## General Description

The MAX8595X/MAX8596X drive up to nine white LEDs with constant current and high efficiency to provide LCD backlighting in cell phones, PDAs, and other handheld devices. The series connection allows the LED currents to be identical for uniform brightness and minimizes the number of traces to the LEDs. The MAX8595X regulates constant LED current over the entire temperature range. The MAX8596X features an ambient-temperature derating function to avoid overdriving the white LEDs during high ambient temperatures, enabling higher drive current below $+42^{\circ} \mathrm{C}$.
A single Dual Mode ${ }^{\text {TM }}$ input provides a simple means of brightness adjustment and on/off control. Fast 1 MHz current-mode PWM operation allows for small input and output capacitors and a small inductor while minimizing ripple on the input supply/battery. Soft-start eliminates inrush current during startup.
The MAX8595X/MAX8596X are available in a spacesaving, $8-\mathrm{pin}, 3 \mathrm{~mm} \times 3 \mathrm{~mm}$ TDFN package.
$\qquad$

## Applications

Cell Phones and Smart Phones
PDAs, Palmtops, and Wireless Handhelds
e-Books and Subnotebooks
White LED Display Backlighting

Dual Mode is a trademark of Maxim Integrated Products, Inc.
Typical Operating Circuit


- Up to Nine LEDs at 25mA
- Temperature Derating Function to Allow Fewer LEDs for Same Light (MAX8596X)
- 86\% Efficiency (PLEDs / Pin)
- 1.7\% Current-Regulation Accuracy
- Output Overvoltage Protection
- Flexible Dimming Control Analog
Direct-PWM Internal Filter
- 1MHz PWM Switching Frequency
- $0.1 \mu \mathrm{~F}$ Output Capacitor
-12mVp-p Low Input Ripple
- Soft-Start Eliminates Inrush Current
- 2.6V to 5.5V Input Range
- $0.3 \mu \mathrm{~A}$ Shutdown Current
- Pin Compatible with the MAX1561 and MAX1599
- TDFN 3mm x 3mm x 0.8mm Package with Exposed Paddle

Ordering Information
\(\left.$$
\begin{array}{|ccc|}\hline \text { PART } & \text { TEMP RANGE PIN-PACKAGE } & \begin{array}{c}\text { PKG } \\
\text { CODE }\end{array}
$$ <br>
\hline MAX8595XETA \& -40^{\circ} \mathrm{C} to+85^{\circ} \mathrm{C} \& 8 TDFN 3 \mathrm{~mm} \times 3 \mathrm{~mm} <br>
\hline T833-1 <br>

\hline MAX8596XETA \& -40^{\circ} \mathrm{C} to+85^{\circ} \mathrm{C} \& 8 TDFN 3 \mathrm{~mm} \times 3 \mathrm{~mm}\end{array}\right]\)| T833-1 |
| :--- |

*Future product-contact factory for availability.
Pin Configuration


# High-Efficiency, 36V Step-Up Converters with TA Derating Option for 2 to 9 White LEDs 

## ABSOLUTE MAXIMUM RATINGS

| IN to GND | -0.3V to +6V |
| :---: | :---: |
| PGND to GND | -0.3V to +0.3 V |
| LX, OUT to GND. | -0.3 V to +40 V |
| CTRL to GND.... | -0.3V to the lower of +6 V or $\left(\mathrm{V}_{\text {IN }}+2 \mathrm{~V}\right)$ |
| COMP, CS to GN | .........................-0.3V to (VIN + 0.3V) |
| Ix | ..1.0ARMS |

