SSTUP32866

1.8 V 25-bit 1 : 1 or 14-bit 1 : 2 configurable registered buffer with parity and programmable output for DDR2-800 RDIMMs

Rev. 02 — 14 September 2006 Product data sheet

1. General description

The SSTUP32866 is a 1.8 V configurable register specifically designed for use on DDR2 memory modules requiring a parity checking function. It is defined in accordance with the JEDEC standard for the SSTUA32866 and SSTUB32866 registered buffers. The register is configurable (using configuration pins C0 and C1) to two topologies: 25-bit 1 : 1 or 14-bit 1 : 2, and in the latter configuration can be designated as Register A or Register B on the DIMM. It offers added features over the JEDEC standard register in that it can be configured for high or normal output drive strength, as well as for operation to 667 MT/s or 800 MT/s, simply by tying two input pins HIGH or LOW as needed.

The SSTUP32866 accepts a parity bit from the memory controller on its parity bit (PAR_IN) input, compares it with the data received on the DIMM-independent D-inputs and indicates whether a parity error has occurred on its open-drain QERR pin (active LOW). The convention is even parity, that is, valid parity is defined as an even number of ones across the DIMM-independent data inputs combined with the parity input bit.

The SSTUP32866 is packaged in a 96-ball, 6×16 grid, 0.8 mm ball pitch LFBGA package (13.5 mm \times 5.5 mm).

2. Features

- Configurable register supporting DDR2 up to 667 MT/s or 800 MT/s Registered DIMM applications
- Configurable to 25-bit 1:1 mode or 14-bit 1:2 mode
- Programmable for normal or high output drive
- Controlled multi-impedance output drivers enable optimal signal integrity and speed
- Programmable for 667 MT/s or 800 MT/s speed
- Excellent propagation delay performance
- Supports up to 450 MHz clock frequency of operation
- Optimized pinout for high-density DDR2 module design
- Chip-selects minimize power consumption by gating data outputs from changing state
- Supports SSTL_18 data inputs
- Checks parity on the DIMM-independent data inputs
- Partial parity output and input allows cascading of two SSTUP32866s for correct parity error processing
- Differential clock (CK and CK) inputs
- Supports LVCMOS switching levels on the control and RESET inputs
- Single 1.8 V supply operation (1.7 V to 2.0 V)



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■ Available in 96-ball, 13.5 mm × 5.5 mm, 0.8 mm ball pitch LFBGA package

3. Applications

■ 667 MT/s to 800 MT/s DDR2 registered DIMMs desiring parity checking functionality

4. Ordering information

Table 1. Ordering information

Type number	Solder process	Package					
		Name	Description	Version			
SSTUP32866EC/G	Pb-free (SnAgCu solder ball compound)	LFBGA96	plastic low profile fine-pitch ball grid array package; 96 balls; body $13.5 \times 5.5 \times 1.05$ mm	SOT536-1			
SSTUP32866EC/S	Pb-free (SnAgCu solder ball compound)	LFBGA96	plastic low profile fine-pitch ball grid array package; 96 balls; body 13.5 \times 5.5 \times 1.05 mm	SOT536-1			

4.1 Ordering options

Table 2. Ordering options

Type number	Temperature range
SSTUP32866EC/G	$T_{amb} = 0 ^{\circ}C \text{ to } +70 ^{\circ}C$
SSTUP32866EC/S	$T_{amb} = 0 ^{\circ}C \text{ to } +85 ^{\circ}C$

5. Functional diagram

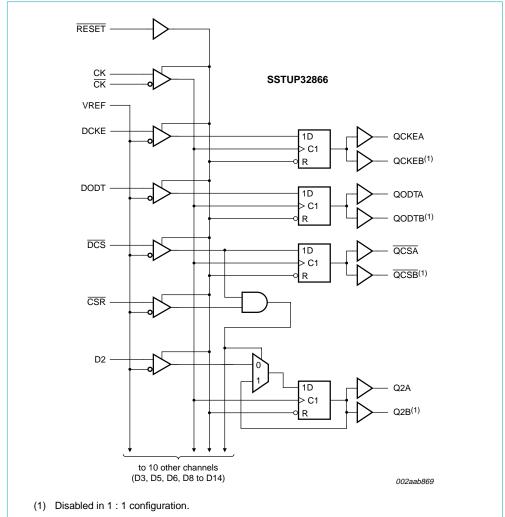
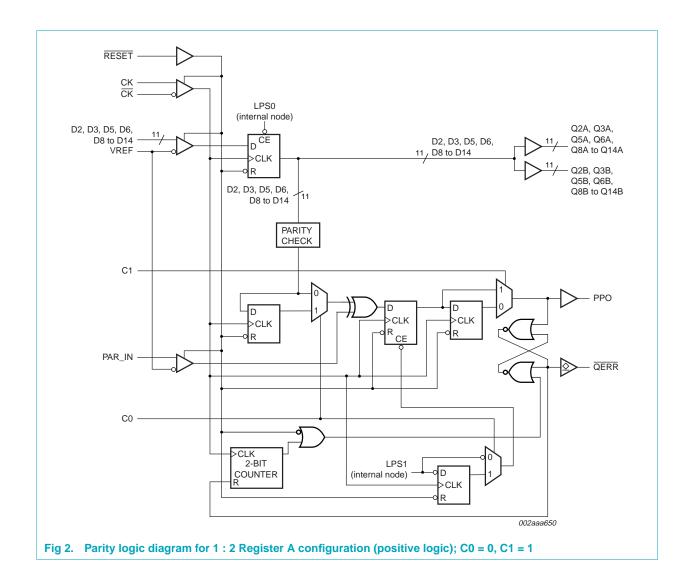


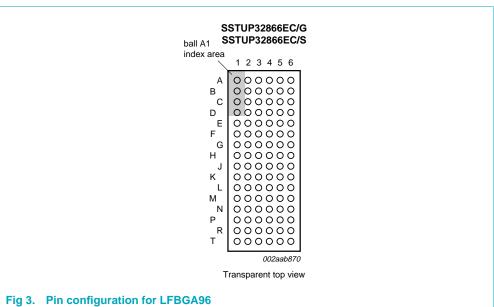
Fig 1. Functional diagram of SSTUP32866; 1 : 2 Register A configuration with C0 = 0 and C1 = 1 (positive logic)

