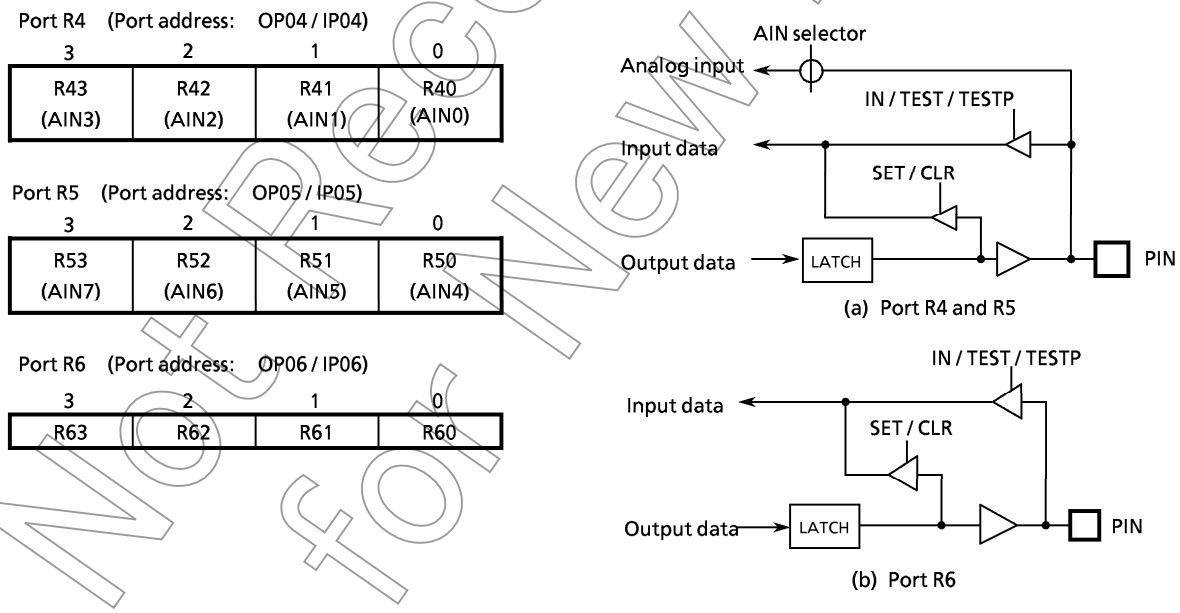


Figure 3-3. Ports R4, R5 and R6

(c) Port R7 (R72 to R70)

Port R7 is a 3-bit I/O port with a latch. R71 is shared by the pulse output and pulse is output when R71 output latch is "0". To use it for an ordinary I/O port, the pulse output must be disabled.

R72/VAREF pin can be selected the I/O port or analog reference voltage supply for AD conversion by the mask option. Note that R73 pin dose not exist actually but "1" is read when an input instruction is executed.

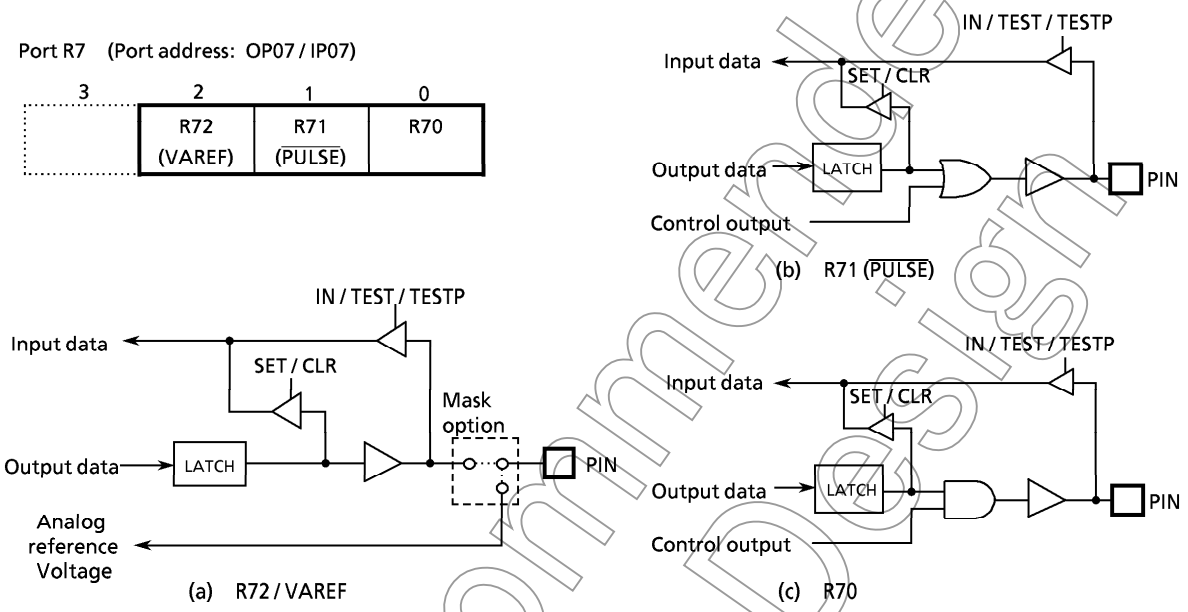


Figure 3-4. Port R7

(2) Port R8 (R83 to R80)

Port R8 is a 4-bit I/O port with a latch. When used as an input port, the latch must be set to "1". The latch is initialized to "1" during reset.

Port R8 is shared with the external interrupt input pin and the timer/counter input pin. To use this port for one of these functional pins, the latch should be set to "1". To use it for an ordinary I/O port, the acceptance of external interrupt should be disabled or the event counter/pulse width measurement modes of the timer/counter should be disabled.

Note: When R82 ($\overline{INT1}$) pin is used for an I/O port, external interrupt 1 occurs upon detection of the falling edge of pin input, and if the interrupt enable master flip-flop is enabled, the interrupt request is always accepted. So that a dummy interrupt processing must be performed (only the interrupt return instruction [RETI] is executed).

With R80 ($\overline{INT2}$) pin, external interrupt 2 occurs like R82 in but bit 0 of the interrupt enable register (EIR_0) is only kept at "0", not accepting the interrupt request.

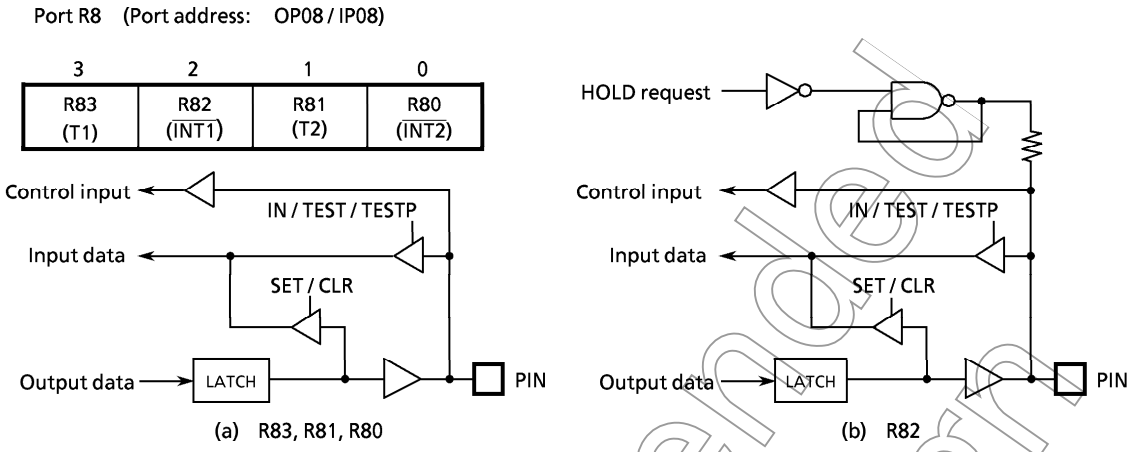


Figure 3-5. Port R8

(3) Port R9 (R92 to R90)

Port R9 is a 3-bit I/O port with a latch. When used as an input, the latch must be set to "1". The latch is initialized to "1" during reset. Port R9 is shared with the serial port. To use port R9 for the serial port, the latch should be set to "1". To use port R9 for an ordinary I/O port, the serial interface must be disabled.

Although R93 pin does not exist actually, the set or clear instruction for R93 ([SET %OP09, 3] or [CLR %OP09, 3]) should not be execution. However, other instructions can be used, in which an undefined value is read upon execution of an input instruction.

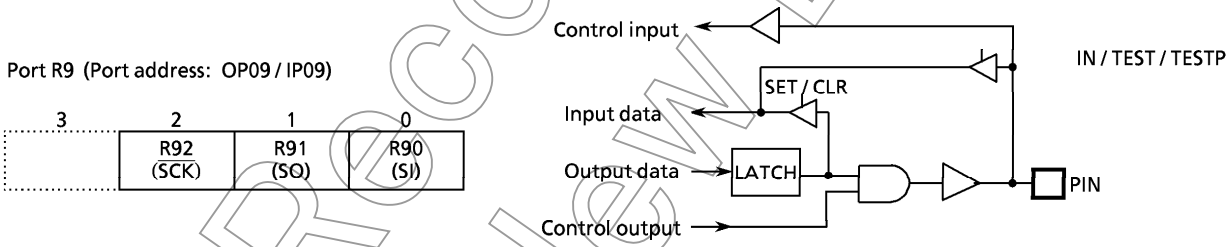


Figure 3-6. Port R9

(4) Port KE ($\overline{KE0}$)

Port KE is a 1-bit sense input port shared with the hold request/release signal input in ($\overline{HOLD}$). This input port is assigned to the least significant bit of Port address IP0E and is processed as the data with inverted polarity. For example, if an input instruction is executed with the pin on the high level, "0" is read. An undefined value is read from bit1 of the IP0E with an input instruction.