Continuous Power Dissipation $\left(\mathrm{T}_{\mathrm{A}}=+70^{\circ} \mathrm{C}\right)$
8 -Pin TDFN $3 \mathrm{~mm} \times 3 \mathrm{~mm}$
(derate $24.4 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ )............................ 1950 mW
Operating Temperature Range ........................... $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Junction Temperature ..................................................... $+150^{\circ} \mathrm{C}$
Storage Temperature Range ............................. $65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Lead Temperature (soldering, 10s) ................................. $+300^{\circ} \mathrm{C}$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(V_{I N}=3.0 \mathrm{~V}, \mathrm{~L}=22 \mu \mathrm{H}, \mathrm{C}_{\text {IN }}=2.2 \mu \mathrm{~F}, \mathrm{COUT}=0.1 \mu \mathrm{~F}, \mathrm{C}\right.$ COMP $=0.1 \mu \mathrm{~F}, \mathrm{R}_{\text {SENSE }}=13 \Omega, \mathrm{~V}_{\mathrm{CTRL}}=1.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Voltage |  |  | 2.6 |  | 5.5 | V |
| UVLO Threshold | VIN rising or falling |  | 2.10 | 2.38 | 2.55 | V |
| UVLO Hysteresis |  |  |  | 30 |  | mV |
| Quiescent Current | No switching |  |  | 0.5 | 0.7 | mA |
| Shutdown Supply Current | $\begin{aligned} & \text { CTRL }=\text { GND }, \\ & \text { VOUT }=\mathrm{V}_{\text {IN }} \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | 0.3 | 2 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{T}_{\mathrm{A}}=+85^{\circ} \mathrm{C}$ | 1 |  |  |  |
| OVLO Threshold | Vout rising |  | 36 | 38 | 40 | V |
| OVLO Hysteresis |  |  |  | 2 |  | V |
| OUT Input Bias Current | VOUT $=32 \mathrm{~V}, \mathrm{~V}_{\text {CTRL }}>0.24 \mathrm{~V}$ |  | 9 | 20 | 35 | $\mu \mathrm{A}$ |
|  | OUT $=1 \mathrm{~N}, \mathrm{CTRL}=\mathrm{GND}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | 0.01 | 1 |  |
|  |  | $\mathrm{T}_{\mathrm{A}}=+85^{\circ} \mathrm{C}$ | 0.1 |  |  |  |
| Output Voltage Range | (Note 2) |  | VIN - $V_{D}$ |  | 36 | V |
| ERROR AMPLIFIER |  |  |  |  |  |  |
| CTRL to CS Regulation | $\begin{aligned} & V_{\mathrm{CTRL}}=1.50 \mathrm{~V}, \\ & \mathrm{~V} \mathrm{IN}=2.6 \mathrm{~V} \text { to } 5.5 \mathrm{~V} \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 0.295 | 0.300 | 0.305 | V |
|  |  | $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 0.292 | 0.300 | 0.308 |  |
|  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 0.290 | 0.300 | 0.310 |  |
| CS Input Bias Current | $\mathrm{V}_{\text {CS }}=\mathrm{V}_{\text {CTRL }} / 5$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | 0.01 | 1 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{T}_{\mathrm{A}}=+85^{\circ} \mathrm{C}$ |  | 0.03 |  |  |
| CS Maximum Brightness Clamp Voltage | MAX8595X, $\mathrm{V}_{\text {CTRL }}=3.0 \mathrm{~V}$ |  | 310 | 330 | 347 | mV |
|  | MAX8596X, <br> $V_{C T R L}=3.0 \mathrm{~V}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+25^{\circ} \mathrm{C}$ | 330 | 345 | 360 |  |
|  |  | $\mathrm{T}_{\mathrm{A}}=+42^{\circ} \mathrm{C}$ |  | 343 |  |  |
|  |  | $\mathrm{T}_{\mathrm{A}}=+85^{\circ} \mathrm{C}$ |  | 106.5 |  |  |
| CTRL Voltage for CS Maximum Brightness Clamp | MAX8595Z |  |  | 1.65 |  | V |
|  | MAX8596Z |  |  | 1.72 |  |  |
| CS Derating Function Start Temperature | MAX8596Z, $\mathrm{V}_{\text {CTRL }}=3.0 \mathrm{~V}$ |  |  | +42 |  | ${ }^{\circ} \mathrm{C}$ |