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Pinning information

6.1 Pinning



	1	2	3	4	5	6
А	DCKE	PPO	VREF	V _{DD}	QCKE	DNU
В	D2	D15	GND	GND	Q2	Q15
С	D3	D16	V _{DD}	V_{DD}	Q3	Q16
D	DODT	QERR	GND	GND	QODT	DNU
E	D5	D17	V_{DD}	V_{DD}	Q5	Q17
F	D6	D18	GND	GND	Q6	Q18
G	PAR_IN	RESET	V_{DD}	V_{DD}	C1	C0
Н	CK	DCS	GND	GND	QCS	DNU
J	CK	CSR	V _{DD}	V _{DD}	SELAB	SELDR
K	D8	D19	GND	GND	Q8	Q19
L	D9	D20	V_{DD}	V_{DD}	Q9	Q20
М	D10	D21	GND	GND	Q10	Q21
N	D11	D22	V _{DD}	V_{DD}	Q11	Q22
Р	D12	D23	GND	GND	Q12	Q23
R	D13	D24	V _{DD}	V _{DD}	Q13	Q24
Т	D14	D25	VREF	V _{DD}	Q14	Q25
						002aab871

Fig 4. Ball mapping, 1 : 1 register (C0 = 0, C1 = 0)

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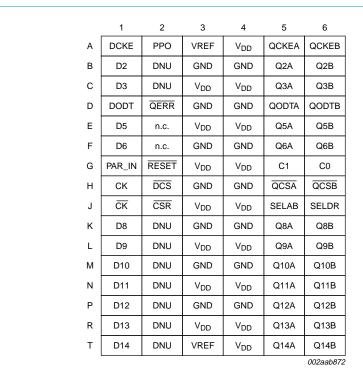


Fig 5. Ball mapping, 1 : 2 Register A (C0 = 0, C1 = 1)

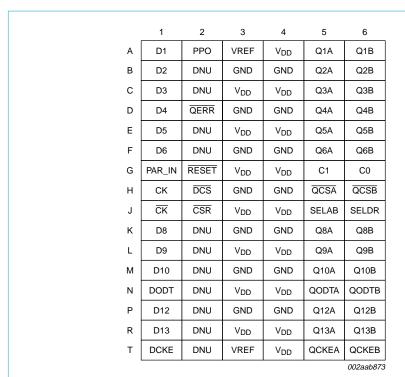


Fig 6. Ball mapping, 1 : 2 Register B (C0 = 1, C1 = 1)

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6.2 Pin description

Table 3. Pin description

Table 3. Pi	n description		
Symbol	Pin	Туре	Description
GND	B3, B4, D3, D4, F3, F4, H3, H4, K3, K4, M3, M4, P3, P4	ground input	ground
V _{DD}	A4, C3, C4, E3, E4, G3, G4, J3, J4, L3, L4, N3, N4, R3, R4, T4	1.8 V nominal	power supply voltage
VREF	A3, T3	0.9 V nominal	input reference voltage
CK	H1	differential input	positive master clock input
CK	J1	differential input	negative master clock input
C0	G6	LVCMOS inputs	Configuration control inputs; Register A
C1	G5		or Register B and 1 : 1 mode or 1 : 2 mode select.
SELDR	J6	LVCMOS input	Selects output drive strength: HIGH for normal drive, LOW for high drive. This pin will default HIGH if left open-circuit (built-in weak pull-up resistor).
SELAB	J5	LVCMOS input	Selects speed grade: HIGH for DDR2-667, LOW for DDR2-800. This pin will default HIGH if left open-circuit (built-in weak pull-up resistor).
RESET	G2	LVCMOS input	Asynchronous reset input (active LOW). Resets registers and disables VREF data and clock.
CSR	J2	SSTL_18 input	Chip select inputs (active LOW). Disables
DCS	H2		D1 to D25 ¹¹ outputs switching when both inputs are HIGH.
D1 to D25	<u>[2]</u>	SSTL_18 input	Data input. Clocked in on the crossing of the rising edge of CK and the falling edge of $\overline{\text{CK}}$.
DODT	<u>[2]</u>	SSTL_18 input	The outputs of this register bit will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.
DCKE	[2]	SSTL_18 input	The outputs of this register bit will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.
PAR_IN	G1	SSTL_18 input	Parity input. Arrives one clock cycle after the corresponding data input.
Q1 to Q25, Q2A to Q14A Q1B to Q14B		1.8 V CMOS outputs	Data outputs that are suspended by the DCS and CSR control ^[3] .
PPO	A2	1.8 V CMOS output	Partial parity out. Indicates odd parity of inputs D1 to D25[1].
$\overline{\text{QCS}}, \overline{\text{QCSA}}, \overline{\text{QCSB}}$	[2]	1.8 V CMOS output	Data output that will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.
QODT, QODT QODTB	A, [2]	1.8 V CMOS output	Data output that will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.

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Table 3.	Pin descrip	tioncontinued
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Symbol	Pin	Туре	Description
QCKE, QCKEA, QCKEB	[2]	1.8 V CMOS output	Data output that will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.
QERR	D2	open-drain output	Output error bit (active LOW). Generated one clock cycle after the corresponding data output
DNU	[2]	-	Do not use. Inputs are in standby-equivalent mode and outputs are driven LOW.

- [1] Data inputs = D2, D3, D5, D6, D8 to D25 when C0 = 0 and C1 = 0.
 Data inputs = D2, D3, D5, D6, D8 to D14 when C0 = 0 and C1 = 1.
 Data inputs = D1 to D6, D8 to D10, D12, D13 when C0 = 1 and C1 = 1.
- [2] Depends on configuration. See Figure 4, Figure 5, and Figure 6 for ball number.
- [3] Data outputs = Q2, Q3, Q5, Q6, Q8 to Q25 when C0 = 0 and C1 = 0.
 Data outputs = Q2, Q3, Q5, Q6, Q8 to Q14 when C0 = 0 and C1 = 1.
 Data outputs = Q1 to Q6, Q8 to Q10, Q12, Q13 when C0 = 1 and C1 = 1.

7. Functional description

The SSTUP32866 is a 25-bit 1 : 1 or 14-bit 1 : 2 configurable registered buffer with parity, designed for 1.7 V to 2.0 V V_{DD} operation. Additionally, the SSTUP32866 can be programmed to deliver either normal or high output drive, and either 600 MT/s or 800 MT/s speeds.

Two programming pins, SELAB and SELDR, allow the user to respectively select speed and drive strength options by tying these pins either LOW or HIGH on the DIMM. The truth table for these options is shown in Table 6.

All clock and data inputs are compatible with the JEDEC standard for SSTL_18. The control and reset (RESET) inputs are LVCMOS. All data outputs are 1.8 V CMOS drivers that have been optimized to drive the DDR2 DIMM load, and meet SSTL_18 specifications. The error (QERR) output is 1.8 V open-drain driver.

The SSTUP32866 operates from a differential clock (CK and $\overline{\text{CK}}$). Data are registered at the crossing of CK going HIGH, and CK going LOW.

The C0 input controls the pinout configuration for the 1 : 2 pinout from A configuration (when LOW) to B configuration (when HIGH). The C1 input controls the pinout configuration from 25-bit 1 : 1 (when LOW) to 14-bit 1 : 2 (when HIGH).

