MCP73831/2

Miniature Single-Cell, Fully Integrated Li-Ion, Li-Polymer Charge Management Controllers

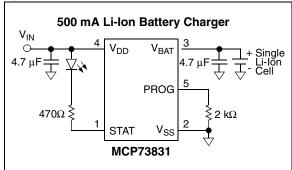
Features:

- Linear Charge Management Controller:
 - Integrated Pass Transistor
 - Integrated Current Sense
 - Reverse Discharge Protection
- High Accuracy Preset Voltage Regulation: ± 0.75%
- · Four Voltage Regulation Options:
 - 4.20V, 4.35V, 4.40V, 4.50V
- Programmable Charge Current
- · Selectable Preconditioning
- · Selectable End-of-Charge Control
- · Charge Status Output
 - Tri-State Output MCP73831
 - Open-Drain Output MCP73832
- · Automatic Power-Down
- · Thermal Regulation
- Temperature Range: -40°C to +85°C
- Packaging:
 - 8-Lead, 2 mm x 3 mm DFN
 - 5-Lead, SOT23

Applications:

- · Lithium-Ion/Lithium-Polymer Battery Chargers
- Personal Data Assistants
- Cellular Telephones
- · Digital Cameras
- · MP3 Players
- · Bluetooth Headsets
- USB Chargers

Typical Application



Description:

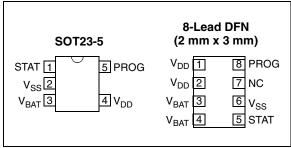
The MCP73831/2 devices are highly advanced linear charge management controllers for use in space-limited, cost-sensitive applications. The MCP73831/2 are available in an 8-Lead, 2 mm x 3 mm DFN package or a 5-Lead, SOT23 package. Along with their small physical size, the low number of external components required make the MCP73831/2 ideally suited for portable applications. For applications charging from a USB port, the MCP73831/2 adhere to all the specifications governing the USB power bus.

The MCP73831/2 employ a constant-current/constant-voltage charge algorithm with selectable preconditioning and charge termination. The constant voltage regulation is fixed with four available options: 4.20V, 4.35V, 4.40V or 4.50V, to accommodate new, emerging battery charging requirements. The constant current value is set with one external resistor. The MCP73831/2 devices limit the charge current based on die temperature during high power or high ambient conditions. This thermal regulation optimizes the charge cycle time while maintaining device reliability.

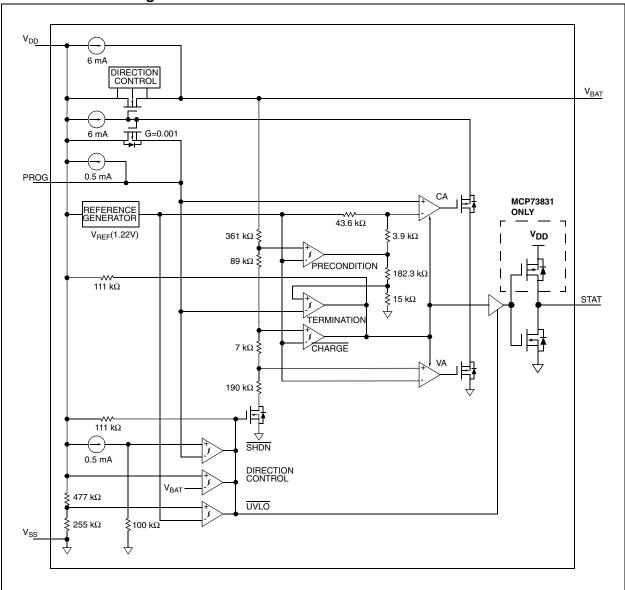
Several options are available for the preconditioning threshold, preconditioning current value, charge termination value and automatic recharge threshold. The preconditioning value and charge termination value are set as a ratio, or percentage, of the programmed constant current value. Preconditioning can be disabled. Refer to Section 1.0 "Electrical Characteristics" for available options and the "Product Identification System" for standard options.

The MCP73831/2 devices are fully specified over the ambient temperature range of -40°C to +85°C.

Package Types



Functional Block Diagram



1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings†

V _{DD} 7.0V
All Inputs and Outputs w.r.t. $\rm V_{SS}$ 0.3 to $\rm (V_{DD}+0.3)V$
$\label{eq:maximum Junction Temperature} \ \ T_J \ \ Internally \ Limited$
Storage temperature65°C to +150°C
ESD protection on all pins:
Human Body Model (1.5 k $\!\Omega$ in Series with 100 pF) $\!\! \geq \! 4$ kV
Machine Model (200 pF, No Series Resistance)400V

† Notice: Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

DC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, all limits apply for V_{DD}= [V_{REG}(typ.) + 0.3V] to 6V, T_A = -40°C to +85°C. Typical values are at +25°C, V_{DD} = [V_{REG} (typ.) + 1.0V]

Parameters Sym. Min. Typ. Max. Units Conditions

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions
Supply Input						
Supply Voltage	V_{DD}	3.75	_	6	V	
Supply Current	I _{SS}	_	510	1500	μΑ	Charging
		_	53	200	μА	Charge Complete, No Battery
		_	25	50	μΑ	PROG Floating
		_	1	5	μΑ	$V_{DD} \le (V_{BAT} - 50 \text{ mV})$
		_	0.1	2	μΑ	V _{DD} < V _{STOP}
UVLO Start Threshold	V _{START}	3.3	3.45	3.6	V	V _{DD} Low-to-High
UVLO Stop Threshold	V _{STOP}	3.2	3.38	3.5	V	V _{DD} High-to-Low
UVLO Hysteresis	V _{HYS}	_	70	_	mV	
Voltage Regulation (Cons	tant-Voltage M	ode)		•		
Regulated Output Voltage	V _{REG}	4.168	4.20	4.232	V	MCP7383X-2
		4.317	4.35	4.383	V	MCP7383X-3
		4.367	4.40	4.433	V	MCP7383X-4
		4.466	4.50	4.534	V	MCP7383X-5
						$V_{DD} = [V_{REG}(Typ)+1V]$ $I_{OUT} = 10 \text{ mA}$ $T_A = -5^{\circ}\text{C to } +55^{\circ}\text{C}$
Line Regulation	$\frac{\left (\Delta V_{BAT}/V_{BAT}\right.}{\left.)/\Delta V_{DD}\right }$	_	0.09	0.30	%/V	$V_{DD} = [V_{REG}(Typ)+1V]$ to 6V $I_{OUT} = 10$ mA
Load Regulation	$ \Delta V_{BAT}/V_{BAT} $	_	0.05	0.30	%	$I_{OUT} = 10 \text{ mA to } 50 \text{ mA}$ $V_{DD} = [V_{REG}(Typ)+1V]$
Supply Ripple Attenuation	PSRR	_	52		dB	I _{OUT} =10 mA, 10Hz to 1 kHz
		_	47	_	dB	I _{OUT} =10 mA, 10Hz to 10 kHz
		_	22	_	dB	I _{OUT} =10 mA, 10Hz to 1 MHz
Current Regulation (Fast	Charge Consta	nt-Current Mo	de)			
Fast Charge Current	I _{REG}	90	100	110	mA	PROG = 10 kΩ
Regulation		450	505	550	mA	PROG = $2.0 \text{ k}\Omega$, Note 1
						$T_A = -5^{\circ}C \text{ to } +55^{\circ}C$

