L6393

Half-bridge gate driver

Features

- High voltage rail up to 600 V
- dV/dt immunity ± 50 V/nsec in full temperature range
- Driver current capability:
  - 270 mA source,
  - 430 mA sink
- Switching times 75/35 nsec rise/fall with 1 nF load
- 3.3 V, 5 V CMOS/TTL inputs comparators with hysteresys
- Integrated bootstrap diode
- Uncommitted comparator
- Adjustable dead-time
- Compact and simplified layout
- Bill of material reduction
- Flexible, easy and fast design

Application

Motor driver for home appliances, factory automation, industrial drives and fans. HID ballasts, power supply units.

SO-14   DIP-14
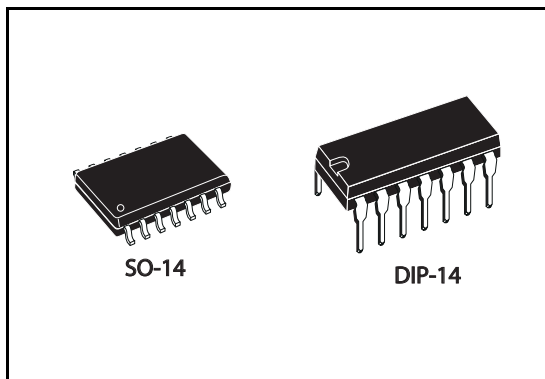
Description

The L6393 is a high-voltage device manufactured with the BCD "OFF-LINE" technology. It has a monolithic half-bridge gate driver for N-channel Power MOSFET or IGBT.

The high side (floating) section is designed to stand a voltage rail up to 600 V.

The logic inputs are CMOS/TTL compatible down to 3.3 V for easy of interfacing µC/DSP.

The IC embeds an uncommited comparator available for protections against over-current, over-temperature, etc.
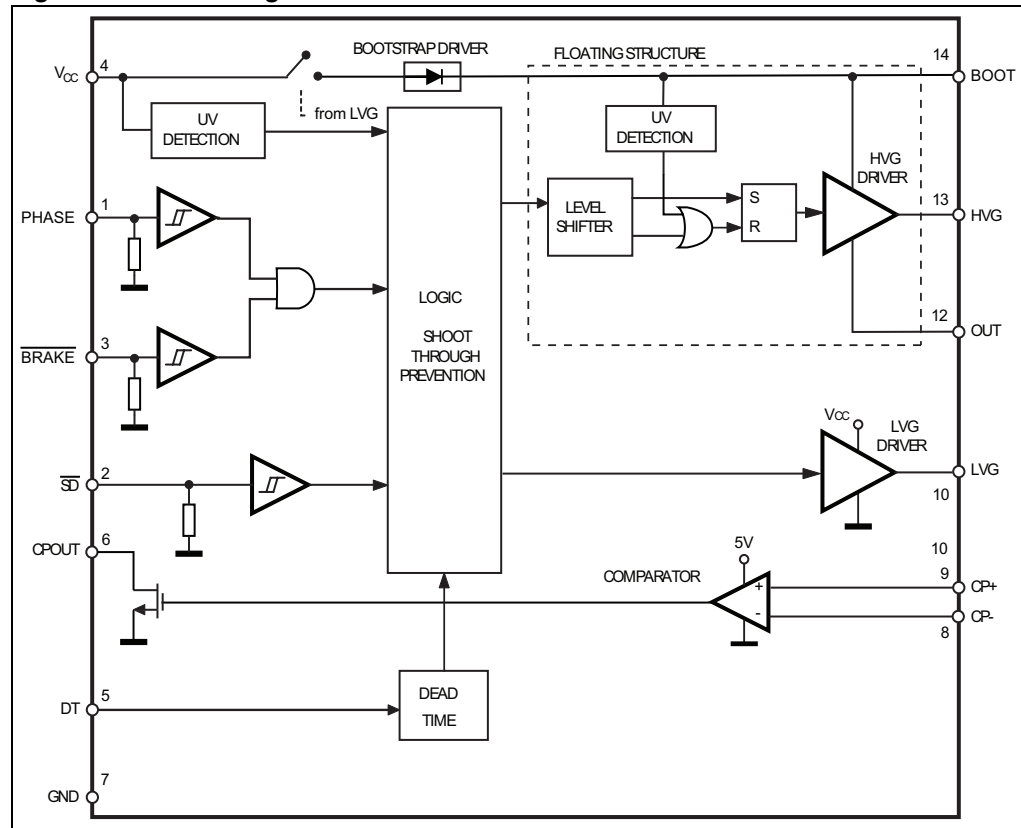
Table 1.   Device summary