The SSTUP32866 accepts a parity bit from the memory controller on its parity bit (PAR_IN) input, compares it with the data received on the DIMM-independent D-inputs and indicates whether a parity error has occurred on its open-drain QERR pin (active LOW). The convention is even parity, that is, valid parity is defined as an even number of ones across the DIMM-independent data inputs combined with the parity input bit.

When used as a single device, the C0 and C1 inputs are tied LOW. In this configuration, parity is checked on the PAR_IN input which arrives one cycle after the input data to which it applies. The Partial-Parity-Out (PPO) and $\overline{\text{QERR}}$ signals are produced three cycles after the corresponding data inputs.

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When used in pairs, the C0 input of the first register is tied LOW and the C0 input of the second register is tied HIGH. The C1 input of both registers are tied HIGH. Parity, which arrives one cycle after the data input to which it applies, is checked on the PAR_IN input of the first device. The PPO and $\overline{\text{QERR}}$ signals are produced on the second device three clock cycles after the corresponding data inputs. The PPO output of the first register is cascaded to the PAR_IN of the second register. The $\overline{\text{QERR}}$ output of the first register is left floating and the valid error information is latched on the $\overline{\text{QERR}}$ output of the second register.

If an error occurs and the $\overline{\text{QERR}}$ output is driven LOW, it stays latched LOW for two clock cycles or until $\overline{\text{RESET}}$ is driven LOW. The DIMM-dependent signals (DCKE, $\overline{\text{DCS}}$, DODT, and $\overline{\text{CSR}}$) are not included in the parity check computation.

The device supports low-power standby operation. When RESET is LOW, the differential input receivers are disabled, and undriven (floating) data, clock and reference voltage (VREF) inputs are allowed. In addition, when RESET is LOW all registers are reset, and all outputs are forced LOW. The LVCMOS RESET input must always be held at a valid logic HIGH or LOW level.

The device also supports low-power active operation by monitoring both system chip select (\overline{DCS} and \overline{CSR}) inputs and will gate the Qn and PPO outputs from changing states when both \overline{DCS} and \overline{CSR} inputs are HIGH. If either \overline{DCS} or \overline{CSR} input is LOW, the Qn and PPO outputs will function normally. The \overline{RESET} input has priority over the \overline{DCS} and \overline{CSR} control and when driven LOW will force the Qn and PPO outputs LOW, and the \overline{QERR} output HIGH. If the \overline{DCS} control functionality is not desired, then the \overline{CSR} input can be hard-wired to ground, in which case, the setup time requirement for \overline{DCS} would be the same as for the other Dn data inputs. To control the low-power mode with \overline{DCS} only, then the \overline{CSR} input should be pulled up to V_{DD} through a pull-up resistor.

To ensure defined outputs from the register before a stable clock has been supplied, RESET must be held in the LOW state during power-up.

In the DDR2 RDIMM application, \overline{RESET} is specified to be completely asynchronous with respect to CK and \overline{CK} . Therefore, no timing relationship can be guaranteed between the two. When entering reset, the register will be cleared and the Qn outputs will be driven LOW quickly, relative to the time to disable the differential input receivers. However, when coming out of reset, the register will become active quickly, relative to the time to enable the differential input receivers. As long as the data inputs are LOW, and the clock is stable during the time from the LOW-to-HIGH transition of \overline{RESET} until the input receivers are fully enabled, the design of the SSTUP32866 must ensure that the outputs will remain LOW, thus ensuring no glitches on the output.

7.1 Function table

Table 4. Function table (each flip-flop)

L = LOW voltage level; H = HIGH voltage level; X = don't care; $\uparrow = LOW$ -to-HIGH transition; $\downarrow = HIGH$ -to-LOW transition.

		In	puts				Outputs[1]	
RESET	DCS	CSR	CK	CK	Dn, DODTn, DCKEn	Qn	QCS	QODT, QCKE
Н	L	L	\uparrow	\downarrow	L	L	L	L
Н	L	L	↑	\downarrow	Н	Н	L	Н
Н	L	L	L or H	L or H	Х	Q_0	Q_0	Q_0
Н	L	Н	↑	\downarrow	L	L	L	L
Н	L	Н	\uparrow	\downarrow	Н	Н	L	Н
Н	L	Н	L or H	L or H	Х	Q_0	Q_0	Q_0
Н	Н	L	1	\	L	L	Н	L
Н	Н	L	\uparrow	\downarrow	Н	Н	Н	Н
Н	Н	L	L or H	L or H	Х	Q_0	Q_0	Q_0
Н	Н	Н	\uparrow	\	L	Q_0	Н	L
Н	Н	Н	↑	\downarrow	Н	Q_0	Н	Н
Н	Н	Н	L or H	L or H	Х	Q_0	Q_0	Q_0
L	X or floating	L	L	L				

^[1] Q_0 is the previous state of the associated output.

Table 5. Parity and standby function table

L = LOW voltage level; H = HIGH voltage level; X = don't care; $\uparrow = LOW$ -to-HIGH transition; $\downarrow = HIGH$ -to-LOW transition.

			Inputs				Outp	outs[1]
RESET	DCS	CSR	CK	CK	Σ of inputs = H (D1 to D25)	PAR_IN[2]	PPO[3]	QERR[4]
Н	L	Χ	1	\downarrow	even	L	L	Н
Н	L	Χ	1	\downarrow	odd	L	Н	L
Н	L	Χ	1	\downarrow	even	Н	Н	L
Н	L	Χ	1	\downarrow	odd	Н	L	Н
Н	Н	L	1	\downarrow	even	L	L	Н
Н	Н	L	1	\downarrow	odd	L	Н	L
Н	Н	L	1	\downarrow	even	Н	Н	L
Н	Н	L	1	\downarrow	odd	Н	L	Н
Н	Н	Н	1	\downarrow	X	Χ	PPO_0	\overline{QERR}_0
Н	Χ	Χ	L or H	L or H	X	Χ	PPO_0	\overline{QERR}_0
L	X or floating	X or floating	L	Н				

^[1] PPO_0 is the previous state of output PPO; \overline{QERR}_0 is the previous state of output \overline{QERR} .

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 ^[2] Data inputs = D2, D3, D5, D6, D8 to D25 when C0 = 0 and C1 = 0.
 Data inputs = D2, D3, D5, D6, D8 to D14 when C0 = 0 and C1 = 1.
 Data inputs = D1 to D6, D8 to D10, D12, D13 when C0 = 1 and C1 = 1.

^[3] PAR_IN arrives one clock cycle (C0 = 0), or two clock cycles (C0 = 1), after the data to which it applies.