Note 1: Not production tested. Ensured by design.

DC CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG}(typ.) + 0.3V]$ to 6V, $T_A = -40$ °C to +85°C. Typical values are at +25°C, $V_{DD} = [V_{REG}(typ.) + 1.0V]$

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions
Preconditioning Current	Regulation (Tri	ckle Charge Co	nstant-Current	Mode)	1	1
Precondition Current	I _{PREG} / I _{REG}	7.5	10	12.5	%	PROG = 2.0 kΩ to 10 kΩ
Ratio		15	20	25	%	PROG = 2.0 kΩ to 10 kΩ
	-	30	40	50	%	PROG = 2.0 kΩ to 10 kΩ
	-	_	100	_	%	No Preconditioning
	-					$T_A = -5^{\circ}C \text{ to } +55^{\circ}C$
Precondition Voltage	V _{PTH} / V _{REG}	64	66.5	69	%	V _{BAT} Low-to-High
Threshold Ratio		69	71.5	74	%	V _{BAT} Low-to-High
Precondition Hysteresis	V _{PHYS}	_	110	_	mV	V _{BAT} High-to-Low
Charge Termination					1	
Charge Termination	I _{TERM} / I _{REG}	3.75	5	6.25	%	PROG = 2.0 kΩ to 10 kΩ
Current Ratio		5.6	7.5	9.4	%	PROG = 2.0 kΩ to 10 kΩ
	-	7.5	10	12.5	%	PROG = 2.0 kΩ to 10 kΩ
	-	15	20	25	%	PROG = 2.0 kΩ to 10 kΩ
	-					$T_A = -5^{\circ}C$ to $+55^{\circ}C$
Automatic Recharge	<u>l</u>		L	L		
Recharge Voltage	V _{RTH} / V _{REG}	91.5	94.0	96.5	%	V _{BAT} High-to-Low
Threshold Ratio	1111	94	96.5	99	%	V _{BAT} High-to-Low
Pass Transistor ON-Resis	stance				ı	DAT 0
ON-Resistance	R _{DSON}	_	350	_	mΩ	V _{DD} = 3.75V, T _J = 105°C
Battery Discharge Curren					ı	7 0
Output Reverse Leakage	IDISCHARGE	_	0.15	2	μА	PROG Floating
Current	DISCHARGE	_	0.25	2	μA	V _{DD} Floating
	-	_	0.15	2	μA	V _{DD} < V _{STOP}
	-	_	-5.5	-15	μA	Charge Complete
Status Indicator – STAT						
Sink Current	I _{SINK}	_	_	25	mA	
Low Output Voltage	V _{OL}	_	0.4	1	V	I _{SINK} = 4 mA
Source Current	I _{SOURCE}	_	_	35	mA	CITAL
High Output Voltage	V _{OH}	_	V _{DD} -0.4	V _{DD} - 1	V	I _{SOURCE} = 4 mA (MCP73831)
Input Leakage Current	I _{LK}	_	0.03	1	μА	High-Impedance
PROG Input	LIX		L	L	<u>'</u>	
Charge Impedance Range	R _{PROG}	2	_	20	kΩ	
Minimum Shutdown Impedance	R _{PROG}	70	_	200	kΩ	
Automatic Power Down	<u>l</u>		<u> </u>	1	1	I .
Automatic Power Down Entry Threshold	V _{PDENTER}	V _{DD} <(V _{BAT} +20mV)	V _{DD} <(V _{BAT} +50mV)	_		$3.5V \le V_{BAT} \le V_{REG}$ V_{DD} Falling
Automatic Power Down Exit Threshold	V _{PDEXIT}		V _{DD} <(V _{BAT} +150mV)	V _{DD} <(V _{BAT} +200mV)		$3.5V \le V_{BAT} \le V_{REG}$ V_{DD} Rising
Thermal Shutdown	<u>l</u>		<u>'</u>	<u>, </u>	1	1 20 0
Die Temperature	T _{SD}		150	_	°C	
	. 3D					<u> </u>

Note 1: Not production tested. Ensured by design.

AC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG} \text{ (typ.)} + 0.3V]$ to 12V, $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$. Typical values are at $+25^{\circ}\text{C}$, $V_{DD} = [V_{REG} \text{ (typ.)} + 1.0V]$										
Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions				
UVLO Start Delay	t _{START}			5	ms	V _{DD} Low-to-High				
Constant-Current Regulation										
Transition Time Out of Preconditioning	^t DELAY	_		1	ms	$V_{BAT} < V_{PTH}$ to $V_{BAT} > V_{PTH}$				
Current Rise Time Out of Preconditioning	t _{RISE}	_		1	ms	I _{OUT} Rising to 90% of I _{REG}				
Termination Comparator Filter	t _{TERM}	0.4	1.3	3.2	ms	Average I _{OUT} Falling				
Charge Comparator Filter	t _{CHARGE}	0.4	1.3	3.2	ms	Average V _{BAT}				
Status Indicator										
Status Output turn-off	t _{OFF}		_	200	μS	I _{SINK} = 1 mA to 0 mA				
Status Output turn-on	t _{ON}	_	_	200	μS	I _{SINK} = 0 mA to 1 mA				

TEMPERATURE SPECIFICATIONS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG} \text{ (typ.)} + 0.3V]$ to 12V. Typical values are at +25°C, $V_{DD} = [V_{REG} \text{ (typ.)} + 1.0V]$										
Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions				
Temperature Ranges										
Specified Temperature Range	T _A	-40	_	+85	°C					
Operating Temperature Range	T _J	-40	_	+125	°C					
Storage Temperature Range	T _A	-65	_	+150	°C					
Thermal Package Resistances										
5-Lead, SOT23	θ_{JA}	_	230	_	°C/W	4-Layer JC51-7 Standard Board, Natural Convection				
8-Lead, 2 mm x 3 mm, DFN	θ_{JA}	_	76	_	°C/W	4-Layer JC51-7 Standard Board, Natural Convection				

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

NOTE: Unless otherwise indicated, $V_{DD} = [V_{REG}(typ.) + 1V]$, $I_{OUT} = 10$ mA and $T_A = +25$ °C, Constant-Voltage mode.