Example: To wait until $\overline{KE0}$ pin goes low.

```
SWAIT   :   TEST    %IP0E,0         ; Waits if KE0 pin = "H".
            B       SWAIT
```

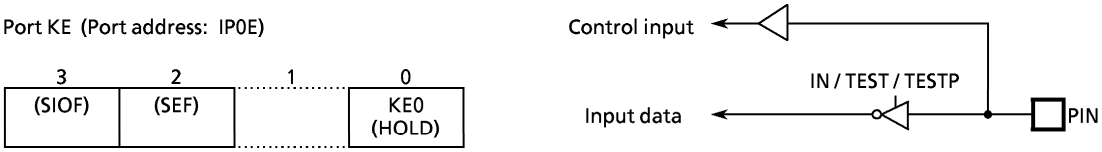


Figure 3-7. Port KE

Table 3-2. Port Address Assignments and Available I/O Instructions

Port address (**)	Port		Input/Output instruction							
	Input (P**)	Output (Op**)	IN %p, A IN %p, @HL	OUT A, %p OUT @HL, %p	OUT #k, %p	OUTB @HL	SET %p, b CLR %p, b	TEST %p, b TESTP %p, b	SET @L CLR @L TEST @L
00_H	———	———	—	—	—	—	—	—	—
01	———	———	—	—	—	—	—	—	—
02	———	———	—	—	—	—	—	—	—
03	———	———	—	—	—	—	—	—	—
04	R4 input port (Analog input)	R4 output port	○	○	○	—	○	○	○
05	R5 input port (Analog input)	R5 output port	○	○	○	○ (Note2)	○	○	○
06	R6 input port	R6 output port	○	○	○	○	○	○	○
07	R7 input port	R7 output port	○	○	○	—	○	○	○
08	R8 input port	R8 output port	○	○	○	—	○	○	—
09	R9 input port	R9 output port	○	○	○	—	○	○	—
0A	———	———	—	—	—	—	—	—	—
0B	———	———	—	—	—	—	—	—	—
0C	———	———	—	—	—	—	—	—	—
0D	———	———	—	—	—	—	—	—	—
0E	SIO, HOLD status	———	—	—	—	—	—	○	—
0F	Serial receive buffer	Serial transmit buffer	○	○	○	—	—	—	—
10_H	———	Hold operating mode control	—	○	—	—	—	—	—
11	———	———	—	—	—	—	—	—	—
12	AD coverted value	AD analog input selector	○	○	—	—	—	—	—
13	AD status input	AD start register	○	○	—	—	—	—	—
14	———	———	—	—	—	—	—	—	—
15	———	Watchdog timer control	—	○	—	—	—	—	—
16	———	———	—	—	—	—	—	—	—
17	———	Pulse output control	—	○	—	—	—	—	—
18	———	———	—	—	—	—	—	—	—
19	———	Interval Timer interrupt control	—	○	—	—	—	—	—
1A	———	———	—	—	—	—	—	—	—
1B	———	———	—	—	—	—	—	—	—
1C	———	Timer/Counter 1 control	—	○	—	—	—	—	—
1D	———	Timer/Counter 2 control	—	○	—	—	—	—	—
1E	———	Serial interface control 1	—	○	—	—	—	—	—
1F	———	Serial interface control 2	—	○	—	—	—	—	—

Note 1: "———" means the reserved state. Unavailable for the user programs.

Note 2: The 5-bit to 8-bit data conversion instruction [OUTB @HL] , automatic access to ports R5 and R6.

3.2 Interval Timer Interrupt (ITMR)

The interval timer can be used to generate an interrupt with a fixed frequency. Interval timer interrupt is control by the command register (OP19). An interval timer interrupt is generated at the first rising edge of the binary counters output after the command has been set. The interval timer is not cleared by command, so that the first interrupt may occur earlier than the preset interrupt period.

Example: To set the interval timer interrupt frequency to $fc/2^{15}$ [Hz].

```
LD      A,  #0110B    ;  OP19←0110B
OUT     A,  %OP19
```

Interval Timer interrupt command register
(Port address: OP19)

3	2	1	0
TMRE		TMRF	

TMRE	Interrupt Enable/disable

00: Stop
01: Enable
1*: (Disable)

TMRF	Interrupt Frequency

Example: At fc = 4. 194304 MHz

00: $fc/2^{11}$ [Hz] ..... 2048 [Hz]
01: $fc/2^{13}$ 512
10: $fc/2^{15}$ 128
11: $fc/2^{17}$ 32

*Note 1: *; Don't care*

Note 2: fc; Basic clock frequency [Hz]

Figure 3-8. Interval Timer Interrupt Command Register

3.3 Timer/Counters (TC1, TC2)

The TMP47C243/443 contain two 12-bit timer/counters (TC1, TC2). RAM addresses are assigned to the count register in unit of 4 bits, permitting the initial value setting and counter reading through the RAM manipulation instruction. When the timer/counter is not used, the mode selection may be set to "stopped" to use the RAM at the address corresponding to the timer/counter for storing the ordinary user-processed data.

	MSB		LSB
	Timer/Counter 1 (TC1)		
RAM address	$TC1_H$	$TC1_M$	$TC1_L$
TMP47C443	($F6_H$)	($F5_H$)	($F4_H$)
TMP47C243	(76_H)	(75_H)	(74_H)

	MSB		LSB
	Timer/Counter 2 (TC2)		
RAM address	$TC2_H$	$TC2_M$	$TC2_L$
TMP47C443	(FA_H)	($F9_H$)	($F8_H$)
TMP47C243	($7A_H$)	(79_H)	(78_H)

Figure 3-9. The Count registers of the Timer/Counters (TC1, TC2)

3.3.1 Functions of Timer/Counters

The timer/counters provide the following functions:

① Event counter
② Programmable timer
③ Pulse width measurement

Timer/Counter 1 control command register
(Port address: OP1C) (Initial value: 0000)

3	2	1	0
TC1MS		IPR1	

TC1MS	Mode selection

00: Stop
01: Event counter mode
10: Timer mode
11: Pulse width measurement mode

IPR1	Internal pulse rate (interval timer output) selection

Example: At fc = 4.194304 MHz

00: $fc/2^6$ [Hz] ..... 65536 [Hz]
01: $fc/2^8$ 16384
10: $fc/2^{10}$ 4096
11: $fc/2^{14}$ 256

Timer/Counter 2 control command register
(Port address: OP1D) (Initial value: 0000)

3	2	1	0
TC2MS		IPR2	

TC2MS	Mode selection

00: Stop
01: Event counter mode
10: Timer mode
11: Pulse width measurement mode

IPR2	Internal pulse rate (interval timer output) selection

Example: At fc = 4.194304 MHz

00: $fc/2^{10}$ [Hz] ..... 4096 [Hz]
01: $fc/2^{14}$ 256
10: $fc/2^{18}$ 16
11: $fc/2^{22}$ 1

Note: fc; Basic clock frequency [Hz]

Figure 3-10. Timer/Counter Control Command Registers

3.3.2 Control of Timer/Counters

The timer/counters are controlled by the command registers. The command register is accessed as port address OP1C for TC1 and port address OP1D for TC2. These registers are initialized to "0" during reset.

The timer/counter increments at the rising edge of each count pulse. Counting starts with the first rising edge of the count pulse generated after the command has been set. Count operation is performed in one instruction cycle after the current instruction execution, during which the execution of a next instruction and the acceptance of an interrupt are delayed. If counting is requested by both TC1 and TC2 simultaneously, the request by TC1 is preferred. The request by TC2 is accepted in the next instruction cycle. Therefore, during count operation, the apparent instruction execution speed drops as counting occurs more frequently.

The timer/counter causes an interrupt upon occurrence of an overflow (a transition of the count value from FFF_H to 000_H). If the timer/counter is in the interrupt enabled state and the overflow interrupt is accepted immediately after its occurrence, the interrupt is processed in the sequence shown in Figure 3-12. Note that counting continues if there is a count request after overflow occurrence.

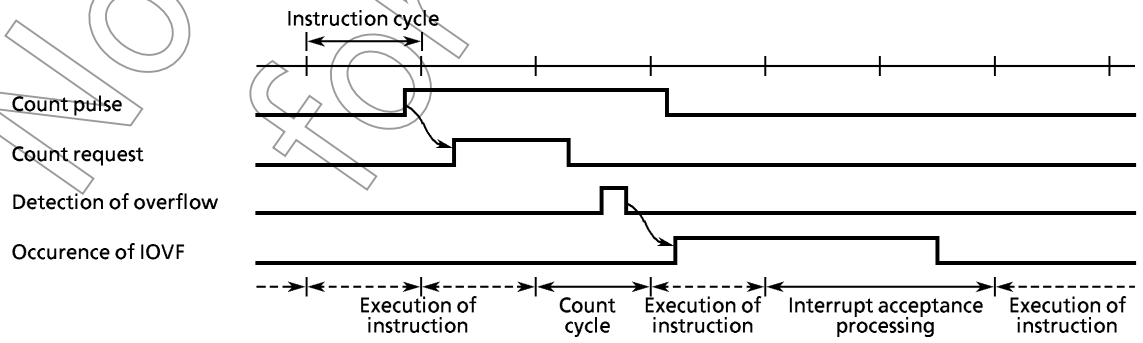


Figure 3-11. Timer/Counter Overflow Interrupt Timing

(1) Event counter mode

In the event counter mode, the timer/counter increments at each rising edge of the external pin (T1, T2) input. T1 and T2 pins are also used as the R83 and R81 pins.

To use these pins as the T/C input, set the output latch of R83 and R81 to "1".

At reset, the output latch is initialized to "1". The maximum applied frequency of the external pin input is fc/32 for the 1-channel operation; for the 2-channel operation, the frequency is fc/32 for TC1 and fc/40 for TC2. The apparent instruction execution speed drops most to (9/11) × 100 = 82% when TC1 and TC2 are operated at the maximum applied frequency because the count operation is inserted once every 4 instruction cycles for TC1 and every 5 cycles for TC2. For example, the instruction execution speed of 2 μs drops to 3.64 μs.

Example: To operate TC2 in the event counter mode

```
LD      A,  #0100B    ;  OP1D←01**B
OUT     A,  %OP1D
```

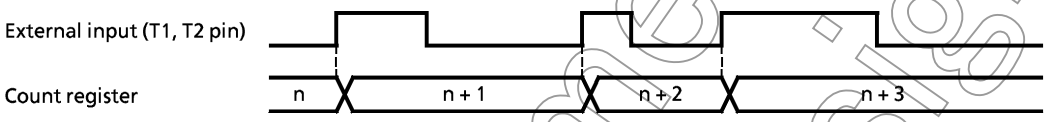


Figure 3-12. Event Counter Mode Timing chart

(2) Timer mode

In the timer mode, the timer/counter increments at the rising edge of the internal pulse generated from the interval timer. One of 4 internal pulse rates can be selected by the command register. The selected rate can be initially set to the timer/counter to generate an overflow interrupt in order to create a desired time interval.

When an internal pulse rate of $fc/2^{10}$ is used, a count operation is inserted once every 128 instruction cycles, so that the apparent instruction execution speed drops by (1/127) × 100 = 0.8%. For example, the instruction execution speed of 2 μs drops to 2.016 μs.

In the timer mode, R83 (T1) and R81 (T2) pins provide the ordinary I/O ports.