^[4] This condition assumes QERR is HIGH at the crossing of CK going HIGH and CK going LOW. If QERR is LOW, it stays latched LOW for two clock cycles or until RESET is driven LOW.

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Table 6. Speed and drive programmability

Inputs		Mode	
SELDR	SELAB		
L	L	DDR2-800; high output drive	
L	Н	DDR2-667; high output drive	
Н	L	DDR2-800; normal output drive	
Н	Н	DDR2-667; normal output drive	

8. Limiting values

Table 7. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V_{DD}	supply voltage		-0.5	+2.5	V
V_{I}	input voltage	receiver	$-0.5^{[1]}$	+2.5 ^[2]	V
V_O	output voltage	driver	-0.5 ^[1]	$V_{DD} + 0.5^{[2]}$	V
I_{IK}	input clamping current	$V_I < 0 \text{ V or } V_I > V_{DD}$	-	-50	mA
I_{OK}	output clamping current	$V_O < 0 \text{ V or } V_O > V_{DD}$	-	±50	mA
Io	output current	continuous; 0 V < $V_O < V_{DD}$	-	±50	mA
I _{CCC}	continuous current through each V_{DD} or GND pin		-	±100	mA
T _{stg}	storage temperature		-65	+150	°C
V_{esd}	electrostatic discharge voltage	Human Body Model (HBM); 1.5 k Ω ; 100 pF	2	-	kV
		Machine Model (MM); 0 Ω ; 200 pF	200	-	V

^[1] The input and output negative voltage ratings may be exceeded if the input and output clamping current ratings are observed.

^[2] This value is limited to 2.5 V maximum.

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9. Recommended operating conditions

Table 8. Recommended operating conditions

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V_{DD}	supply voltage			1.7	-	2.0	V
V_{ref}	reference voltage			$0.49 \times V_{DD}$	$0.50 \times V_{DD}$	$0.51 \times V_{DD}$	V
V_{T}	termination voltage			$V_{ref} - 0.040$	V_{ref}	$V_{ref} + 0.040$	V
V_{I}	input voltage			0	-	V_{DD}	V
$V_{\text{IH(AC)}}$	AC HIGH-level input voltage	data (Dn), CSR, and PAR_IN inputs		V _{ref} + 0.250	-	-	V
$V_{\text{IL}(AC)}$	AC LOW-level input voltage	data (Dn), CSR, and PAR_IN inputs		-	-	V _{ref} – 0.250	V
$V_{\text{IH(DC)}}$	DC HIGH-level input voltage	data (Dn), CSR, and PAR_IN inputs		V _{ref} + 0.125	-	-	V
$V_{\text{IL}(DC)}$	DC LOW-level input voltage	data (Dn), CSR, and PAR_IN inputs		-	-	V _{ref} – 0.125	V
V_{IH}	HIGH-level input voltage	RESET, Cn, SELAB, SELDR	<u>[1]</u>	$0.65 \times V_{DD}$	-	-	V
V _{IL}	LOW-level input voltage	RESET, Cn, SELAB, SELDR	<u>[1]</u>	-	-	$0.35 \times V_{DD}$	V
V_{ICR}	common mode input voltage range	CK, $\overline{\text{CK}}$	[2]	0.675	-	1.125	V
V_{ID}	differential input voltage	CK, CK	[2]	600	-	-	mV
I _{OH}	HIGH-level output current	SELDR either HIGH or LOW		-	-	-8	mA
I _{OL}	LOW-level output current	SELDR either HIGH or LOW		-	-	8	mA
T _{amb}	ambient temperature	operating in free air					
		SSTUP32866EC/G		0	-	70	°C
		SSTUP32866EC/S		0	-	85	°C

^[1] The RESET and Cn inputs of the device must be held at valid levels (not floating) to ensure proper device operation.

^[2] The differential inputs must not be floating, unless RESET is LOW.

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10. Characteristics

Table 9. Characteristics

At recommended operating conditions (see Table 8); unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{OH}	HIGH-level output voltage	$I_{OH} = -6 \text{ mA}; V_{DD} = 1.7 \text{ V}$	1.2	-	-	V
V_{OL}	LOW-level output voltage	$I_{OL} = 6 \text{ mA}; V_{DD} = 1.7 \text{ V}$	-	-	0.5	V
I _I	input current	all inputs; $V_I = V_{DD}$ or GND; $V_{DD} = 2.0 \text{ V}$	-	-	±5	μΑ
I_{DD}	supply current	static Standby mode; $\overline{RESET} = GND$; I _O = 0 mA; V _{DD} = 2.0 V	-	-	100	μΑ
			-	-	40	mA
I _{DDD}	dynamic operating current per MHz	clock only; $\overline{RESET} = V_{DD}$; $V_I = V_{IH(AC)}$ or $V_{IL(AC)}$; CK and \overline{CK} switching at 50 % duty cycle; $I_O = 0$ mA; $V_{DD} = 1.8$ V	-	16	-	μΑ
		per each data input, 1 : 1 mode; $\overline{RESET} = V_{DD}; \ V_I = V_{IH(AC)} \ or \ V_{IL(AC)};$ CK and \overline{CK} switching at 50 % duty cycle; one data input switching at half clock frequency, 50 % duty cycle; $I_O = 0$ mA; $V_{DD} = 1.8 \ V$	-	11	-	μΑ
		per each data input, 1 : 2 mode; $\overline{RESET} = V_{DD}$; $V_I = V_{IH(AC)}$ or $V_{IL(AC)}$; CK and \overline{CK} switching at 50 % duty cycle; one data input switching at half clock frequency, 50 % duty cycle; $I_O = 0$ mA; $V_{DD} = 1.8$ V	-	19	-	μА
Ci	input capacitance	data and $\overline{\text{CSR}}$ inputs; V _I = V _{ref} ± 250 mV; V _{DD} = 1.8 V	2.5	-	3.5	pF
		CK and $\overline{\text{CK}}$ inputs; $V_{\text{ICR}} = 0.9 \text{ V}$; $V_{\text{i(p-p)}} = 600 \text{ mV}$; $V_{\text{DD}} = 1.8 \text{ V}$	2	-	3	pF
		\overline{RESET} input; $V_I = V_{DD}$ or GND; $V_{DD} = 1.8 \text{ V}$	3	-	4	pF
Z _o	output impedance	normal drive; instantaneous	[1] -	15	-	Ω
		normal drive; steady-state	-	53	-	Ω
		high drive; instantaneous	<u>[1]</u> _	7	-	Ω
		high drive; steady-state	-	53	-	Ω

^[1] Instantaneous is defined as within < 2 ns following the output data transition edge.