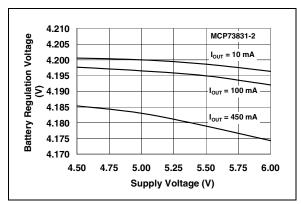


FIGURE 2-1: Battery Regulation Voltage (V_{BAT}) vs. Supply Voltage (V_{DD}) .

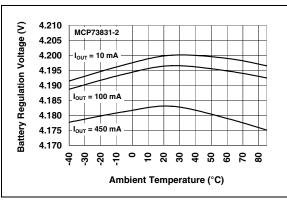


FIGURE 2-2: Battery Regulation Voltage (V_{BAT}) vs. Ambient Temperature (T_A) .

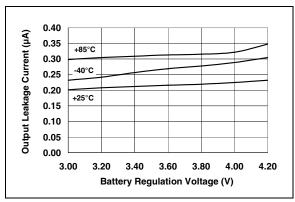


FIGURE 2-3: Output Leakage Current $(I_{DISCHARGE})$ vs. Battery Regulation Voltage (V_{BAT}) .

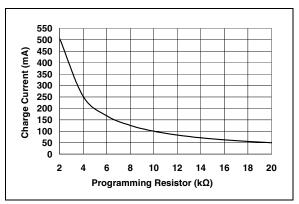


FIGURE 2-4: Charge Current (I_{OUT}) vs. Programming Resistor (R_{PROG}) .

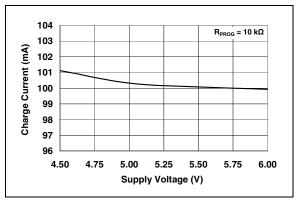


FIGURE 2-5: Charge Current (I_{OUT}) vs. Supply Voltage (V_{DD}).

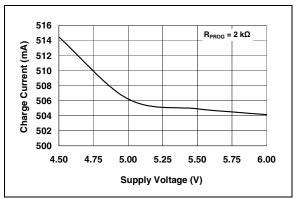


FIGURE 2-6: Charge Current (I_{OUT}) vs. Supply Voltage (V_{DD}).

TYPICAL PERFORMANCE CURVES (CONTINUED)

NOTE: Unless otherwise indicated, $V_{DD} = [V_{REG}(typ.) + 1V]$, $I_{OUT} = 10$ mA and $T_A = +25$ °C, Constant-Voltage mode.

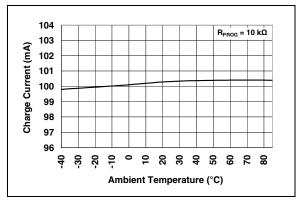


FIGURE 2-7: Charge Current (I_{OUT}) vs. Ambient Temperature (T_A) .

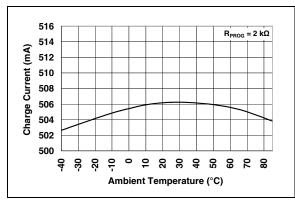


FIGURE 2-8: Charge Current (I_{OUT}) vs. Ambient Temperature (T_{Δ}) .

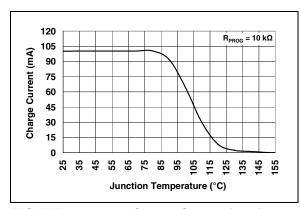


FIGURE 2-9: Charge Current (I_{OUT}) vs. Junction Temperature (T_J) .

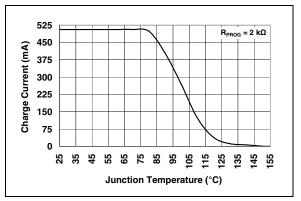


FIGURE 2-10: Charge Current (I_{OUT}) vs. Junction Temperature (T_1) .

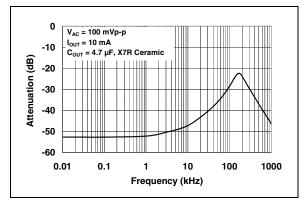


FIGURE 2-11: Power Supply Ripple Rejection (PSRR).

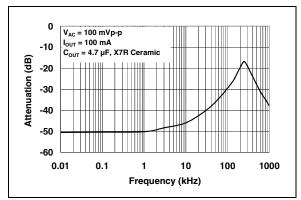


FIGURE 2-12: Power Supply Ripple Rejection (PSRR).

TYPICAL PERFORMANCE CURVES (CONTINUED)

NOTE: Unless otherwise indicated, $V_{DD} = [V_{REG}(typ.) + 1V]$, $I_{OUT} = 10$ mA and $T_A = +25$ °C, Constant-Voltage mode.

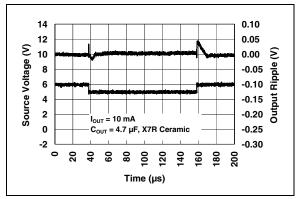


FIGURE 2-13: Line Transient Response.

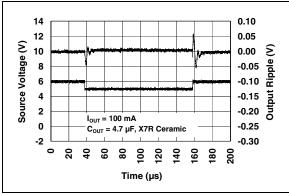


FIGURE 2-14: Line Transient Response.

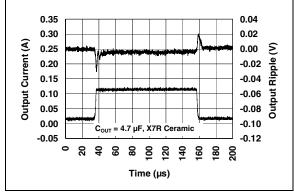


FIGURE 2-15: Load Transient Response.

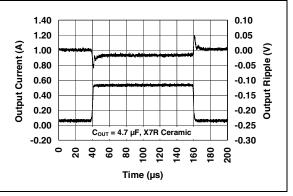


FIGURE 2-16: Load Transient Response.