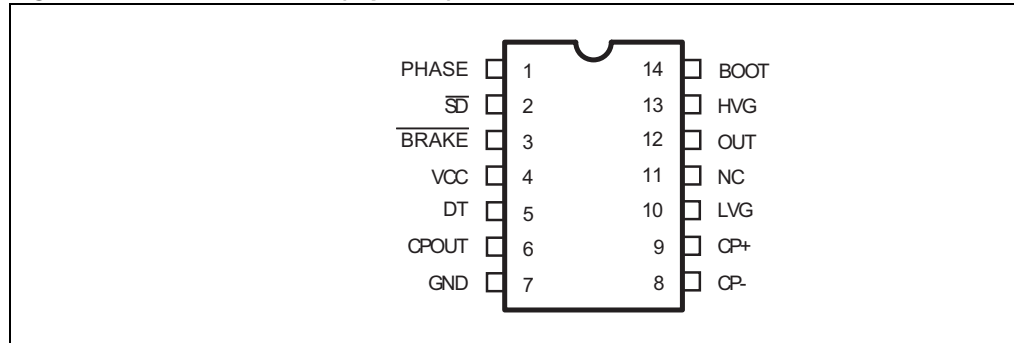| Order codes	Package	Packaging
L6393	DIP-14	Tube
L6393D	SO-14	Tube
L6393D013TR	SO-14	Tape and reel

Contents

1 Block diagram

Figure 1.   Block diagram

2 Pin connection

Figure 2. Pin connection (top view)

2.1 Pin description

Table 2. Pin description

Pin N#	Pin name	Type	Function
1	PHASE	I	Driver logic input (active high)
2	$\overline{SD}$ [1]	I	Shut down input (active low)
3	$\overline{BRAKE}$	I	Driver logic input (active low)
4	VCC	P	Lower section supply voltage
5	DT	I	Dead time setting
6	CPOUT	O	Comparator output (open drain)
7	GND	P	Ground
8	CP-	I	Comparator negative input
9	CP+	I	Comparator positive input
10	LVG [1]	O	Low side driver output
11	NC		Not connected
12	OUT	P	High side (floating) common voltage
13	HVG [1]	O	High side driver output
14	BOOT	P	Bootstrapped supply voltage

1. The circuit guarantees less than 1 V on the LVG and HVG pins (@ Isink = 10 mA), with V_{CC} > 3 V. This allows omitting the "bleeder" resistor connected between the gate and the source of the external MOSFET normally used to hold the pin low; the gate driver assures low impedance also in SD condition.

3 Truth table

Table 3.   Truth table

INPUTS			OUTPUTS	
$\overline{SD}$	PHASE	$\overline{BRAKE}$	LVG	HVG
L	X	X	L	L
H	L	L	H	L
H	L	H	H	L
H	H	L	H	L
H	H	H	L	H

Note: *X: don't care*

In the L6393 IC the two input signals PHASE and $\overline{BRAKE}$ are fed into an AND logic port and the resulting signal is in phase with the high side output HVG and in opposition of phase with the low side output LVG. This means that if $\overline{BRAKE}$ is kept to high level, the PHASE signal drives the half-bridge in phase with the HVG output and in opposition of phase with the LVG output. If $\overline{BRAKE}$ is set to low level the low side output LVG is always ON and the high side output HVG is always OFF, whatever the PHASE signal. This kind of logic interface provides the possibility to control the power stages using the PHASE signal to select the current direction in the bridge and the $\overline{BRAKE}$ signal to perform current slow decay on the low sides.

From the point of view of the logic operations the two signals PHASE and $\overline{BRAKE}$ are completely equivalent, that means the two signals can be exchanged without any change in the behavior on the resulting output signals (see the Block Diagram in Fig.1).

Note: the dead time between the turn OFF of one power switch and the turn ON of the other power switch is defined by the resistor connected between DT pin and the ground.