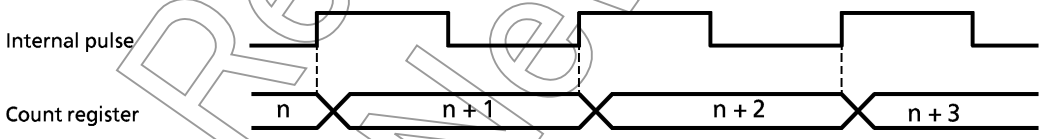


Figure 3-13. Timer Mode Timing chart

Example: To generate an overflow interrupt (at fc = 4 MHz) by the TC1 after 100 ms.

```
LD      HL,  #0F4H    ;  TC1←E79H (setting of the count register)
ST      #9,  @HL+
ST      #7,  @HL+
ST      #0EH,  @HL+
LD      A,  #1000B    ;  OP1C←1000B
OUT     A,  %OP1C
LD      A,  #0100B    ;  EIR←0100B (enables interrupt)
XCH     A,  EIR
EICLR   IL,  110111B  ;  EIF←1, IL3←0
```

* ***Calculating the preset value of the counter register***

The preset value of the count register is obtained from the following relation:

2^{12} – (interrupt setting time) × (internal pulse rate)

For example, to generate an overflow interrupt after 100 ms at fc = 4 MHz with the internal pulse rate of fc/2^{10}, set the following value to the count register as the preset value:

$2^{12} - (100 \times 10^{-3}) \times (4 \times 10^{6}/2^{10}) = 3705 = E79_{H}$

* The drop of the apparent instruction execution speed can be calculated by the following way.

$$1 \div \{ \frac{(\text{basical clock freq.}) / 8}{(\text{Internal pulse rate})} - 1 \} \times 100 \quad [\%]$$

Table 3-3. Internal Pulse Rate Selection

Internal pulse rate	Max. setting time	Example: At fc = 4.194304 MHz	
		Internal pulse rate	Max. setting time
fc/2^{6} [Hz]	2^{18} / fc [s]	65536 [Hz]	0.0625 [s]
fc/2^{8}	2^{20} / fc	16384	0.25
fc/2^{10}	2^{22} / fc	4096	1
fc/2^{14}	2^{26} / fc	256	16
fc/2^{18}	2^{30} / fc	16	256
fc/2^{22}	2^{34} / fc	1	4096

(3) Pulse width measurement mode

In the pulse width measurement mode, the timer/counter increments with the pulse obtained by sampling the external pins (T1, T2) by the internal pulse. As shown in Figure 3-14, the timer/counter increments only while the external pin input is high. The maximum applied frequency to the external pin input must be one that is enough for analyzing the count value. Normally, a frequency sufficient slower than the internal pulse rate setting is applied to the external pin.

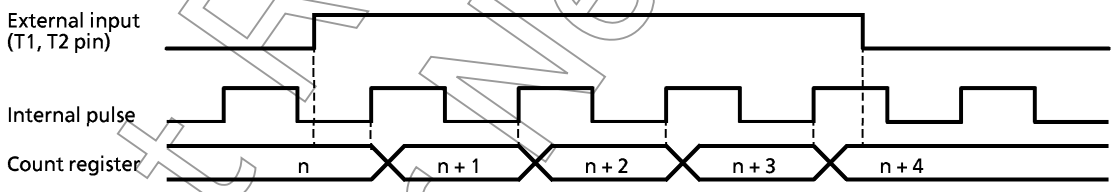


Figure 3-14. Pulse Width Measurement Mode Timing chart

3.4 AD Converter

TMP47C243/443 have a 8-bit successive approximate type AD converter and is capable of processing 8 analog inputs. Analog reference voltage can select VDD or R72 pin by mask option. In the hold mode, analog reference voltage is cut off automatically.

3.4.1 Circuit Configuration

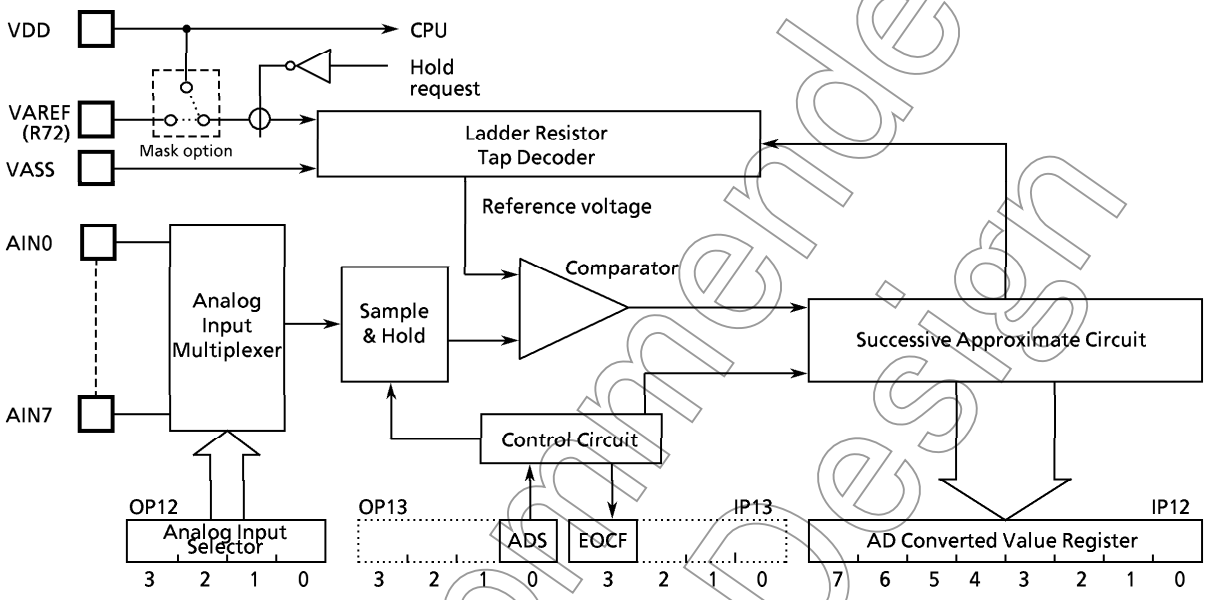


Figure 3-15. Block Diagram of AD Converter

3.4.2 Control of AD converter

The operation of AD converter is controlled by a command. The command register is accessed as port addresses OP12 and OP13. AD converted value and end of conversion flag (EOCF) can be known by accessing port addresses IP12 and IP13.

(1) Analog input selector (OP12)

Analog inputs (AIN0 through AIN7) are selected by this register.

Analog input select command register
(Port address: OP12)

3	2	1	0
SAIN			

(Initial value: 0000)

SAIN	Analog input selection
0000:	R40 (AIN0)
0001:	R41 (AIN1)
0010:	R42 (AIN2)
0011:	R43 (AIN3)
0100:	R50 (AIN4)
0101:	R51 (AIN5)
0110:	R52 (AIN6)
0111:	R53 (AIN7)
1***:	Analog input is not selected.

*Note: *; Don't care*

Figure 3-16. Analog input selector

(2) Start of AD conversion (OP13)

AD conversion is started when ADS is set to "1". After the conversion is started, ADS is cleared by hardware. If the restart is requested during the conversion, the conversion is started again at the time. Analog input voltage is hold by the sample hold circuit.

AD conversion start command register
(Port address: OP13)

3	2	1	
			ADS

(Initial value: 0000)

ADS	AD conversion enable

1: AD conversion is started (clears after starting)

Figure 3-17. AD conversion start register

(3) AD converter and register (IP13)

End of conversion flag (EOCF) is a single bit flag showing the end of conversion and is set to "1" when conversion ended. When both upper 4 bits and lower 4 bits of a converted value are read or AD conversion is started, EOCF is cleared to "0".

AD converter status register
(Port address: IP13)

3	2	1	0
EOCF			

EOCF	End of conversion flag

0: Under AD conversion or before AD conversion
1: End of AD conversion

Figure 3-18. AD converter status register

(4) AD converted value register (IP12)

An AD converted value is read by accessing port address IP12. An AD converted value is read by splitting into upper 4 bits and lower 4 bits by a value of LR_0 (LSB of the L registers).

AD converted value register
(Port address: IP12)

3	2	1	0
D3 / D7	D2 / D6	D1 / D5	D0 / D4

D7 to D0	Reading of an AD converted value

When LR_0 = 0, lower 4 bits of the converted value is read.
When LR_0 = 1, upper 4 bits of the converted value is read.

Figure 3-19. AD converted value register

3.4.3 How to use AD converter

Apply positive of analog reference voltage to the VDD pin and negative to the VSS pin. The AD conversion is carried out by splitting reference voltage between VAREF and VASS to bit corresponding voltage by a ladder resistor and making a judgement by comparing it with analog input voltage. R72 pin can use as the analog reference voltage with the mask option.

(1) Start of AD conversion

Prior to conversion, select one of the analog input AIN0 througn AIN7 by the analog input selector. Place output of the analog input, which is to be AD converted, in the high impedance state by setting "1". If other port is used as an output port, be careful not to execute the output instruction for any port during conversion in order to keep accuracy of conversion.

AD conversion is started by setting ADS (bit1 of the AD conversion start register). When conversion ends after 24 instruction cycles, EOCF showing the end of conversion is set to "1".

Analog input voltage is sampled during the following 2 instruction cycles after setting ADS.

Note: The sample and hold circuit has capacitor (C_A = 12 pF typ.) with resistor (R_A = 5 kΩ typ.). See I/O circuitry table. This capacitor should be charged or discharged within 2 instruction cycles.

(2) Reading of an AD converted value

After the end of conversion, read an AD converted value is read by splitting into lower 4 bits and upper 4 bits by the AD converted value register (IP12). Lower 4 bits of the AD converted value can be read when LR_0 = 0 and upper 4 bits when LR_0 = 1. Usually an AD converted value is stored in RAM by an instruction [IN %p, @HL]. Further, if an AD converted value is read during conversion, it becomes an indefinite value.

(3) AD conversion with HOLD operation

When the HOLD operation is started during the conversion, the conversion is terminated and an AD converted value becomes indefinite. Therefore, EOCF is kept clear to "0" after release from the HOLD operation. However, if the HOLD operation is started after the end of conversion (after EOCF has been set), AD converted value and status of EOCF are held.

Example: Selecting analog input (AIN4), starting AD conversion, monitoring EOCF, and storing lower 4 bits and upper 4 bits of a converted value to RAM [10_H] and RAM [11_H] respectively.

```
        LD      A, #4H          ; Selecting analog input (AIN4)
        OUT     A, %OP12
        LD      A, #1H          ; Start of AD conversion
        OUT     A, %OP13
SLOOP : IN      %IP13, A        ; To wait until EOCF goes to "1"
        TEST    A, 3
        B       SLOOP
        LD      HL, #10H        ; HL ← 10H
        IN      %IP12, @HL      ; RAM [10H] ← Lower 4 bits
        INC     L               ; Increment of L registers
        IN      %IP12, @HL      ; RAM [11H] ← Upper 4 bits
```

3.5 Pulse output

Pulse output is used for buzzer drive and remote control carrier. Pulse output is shared with the R71 pin. Pulse output is asynchronous.

3.5.1 Circuit Configuration

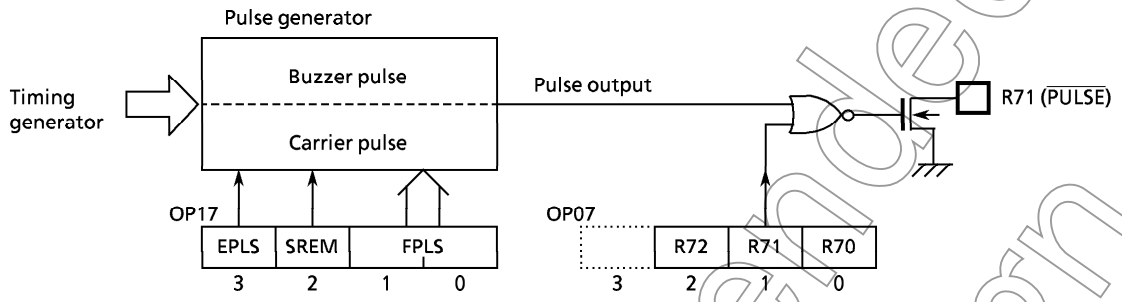


Figure 3-20. Pulse Generator

3.5.2 Control of pulse

The pulse output is controlled by the command register (OP17) and R71 output latch data (bit 1 of OP07). At reset, the OP17 is initialized to "0000_B" and pulse output is disabled. To use the pulse output, instruct start/stop of pulse after pulse output is enabled by the OP17.