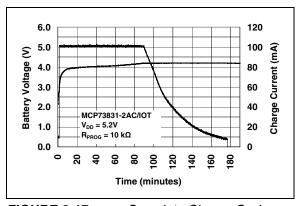


FIGURE 2-17: Complete Charge Cycle (180 mAh Li-Ion Battery).

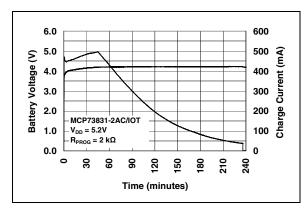


FIGURE 2-18: Complete Charge Cycle (1000 mAh Li-Ion Battery).

3.0 PIN DESCRIPTION

The descriptions of the pins are listed in Table 3-1.

TABLE 3-1: PIN FUNCTION TABLES

Pin	Pin No.		Function				
DFN	SOT23-5	Symbol	runction				
1	4	V_{DD}	Battery Management Input Supply				
2	_	V_{DD}	Battery Management Input Supply				
3	3	V_{BAT}	Battery Charge Control Output				
4	_	V_{BAT}	Battery Charge Control Output				
5	1	STAT	Charge Status Output				
6	2	V _{SS}	Battery Management 0V Reference				
7	_	NC	No Connection				
8	5	PROG	Current Regulation Set and Charge Control Enable				

3.1 Battery Management Input Supply (V_{DD})

A supply voltage of [V_{REG} (typ.) + 0.3V] to 6V is recommended. Bypass to V_{SS} with a minimum of 4.7 $\mu F.$

3.2 Battery Charge Control Output (V_{BAT})

Connect to positive terminal of battery. Drain terminal of internal P-channel MOSFET pass transistor. Bypass to V_{SS} with a minimum of 4.7 μF to ensure loop stability when the battery is disconnected.

3.3 Charge Status Output (STAT)

STAT is an output for connection to an LED for charge status indication. Alternatively, a pull-up resistor can be applied for interfacing to a host microcontroller.

STAT is a tri-state logic output on the MCP73831 and an open-drain output on the MCP73832.

3.4 Battery Management 0V Reference (Vss)

Connect to negative terminal of battery and input supply.

3.5 Current Regulation Set (PROG)

Preconditioning, fast charge and termination currents are scaled by placing a resistor from PROG to $V_{\rm SS}$.

The charge management controller can be disabled by allowing the PROG input to float.

4.0 DEVICE OVERVIEW

The MCP73831/2 are highly advanced linear charge management controllers. Figure 4-1 depicts the operational flow algorithm from charge initiation to completion and automatic recharge.

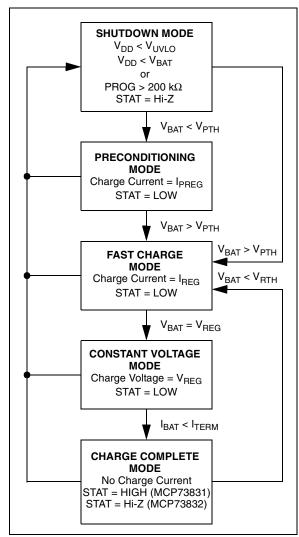


FIGURE 4-1: Flowchart.

4.1 Under Voltage Lockout (UVLO)

An internal UVLO circuit monitors the input voltage and keeps the charger in Shutdown mode until the input supply rises above the UVLO threshold. The UVLO circuitry has a built in hysteresis of 100 mV.

In the event a battery is present when the input power is applied, the input supply must rise 150 mV above the battery voltage before the MCP73831/2 becomes operational.

The UVLO circuit places the device in Shutdown mode if the input supply falls to within +50 mV of the battery voltage. Again, the input supply must rise to a level 150 mV above the battery voltage before the MCP73831/2 become operational.

The UVLO circuit is always active. At any time the input supply is below the UVLO threshold or within +50~mV of the voltage at the V_{BAT} pin, the MCP73831/2 are placed in a Shutdown mode.

During any UVLO condition, the battery reverse discharge current shall be less than 2 μA .

4.2 Charge Qualification

For a charge cycle to begin, all UVLO conditions must be met and a battery or output load must be present. A charge current programming resistor must be connected from PROG to V_{SS} . If the PROG pin is open or floating, the MCP73831/2 are disabled and the battery reverse discharge current is less than 2 μ A. In this manner, the PROG pin acts as a charge enable and can be used as a manual shutdown.

4.3 Preconditioning

If the voltage at the V_{BAT} pin is less than the preconditioning threshold, the MCP73831/2 enter a preconditioning or Trickle Charge mode. The preconditioning threshold is factory set. Refer to **Section 1.0** "**Electrical Characteristics**" for preconditioning threshold options and the Product Identification System for standard options.

In this mode, the MCP73831/2 supply a percentage of the charge current (established with the value of the resistor connected to the PROG pin) to the battery. The percentage or ratio of the current is factory set. Refer to **Section 1.0** "**Electrical Characteristics**" for preconditioning current options and the Product Identification System for standard options.

When the voltage at the V_{BAT} pin rises above the preconditioning threshold, the MCP73831/2 enter the Constant-Current or Fast Charge mode.

4.4 Fast Charge Constant-Current Mode

During the Constant-Current mode, the programmed charge current is supplied to the battery or load. The charge current is established using a single resistor from PROG to V_{SS} . Constant-Current mode is maintained until the voltage at the V_{BAT} pin reaches the regulation voltage, V_{REG} .

4.5 Constant-Voltage Mode

When the voltage at the V_{BAT} pin reaches the regulation voltage, V_{REG} constant voltage regulation begins. The regulation voltage is factory set to 4.2V, 4.35V, 4.40V, or 4.50V with a tolerance of $\pm 0.75\%$.

4.6 Charge Termination

The charge cycle is terminated when, during Constant-Voltage mode, the average charge current diminishes below a percentage of the programmed charge current (established with the value of the resistor connected to the PROG pin). A 1 ms filter time on the termination comparator ensures that transient load conditions do not result in premature charge cycle termination. The percentage or ratio of the current is factory set. Refer to **Section 1.0 "Electrical Characteristics"** for charge termination current options and the "**Product Identification System**" for standard options.

The charge current is latched off and the MCP73831/2 enter a Charge Complete mode.