4 Electrical data

4.1 Absolute maximum ratings

Table 4. Absolute maximum rating

Symbol	Parameter	Value	Unit
V_{out}	Output voltage	V_{boot} - 21 to V_{boot} + 0.3	V
V_{cc}	Supply voltage	- 0.3 to + 21	V
V_{cp-}	Comparator negative input voltage	-0.3 to V_{CC} + 0.3	V
V_{cp+}	Comparator positive input voltage	-0.3 to V_{CC} + 0.3	V
V_{boot}	Floating supply voltage	V_{CC} - 0.3 to 620	V
V_{hvg}	High side gate output voltage output voltage	V_{out} - 0.3 to V_{boot} + 0.3	V
V_{lvg}	High side gate output voltage output voltage	-0.3 to V_{cc} +0.3	V
V_i	Logic input voltage	-0.3 to 15	V
V_{cpout}	Open drain voltage	-0.3 to 15	V
dV_{out}/dt	Allowed output slew rate	50	V/ns
P_{tot}	Total power dissipation (T_A = 85 °C)	TBD	mW
T_J	Junction temperature	150	°C
T_{stg}	Storage temperature	-50 to 150	°C

Note: *ESD immunity for pins 12, 13 and 14 is guaranteed up to V (Human Body Model)*

4.2 Thermal data

Table 5. Thermal data

Symbol	Parameter	SO-14	DIP-14	Unit
$R_{th(JA)}$	Thermal resistance junction to ambient max.	165	100	°C/W

4.3   Recommended operating conditions

Table 6.   Recommended operating conditions

Symbol	Pin	Parameter	Test condition	Min	Max	Unit
V_{out}	12	Output voltage [1]			580	V
V_{BS} [2]	14	Floating supply voltage [1]		TBD	TBD	V
f_{sw}		Switching frequency	HVG, LVG load C_L = 1 nF		800	kHz
V_{cc}	4	Supply Voltage		TBD	TBD	V
T_j		Junction Temperature		40	125	°C

1. If the condition TBD V < V_{boot} - V_{out} < TBD V and V_{boot} < TBD V are guaranteed, V_{out} can range from TBD V to 580 V
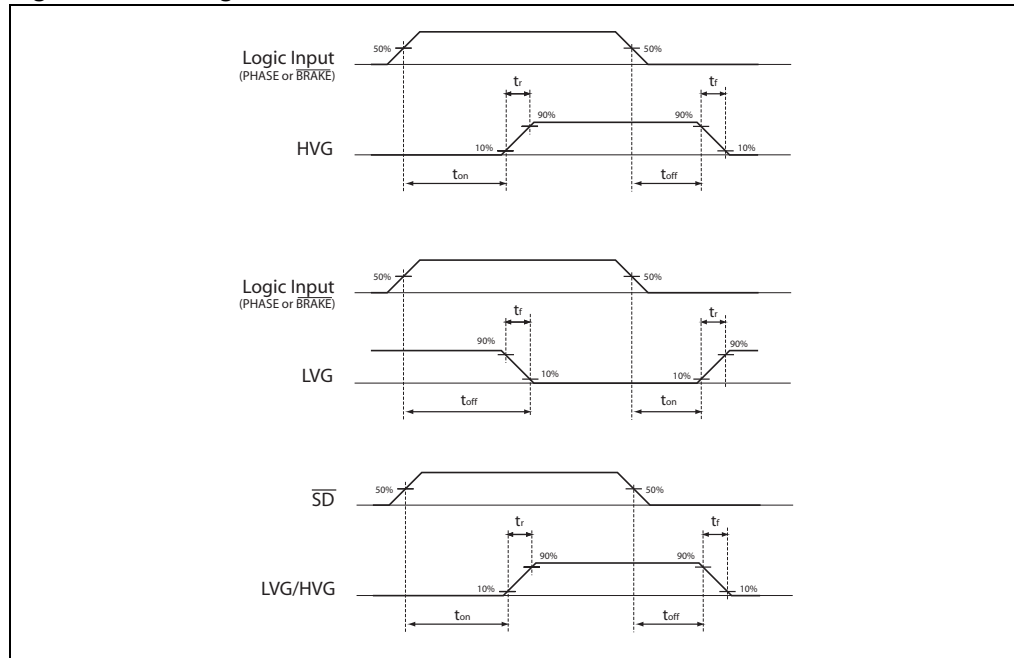
2. V_{BS} = V_{boot} -V_{out}