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Table 10. Timing requirements

At recommended operating conditions (see Table 8), unless otherwise specified. See Section 11.1.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
f_{clock}	clock frequency			-	-	450	MHz
t_W	pulse width	CK, CK HIGH or LOW		1	-	-	ns
t_{ACT}	differential inputs active time		[1][2]	-	-	10	ns
t _{INACT}	differential inputs inactive time		[1][3]	-	-	15	ns
t _{su} setup time	setup time	$\overline{\text{DCS}}$ before CK \uparrow , $\overline{\text{CK}}\downarrow$, $\overline{\text{CSR}}$ HIGH; $\overline{\text{CSR}}$ before CK \uparrow , $\overline{\text{CK}}\downarrow$, $\overline{\text{DCS}}$ HIGH		0.6	-	-	ns
		$\overline{\text{DCS}}$ before CK \uparrow , $\overline{\text{CK}}\downarrow$, $\overline{\text{CSR}}$ LOW		0.5	-	-	ns
		DODT, DCKE and data (Dn) before CK \uparrow , $\overline{\text{CK}}\downarrow$		0.5	-	-	ns
		PAR_IN before CK \uparrow , $\overline{\text{CK}}\downarrow$		0.5	-	-	ns
t _h	hold time	DCS, DODT, DCKE and data (Dn) after CK1, $\overline{\text{CK}}\downarrow$		0.4	-	-	ns
		PAR_IN after CK↑, CK↓		0.4	-	-	ns

^[1] This parameter is not necessarily production tested.

Table 11. Switching characteristics (667 mode, SELAB = HIGH)

At recommended operating conditions (see Table 8), unless otherwise specified. See Section 11.1.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f_{max}	maximum input clock frequency		450	-	-	MHz
t _{PDM}	peak propagation delay	single bit switching; from CK \uparrow and $\overline{\text{CK}} \downarrow$ to Qn	<u>1</u> 1.2	-	1.8	ns
t_{PD}	propagation delay	from CK↑ and $\overline{\text{CK}}$ to PPO	0.5	-	1.8	ns
t_{LH}	LOW-to-HIGH delay	from CK \uparrow and $\overline{\text{CK}} \downarrow$ to $\overline{\text{QERR}}$	1.2	-	3	ns
t_{HL}	HIGH-to-LOW delay	from CK↑ and $\overline{CK} \downarrow$ to \overline{QERR}	1	-	2.4	ns
t _{PDMSS}	simultaneous switching peak propagation delay	from CK \uparrow and $\overline{\text{CK}} \downarrow$ to Qn	[1][2]	-	2.0	ns
t_{PHL}	HIGH-to-LOW propagation delay	from RESET↓ to Qn↓	-	-	3	ns
		from $\overline{RESET} \downarrow$ to $PPO \downarrow$	-	-	3	ns
t _{PLH}	LOW-to-HIGH propagation delay	from $\overline{RESET} \downarrow$ to $\overline{QERR} \uparrow$	-	-	3	ns

^[1] Includes 350 ps of test load transmission line delay.

^[2] VREF must be held at a valid input voltage level and data inputs must be held LOW for a minimum time of t_{ACT(max)} after RESET is taken HIGH.

^[3] VREF, data and clock inputs must be held at valid levels (not floating) a minimum time of t_{INACT(max)} after RESET is taken LOW.

^[2] This parameter is not necessarily production tested.

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Table 12. Switching characteristics (800 mode, SELAB = LOW)

At recommended operating conditions (see Table 8), unless otherwise specified. See Section 11.1.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f_{max}	maximum input clock frequency		450	-	-	MHz
t _{PDM}	peak propagation delay	single bit switching; from $CK \uparrow$ and $\overline{CK} \downarrow$ to Qn	<u>11</u> 1.1	-	1.5	ns
t _{PD}	propagation delay	from CK $↑$ and $\overline{\text{CK}} ↓$ to PPO	0.5	-	1.7	ns
t _{LH}	LOW-to-HIGH delay	from CK \uparrow and $\overline{\text{CK}} \downarrow$ to $\overline{\text{QERR}}$	1.2	-	3	ns
t _{HL}	HIGH-to-LOW delay	from CK \uparrow and $\overline{\text{CK}} \downarrow$ to $\overline{\text{QERR}}$	1	-	2.4	ns
t _{PDMSS}	simultaneous switching peak propagation delay	from CK↑ and $\overline{\text{CK}} \downarrow$ to Qn	[1][2] -	-	1.6	ns
t _{PHL}	HIGH-to-LOW propagation delay	from $\overline{RESET} \downarrow$ to $Qn \downarrow$	-	-	3	ns
		from $\overline{RESET} \downarrow$ to $PPO \downarrow$	-	-	3	ns
t _{PLH}	LOW-to-HIGH propagation delay	from $\overline{RESET} \downarrow$ to $\overline{QERR} \uparrow$	-	-	3	ns

^[1] Includes 350 ps of test load transmission line delay.

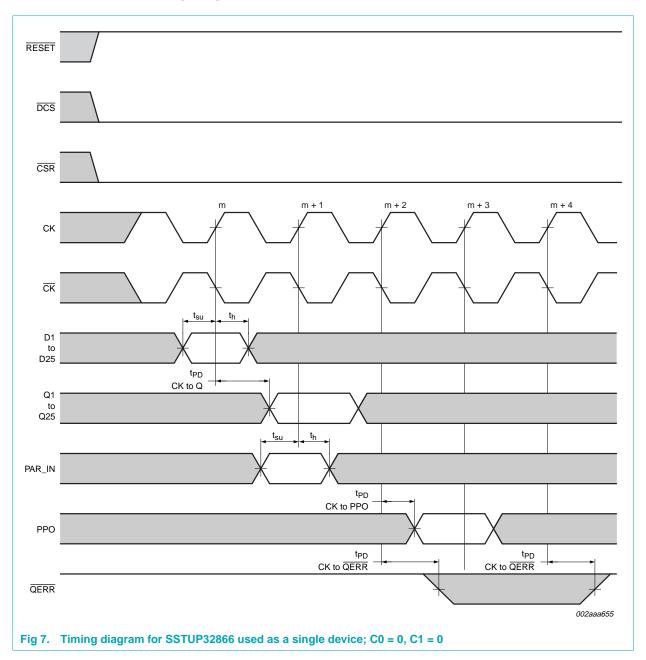
Table 13. Data output edge rates

At recommended operating conditions (see Table 8), unless otherwise specified. See Section 11.2.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
dV/dt_r	rising edge slew rate	from 20 % to 80 %	1	-	4	V/ns
dV/dt_f	falling edge slew rate	from 80 % to 20 %	1	-	4	V/ns
dV/dt_Δ	absolute difference between dV/dt_r and dV/dt_f	from 20 % or 80 % to 80 % or 20 %	-	-	1	V/ns

^[2] This parameter is not necessarily production tested.