4.7 Automatic Recharge

The MCP73831/2 continuously monitor the voltage at the V_{BAT} pin in the Charge Complete mode. If the voltage drops below the recharge threshold, another charge cycle begins and current is once again supplied to the battery or load. The recharge threshold is factory set. Refer to **Section 1.0** "Electrical Characteristics" for recharge threshold options and the Product Identification System for standard options.

4.8 Thermal Regulation

The MCP73831/2 limit the charge current based on the die temperature. The thermal regulation optimizes the charge cycle time while maintaining device reliability. Figure 4-2 depicts the thermal regulation for the MCP73831/2.

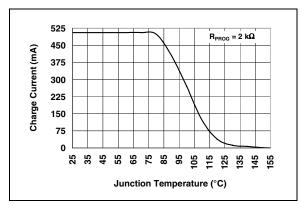


FIGURE 4-2: Thermal Regulation.

4.9 Thermal Shutdown

The MCP73831/2 suspend charge if the die temperature exceeds 150°C. Charging will resume when the die temperature has cooled by approximately 10°C.

5.0 DETAILED DESCRIPTION

5.1 Analog Circuitry

5.1.1 BATTERY MANAGEMENT INPUT SUPPLY (VDD)

The V_{DD} input is the input supply to the MCP73831/2. The MCP73831/2 automatically enter a Power-Down mode if the voltage on the V_{DD} input falls below the UVLO voltage (V_{STOP}). This feature prevents draining the battery pack when the V_{DD} supply is not present.

5.1.2 CURRENT REGULATION SET (PROG)

Fast charge current regulation can be scaled by placing a programming resistor (R_{PROG}) from the PROG input to V_{SS} . The program resistor and the charge current are calculated using the following equation:

$$I_{REG} = \frac{1000 V}{R_{PROG}}$$
 Where:
$$R_{PROG} = kOhms$$

$$I_{REG} = milliampere$$

The preconditioning trickle charge current and the charge termination current are ratiometric to the fast charge current based on the selected device options.

5.1.3 BATTERY CHARGE CONTROL OUTPUT (V_{BAT})

The battery charge control output is the drain terminal of an internal P-channel MOSFET. The MCP73831/2 provide constant current and voltage regulation to the battery pack by controlling this MOSFET in the linear region. The battery charge control output should be connected to the positive terminal of the battery pack.

5.2 Digital Circuitry

5.2.1 STATUS INDICATOR (STAT)

The charge status output of the MCP73831 has three different states: High (H), Low (L), and High-Impedance (Hi-Z). The charge status output of the MCP73832 is open-drain, and, as such, has two different states: Low (L), and High-Impedance (Hi-Z). The charge charge status output can be used to illuminate 1, 2, or tri-color LEDs. Optionally, the charge status output can be used as an interface to a host microcontroller.

Table 5-1 summarize the state of the status output during a charge cycle..

TABLE 5-1: STATUS OUTPUT

Charge Cycle State	STAT1				
Charge Cycle State	MCP73831	MCP73832			
Shutdown	Hi-Z	Hi-Z			
No Battery Present	Hi-Z	Hi-Z			
Preconditioning	L	L			
Constant-Current Fast Charge	L	L			
Constant Voltage	L	L			
Charge Complete – Standby	Н	Hi-Z			

5.2.2 DEVICE DISABLE (PROG)

The current regulation set input pin (PROG) can be used to terminate a charge at any time during the charge cycle, as well as to initiate a charge cycle or initiate a recharge cycle.

Placing a programming resistor from the PROG input to V_{SS} enables the device. Allowing the PROG input to float or by applying a logic-high input signal, disables the device and terminates a charge cycle. When disabled, the device's supply current is reduced to 25 μ A, typically.

6.0 APPLICATIONS

The MCP73831/2 are designed to operate in conjunction with a host microcontroller or in a stand-alone application. The MCP73831/2 provide the preferred charge algorithm for Lithium-Ion and Lithium-Polymer

cells constant current followed by constant voltage. Figure 6-1 depicts a typical stand-alone application circuit, while Figures 6-2 and 6-3 depict the accompanying charge profile.

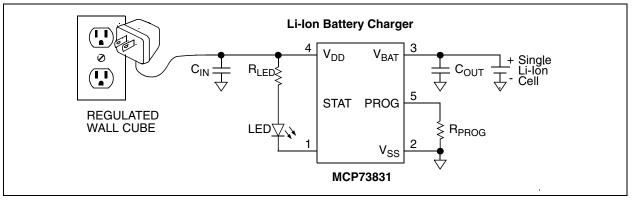


FIGURE 6-1: Typical Application Circuit.

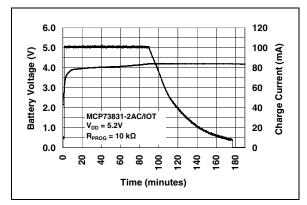


FIGURE 6-2: Typical Charge Profile (180 mAh Battery).

6.0 600 500 5.0 Battery Voltage (V) 400 4.0 300 3.0 200 2.0 MCP73831-2AC/IOT $V_{DD} = 5.2V$ 1.0 100 දි $R_{PPOG} = 2 k\Omega$ 0 0.0 8 8 210 240 20 20 Time (minutes)

FIGURE 6-3: Typical Charge Profile in Thermal Regulation (1000 mAh Battery).

6.1 Application Circuit Design

Due to the low efficiency of linear charging, the most important factors are thermal design and cost, which are a direct function of the input voltage, output current and thermal impedance between the battery charger and the ambient cooling air. The worst-case situation is when the device has transitioned from the Preconditioning mode to the Constant-Current mode. In this situation, the battery charger has to dissipate the maximum power. A trade-off must be made between the charge current, cost and thermal requirements of the charger.

6.1.1 COMPONENT SELECTION

Selection of the external components in Figure 6-1 is crucial to the integrity and reliability of the charging system. The following discussion is intended as a guide for the component selection process.

6.1.1.1 Current Programming Resistor (R_{PROG})

The preferred fast charge current for Lithium-Ion cells is at the 1C rate, with an absolute maximum current at the 2C rate. For example, a 500 mAh battery pack has a preferred fast charge current of 500 mA. Charging at this rate provides the shortest charge cycle times without degradation to the battery pack performance or life.