5 Electrical characteristics

5.1 AC operation

Table 7. AC operation electrical characteristics (V_{CC} = 15 V, T_J = +25 °C)

Symbol	Pin	Parameter	Test condition	Min	Typ	Max	Unit
AC operation							
t_{on}	1,3 vs 10, 13	High/low side driver turn-on propagation delay	V_{out} = 0 V V_{boot} = V_{cc} C_L = 1 nF V_i = 0 to 3.3 V see *Figure 3 on page 9*		125		ns
t_{off}		High/low side driver turn-off propagation delay			125		ns
t_{sd}	2 vs 10, 13	Shut down to high/low side propagation delay			125		ns
MT		Delay matching, HS & LS turn-ON/OFF				40	ns
dt	5	Dead time setting range	R_{dt} = 0, C_L = 1 nF, C_{DT} = 100 nF R_{dt} = 37 kΩ, C_L = 1 nF, C_{DT} = 100 nF R_{dt} = 136 kΩ, C_L = 1 nF, C_{DT} = 100 nF R_{dt} = 260 kΩ, C_L = 1 nF, C_{DT} = 100 nF		0.15 0.5 1.5 2.8		µs µs µs µs
MDT		Matching dead time	R_{dt} = 0 Ω; C_L = 1 nF; C_{DT} = 100 nF R_{dt} = 37 kΩ;C_L = 1 nF;C_{DT} = 100 nF R_d = 136 kΩ;C_L = 1 nF;C_{DT} = 100 nF R_{dt} = 260 kΩ;C_L = 1 nF;C_{DT} = 100 nF			60 TBD TBD TBD	ns ns ns ns
t_r	10, 13	Rise time	C_L = 1000 pF		75		ns
t_f		Fall time	C_L = 1000 pF		35		ns

Figure 3.　Timing

5.2 DC operation

Table 8. DC opereation electrical characteristics (V_{CC} = 15 V; T_J = +25 °C)

Symbol	Pin	Parameter	Test condition	Min	Typ	Max	Unit
Low supply voltage section							
V_{cc_hys}		V_{cc} UV hysteresis		600	1500		mV
V_{cc_thON}		V_{cc} UV Turn ON threshold			9.5		V
V_{cc_thOFF}		V_{cc} UV Turn OFF threshold			8.0		V
I_{qccu}	4	Undervoltage quiescent supply current	V_{CC} = 8 V $\overline{SD}$ = 5 V; PHASE and $\overline{BRAKE}$ = GND; R_{DT} = 0 Ω; CP + = GND; CP - = 0.5 V		110	150	μA
I_{qcc}		Quiescent current	V_{CC} = 15 V $\overline{SD}$ = 5 V; PHASE and $\overline{BRAKE}$ = GND; R_{DT} = 0 Ω; CP + = GND; CP - = 0.5 V		600	1000	μA
Bootstrapped supply voltage section							
V_{BS_hys}		V_{BS} UV hysteresis		600	1000		mV
V_{BS_thON}		V_{BS} UV turn ON Threshold			9.1		V
V_{BS_thOFF}		V_{BS} UV turn OFF Threshold			8.1		V
I_{QBSU}	14	Undervoltage V_{boot} quiescent current	V_{BS} = 7 V $\overline{SD}$ = 5 V; PHASE and $\overline{BRAKE}$ = 5 V; R_{DT} = 0 Ω; CP + = GND; CP - = 0.5 V		60	110	μA
I_{QBS}		V_{boot} quiescent current	V_{BS} = 15 V SD = 5 V; PHASE and $\overline{BRAKE}$ = 5 V; R_{DT} = 0 Ω; CP + = GND; CP - = 0.5 V		140	210	μA
I_{LK}		High voltage leakage current	V_{hvg} = V_{out} = V_{boot} = 600 V			10	μA
R_{dson}		Bootstrap driver on resistance [1]	LVG ON		120		Ω

Table 8.   DC opereation electrical characteristics (V_{CC} = 15 V; T_J = +25 °C) (continued)