Also, pulse output is "L" level (the OP17 is cleared to "0000_B") during the HOLD operating mode. External LED and so foth may be destroyed if HOLD operation is executed during output of pulse. Therefore, HOLD operating mode should be execute after pulse is stopped (after R71 output latch set to "1").

Example: Buzzer pulse of 2 kHz is output (fc = 4 MHz)

```
LD      A, #1001B
OUT     A, %OP17  ;  OP17←1001B
  :
CLR     %OP07, 1  ;  Pulse start
  :
SET     %OP07, 1  ;  Pulse stop
```

(Port address: OP17) (Initial value: 0000)

3	2	1	0
EPLS	SREM	FPLS	

EPLS	Pulse enable/disable

0: Disable
1: Enable

SREM	Selects output mode

0: Buzzer pulse (1/2 duty)
1: Remo-com carrier (1/3 duty)

PFLS	Selects output frequency

00: $fc/2^{12}$, $fc/(2^2 \times 3)$ [Hz]
01: $fc/2^{11}$, $fc/(2^4 \times 3)$
10: $fc/2^{10}$, $fc/(2^5 \times 3)$
11: $fc/2^9$, (Unused)

Buzzer Carrier

Figure 3-21. Pulse Output Control

Table 3-4. Pulse Output Frequency

FPLS	Buzzer pulse		Carrier pulse	
	Pulse rate	at fc = 4.19 MHz	Pulse rate	frequency
00	$fc/2^{12}$ [Hz]	1.024 [kHz]	$fc/(2^2 \times 3)$ [Hz]	37.9 [kHz] (fc = 455 kHz)
01	$fc/2^{11}$	2.048	$fc/(2^4 \times 3)$	37.5 (fc = 1.8 MHz)
10	$fc/2^{10}$	4.096	$fc/(2^5 \times 3)$	37.5 (fc = 3.6 MHz)
11	$fc/2^9$	8.192	Don't use	–

(1) Buzzer pulse

The buzzer pulse can be selected one of the four pulse rates by the program. The buzzer pulse is output only when the R71 output latch is "0". "H" level is output when the output latch is "1".

Note: When a piezoelectric buzzer is connected to the pin, voltage may be generated by the buzzer due to thermal or mechanical shock. In such cases, there is danger of the pin being destroyed so a zener diode should always be connected for protection.

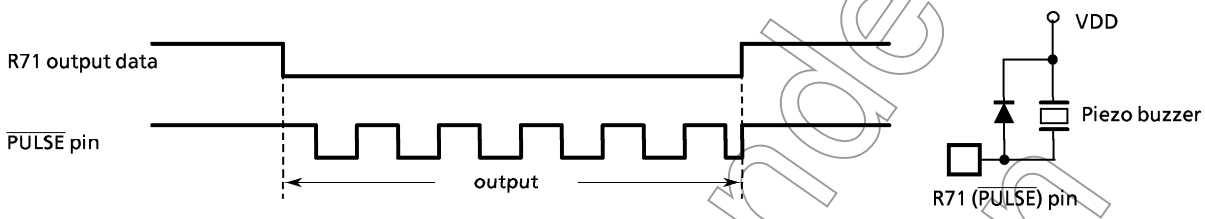


Figure 3-22. Circuit Example of Buzzer Pulse and Timing

(2) Carrier pulse for remote control signal transmitter

The remote control transmitting carrier has a frequency in Table 3-4, which is the basic clock (fc) divided by 12, 48 or 96. Also, the remote control transmitting carrier is output only when the R71 output latch is "0". "H" level is output when the output latch is "1".

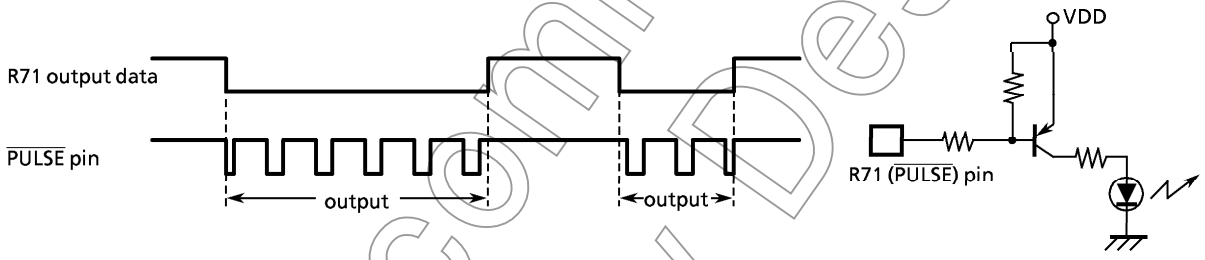


Figure 3-23. Circuit Example of Carrier Pulse and Timing

3.6 Zero-cross detector

R82 pin is used for zero-cross detection input (ZIN) and zero-cross detection can be performed by connecting and external capacitor. To use the zero-cross detector, the R82 output latch must be set to "1" (it is set to "1" during reset).

This function can be used for commercial power supply frequency input, and time base or triac control. ZIN pin is shared by the external interrupt 1. The INT1 interrupt occurs at the falling edge of the pin input by setting interrupt enable master flip-flop (EIF) to "1".

The zero-cross detector is disable and R82 pin is set to "H" level during the hold operating mode. When driving R82 pin directly without using an external capacitor, R82 is used for normal digital input or interrupt input.

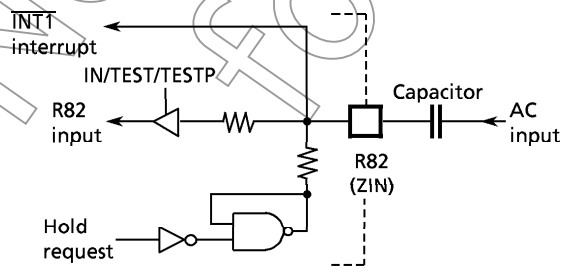


Figure 3-24. Zero-cross Detector

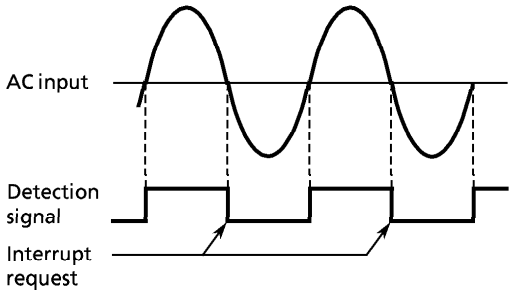


Figure 3-25. Detection Signal

3.7 Serial Interface (SIO)

The TMP47C243/443 have a serial interface with an 8-bit buffer. 4-bit/8-bit tramsfer mode can be selected. In the 8-bit transfer mode, data may be transmitted and received simultaneously. The serial interface is connected to the exterenal device via 3 pins (the serial port): R92 ($\overline{SCK}$), R91 (SO), and R90 (SI). The serial port is shared by port R9. For the serial port, the output latch of port R9 must be set to "1". In the transmit mode, R90 pin provides the I/O port; in the receive mode, R91 pin provides the I/O port.

3.7.1 Configuration of Serial Interface

Figure 3-26 shows configuration of serial interface.

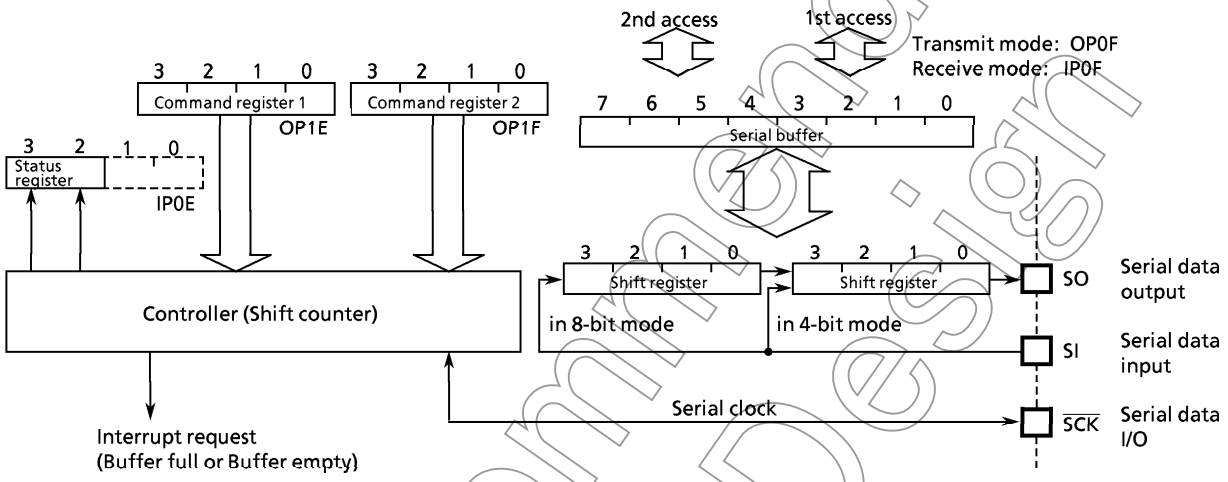


Figure 3-26. Configuration of Serial Interface

3.7.2 Control of Serial Interface

The serial interface is controlled by command registers (OP1E, OP1F). The operating states of the serial interface can be monitored by the status register (IP0E).

Serial interface control command register1
(Port address: OP1E) (Initial value: 0001)

3	2	1	0
INH	SBIT	CKR	

INH	Forcible stop of serial transfer

0: Transfer continue
1: Automatically clearing after stopped

SBIT	Transfer bit number

0: 4-bit serial transfer
1: 8-bit serial transfer

CKR	Select serial clock frequency

00: $fc/2^6$ [Hz]
01: $fc/2^7$
10: $fc/2^9$
11: $fc/2^{12}$

Note 1: When setting the transfer mode, ESIO must be "0"
Note 2: Transmit/Receive mode: Don't cara Transmit: LM = 1

Serial interface control command register 2
(Port address: OP1F) (Initial value: 0000)

3	2	1	0
ESIO	RM	LM	ECKM

ESIO	Instructs serial transfer start/end

0: Instructs serial transfer end
1: Instructs serial transfer start

RM	Select transfer mode	(Note 1)

	4 bit transfer	8 bit transfer
0:	Transmit mode	Transmit mode
1:	Receive mode	Transmit/Receive mode

LM	Select shift edge	(Note 2)

0: Shift at the trailing edge of serial clock
1: Shift at the leading edge of serial clock

ECKM	Select shift clock

0: Internal clock (output to $\overline{SCK}$ pin)
1: External clock (input from $\overline{SCK}$ pin)

Figure 3-27. Serial Interface Control Command Register

Serial interface status register
(Port address: IP0E).

3	2	1	0
SIOF	SEF		(HOLD)

SIOF	Monitor serial transfer operation state

0: Transfer is terminated
1: Transfer is in progress

SEF	Monitors shift operation status

0: Shift operation is terminated
1: Shift operation is in progress

Figure 3-28. Serial Interface Status Register

3.7.3 Serial clock

For the serial clock, one of the following can be selected according to the contents of the command registers:

(1) Clock source selection

a. Internal clock

The serial clock frequency is selected by command register 1.