6.1.1.2 Thermal Considerations

The worst-case power dissipation in the battery charger occurs when the input voltage is at the maximum and the device has transitioned from the Preconditioning mode to the Constant-Current mode. In this case, the power dissipation is:

$$PowerDissipation = (V_{DDMAX} - V_{PTHMIN}) \times I_{REGMAX}$$

Where:

V_{DDMAX} = the maximum input voltage

I_{REGMAX} = the maximum fast charge current

V_{PTHMIN} = the minimum transition threshold

voltage

Power dissipation with a 5V, $\pm 10\%$ input voltage source is:

$$PowerDissipation = (5.5V - 2.7V) \times 550mA = 1.54W$$

This power dissipation with the battery charger in the SOT23-5 package will cause thermal regulation to be entered as depicted in Figure 6-3. Alternatively, the 2mm x 3mm DFN package could be utilized to reduce charge cycle times.

6.1.1.3 External Capacitors

The MCP73831/2 are stable with or without a battery load. In order to maintain good AC stability in the Constant-Voltage mode, a minimum capacitance of 4.7 μF is recommended to bypass the V_{BAT} pin to V_{SS} . This capacitance provides compensation when there is no battery load. In addition, the battery and interconnections appear inductive at high frequencies. These elements are in the control feedback loop during Constant-Voltage mode. Therefore, the bypass capacitance may be necessary to compensate for the inductive nature of the battery pack.

Virtually any good quality output filter capacitor can be used, independent of the capacitor's minimum Effective Series Resistance (ESR) value. The actual value of the capacitor (and its associated ESR) depends on the output load current. A 4.7 μF ceramic, tantalum or aluminum electrolytic capacitor at the output is usually sufficient to ensure stability for output currents up to a 500 mA.

6.1.1.4 Reverse-Blocking Protection

The MCP73831/2 provide protection from a faulted or shorted input. Without the protection, a faulted or shorted input would discharge the battery pack through the body diode of the internal pass transistor.

6.1.1.5 Charge Inhibit

The current regulation set input pin (PROG) can be used to terminate a charge at any time during the charge cycle, as well as to initiate a charge cycle or initiate a recharge cycle.

Placing a programming resistor from the PROG input to V_{SS} enables the device. Allowing the PROG input to float or by applying a logic-high input signal, disables the device and terminates a charge cycle. When disabled, the device's supply current is reduced to 25 μ A, typically.

6.1.1.6 Charge Status Interface

A status output provides information on the state of charge. The output can be used to illuminate external LEDs or interface to a host microcontroller. Refer to Table 5-1 for a summary of the state of the status output during a charge cycle.

6.2 PCB Layout Issues

For optimum voltage regulation, place the battery pack as close as possible to the device's V_{BAT} and V_{SS} pins. This is recommended to minimize voltage drops along the high current-carrying PCB traces.

If the PCB layout is used as a heatsink, adding many vias in the heatsink pad can help conduct more heat to the backplane of the PCB, thus reducing the maximum junction temperature. Figures 6-4 and 6-5 depict a typical layout with PCB heatsinking.

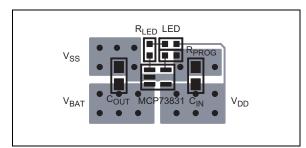


FIGURE 6-4: Typical Layout (Top).

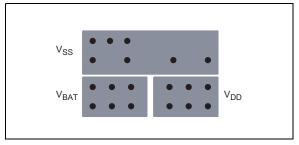
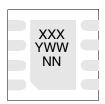


FIGURE 6-5: Typical Layout (Bottom).

7.0 PACKAGING INFORMATION

7.1 **Package Marking Information**

8-Lead DFN (2 mm x 3 mm)



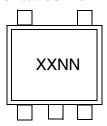
Device	Code
MCP73831T-2ACI/MC	AAE
MCP73831T-2ATI/MC	AAF
MCP73831T-2DCI/MC	AAG
MCP73831T-3ACI/MC	AAH
MCP73831T-4ADI/MC	AAJ
MCP73831T-5ACI/MC	AAK
MCP73832T-2ACI/MC	AAL
MCP73832T-2ATI/MC	AAM
MCP73832T-2DCI/MC	AAP
MCP73832T-3ACI/MC	AAQ
MCP73832T-4ADI/MC	AAR
MCP73832T-5ACI/MC	AAS
Note: Applies to 8-1 ead	DEN

Note: Applies to 8-Lead DFN

Example:



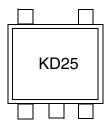
5-Lead SOT-23



Code
KDNN
KENN
KFNN
KGNN
KHNN
KJNN
KKNN
KLNN
KMNN
KPNN
KQNN
KRNN

Note: Applies to 5-Lead SOT-23

Example:



Legend: XX...X Customer-specific information

> Year code (last digit of calendar year) Υ YYYear code (last 2 digits of calendar year) ww Week code (week of January 1 is week '01') NNN Alphanumeric traceability code

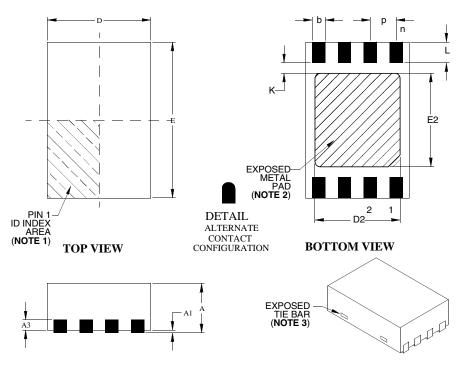
(e3) Pb-free JEDEC designator for Matte Tin (Sn)

This package is Pb-free. The Pb-free JEDEC designator (e3)

can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

8-Lead Plastic Dual-Flat, No-Lead Package (MC) 2x3x0.9 mm Body (DFN) - Saw Singulated



	Units		INCHES		MILLIMETERS*			
Dimension Limi	ts	MIN	NOM	MAX	MIN	NOM	MAX	
Number of Pins	n		8			8		
Pitch	е		.020 BSC			0.50 BSC		
Overall Height	Α	.031	.035	.039	0.80	0.90	1.00	
Standoff	A1	.000	.000 .001 .002			0.02	0.05	
Contact Thickness	А3		.008 REF.			0.20 REF.		
Overall Length	D		.079 BSC		2.00 BSC			
Overall Width	E		.118 BSC		3.00 BSC			
Exposed Pad Length	D2	.051	_	.069	1.30**	_	1.75	
Exposed Pad Width	E2	.059	.059 – .075			_	1.90	
Contact Length §	L	.012 .016 .020			0.30	0.40	0.50	
Contact-to-Exposed Pad §	K	.008	-	_	0.20	_	_	
Contact Width	b	.008	.010	.012	0.20	0.25	0.30	