Symbol	Pin	Parameter	Test condition	Min	Typ	Max	Unit
Driving buffers section							
I_{so}	10, 13	High/low side source short circuit current	$V_{IN} = V_{ih}$ ($t_p < 10$ μs)		270		mA
I_{si}		High/low side sink short circuit current	$V_{IN} = V_{il}$ (tp < 10 μs)		430		mA
Logic inputs							
V_{il}	1, 2, 3	Low level logic threshold voltage				0.83	V
V_{ih}		High level logic threshold voltage		2.21			V
I_{PHASEh}	1	PHASE logic "1" input bias current	PHASE = 15 V		175	260	μA
I_{PHASEl}		PHASE logic "0" input bias current	PHASE = 0 V			1	μA
I_{BRAKEh}	3	$\overline{BRAKE}$ logic "1" input bias current	$\overline{BRAKE}$ = 15 V		175	260	μA
I_{BRAKEl}		$\overline{BRAKE}$ logic "0" input bias current	$\overline{BRAKE}$ = 0 V			1	μA
I_{SDh}	2	$\overline{SD}$ logic "1" input bias current	$\overline{SD}$ = 15 V		30	100	μA
I_{SDl}		$\overline{SD}$ logic "0" input bias current	$\overline{SD}$ = 0 V			1	μA

1. R_{DSon} is tested in the following way:
 $R_{DSon} = [(V_{CC} - V_{CBOOT1}) - (V_{CC} - V_{CBOOT2})] / [I_1(V_{CC}, V_{CBOOT1}) - I_2(V_{CC}, V_{CBOOT2})]$
    where I_1 is pin 14 current when $V_{CBOOT} = V_{CBOOT1}$, I_2 when $V_{CBOOT} = V_{CBOOT2}$.

Table 9.   Sense comparator

| Symbol	Pin	Parameter	Test conditions	Min	Typ	Max	Unit
V_{io}	8, 9	Input offset voltage			±10	TBD	mV
I_{ib}		Input bias current				1	μA
V_{ol}	6	Open drain low level Output voltage	I_{od} = - 3 mA			0.5	V
t_{d_comp}		Comparator delay	C_{POUT} pulled to 5 V through 100kΩ resistor		110	210	ns
SR	6	Slew rate	C_L = 180 nF, R_{pu} = 5 kΩ		TBD		V/μs

6   Waveforms definition

Figure 4.   Dead time waveform definition
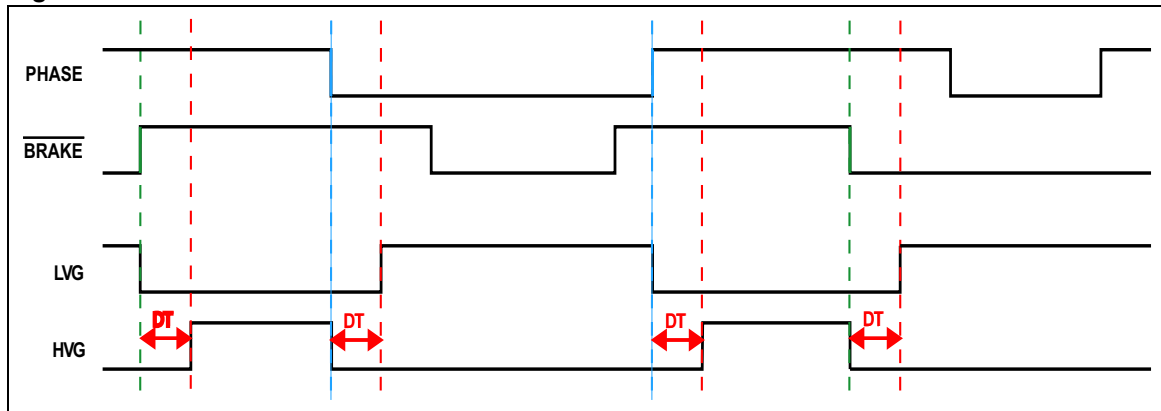
7 Typical application diagram