## High-Efficiency, 36V Step-Up Converters with TA Derating Option for 2 to 9 White LEDs

## ELECTRICAL CHARACTERISTICS (continued)

$\left(\mathrm{V}_{\text {IN }}=3.0 \mathrm{~V}, \mathrm{~L}=22 \mu \mathrm{H}, \mathrm{CIN}_{\mathrm{IN}}=2.2 \mu \mathrm{~F}, \mathrm{COUT}=0.1 \mu \mathrm{~F}, \mathrm{C}_{\text {COMP }}=0.1 \mu \mathrm{~F}, \mathrm{R}_{\text {SENSE }}=13 \Omega, \mathrm{~V}_{\mathrm{CTRL}}=1.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}\right.$ to $+85^{\circ} \mathrm{C}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CS Derating Function Slope | MAX8596X, V $\mathrm{V}_{\text {CTRL }}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+42^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |  |  | -5.5 |  | $\mathrm{mV} /{ }^{\circ} \mathrm{C}$ |
| CTRL Input Resistance | $\mathrm{V}_{\text {CTRL }} \leq 1.5 \mathrm{~V}$ |  | 250 | 500 | 780 | k ת |
| CTRL Dual-Mode Threshold |  |  | 100 | 170 | 240 | mV |
| CTRL Dual-Mode Hysteresis |  |  |  | 5 |  | mV |
| CTRL Shutdown Enable Delay | (Note 3) |  | 6.0 | 8.2 | 10.5 | ms |
| CS to COMP Transconductance | $\mathrm{V}_{\text {COMP }}=1.5 \mathrm{~V}$ |  | 32 | 50 | 82 | $\mu \mathrm{S}$ |
| COMP Input Resistance to Ground | In shutdown, UVLO or OVLO |  |  | 20 |  | k $\Omega$ |
| OSCILLATOR |  |  |  |  |  |  |
| Operating Frequency |  |  | 0.75 | 1.0 | 1.25 | MHz |
| Minimum Duty Cycle | PWM mode |  |  | 12 |  | \% |
|  | Pulse skipping |  |  | 0 |  |  |
| Maximum Duty Cycle | CTRL = IN, CS = GND |  | 94 | 95 |  | \% |
| n-CHANNEL SWITCH |  |  |  |  |  |  |
| LX On-Resistance | $\mathrm{LLX}=190 \mathrm{~mA}$ |  |  | 0.8 | 1.35 | $\Omega$ |
| LX Leakage Current | VLX $=36 \mathrm{~V}, \mathrm{CTRL}=\mathrm{GND}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | 0.01 | 5 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{T}_{\mathrm{A}}=+85^{\circ} \mathrm{C}$ |  | 1 |  |  |
| LX Current Limit | Duty cycle $=90 \%$ |  | 500 | 700 | 900 | mA |

Note 1: Parameters are $100 \%$ production tested at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$. Limits over the operating temperature range are regulated by design and characterization.
Note 2: $V_{D}$ is the forward-voltage drop of the Schottky diode in Figure 1
Note 3: Time from CTRL going below the Dual-Mode threshold to IC shutdown.

## High-Efficiency, 36V Step-Up Converters with TA Derating Option for 2 to 9 White LEDs

(Circuit of Figure 1, $\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}$, $\operatorname{LLED}=25 \mathrm{~mA}, \mathrm{~L}=22 \mu \mathrm{H}, \mathrm{CIN}=2.2 \mu \mathrm{~F}$, COUT $=0.1 \mu \mathrm{~F}, \mathrm{C} C O M P=0.1 \mu \mathrm{~F}$, RSENSE $=13 \Omega, 4 \mathrm{LEDs}, \mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)


LED CURRENT
vs. DIRECT-PWM DIMMING



500ns/div

EFFICIENCY
vs. LED CURRENT


LED CURRENT
vs. AMBIENT TEMPERATURE



# High-Efficiency, 36V Step-Up Converters with TA Derating Option for 2 to 9 White LEDs 

## Typical Operating Characteristics (continued)

(Circuit of Figure 1, V IN $=3.6 \mathrm{~V}, \operatorname{ILED}=25 \mathrm{~mA}, \mathrm{~L}=22 \mu \mathrm{H}, \mathrm{CIN}=2.2 \mu \mathrm{~F}, \mathrm{COUT}=0.1 \mu \mathrm{~F}, \mathrm{C} C O M P=0.1 \mu \mathrm{~F}$, RSENSE $=13 \Omega, 4 \mathrm{LEDs}, \mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)