10.1 Timing diagrams



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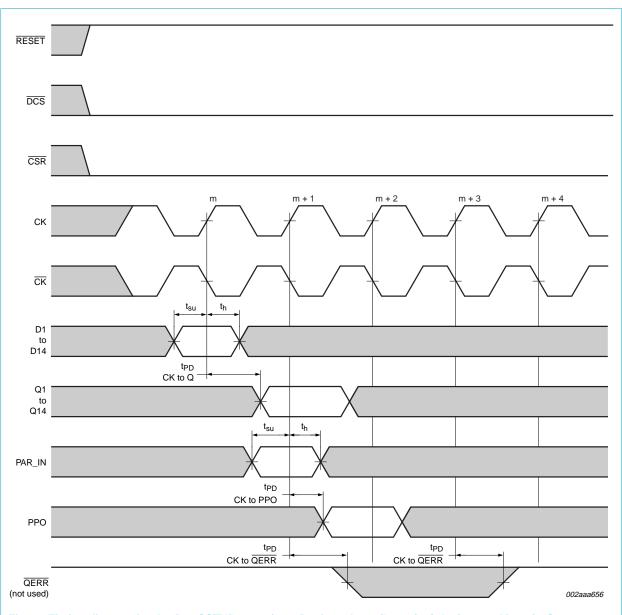
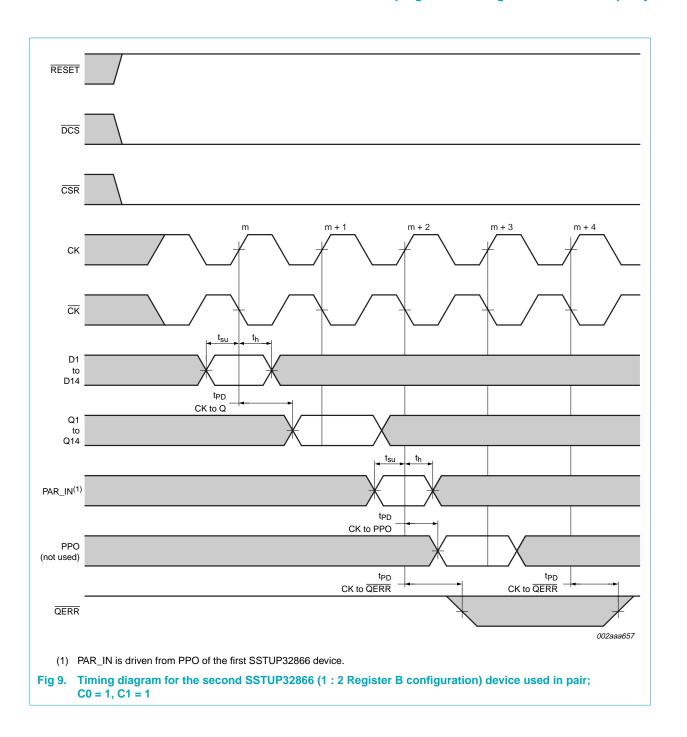


Fig 8. Timing diagram for the first SSTUP32866 (1 : 2 Register A configuration) device used in pair; C0 = 0, C1 = 1

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11. Test information

11.1 Parameter measurement information for data output load circuit

 $V_{DD} = 1.8 \text{ V} \pm 0.1 \text{ V}.$

All input pulses are supplied by generators having the following characteristics: PRR \leq 10 MHz; $Z_0 = 50 \Omega$; input slew rate = 1 V/ns \pm 20 %, unless otherwise specified.

The outputs are measured one at a time with one transition per measurement.

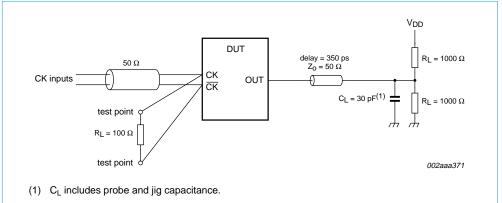
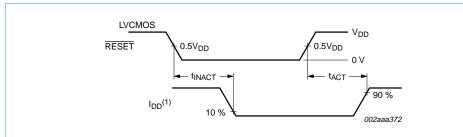
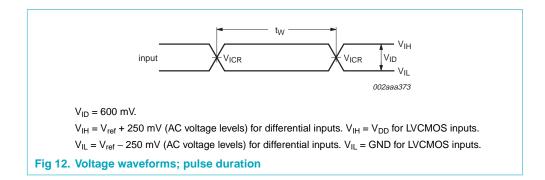


Fig 10. Load circuit, data output measurements



(1) I_{DD} tested with clock and data inputs held at V_{DD} or GND, and I_{O} = 0 mA.

Fig 11. Voltage and current waveforms; inputs active and inactive times

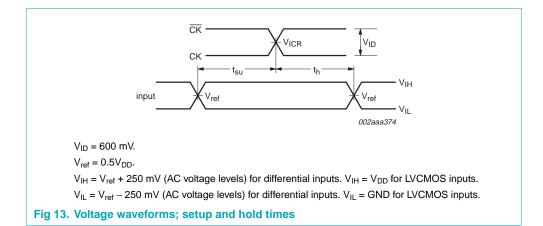


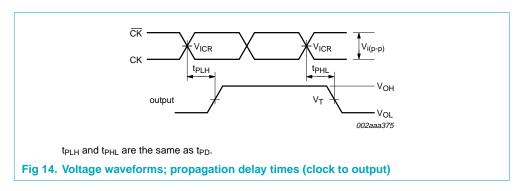
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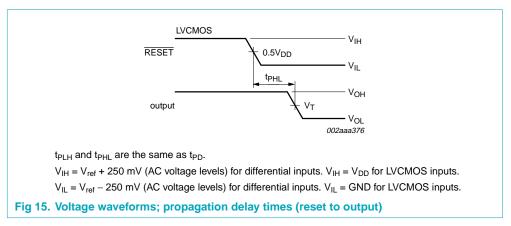
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11.2 Data output slew rate measurement information

 $V_{DD} = 1.8 \text{ V} \pm 0.1 \text{ V}.$

All input pulses are supplied by generators having the following characteristics: PRR \leq 10 MHz; $Z_0 = 50 \Omega$; input slew rate = 1 V/ns \pm 20 %, unless otherwise specified.