Figure 5.   **Application diagram**
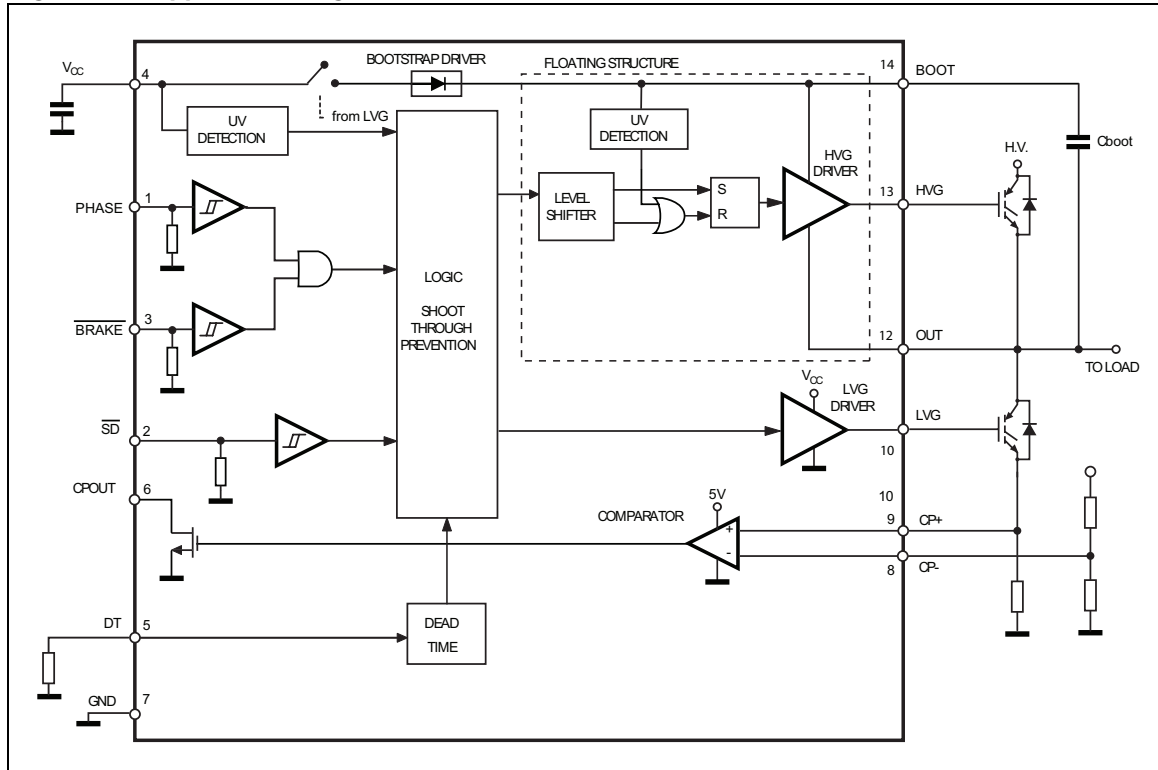
8 Bootstrap driver

A bootstrap circuitry is needed to supply the high voltage section. This function is normally accomplished by a high voltage fast recovery diode (*Figure 6.*a). In the L6393 a patented integrated structure replaces the external diode. It is realized by a high voltage DMOS, driven synchronously with the low side driver (LVG), with diode in series, as shown in *Figure 6.*b.

An internal charge pump (*Figure 6.*b) provides the DMOS driving voltage.

The diode connected in series to the DMOS has been added to avoid undesirable turn on of it.

8.1 C_{BOOT} selection and charging

To choose the proper C_{BOOT} value the external MOS can be seen as an equivalent capacitor. This capacitor C_{EXT} is related to the MOS total gate charge:

$$C_{EXT} = \frac{Q_{gate}}{V_{gate}}$$

The ratio between the capacitors C_{EXT} and C_{BOOT} is proportional to the cyclical voltage loss. It has to be:

$$C_{BOOT} >>> C_{EXT}$$

e.g.: if Q_{gate} is 30nC and V_{gate} is 10 V, C_{EXT} is 3 nF. With C_{BOOT} = 100 nF the drop would be 300 mV.

If HVG has to be supplied for a long time, the C_{BOOT} selection has to take into account also the leakage and quiescent losses.

e.g.: HVG steady state consumption is lower than 200 μA, so if HVG T_{ON} is 5 ms, C_{BOOT} has to supply 1 μC to C_{EXT}. This charge on a 1 μF capacitor means a voltage drop of 1 V.