The serial clock is output on the $\overline{SCK}$ pin. Note that the start of transfer, the $\overline{SCK}$ pin output goes high. When writing (transmitting) or reading (receiving) data in a program does not catch up the serial clock rate, this device stops the serial clock autmatically. Additionally it provides the wait function in which the shift is not occurred until this processing is completed.

The highest transfer rate based on the internal clock is 93750 bits/second (at fc = 6 MHz).

b. External clock

The signal obtained by the clock supplied to the $\overline{SCK}$ pin from the outside is used for the serial clock. In this case, the output latch of R92 ($\overline{SCK}$) must be set to "1" beforehand. For the shift operation to be performed correctly, each of the serial clock high and low levels needs 2 instruction cycles or more to be completed.

(2) Shift edge selection

a. Leading edge

Date is shifted at the leading edge (the falling edge of $\overline{SCK}$ pin input) of the serial clock.

b. Trailing edge

Data is shifted at the trailing edge (the rising edge of $\overline{SCK}$ pin input) of the serial clock. However, in the transmit mode, the trailing-edge shift is not supported.

3.7.4 Transfer bit number

SBIT (bit 2 of the command register 2) can select 4-bit/8-bit serial transfer.

(1) 4-bit serial transfer

In this mode, transmission/reception is performed on 4-bit basis. ISIO interrupt is generated every 4-bit transfer. Transmit/receive data is written/read by accessing the buffer register (OP0F/IP0F) respectively.

(2) 8-bit serial transfer

In this mode, transmission/reception is performed on 8-bit basis. ISIO interrupt is generated every 8-bit transfer. Transmit /receive data is written / read by accessing the buffer register (OP0F / IP0F) twice.

At the first access after setting transfer mode or generating the interrupt request, the write/read operation of lower 4-bit is performed to from the buffer register. At the second access, that of upper 4-bit is performed.

3.7.5 Transfer modes

Selection between the transmit mode, the receive mode (at transferring 4 bit) and the transmit and receive mode (at transferring 8 bit) is performed by RM (bit 2 of the command register 2). Switching the transfer modes should be implemented after specifying the end of transferring (clears ESIO to "0") and conferring the end of transferring (SIOF).

(1) Transmit mode

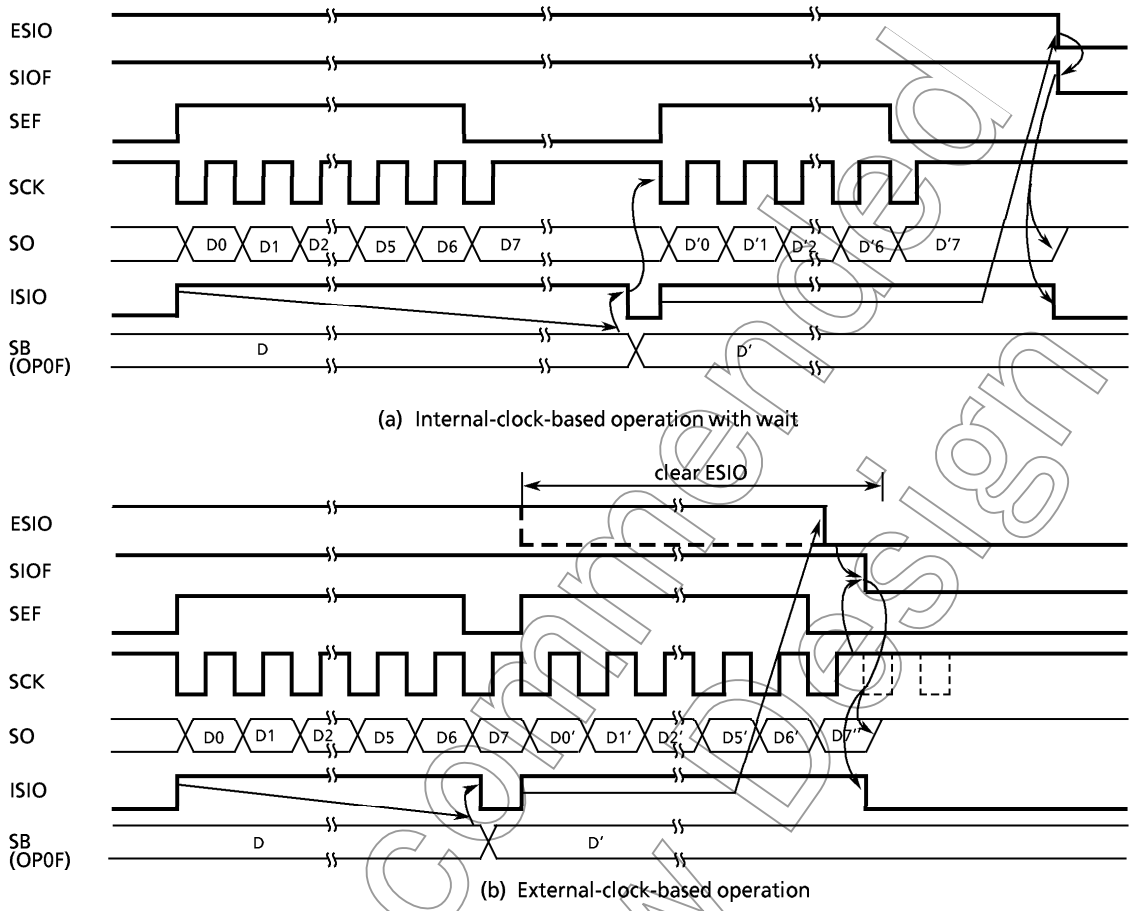
The transmit mode is set to the command register than writes the first transmit data (4 bits or 8 bits) is written to the buffer register (OP0F). (If the transmit mode is not set, the data is not written to the buffer register). In the 8-bit transfer mode, the 8-bit data is wirtten by accessing the buffer register (OP0F) twice. The transmit data is written after the 8-bit transfer mode is set or an interrupt request occurs: the lower 4 bits are written by the first access and the upper 4 bits by the next access. Then, setting ESIO to "1" starts transmission. The transmit data is output to the SO pin in synchronization with the serial clock from the LSB side sequentially. When the LSB is output, the transmit data is moved from the buffer register to the shift register. When the buffer register becomes empty, the buffer empty interrupt (ISIO) to request for the next transmit data is generated. In the interrupt service program, when the nexttransmit data tis written o the buffer register, the interrupt request is reset.

In the operation based on the internal clock, if no more data is set after the transmission of the 4-bit or 8-bit data, the serial clock is stopped and the wait state sets in. In the operation based on the external clock, the data must be set in the buffer register by the time the next data shift operation starts. Therefore, the transfer rate is determined by the maximum delay time between the occurrence of the interrupt request and the writing of data to the buffer register by the interrupt serviced program.

To end transmission, ESIO is cleared to "0" instead of writing the next transmit data by the buffer empty interrupt service program. When ESIO is cleared,transmission stops upon termination of the currently shifted-out data. The transmission end can be known by the SIOF state (SIOF goes "0" upon transmission end). In the operation based on the external clock, ESIO must be cleared to "0" before the next data is shifted out. If ESIO is not cleared before, the transmission stops upon sending the next 4-bit or 8-bit data (dummy).

Example: Transfers the data stored in the data memory (specified by DMB, HL register pair) in the internal clock operation ($fc/2^7$). (8 bit serial transferring).

```
LD      A,  #0101B
OUT     A,  %OP1E
LD      A,  #0010B
OUT     A,  %OP1F
OUT     @HL,%OP0F
INC     L
OUT     @HL,%OP0F
LD      A,  #1010B
OUT     A,  %OP1F
```



(a) Internal-clock-based operation with wait

(b) External-clock-based operation

Figure 3-29. Transmit Mode

(2) 4-bit receive mode

Data can be received when ESIO is set to "1" after setting the receive mode to the command register. The data is put from the SI pin to the shift register in synchronization with the serial clock. Then the 4/8-bit data is transferred from the shift register to the buffer register (IP0F),upon which the (buffer full)interrupt (ISIO) to reguest for readingreceived data is generated. The receive data is read from the buffer register by the interrupt service program. When the data has been read, the interrupt request is reset and the next data is put in the shift register to be transferred to the buffer register.

In the operation based on the internal clock, if the previous receive data has not been read from the buffer register at the end of capturing the next data, the serial clock is stopped and the wait operation is performed until the data has been read. In the operation based on the external clock, the shift operation is performed in synchronization with the externally-suppliedclock, so that the data must be read from the buffer register before the next receive data is transferred to it. The maximum transfer rate in the external-clock-based operation is determined by the maximum delay time between the generation of interrupt request and the reading of receive data. In the receive mode, the shift operation may be performed at either the leading edge or the trasiling edge. In the leadingedge shift operation,data is captured at the leading edge of the serial clock, so that the first shift data must be put in the SI pin before the first serial clock is applied at the start of transfer.

To end the receive operation, ESIO must be cleared to "0". When ESIO is cleared, the completion of the transfer of the current 4-bit data to the buffer register terminates the receive operation. To confirm the end of the receive operation by program, SIOF (bit 3 of the status register) must be sensed. SIOF goes "0" upon the end of receive operation.

Note: If the transfer modes are changed, the contents of the buffer register are lost. Therefore, the modes should not be changed until the last received data is read even after the end of reception is instructed (by clearing ESIO to "0").

The receive operation can be terminated in one of the following approaches determined by the transfer rate:

a. When the transfer rate is sufficiently low (the external-clock-based operation):
If ESIO can be cleared to "0" before the next serial clock is applied upon occurrence of buffer full interrupt in the external-clock-based operation, ESIO is cleared to "0" by the interrupt service program, then the last received data is read.

Example: To instruct receive end with sufficien flow transfer rate (the leading edge shift).

```
LD      A,  #0111B      ;  ESIO ← 0 (Instruct reception end)
OUT     A,  %OP1E
IN      %IP0F,A         ;  Acc ← IP0E (Reads received data)
```

b. When the transfer rate is high (the internal/external clock-based operation):
If the transfer rate is high and,therefore, it is possible that the capture of the next data starts before ESIO is cleared to "0" upon acceptance of any interrupt, ESIO must be cleared to "0" by confirming that SEF (bit 2 of the status register) is set at reading the data proceeding the last data. Then, the data is read. In the interrupt serevicing following the reception of the last data, no operation is needed for termination; only the reading of the received data is performed. This method is generally employed for the internal-clock-based operations. For an external-clock-based operation, ESIO must be cleared and the received data must be read before the last data is transferred to the buffer register.

Example: To instruct reception end when transfer rate is high (the internal clock, leading-edge shift).