^{*} Controlling Parameter

Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- ${\bf 2.}\ {\bf Exposed}\ {\bf pad}\ {\bf may}\ {\bf vary}\ {\bf according}\ {\bf to}\ {\bf die}\ {\bf attach}\ {\bf paddle}\ {\bf size}.$
- 3. Package may have one or more exposed tie bars at ends.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

See ASME Y14.5M

REF: Reference Dimension, usually without tolerance, for information purposes only. See ASME Y14.5M

JEDEC Equivalent MO-229 VCED-2

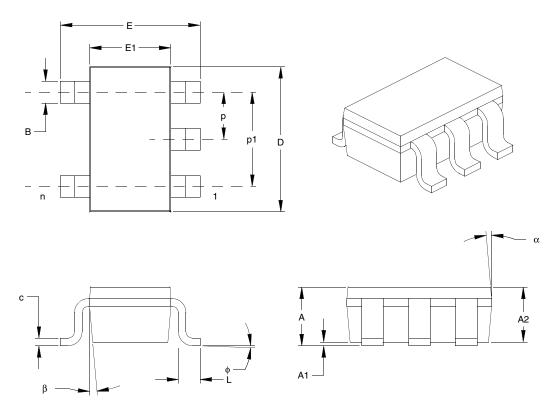
DWG No. C04-123

Revised 09-12-05

^{**} Not within JEDEC parameters

[§] Significant Characteristic

5-Lead Plastic Small Outline Transistor (OT) (SOT-23)



	Units		INCHES*		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		5			5	
Pitch	р		.038			0.95	
Outside lead pitch (basic)	p1		.075			1.90	
Overall Height	Α	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	E	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	L	.014	.018	.022	0.35	0.45	0.55
Foot Angle	f	0	5	10	0	5	10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	а	0	5	10	0	5	10
Mold Draft Angle Bottom	b	0	5	10	0	5	10

^{*} Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side.

EIAJ Equivalent: SC-74A

Drawing No. C04-091

Revised 09-12-05

V	IC	P	73	8	3	1	12
	•			\sim	•		

NOTES:

APPENDIX A: REVISION HISTORY

Revision B (March 2006)

• Added MCP73832 through document.

Revision A (November 2005)

• Original Release of this Document.

N	C	D.	73	R	3	1	12
IVI	V		ľ	v	v	- 1 /	

NOTES:

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	- <u>X</u> XX X /XX	Examples: *		
Device	V _{REG} Options Temperature Package Range	a) MCP73831-2ACI/OT: 4.20V V _{REG} , Options AC, 5LD SOT23 Pkg b) MCP73831T-2ACI/OT: Tape and Reel, 4.20V V _{REG} , Options AC, 5LD SOT23 Pkg		
Device:	MCP73831: Single-Cell Charge Controller MCP73831T: Single-Cell Charge Controller (Tape and Reel) MCP73832 Single-Cell Charge Controller	c) MCP73832-2ACI/MC: 4.20V V _{REG} , Options AC, 8LD DFN Package d) MCP73832T-2ACI/MC: Tape and Reel, 4.20V V _{REG} , Options AC, 8LD DFN Package		
	MCP73832T: Single-Cell Charge Controller (Tape and Reel)	a) MCP73831-2ATI/OT: 4.20V V _{REG} , Options AT, 5LD SOT23 Pkg b) MCP73831T-2ATI/OT: Tape and Reel,		
Regulation Voltage:	Code V _{REG}	4.20V V _{REG} . Options AT, 5LD SOT23 Pkg c) MCP73832-2ATI/MC: 4.20V V _{REG} . Options AT, 8LD DFN Package		
	2 = 4.20V 3 = 4.35V 4 = 4.40V	d) MCP73832T-2ATI/MC: Tape and Reel, 4.20V V _{REG} , Options AT, 8LD DFN Package		
	5 = 4.50V	 a) MCP73831-2DCI/OT: 4.20V V_{REG}, Options DC, 5LD SOT23 Pkg b) MCP73831T-2DCI/OT: Tape and Reel, 4.20V V_{REG}, Options DC, 5LD SOT23 Pkg 		
Options: *	Code I _{PREG} /I _{REG} V _{PTH} /V _{REG} I _{TERM} /I _{REG} V _{RTH} /V _{REG} AC 10 66.5 7.5 96.5 AD 10 66.5 7.5 94 AT 10 71.5 20 94	c) MCP73832-2DCI/MC: 4.20V V _{REG} , Options DC, 8LD DFN Package d) MCP73832T-2DCI/MC: Tape and Reel, 4.20V V _{REG} , Options DC, 8LD DFN Package		
	DC 100 x 7.5 96.5 * Consult Factory for Alternative Device Options	a) MCP73831-3ACI/OT: 4.35V V _{REG} , Options AC, 5LD SOT23 Pkg		
Temperature Range:	I = -40°C to +85°C (Industrial)	 b) MCP73831T-3ACI/OT: Tape and Reel, 4.35V V_{REG}. Options AC, 5LD SOT23 Pkg c) MCP73832-3ACI/MC: 4.35V V_{REG}. 		
Package:	MC = Dual-Flat, No-Lead (2x3 mm body), 8-Lead	Options AC, 8LD DFN Package d) MCP73832T-3ACI/MC: Tape and Reel, 4.35V V _{REG} . Options AC, 8LD DFN Package		
	OT = Small Outline Transistor (SOT23), 5-Lead	a) MCP73831-4ADI/OT: 4.40V V _{REG} , Options AD, 5LD SOT23 Pkg		
		 b) MCP73831T-4ADI/OT: Tape and Reel, 4.40V V_{REG}. Options AD, 5LD SOT23 Pkg c) MCP73832-4ADI/MC: 4.40V V_{REG}. 		
		Options AD, 8LD DFN Package d) MCP73832T-4ADI/MC: Tape and Reel, 4.40V V _{REG} . Options AD, 8LD DFN Package		
		a) MCP73831-5ACI/OT: 4.50V V _{REG} , Options AC, 5LD SOT23 Pkg		
		b) MCP73831T-5ACI/OT: Tape and Reel, 4.50V V _{REG} . Options AC, 5LD SOT23 Pkg c) MCP73832-5ACI/MC: 4.50V V _{REG} .		
		Options AC, 8LD DFN Package d) MCP73832T-5ACI/MC: Tape and Reel, 4.50V V _{REG} , Options AC, 8LD DFN Package		

* Consult Factory for Alternate Device Options

١/	CP	72	Ω	1	12
IVI			UJ	•	

NOTES:

Note the following details of the code protection feature on Microchip devices:

- · Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the
 intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WAR-RANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE. MERCHANTABILITY OR FITNESS FOR PURPOSE. Microchip disclaims all liability arising from this information and its use. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights.