The internal bootstrap driver gives a great advantage: the external fast recovery diode can be avoided (it usually has great leakage current).

This structure can work only if V_{OUT} is close to GND (or lower) and in the meanwhile the LVG is on. The charging time (T_{charge}) of the C_{BOOT} is the time in which both conditions are fulfilled and it has to be long enough to charge the capacitor.

The bootstrap driver introduces a voltage drop due to the DMOS R_{DSon} (typical value: 120 Ω). At low frequency this drop can be neglected. Anyway increasing the frequency it must be taken in to account.

The following equation is useful to compute the drop on the bootstrap DMOS:

$$V_{drop} = I_{charge}R_{dson} \rightarrow V_{drop} = \frac{Q_{gate}}{T_{charge}}R_{dson}$$

where Q_{gate} is the gate charge of the external power MOS, R_{DSon} is the on resistance of the bootstrap DMOS, and T_{charge} is the charging time of the bootstrap capacitor.

For example: using a power MOS with a total gate charge of 30 nC the drop on the bootstrap DMOS is about 1 V, if the T_{charge} is 5 µs. In fact:

$$V_{drop} = \frac{30nC}{5\mu s} \cdot 120\Omega \sim 0.7V$$

V_{drop} has to be taken into account when the voltage drop on C_{BOOT} is calculated: if this drop is too high, or the circuit topology doesn't allow a sufficient charging time, an external diode can be used.

Figure 6.   Bootstrap driver
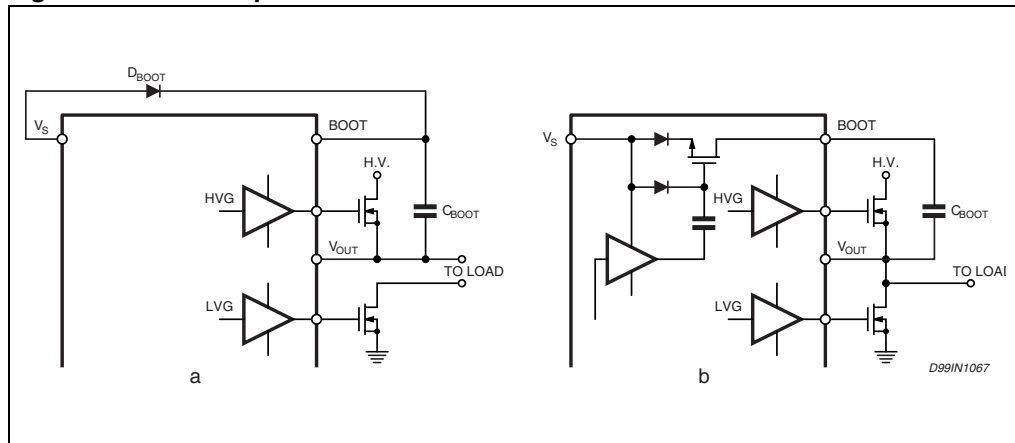
9 Package mechanical data

In order to meet environmental requirements, ST offers these devices in ECOPACK® packages. These packages have a Lead-free second level interconnect . The category of second level interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com

Figure 7. DIP-14 mechanical data and package dimensions

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
a1	0.51			0.020		
B	1.39		1.65	0.055		0.065
b		0.5			0.020	
b1		0.25			0.010	
D			20			0.787
E		8.5			0.335	
e		2.54			0.100	
e3		15.24			0.600	
F			7.1			0.280
I			5.1			0.201
L		3.3			0.130	
Z	1.27		2.54	0.050		0.100