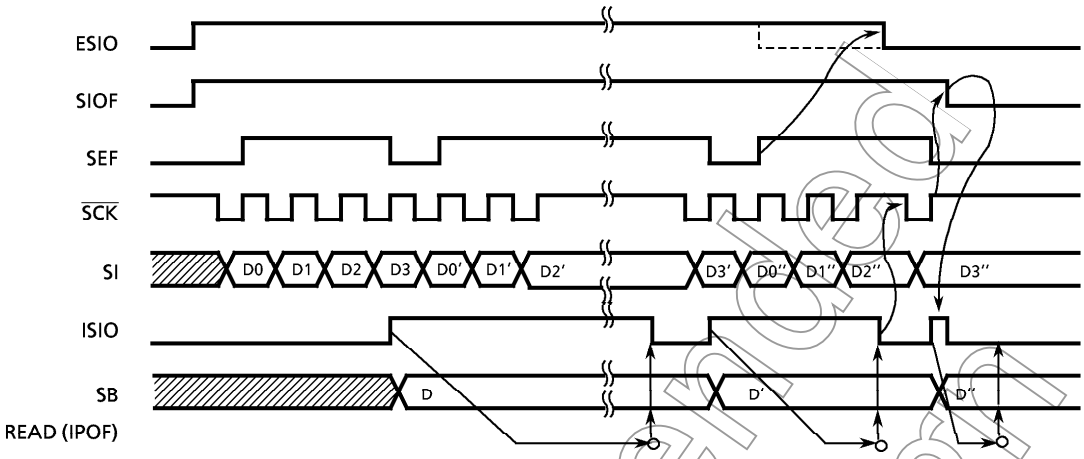
```
SSEF0 :  TEST    %IP0E,  2      ; Waits until SEF = "1"
         B       SSEF0
         LD      A,  #0110B     ; ESIO ← 0
         OUT     A,  %OP1F
         IN      %IP0F,  A      ; Acc ← IP0F (Reads received data)
```

c. One-word reception

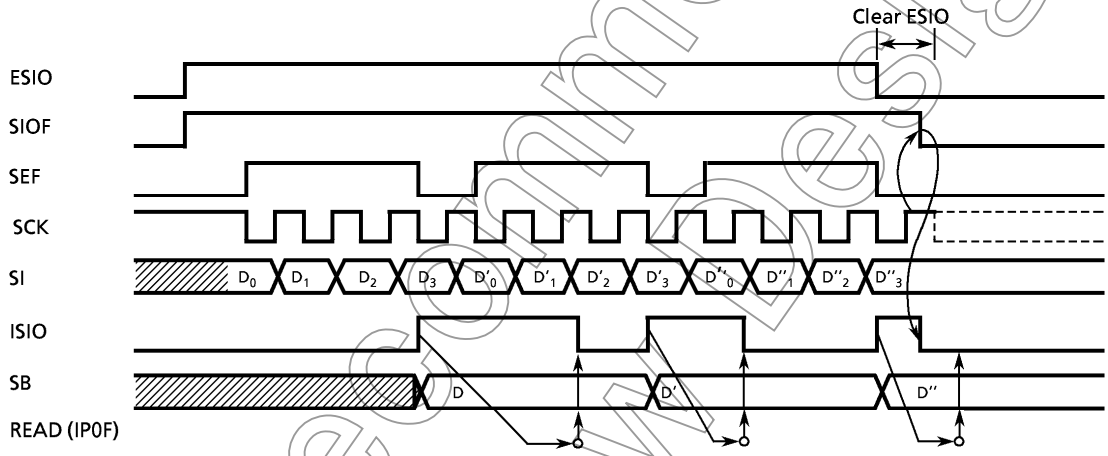
When receiving only 1 word, ESIO is set to "1" then it is cleared to "0" after confirming that SEF has gone "1". In this case, buffer full interrupt is caused only once, so that the received data is read by the interrupt service program.

Example: To instruct the start/end of 1-word reception (the internal clock, the trailing edge shift).

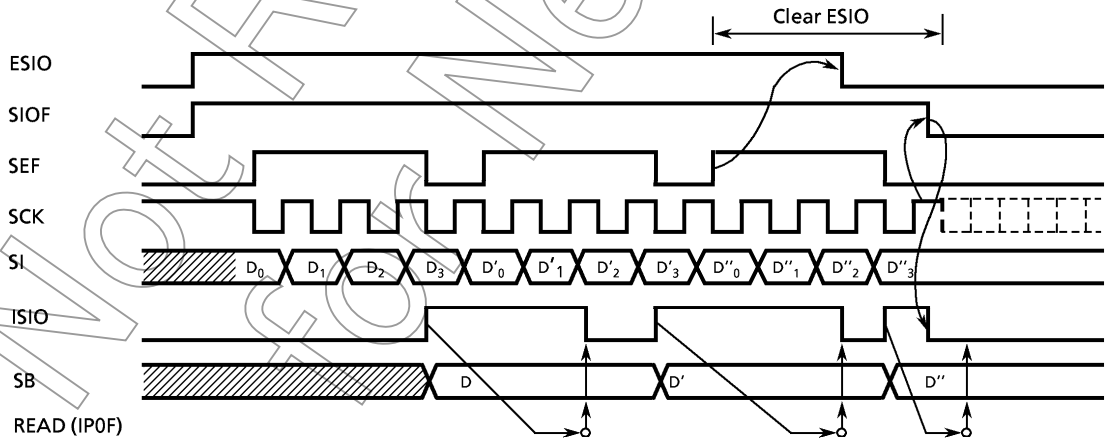
```
        LD      A,  #0100B      ; OP1F ← 0100B (Sets in the receive mode)
        OUT     A,  %OP1F
        EI                      ; EIF ← 1 (Enables interrupt)
        LD      A,  #1110B      ; ESIO ← 1 (Instructs reception start)
        OUT     A,  %OP1F
SSEF0 : TEST    %IP0E,  2       ; Confirms that SEF = "1"
        B       SSEF0
        LD      A,  #0100B      ; ESIO ← 0 (Instructs reception end)
        OUT     A,  %OP1F
```



(a) Internal-clock-based operation, trailing-edge-shift (with wait)



(b) External-clock-based operation, leading-edge shift (when transfer rate is low)



(c) Internal-clock-based operation, leading-edge-shift (when transfer rate is high)

Figure3-30. 4-bit Receive Mode

(3) 8-bit Transmit/Receive Mode

After setting the transmition/reception mode to the command register, write first transmit data into the buffer register. Then, when "1" is set to ESIO, data transmition/reception becomes possible. The transmit data is output to the SO pin at the leading edge of serial clock and the receive data is input from the SI pin at the trailing edge. If the shift register is filled with the receive data, the data is transferred to the buffer register and ISIO (buffer full) interrupt is generated to request data read. The received data is read from the buffer register by the interrupt service program, and then write the transmit data to the buffer register.

Lower order 4 bits of both transmit and receive data are read/written from/into the buffer register by first access after setting of transmition/reception mode or generation of ISIO and higher 4 bits by next access.

In the operation based on the internal clock, SIO becomes the wait state until the received data are read out and the next data to be transmitted are written.

In the operation based on the external clock, the shift operation is synchronized with the external clock ; therefore, it is necessary to read the data received and to write data to be sent next before starting the next shift operation. The maximum transfer rate using an external clock is determined by the maximum delay time between the generation of the interrupt request and the writing of the data to be transmitted after the reading of the received data.

Also, the buffer register is used for both transmission and reception, therefore, the data must be written after reading 8 bits of receive data.

This operation is ended by clearing ESIO to "0". When ESIO is cleared, this operation is ended after transfer of the current 8 bits od data to the buffer register is completed. Programs can confirm that the operation has been completed by sensing SIOF (bit 3 of the status register) because SIOF is cleared to "0" when the operation is completed.

Example 1: To write data to be transmitted and to instruct the transmit/receive start.

```
LD      A, #0110B       ; Sets the 8-bit transfer and serial clock frequency.
OUT     A, %OP1E
LD      A, #0110B       ; Sets the transmit/receive mode of internal clock
                          operation
OUT     A, %OP1F
LD      HL, #20H        ; OP0F←RAM[20H] (Writes lower 4-bit data to be
                          transmitted)
OUT     @HL, %OP0F
INC     L               ; OP0F←RAM[21H] (Writes upper 4-bit data to be
                          transmitted)
OUT     @HL, %OP0F
LD      A, #1110B       ; ESIO ← 1 (Instructs serial transfer start)
OUT     A, %OP1F
  ⋮                     ; Data transfer
```

Example 2: To read data received and to write next data to be transmitted.

```
LD      HL, #30H        ; Stores lower 4-bit data received in RAM[30H].
IN      %IP0F, @HL
INC     L               ; Stores upper 4-bit received in RAM[31H].
IN      %IP0F, @HL
LD      HL, #22H        ; Writes next lower 4-bit data to be transmitted.
OUT     @HL, %OP0F
INC     L               ; Writes next upper 4-bit data to be transmitted.
OUT     @HL, %OP0F
```

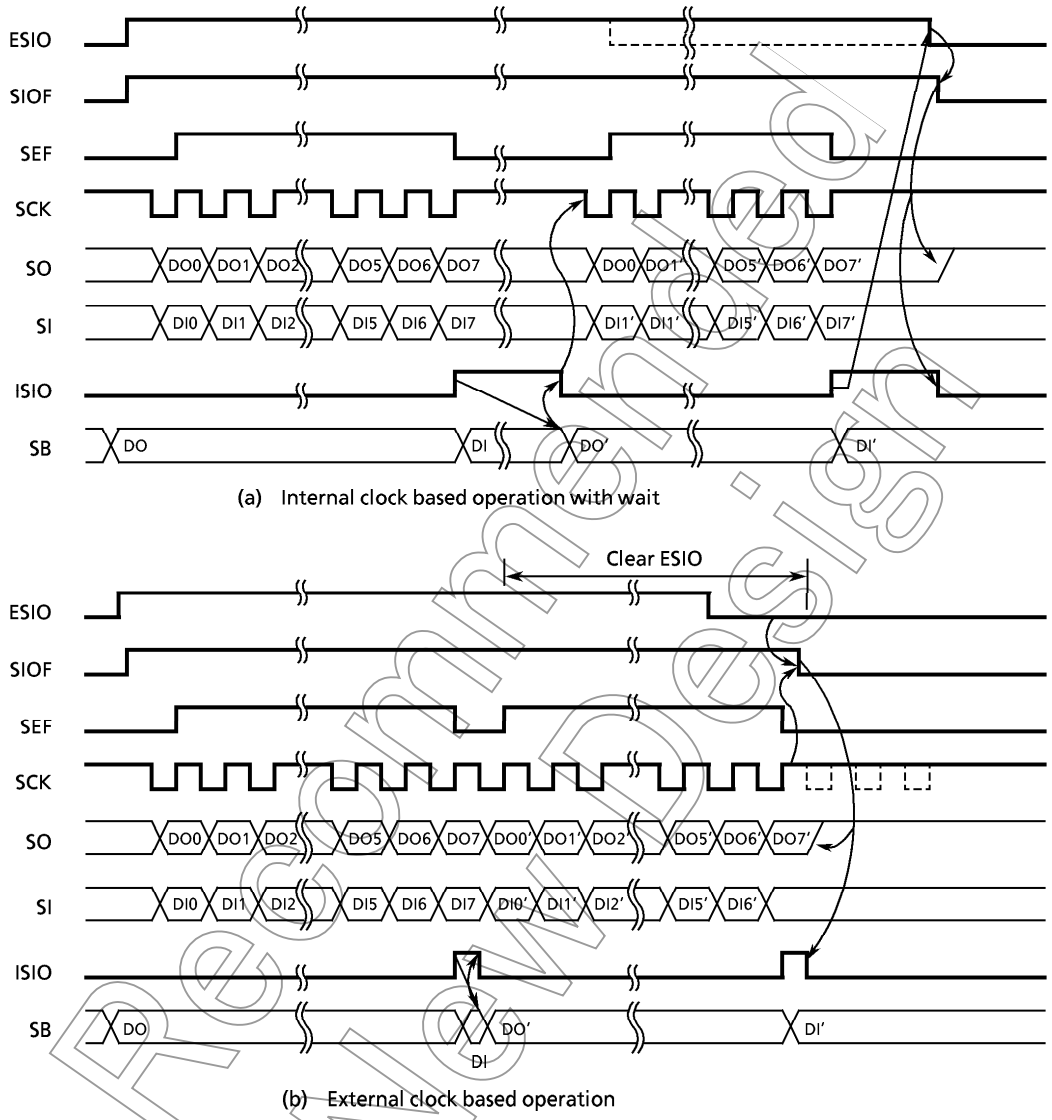


Figure 3-31. 8-bit Transmit/Receive Mode

3.7.6 Stopping serial transfer

A serial transfer operation can be stopped forcibly.