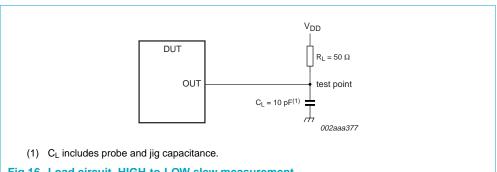
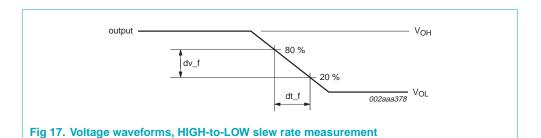
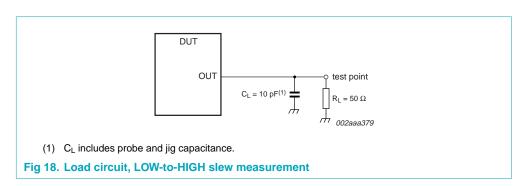
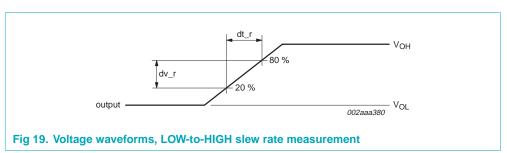


Fig 16. Load circuit, HIGH-to-LOW slew measurement







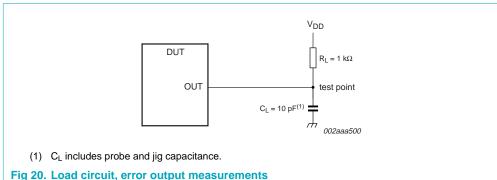
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11.3 Error output load circuit and voltage measurement information

 $V_{DD} = 1.8 \text{ V} \pm 0.1 \text{ V}.$

All input pulses are supplied by generators having the following characteristics: PRR \leq 10 MHz; $Z_0 = 50 \Omega$; input slew rate = 1 V/ns \pm 20 %, unless otherwise specified.



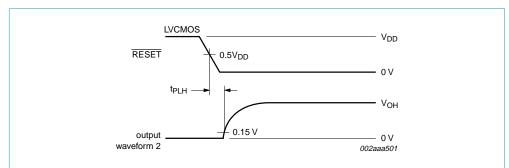


Fig 21. Voltage waveforms, open-drain output LOW-to-HIGH transition time with respect to **RESET** input

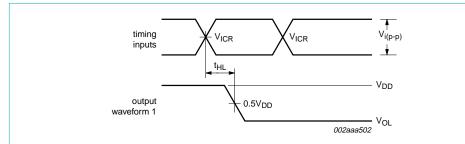


Fig 22. Voltage waveforms, open-drain output HIGH-to-LOW transition time with respect to clock inputs

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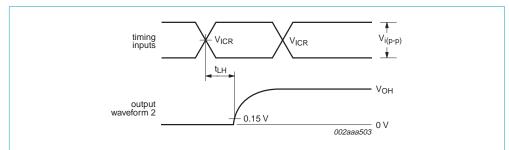


Fig 23. Voltage waveforms, open-drain output LOW-to-HIGH transition time with respect to clock inputs

11.4 Partial parity out load circuit and voltage measurement information

 $V_{DD} = 1.8 \text{ V} \pm 0.1 \text{ V}.$

All input pulses are supplied by generators having the following characteristics: PRR \leq 10 MHz; Z_0 = 50 Ω ; input slew rate = 1 V/ns \pm 20 %, unless otherwise specified.

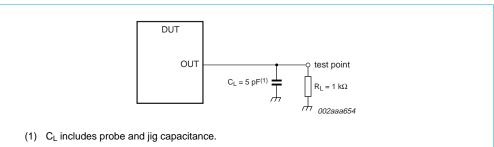


Fig 24. Partial parity out load circuit

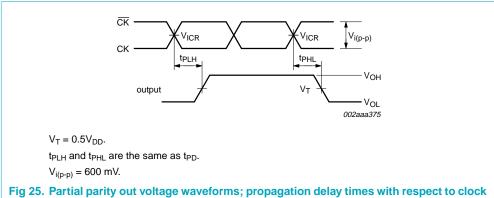
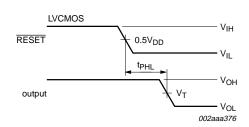


Fig 25. Partial parity out voltage waveforms; propagation delay times with respect to clock inputs

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 $V_T = 0.5V_{DD}$.

 t_{PLH} and t_{PHL} are the same as t_{PD} .

 V_{IH} = V_{ref} + 250 mV (AC voltage levels) for differential inputs. V_{IH} = V_{DD} for LVCMOS inputs.

 V_{IL} = V_{ref} – 250 mV (AC voltage levels) for differential inputs. V_{IL} = GND for LVCMOS inputs.

Fig 26. Partial parity out voltage waveforms; propagation delay times with respect to RESET input

12. Package outline

LFBGA96: plastic low profile fine-pitch ball grid array package; 96 balls; body 13.5 x 5.5 x 1.05 mm SOT536-1

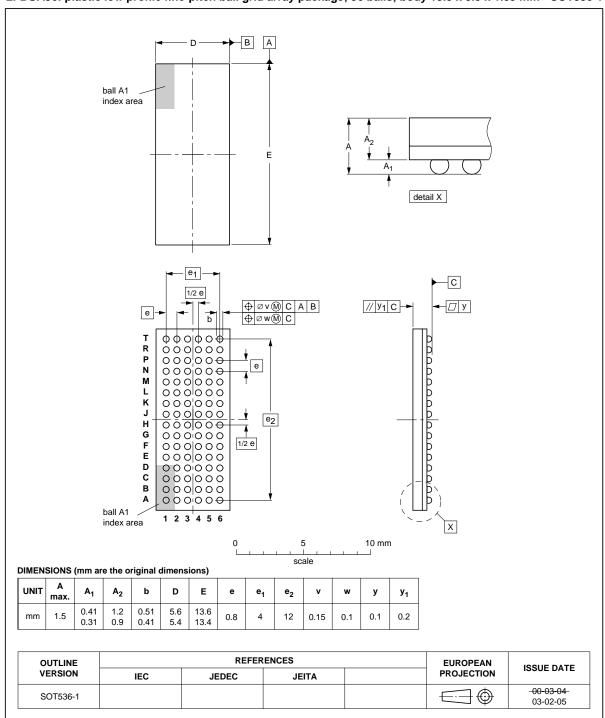


Fig 27. Package outline SOT536-1 (LFBGA96)

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13. Soldering

13.1 Introduction to soldering surface mount packages

There is no soldering method that is ideal for all surface mount IC packages. Wave soldering can still be used for certain surface mount ICs, but it is not suitable for fine pitch SMDs. In these situations reflow soldering is recommended.

13.2 Reflow soldering

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement. Driven by legislation and environmental forces the worldwide use of lead-free solder pastes is increasing.

Several methods exist for reflowing; for example, convection or convection/infrared heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 100 seconds and 200 seconds depending on heating method.