Pin Description

| PIN | NAME | FUNCTION |
| :---: | :---: | :---: |
| 1 | OUT | Overvoltage Sense. When Vout is greater than 38 V (typ), the internal n-channel MOSFET turns off until Vout drops below 36V (typ), then the IC reenters soft-start. Connect a $0.1 \mu \mathrm{~F}$ ceramic capacitor from OUT to ground. In shutdown, VOUT is one diode drop below VIN. |
| 2 | IN | Input Voltage Supply. The input voltage range is 2.6 V to 5.5 V . Connect a $2.2 \mu \mathrm{~F}$ ceramic capacitor from IN to GND. |
| 3 | CTRL | Brightness Control Input. The voltage applied to CTRL controls LED brightness. Varying the voltage from 0.24 V to 1.65 V ( 1.72 V for the MAX8596X) adjusts the brightness from dim to $100 \%$ brightness, respectively. Any voltage above 1.65 V (1.72V) does not increase brightness. Hold CTRL below 100 mV to shut down the IC after an 8.2 ms delay. |
| 4 | CS | Current-Sense Feedback Input. Connect a resistor from CS to GND to set the LED bias current. The voltage at CS regulates to $\mathrm{V}_{\text {CTRL }} / 5$ or 0.330 V ( 0.343 V for the MAX8596X), whichever is lower. |
| 5 | COMP | Compensation Input. Connect a $0.1 \mu \mathrm{~F}$ ceramic capacitor (CCOMP) from COMP to GND. CcOMP stabilizes the converter, controls soft-start, and lowpass filters direct PWM dimming at CTRL. CCOMP discharges to 0 V through an internal $20 \mathrm{k} \Omega$ resistor in shutdown. |
| 6 | GND | Ground. Connect to PGND and the exposed pad directly under the IC. |
| 7 | PGND | Power Ground. Connect to GND and the exposed pad directly under the IC. |
| 8 | LX | Inductor Connection. Connect LX to the node between the inductor and the Schottky diode. LX is high impedance in shutdown. |
| - | EP | Exposed Pad. Connect to a large ground plane for maximum package heat dissipation. Connect directly to GND and PGND under the IC. |

# High-Efficiency, 36V Step-Up Converters with TA Derating Option for 2 to 9 White LEDs 


#### Abstract

Detailed Description The high efficiency and small size of the MAX8595X/ MAX8596X make them ideally suited to drive up to nine series-connected LEDs. These devices operate as a boost DC-DC converter that regulates output current rather than voltage. The MAX8595X/MAX8596X provide even illumination by sourcing the same output current through each LED, eliminating the need for expensive factory calibration. The fast 1 MHz internal oscillator allows for a small inductor and small input and output capacitors while minimizing input and output ripple. The single analog control input (CTRL) allows easy adjustment of LED brightness and on/off control. This allows simple logic-level on/off control, analog voltage control, or PWM duty-cycle control of both brightness and shutdown. In shutdown, supply current is reduced to a low $0.3 \mu \mathrm{~A}$ (typ). A soft-start gradually illuminates the LEDs, eliminating the inrush current during startup. The MAX8596X has the additional feature of derating LED current as ambient temperature rises. Above $+42^{\circ} \mathrm{C}$, the CS regulation voltage is reduced at a rate of $5.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$, thus reducing the LED current.


## Soft-Start

The MAX8595X/MAX8596X attain soft-start by charging Ccomp gradually with a current source. When Vcomp rises above 1.25 V , the internal MOSFET begins switching at a reduced duty cycle. When Vcomp rises above 2.25 V , the duty cycle is at its maximum. See the Typical Operating Characteristics for an example of soft-start operation.

## Shutdown

The MAX8595X/MAX8596X enter shutdown when VCTRL is less than 100 mV for more than 8.2 ms . In shutdown, supply current is reduced to $0.3 \mu \mathrm{~A}$ (typ) by powering down the entire IC except for the CTRL voltage-detection circuitry. CCOMP is discharged during shutdown, allowing the device to reinitiate soft-start when it is enabled. Although the internal n-channel MOSFET does not switch in shutdown, there is still a DC current path between the input and the LEDs through the inductor and Schottky diode. The minimum forward voltage of the LED array must exceed the maximum input voltage to ensure that the LEDs remain off in shutdown. However, with two or more LEDs, the forward voltage is large enough to keep leakage current low, less than $1 \mu \mathrm{~A}$ (typ). Typical shutdown timing characteristics are shown in the Typical Operating Characteristics.

## Overvoltage Protection

Overvoltage lockout (OVLO) occurs when Vout is above 38 V (typ). The protection circuitry stops the inter-


Figure 1. Functional Diagram and Typical Application Circuit
nal MOSFET from switching and causes $V_{C O M P}$ to decay towards OV. The device comes out of OVLO and into soft-start when Vout falls below 36V (typ).

## Ambient Temperature Derating Function

(MAX8596X)
The MAX8596X limits the maximum LED current depending on the die temperature. VCS is limited to 343 mV up to $+42^{\circ} \mathrm{C}$. Once the temperature reaches $+42^{\circ} \mathrm{C}$, the maximum VCs declines by $5.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ until the minimum 106.5 mV threshold is reached at $+85^{\circ} \mathrm{C}$. Due to the package's exposed paddle, the die temperature is always very close to the PC board temperature.
The temperature derating function allows the LED current to be safely set higher at normal operating temperatures, thereby allowing either a brighter display or fewer LEDs to be used for normal display brightness. See the Typical Operating Characteristics for LED Current vs. Ambient Temperature.