OUTLINE AND MECHANICAL DATA

DIP-14
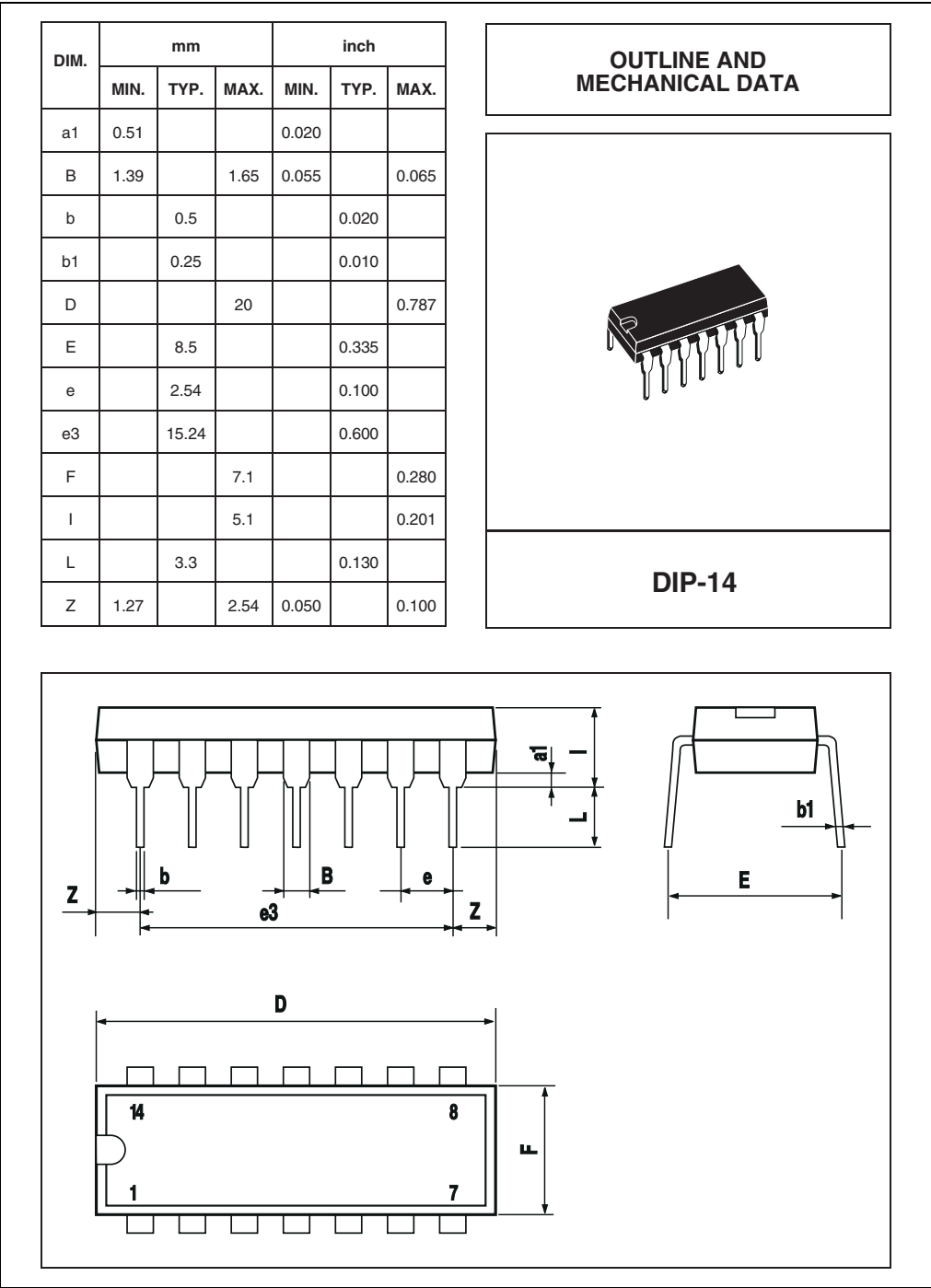
Figure 8.   SO-14 mechanical data and package dimensions

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	1.35		1.75	0.053		0.069
A1	0.10		0.30	0.004		0.012
A2	1.10		1.65	0.043		0.065
B	0.33		0.51	0.013		0.020
C	0.19		0.25	0.007		0.01
D [1]	8.55		8.75	0.337		0.344
E	3.80		4.0	0.150		0.157
e		1.27			0.050	
H	5.8		6.20	0.228		0.244
h	0.25		0.50	0.01		0.02
L	0.40		1.27	0.016		0.050
k	0° (min.), 8° (max.)					
ddd			0.10			0.004

(1) "D" dimension does not include mold flash, protusions or gate burrs. Mold flash, protusions or gate burrs shall not exceed 0.15mm per side.

OUTLINE AND MECHANICAL DATA

SO-14

0016019 D

10   Revision history

Table 10.   Document revision history

Date	Revision	Changes
03-Mar-2008	1	Initial release
18-Mar-2008	2	Cover page updated

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