It is stopped by setting INH (bit 3 of command register 1) to "1", clearing the shift counter. When the serial transfer is over, INH is automatically cleared to "0" with no other bits of command register affected. In the transmit mode of this case, $\overline{SCK}$ and SO output are initialized to "H" level whereas the shift register is not cleared. Therefore, after the resumption of transmit, SO holds the data just before forcible stop via the shift register until the 1st shift data comes to SO.

Port Condition by RESET Operation

The transition of Port condition by RESET operation is shown as below.

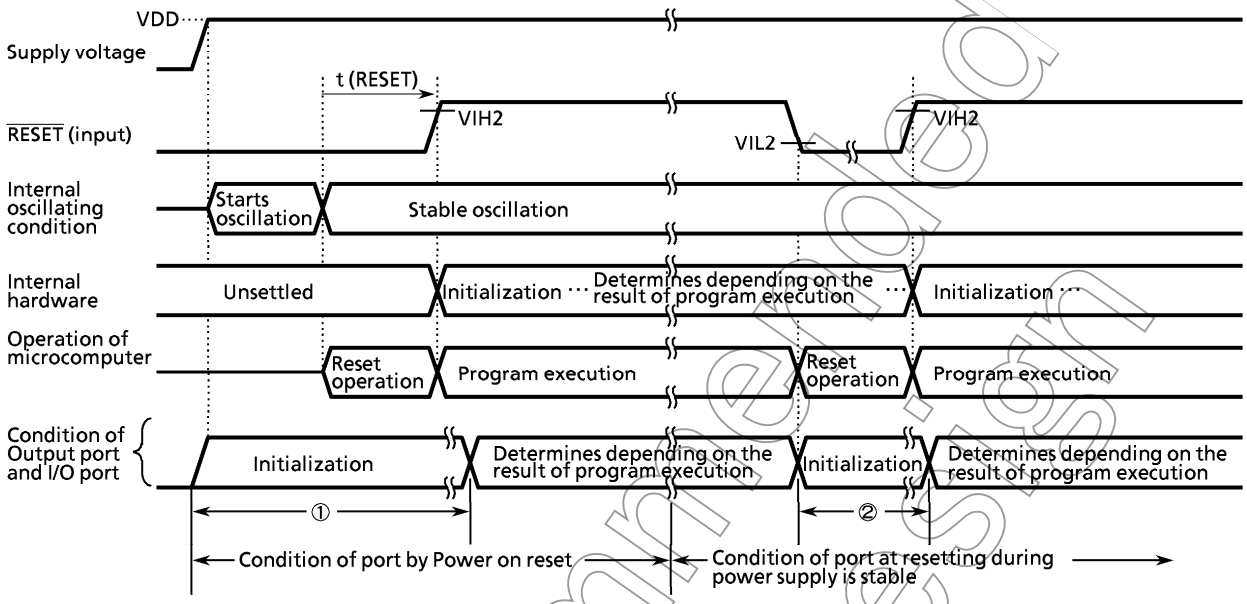


Figure 3-32. Port condition by Reset operation

Note 1: t (RESET) > 24/fc

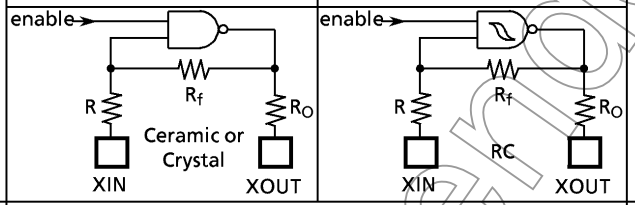
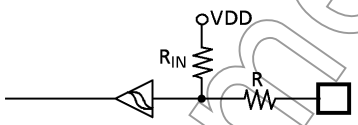
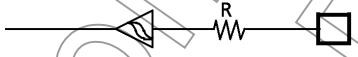
Note 2: VIL2: Stands for low level input voltage of $\overline{\text{RESET}}$ *pin.*
VIH2: Stands for high level input voltage of $\overline{\text{RESET}}$ *pin.*

Note 3: The term from power on reset to the time program is executed (above ①) and also the term starting from reset operation during power supply is stable to the program is executed (above ②), the port is on the initial condition. The initial condition of Port differs from I/O circuit by each port, refer to the section of "INPUT/OUTPUT CIRCUITRY". Thus, when using Port as an output pin, in the term of the above ① and ②, the voltage level on the signal that connects with the output pin of Port to the input pin of external application circuit should be determined by the external circuitry such as pull-up resistor and / or pull-down resistor.

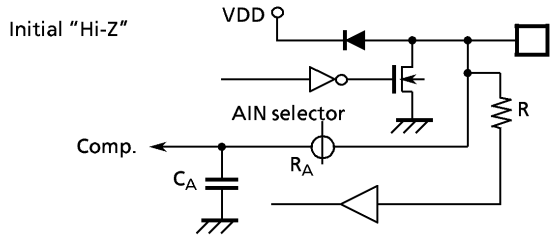
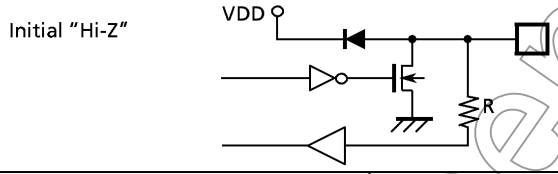
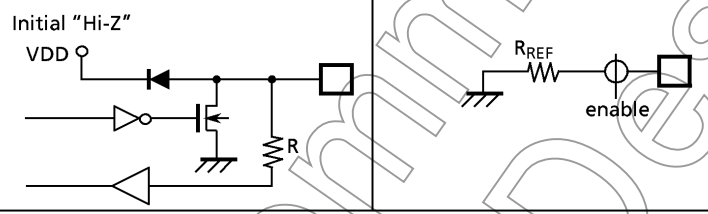
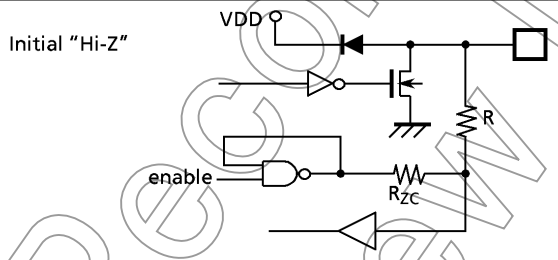
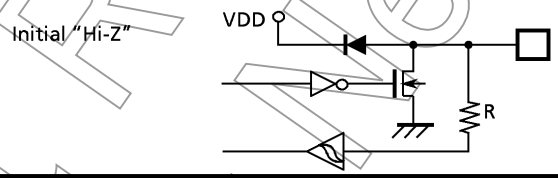
Input / Output Circuitry

The input/output circuitries of the TMP47C243/443 are shown as below, any one of the circuitries can be chosen by a code (SA, SD or SG) as a mask option.

(1) Control pins

Control Pin	I/O	Circuitry and Code	Remarks
XIN XOUT	Input Output	SA, SG / SD 	Resonator connecting pins R = 1 kΩ (typ.) R_f = 1.5 MΩ (typ.) R_O = 2 kΩ (typ.)
$\overline{\text{RESET}}$	Input		Hysteresis input Pull-up resistor R_{IN} = 220 kΩ (typ.) R = 1 kΩ (typ.)
$\overline{\text{HOLD}}$ ($\overline{\text{KE0}}$)	Input (Input)		Hysteresis input R = 1 kΩ (typ.)

(2) I/O ports

Port	I/O	Input / Output Circuitry and Code	Remarks
R4 R5	I/O		Sink open drain output R = 1 kΩ (typ.) High current (R5) I_{OL} = 20 mA (typ.) Analog input R_A = 5 kΩ (typ.) C_A = 12 pF (typ.)
R6 R70 R71	I/O		Sink open drain output R = 1 kΩ (typ.) High current (R6) I_{OL} = 20 mA (typ.)
R72	I/O	SA, SD / SG 	Sink open drain output R = 1 kΩ (typ.) Analog reference voltage suply R_{REF} = 10 KΩ(typ.)
R82	I/O		Sink open drain output R = 1 kΩ (typ.) Zero-cross input R_{ZC} = 1 MΩ (typ.)
R80 R81 R83 R9	I/O		Sink open drain output Hysteresis input R = 1 kΩ (typ.)

Electrical Characteristics

Absolute Maximum Ratings (V_{SS} = 0 V)

Parameter	Symbol	Pins		Ratings	Unit
Supply Voltage	V_{DD}			– 0.3 to 6.5	V
Input Voltage	V_{IN}			– 0.3 to V_{DD} + 0.3	V
Output Voltage	V_{OUT}			– 0.3 to V_{DD} + 0.3	V
Output Current (Per 1 pin)	I_{OUT1}	Port R5, R6		30	mA
	I_{OUT2}	Ports R4, R7, R8, R9		3.2	
Output Current (Total)	ΣI_{OUT}	Port R5, R6		120	mA
Power Dissipation [Topr = 70°C]	PD		DIP	300	mW
			SOP	180	
			SSOP	145	
Soldering Temperature (time)	Tsld			260 (10 s)	°C
Storage Temperature	Tstg			– 55 to 125	°C
Operating Temperature	Topr			– 30 to 70	°C

Note: The absolute maximum ratings are rated values which must not be exceeded during operation, even for an instant. Any one of the ratings must not be exceeded. If any absolute maximum rating is exceeded, a device may break down or its performance may be degraded, causing it to catch fire or explode resulting in injury to the user. Thus, when designing products which include this device, ensure that no absolute maximum rating value will ever be exceeded.

