NOT RECOMMENDED FOR NEW DESIGNS January 22, 2004

FN7196

Triple 400MHz Fixed Gain Amplifier with Enable



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The EL5396A is a triple channel, fixed gain amplifier with a bandwidth of 400MHz, making these amplifiers ideal

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for today's high speed video and monitor applications. The EL5396A features internal gain setting resistors and can be configured in a gain of +1, -1 or +2. The same bandwidth is seen in both gain-of-1 and gain-of-2 applications.

With a supply current of just 9mA per amplifier and the ability to run from a single supply voltage from 5V to 10V, these amplifiers are also ideal for hand held, portable or battery powered equipment.

The EL5396A also incorporates an enable and disable function to reduce the supply current to 100µA typical per amplifier. Allowing the CE pin to float or applying a low logic level will enable the amplifier.

For applications where board space is critical, the EL5396A is offered in the 16-pin QSOP package, as well as a 16-pin SO (0.150"). The EL5396A is specified for operation over the full industrial temperature range of -40°C to +85°C.

Ordering Information

PART NUMBER	PACKAGE	TAPE & REEL	PKG. NO.
EL5396ACS	16-Pin SO (0.150")	-	MDP0027
EL5396ACS-T7	16-Pin SO (0.150")	7"	MDP0027
EL5396ACS-T13	16-Pin SO (0.150")	13"	MDP0027
EL5396ACU	16-Pin QSOP	-	MDP0040
EL5396ACU-T13	16-Pin QSOP	13"	MDP0040

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Features

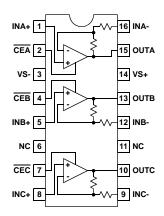
- Gain selectable (+1, -1, +2)
- 400MHz -3dB bandwidth (A_V = 1, 2)
- 9mA supply current (per amplifier)
- Single and dual supply operation, from 5V to 10V or ±2.5V to ±5V
- · Fast enable/disable
- Power-down
- Available in 16-pin QSOP package
- Single (EL5196) available
- 200MHz, 3mA products available (EL5197 & EL5397)

Applications

- · Video amplifiers
- Cable drivers
- **RGB** amplifiers
- Test equipment
- Instrumentation
- Current to voltage converters

Pinout





Absolute Maximum Ratings $(T_A = 25^{\circ}C)$

Supply Voltage between V _S + and V _S 11V	
Maximum Continuous Output Current	
Operating Junction Temperature	
Differential Input Voltage3V	

Power Dissipation	See Curves
Pin Voltages	V_{S} 0.5V to V_{S} + +0.5V
Storage Temperature	65°C to +150°C
Operating Temperature	40°C to +85°C

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

PARAMETER	DESCRIPTION	CONDITIONS	MIN	ТҮР	MAX	UNIT
AC PERFORMA	NCE		l			
BW	-3dB Bandwidth	A _V = +1		400		MHz
		A _V = +2		400		MHz
		A _V = -1		400		MHz
BW1	0.1dB Bandwidth			35		MHz
SR	Slew Rate	$V_{O} = -2.5V$ to +2.5V, $A_{V} = +2$	2400	2600		V/µs
t _S	0.1% Settling Time	V _{OUT} = -2.5V to +2.5V, AV = -1		9		ns
C _S	Channel Separation	f = 5MHz		68		dB
e _N	Input Voltage Noise			3.8		nV/√Hz
i _N -	IN- Input Current Noise			25		pA/√Hz
i _N +	IN+ Input Current Noise			55		pA/√Hz
dG	Differential Gain Error (Note 1)	A _V = +2		0.035		%
dP	Differential Phase Error (Note 1)	A _V = +2		0.04		٥
DC PERFORMA	NCE			1	1	1
V _{OS}	Offset Voltage		-15	1	15	mV
T _C V _{OS}	Input Offset Voltage Temperature Coefficient	Measured from T _{MIN} to T _{MAX}		5		µV/°C
A _E	Gain Error	$V_{O} = -3V$ to $+3V$	-2	1.3	2	%
R _F , R _G	Internal R_F and R_G		320	400	480	Ω
INPUT CHARAC	TERISTICS					
CMIR	Common Mode Input Range		±3V	±3.3V		V
+I _{IN}	+ Input Current		-120	40	120	μA
-I _{IN}	- Input Current		-40	4	40	μA
R _{IN}	Input Resistance			27		kΩ
C _{IN}	Input Capacitance			0.5		pF
OUTPUT CHAR	ACTERISTICS					
V _O	Output Voltage Swing	$R_L = 150\Omega$ to GND	±3.4V	±3.7V		V
		$R_L = 1k\Omega$ to GND	±3.8V	±4.0V		V
IOUT	Output Current	$R_L = 10\Omega$ to GND	95	120		mA
ENABLE (SELE	CTED PACKAGES ONLY)					
t _{EN}	Enable Time			40		ns
t _{DIS}	Disable Time (Note 2)			600		ns

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EL5396A

PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
IIHCE	CE pin Input High Current	CE = V _S +		0.8	6	μA
I _{ILCE}	CE pin Input Low Current	CE = V _S -		0	-0.1	μA
VIHCE	CE pin Input High Voltage for Power Down		V _S + 1			V
VILCE	CE pin Input Low Voltage for Power Up				V _S + -3	V
SUPPLY			I			
I _{SON}	Supply Current - Enabled (per amplifier)	No load, $V_{IN} = 0V$, $\overline{CE} = -5V$	8	9	11	mA
ISOFF	Supply Current - Disabled (per amplifier)	No load, $V_{IN} = 0V$, $\overline{CE} = +4.5V$		100	150	μA
PSRR	Power Supply Rejection Ratio	DC, $V_{S} = \pm 4.75V$ to $\pm 5.25V$	55	75		dB
-IPSR	- Input Current Power Supply Rejection	DC, V _S = ±4.75V to ±5.25V	-2		2	µA/V