Typical reflow temperatures range from 215 °C to 260 °C depending on solder paste material. The peak top-surface temperature of the packages should be kept below:

Table 14. SnPb eutectic process - package peak reflow temperatures (from *J-STD-020C* July 2004)

Package thickness	Volume mm ³ < 350	Volume mm³ ≥ 350
< 2.5 mm	240 °C + 0/–5 °C	225 °C + 0/–5 °C
≥ 2.5 mm	225 °C + 0/–5 °C	225 °C + 0/-5 °C

Table 15. Pb-free process - package peak reflow temperatures (from *J-STD-020C* July 2004)

Package thickness	Volume mm ³ < 350	Volume mm ³ 350 to 2000	Volume mm ³ > 2000
< 1.6 mm	260 °C + 0 °C	260 °C + 0 °C	260 °C + 0 °C
1.6 mm to 2.5 mm	260 °C + 0 °C	250 °C + 0 °C	245 °C + 0 °C
≥ 2.5 mm	250 °C + 0 °C	245 °C + 0 °C	245 °C + 0 °C

Moisture sensitivity precautions, as indicated on packing, must be respected at all times.

13.3 Wave soldering

Conventional single wave soldering is not recommended for surface mount devices (SMDs) or printed-circuit boards with a high component density, as solder bridging and non-wetting can present major problems.

To overcome these problems the double-wave soldering method was specifically developed.

If wave soldering is used the following conditions must be observed for optimal results:

- Use a double-wave soldering method comprising a turbulent wave with high upward pressure followed by a smooth laminar wave.
- For packages with leads on two sides and a pitch (e):

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 larger than or equal to 1.27 mm, the footprint longitudinal axis is preferred to be parallel to the transport direction of the printed-circuit board;

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 smaller than 1.27 mm, the footprint longitudinal axis must be parallel to the transport direction of the printed-circuit board.

The footprint must incorporate solder thieves at the downstream end.

 For packages with leads on four sides, the footprint must be placed at a 45° angle to the transport direction of the printed-circuit board. The footprint must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Typical dwell time of the leads in the wave ranges from 3 seconds to 4 seconds at 250 °C or 265 °C, depending on solder material applied, SnPb or Pb-free respectively.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

13.4 Manual soldering

Fix the component by first soldering two diagonally-opposite end leads. Use a low voltage (24 V or less) soldering iron applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C.

When using a dedicated tool, all other leads can be soldered in one operation within 2 seconds to 5 seconds between 270 °C and 320 °C.

13.5 Package related soldering information

Table 16. Suitability of surface mount IC packages for wave and reflow soldering methods

Package[1]	Soldering method		
	Wave	Reflow[2]	
BGA, HTSSONT ³ , LBGA, LFBGA, SQFP, SSOPT ³ , TFBGA, VFBGA, XSON	not suitable	suitable	
DHVQFN, HBCC, HBGA, HLQFP, HSO, HSOP, HSQFP, HSSON, HTQFP, HTSSOP, HVQFN, HVSON, SMS	not suitable[4]	suitable	
PLCC[5], SO, SOJ	suitable	suitable	
LQFP, QFP, TQFP	not recommended[5][6]	suitable	
SSOP, TSSOP, VSO, VSSOP	not recommended[7]	suitable	
CWQCCNL[8], PMFP[9], WQCCNL[8]	not suitable	not suitable	

^[1] For more detailed information on the BGA packages refer to the (*LF*)BGA Application Note (AN01026); order a copy from your Philips Semiconductors sales office.

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^[2] All surface mount (SMD) packages are moisture sensitive. Depending upon the moisture content, the maximum temperature (with respect to time) and body size of the package, there is a risk that internal or external package cracks may occur due to vaporization of the moisture in them (the so called popcorn effect). For details, refer to the Drypack information in the Data Handbook IC26; Integrated Circuit Packages; Section: Packing Methods.

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- [3] These transparent plastic packages are extremely sensitive to reflow soldering conditions and must on no account be processed through more than one soldering cycle or subjected to infrared reflow soldering with peak temperature exceeding 217 °C ± 10 °C measured in the atmosphere of the reflow oven. The package body peak temperature must be kept as low as possible.
- [4] These packages are not suitable for wave soldering. On versions with the heatsink on the bottom side, the solder cannot penetrate between the printed-circuit board and the heatsink. On versions with the heatsink on the top side, the solder might be deposited on the heatsink surface.
- [5] If wave soldering is considered, then the package must be placed at a 45° angle to the solder wave direction. The package footprint must incorporate solder thieves downstream and at the side corners.
- [6] Wave soldering is suitable for LQFP, QFP and TQFP packages with a pitch (e) larger than 0.8 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.65 mm.
- [7] Wave soldering is suitable for SSOP, TSSOP, VSO and VSSOP packages with a pitch (e) equal to or larger than 0.65 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.5 mm.
- [8] Image sensor packages in principle should not be soldered. They are mounted in sockets or delivered pre-mounted on flex foil. However, the image sensor package can be mounted by the client on a flex foil by using a hot bar soldering process. The appropriate soldering profile can be provided on request.
- [9] Hot bar soldering or manual soldering is suitable for PMFP packages.

14. Abbreviations

Table 17. Abbreviations

Acronym	Description
CMOS	Complementary Metal Oxide Semiconductor
DDR	Double Data Rate
DIMM	Dual In-line Memory Module
LVCMOS	Low Voltage Complementary Metal Oxide Semiconductor
PPO	Partial Parity Out
PRR	Pulse Repetition Rate
RDIMM	Registered Dual In-line Memory Module
SSTL	Stub Series Terminated Logic

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15. Revision history

Table 18. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes				
SSTUP32866_2	20060914	Product data sheet	-	SSTUP32866_1				
Modifications:	Table 1 "Orde	Table 1 "Ordering information": added Type number SSTUP32866EC/S						
	 added <u>Section</u> 	added Section 4.1 "Ordering options"						
	• Figure 3 "Pin	configuration for LFBGA96"	modified (added Type nui	mber SSTUP32866EC/S)				
	 <u>Table 5 "Parity and standby function table"</u>: added reference to <u>Table note 4</u> at column head "QERR" 							
	 <u>Table 8 "Recommended operating conditions"</u>, Symbol T_{amb}: added separate specifications f SSTUP32866EC/G and SSTUP32866EC/S 							
	• <u>Table 10 "Timing requirements"</u> , description below table title: changed "See <u>Figure 2</u> ." to "See Section 11.1."							
		tching characteristics (667 n s (800 mode, SELAB = LOW		nd Table 12 "Switching				
	 changed parameter description of t_{PDM} from "propagation delay, single bit switching" to "propagation delay" (moved "single bit switching" to Conditions column) 							
	 changed parameter description of t_{PDMSS} from "propagation delay, simultaneous switching" "simultaneous switching peak propagation delay" 							
 Figure 26 "Partial parity out voltage waveforms; propagation delay times with respe- input", 4th note: changed "V_{IL} = V_{DD}" to "V_{IL} = GND" 								
SSTUP32866_1	20060406	Product data sheet	-	-				

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16. Legal information

16.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
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