Recommended Operating Conditions (V_{SS} = 0 V, Topr = – 30 to 70°C)

Parameter	Symbol	Pins	Conditions	Min	Max	Unit
Supply Voltage	V_{DD}		fc = 8.0 MHz	2.7	5.5	V
			fc = 4.2 MHz	2.2		
			In the HOLD mode	2.0		
Input High Voltage	V_{IH1}	Except Hysteresis Input	In the normal operating area	V_{DD} × 0.7	V_{DD}	V
	V_{IH2}	Hysteresis Input		V_{DD} × 0.75		
	V_{IH3}		In the HOLD mode	V_{DD} × 0.9		
Input Low Voltage	V_{IL1}	Except Hysteresis Input	In the normal operating area	0	V_{DD} × 0.3	V
	V_{IL2}	Hysteresis Input			V_{DD} × 0.25	
	V_{IL3}		In the HOLD mode		V_{DD} × 0.1	
Clock Frequency	fc	XIN, XOUT	V_{DD} = 2.7 to 5.5 V	0.4	8.0	MHz
			V_{DD} = 2.2 to 5.5 V		4.2	
			In the RC oscillation		2.5	

Note: The recommended operating conditions for a device are operating conditions under which it can be guaranteed that the device will operate as specified. If the device is used under operating conditions other than the recommended operating conditions (supply voltage, operating temperature range, specified AC/DC values etc.), malfunction may occur. Thus, when designing products which include this device, ensure that the recommended operating conditions for the device are always adhered to.

DC Characteristics

(V_{SS} = 0 V, Topr = – 30 to 70°C)

Parameter	Symbol	Pins	Conditions	Min	Typ.	Max	Unit
Hysteresis Voltage	V_{HS}	Hysteresis Input		–	0.7	–	V
Input Current	I_{IN1}	$\overline{RESET}$, $\overline{HOLD}$	V_{DD} = 5.5 V, V_{IN} = 5.5 V / 0 V	–	–	± 2	μA
	I_{IN2}	Open drain output ports					
Input Resistance	R_{IN}	$\overline{RESET}$		100	220	450	kΩ
Output Leakage Current	I_{LO}	Open drain output ports	V_{DD} = 5.5 V, V_{OUT} = 5.5 V	–	–	2	μA
Output Low Voltage	V_{OL}	Port R4, R7, R8, R9	V_{DD} = 4.5 V, I_{OL} = 1.6 mA	–	–	0.4	V
			V_{DD} = 2.2 V, I_{OL} = 20 μA	–	–	0.1	
Output Low Current	I_{OL1}	Port R5, R6	V_{DD} = 4.5 V, V_{OL} = 1.0 V	7	20	–	mA
Supply Current (in the Normal operating mode)	I_{DD}		V_{DD} = 5.5 V, fc = 4 MHz	–	2	4	mA
			V_{DD} = 3.0 V, fc = 4 MHz	–	1	2	
			V_{DD} = 3.0 V, fc = 400 kHz	–	0.5	1	
Supply Current (in the HOLD operating mode)	I_{DDH}		V_{DD} = 5.5 V	–	0.5	10	μA

Note 1: Typ. values show those at Topr = 25°C, V_{DD} = 5 V.

Note 2: Input Current I_{IN1}: The current through resistor is not included.

Note 3: Supply Current: V_{IN} = 5.3 V / 0.2 V (V_{DD} = 5.5 V), 2.8 V / 0.2 V (V_{DD} = 3.0 V)

AD Conversion Characteristics

(Topr = – 30 to 70°C)

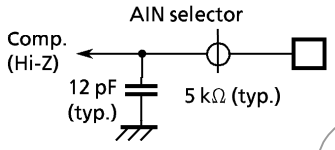
Parameter	Symbol	Conditions	Min	Typ.	Max	Unit
Analog Reference Voltage	V_{AREF}	(Mask option)	V_{DD} – 1.5	—	V_{DD}	V
Analog Reference Voltage Range	ΔV_{AREF}	V_{AREF} – V_{SS}	2.7	—	—	V
Analog Input Voltage	V_{AIN}		V_{SS}	—	V_{DD}	V
Analog Supply current	I_{REF}		—	0.5	1.0	mA
Nonlinearity Error		V_{DD} = 2.7 to 5.5 V V_{AREF} = V_{DD} ± 0.001 V V_{SS} = ± 0.001 V	—	—	± 1	LSB
Zero Point Error			—	—	± 1	
Full Scale Error			—	—	± 1	
Total Error			—	—	± 2	

AC Characteristics

(V_{SS} = 0 V, Topr = – 30 to 70°C)

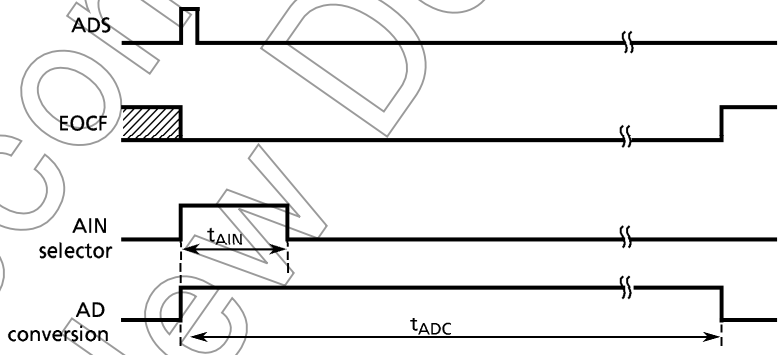
Parameter	Symbol	Conditions		Min	Typ.	Max	Unit
Instruction Cycle Time	tcy		V_{DD} = 2.7 to 5.5 V	1.0	–	20	μs
			V_{DD} = 2.2 to 5.5 V	1.9			
			RC oscillation	3.2			
High level clock pulse width	t_{WCH}	For external clock (XIN input)	$V_{DD} \geqq 2.7$ V	60	–	–	ns
			$V_{DD} < 2.7$ V	120			
Low level clock pulse width	t_{WCL}		$V_{DD} \geqq 2.7$ V	60			
			$V_{DD} < 2.7$ V	120			
AD Conversion Time	t_{ADC}			–	24 tcy	–	μs
AD Sampling Time	t_{AIN}			–	2 tcy	–	
Shift data Hold Time	t_{SDH}			0.5 tcy – 0.3	–	–	μs

Note 1: AD conversion timing:
Internal circuit for pins AIN0 to 7

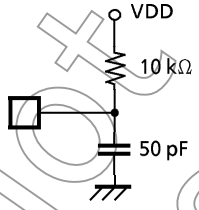


* Electrical change inust be loaded into the buit-in condensen during t_{AIN} for normal AD conversion.

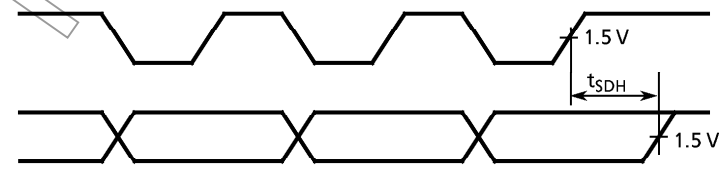
AD conversion timing



Note 2: Shift data Hold Time:
External circuit for pins
SCK and SO



Serial port (completed of transmission)

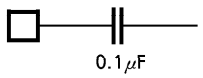


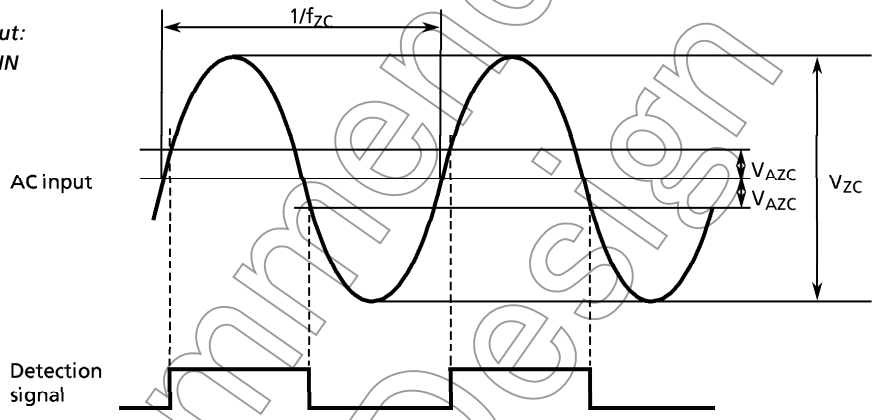
Zero-Cross Detection Characteristics

(Vss = 0 V, Topr = – 30 to 70℃)

Parameter	Symbol	Conditions	Min	Typ.	Max	Unit
Zero-cross Input Voltage	V_{ZC}	AC coupling (C = 0.1 μF)	1.0	—	3.0	V_{P-P}
Zero-cross Accuracy	V_{AZC}	f_{ZC} = 50 to 60 Hz (sine curve)	—	—	± 135	mV
Zero-cross input frequensy	f_{ZC}		40	—	1000	Hz

Note 3: Zero-cross detection input:
External circuit for pin ZIN

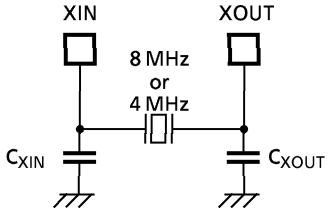




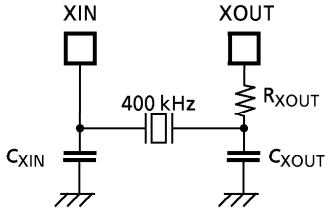
Recommended Oscillating Conditions

(V_{SS} = 0 V, V_{DD} = 2.5 to 5.5 V, Topr = – 30 to 70°C)

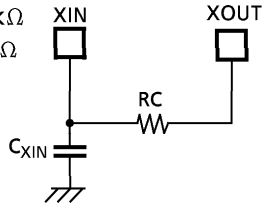
(1) 8 MHz (V_{SS} = 0 V, V_{DD} = 2.7 to 5.5 V, Topr = – 30 to 70°C)
Ceramic Resonator
KBR-800M (KYOCERA) $C_{XIN} = C_{XOUT}$ = 30 pF
Crystal Oscillator
210B-6F 8.0000 (TOYOCOM) $C_{XIN} = C_{XOUT}$ = 20 pF



(2) 4 MHz (V_{SS} = 0 V, V_{DD} = 2.2 to 5.5 V, Topr = – 30 to 70°C)
Ceramic Resonator
CSA4.00MG (MURATA) $C_{XIN} = C_{XOUT}$ = 30 pF
KBR-4.00MS (KYOCERA) $C_{XIN} = C_{XOUT}$ = 30 pF
EFOEC4004A4 (NATIONAL) $C_{XIN} = C_{XOUT}$ = 30 pF
Crystal Oscillator
204B-6F 4.0000 (TOYOCOM) $C_{XIN} = C_{XOUT}$ = 20 pF



(3) 400 kHz (V_{SS} = 0 V, V_{DD} = 2.2 to 5.5 V, Topr = – 30 to 70°C)
Ceramic Resonator
CSB400B (MURATA) $C_{XIN} = C_{XOUT}$ = 220 pF, R_{XOUT} = 6.8 kΩ
KBR-400B (KYOCERA) $C_{XIN} = C_{XOUT}$ = 100 pF, R_{XOUT} = 10 kΩ
EFOA400K04B (NATIONAL) $C_{XIN} = C_{XOUT}$ = 470 pF, R_{XOUT} = 0 Ω



(4) RC Oscillation (V_{SS} = 0 V, V_{DD} = 5.0 V, Topr = 25°C)
2 MHz (Typ.) C_{XIN} = 33 pF, R_X = 10 kΩ
400 kHz (Typ.) C_{XIN} = 100 pF, R_X = 30 kΩ

Typical Characteristics