# High-Efficiency, 36V Step-Up Converters with TA Derating Option for 2 to 9 White LEDs 

## Design Procedure

## Adjusting LED Current

Adjusting the output current of the MAX8595X/ MAX8596X changes the brightness of the LEDs. An analog input (CTRL) and the sense-resistor value set the output current. Output current is given by:

$$
\text { ILED }=\text { VCTRL } /(5 \times \text { RSENSE })
$$

The VCTRL voltage range for adjusting output current is 0.24 V to 1.65 V (or 1.72 V for the MAX8596X). To set the maximum current, calculate RSENSE when VCTRL is at its maximum as follows:

For the MAX8595X, RSENSE $=1.65 /(5 \times \operatorname{lLED}(\mathrm{MAX}))$
For the MAX8596X, RSENSE $=1.72 /(5 \times \operatorname{lLED}(M A X))$
Power dissipation in RSENSE is typically less than 10 mW , allowing the use of a small surface-mount resistor.

## PWM Dimming Control

CTRL is also used as a digital input allowing LED brightness control with a logic-level PWM signal applied directly to CTRL. The frequency range is from 200 Hz to 200 kHz , while $0 \%$ duty cycle corresponds to zero current and 100\% duty cycle corresponds to full current. The error amplifier and compensation capacitor form a lowpass filter so PWM dimming results in DC current to the LEDs without the need for any additional RC filters; see the Typical Operating Characteristics.

## Capacitor Selection

Ceramic capacitors with X5R, X7R, or better dielectric are recommended for stable operation over the entire operating temperature range. The exact values of input and output capacitors are not critical. The typical value for the input capacitor is $2.2 \mu \mathrm{~F}$, and the typical value for the output capacitor is $0.1 \mu \mathrm{~F}$. Higher value capacitors can be used to reduce input and output ripple, but at the expense of size and higher cost. CCOMP stabilizes the converter and controls soft-start. Connect a $0.1 \mu \mathrm{~F}$ capacitor from COMP to GND. For stable operation, Cout must not exceed 10 times Ccomp.

## Inductor Selection

Inductor values range from $10 \mu \mathrm{H}$ to $47 \mu \mathrm{H}$. A $22 \mu \mathrm{H}$ inductor optimizes the efficiency for most applications while maintaining low 12 mV P-P input ripple. With input voltages near 5V, a larger value of inductance can be
more efficient. To prevent core saturation, ensure that the inductor-saturation current rating exceeds the peak inductor current for the application. Calculate the peak inductor current with the following formula:

$$
I_{\text {PEAK }}=\frac{\mathrm{V}_{\mathrm{OUT}(\mathrm{MAX})} \times \mathrm{I}_{\mathrm{LED}(\mathrm{MAX})}}{0.9 \times \mathrm{V}_{\operatorname{IN}(\mathrm{MIN})}}+\frac{\mathrm{V}_{\mathrm{IN}(\mathrm{MIN})} \times 0.9 \mu \mathrm{~s}}{2 \times \mathrm{L}}
$$

Schottky Diode Selection
The high switching frequency of the MAX8595X/ MAX8596X demands a high-speed rectification diode (D1) for optimum efficiency. A Schottky diode is recommended due to its fast recovery time and low forwardvoltage drop. Ensure that the diode's average and peak current rating exceed the average output current and peak inductor current. In addition, the diode's reverse breakdown voltage must exceed VOUT. The RMS diode current can be approximated from:

$$
\operatorname{IDIODE(RMS)}=\sqrt{\text { IOUT } \times I_{\text {PEAK }}}
$$

## Applications Information

## PC Board Layout

Due to fast switching waveforms and high-current paths, careful PC board layout is required. An evaluation kit (MAX8596XEVKIT) is available to speed design.
When laying out a board, minimize trace lengths between the IC and RSENSE, the inductor, the diode, the input capacitor, and the output capacitor. Keep traces short, direct, and wide. Keep noisy traces, such as the LX node trace, away from CS. The IN bypass capacitor (CIN) should be placed as close to the IC as possible. PGND and GND should be connected directly to the exposed paddle underneath the IC. The ground connections of CIN and COUT should be as close together as possible. The traces from IN to the inductor and from the Schottky diode to the LEDs can be longer.

## Chip Information

TRANSISTOR COUNT: 2143
PROCESS: BiCMOS

## High-Efficiency, 36V Step-Up Converters with TA Derating Option for 2 to 9 White LEDs

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)


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