Trademarks

The Microchip name and logo, the Microchip logo, Accuron, dsPIC, KEELOQ, microID, MPLAB, PIC, PICmicro, PICSTART, PRO MATE, PowerSmart, rfPIC, and SmartShunt are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

AmpLab, FilterLab, Migratable Memory, MXDEV, MXLAB, SEEVAL, SmartSensor and The Embedded Control Solutions Company are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Analog-for-the-Digital Age, Application Maestro, dsPICDEM, dsPICDEM.net, dsPICworks, ECAN, ECONOMONITOR, FanSense, FlexROM, fuzzyLAB, In-Circuit Serial Programming, ICSP, ICEPIC, Linear Active Thermistor, MPASM, MPLIB, MPLINK, MPSIM, PICkit, PICDEM, PICDEM.net, PICLAB, PICtail, PowerCal, PowerInfo, PowerMate, PowerTool, REAL ICE, rfLAB, rfPICDEM, Select Mode, Smart Serial, SmartTel, Total Endurance, UNI/O, WiperLock and Zena are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

SQTP is a service mark of Microchip Technology Incorporated in the U.S.A.

All other trademarks mentioned herein are property of their respective companies.

© 2006, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

Printed on recycled paper.

QUALITY MANAGEMENT SYSTEM

CERTIFIED BY DNV

ISO/TS 16949:2002

Microchip received ISO/TS-16949:2002 quality system certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona and Mountain View, California in October 2003. The Company's quality system processes and procedures are for its PICmicro® 8-bit MCUs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.



WORLDWIDE SALES AND SERVICE

AMERICAS

Corporate Office

2355 West Chandler Blvd. Chandler, AZ 85224-6199 Tel: 480-792-7200

Fax: 480-792-7277 Technical Support:

http://support.microchip.com

Web Address: www.microchip.com

Atlanta

Alpharetta, GA Tel: 770-640-0034 Fax: 770-640-0307

Boston

Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca. IL

Tel: 630-285-0071 Fax: 630-285-0075

Dallas

Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

Detroit

Farmington Hills, MI Tel: 248-538-2250 Fax: 248-538-2260

Kokomo

Kokomo, IN Tel: 765-864-8360 Fax: 765-864-8387

Los Angeles

Mission Viejo, CA Tel: 949-462-9523 Fax: 949-462-9608

San Jose

Mountain View, CA Tel: 650-215-1444 Fax: 650-961-0286

Toronto

Mississauga, Ontario,

Canada

Tel: 905-673-0699 Fax: 905-673-6509

ASIA/PACIFIC

Australia - Sydney

Tel: 61-2-9868-6733 Fax: 61-2-9868-6755

China - Beijing

Tel: 86-10-8528-2100 Fax: 86-10-8528-2104

China - Chengdu

Tel: 86-28-8676-6200 Fax: 86-28-8676-6599

China - Fuzhou

Tel: 86-591-8750-3506 Fax: 86-591-8750-3521

China - Hong Kong SAR

Tel: 852-2401-1200 Fax: 852-2401-3431

China - Qingdao

Tel: 86-532-8502-7355 Fax: 86-532-8502-7205

China - Shanghai

Tel: 86-21-5407-5533 Fax: 86-21-5407-5066

China - Shenyang

Tel: 86-24-2334-2829 Fax: 86-24-2334-2393

China - Shenzhen

Tel: 86-755-8203-2660 Fax: 86-755-8203-1760

China - Shunde

Tel: 86-757-2839-5507 Fax: 86-757-2839-5571

China - Wuhan

Tel: 86-27-5980-5300 Fax: 86-27-5980-5118

China - Xian

Tel: 86-29-8833-7250 Fax: 86-29-8833-7256

ASIA/PACIFIC

India - Bangalore

Tel: 91-80-4182-8400 Fax: 91-80-4182-8422

India - New Delhi

Tel: 91-11-5160-8631 Fax: 91-11-5160-8632

India - Pune

Tel: 91-20-2566-1512 Fax: 91-20-2566-1513

Japan - Yokohama

Tel: 81-45-471- 6166 Fax: 81-45-471-6122

Korea - Gumi

Tel: 82-54-473-4301 Fax: 82-54-473-4302

Korea - Seoul

Tel: 82-2-554-7200 Fax: 82-2-558-5932 or

82-2-558-5934

Malaysia - Penang Tel: 60-4-646-8870 Fax: 60-4-646-5086

Philippines - Manila

Tel: 63-2-634-9065 Fax: 63-2-634-9069

Singapore

Tel: 65-6334-8870 Fax: 65-6334-8850 Taiwan - Hsin Chu

Tel: 886-3-572-9526

Fax: 886-3-572-6459

Taiwan - Kaohsiung

Tel: 886-7-536-4818 Fax: 886-7-536-4803

Taiwan - Taipei

Tel: 886-2-2500-6610 Fax: 886-2-2508-0102

Thailand - Bangkok

Tel: 66-2-694-1351 Fax: 66-2-694-1350

EUROPE

Austria - Wels

Tel: 43-7242-2244-399 Fax: 43-7242-2244-393

Denmark - Copenhagen

Tel: 45-4450-2828 Fax: 45-4485-2829

France - Paris

Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

Germany - Munich

Tel: 49-89-627-144-0 Fax: 49-89-627-144-44

Italy - Milan

Tel: 39-0331-742611 Fax: 39-0331-466781

Netherlands - Drunen

Tel: 31-416-690399 Fax: 31-416-690340

Spain - Madrid

Tel: 34-91-708-08-90 Fax: 34-91-708-08-91

UK - Wokingham Tel: 44-118-921-5869

Fax: 44-118-921-5820

02/16/06