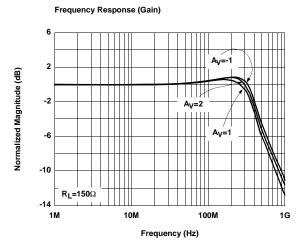
Electrical Specifications V_{S} + = +5V, V_{S} - = -5V, R_{L} = 150 Ω , T_{A} = 25°C unless otherwise specified. (Continued)

NOTES:

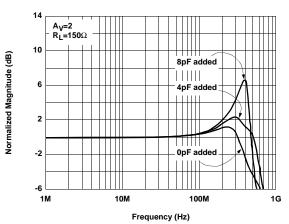
1. Standard NTSC test, AC signal amplitude = $286mV_{P-P}$, f = 3.58MHz

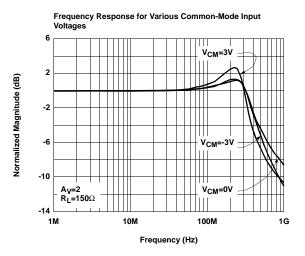
2. Measured from the application of CE logic signal until the output voltage is at the 50% point between initial and final values

Typical Performance Curves

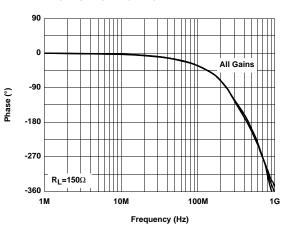


Frequency Response for Various CL

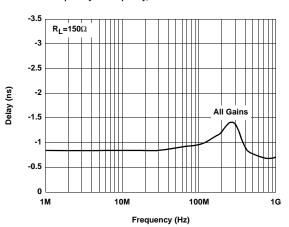


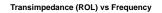


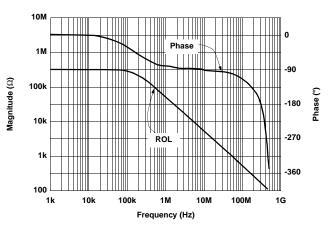
Frequency Response (Phase)



Group Delay vs Frequency, All Gains



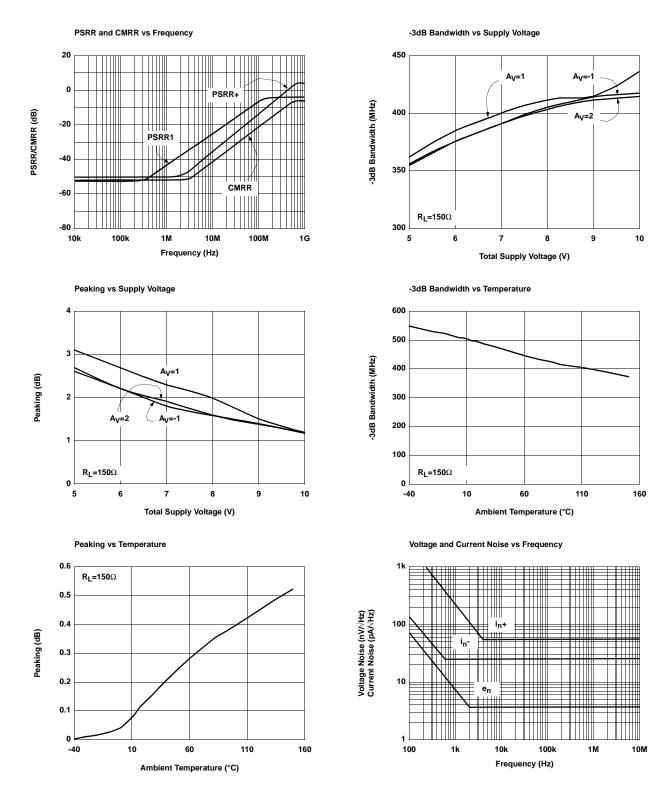




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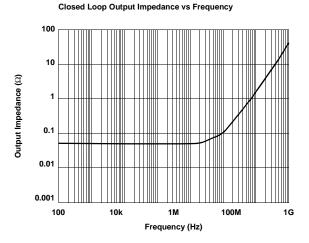
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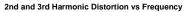
Typical Performance Curves (Continued)

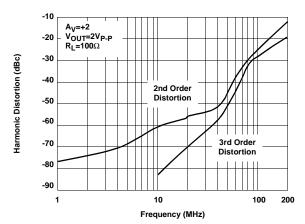


Input Power Intercept (dBm)

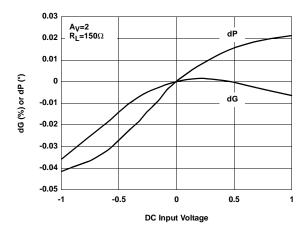
Typical Performance Curves (Continued)



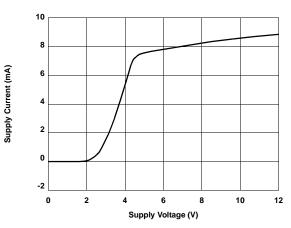


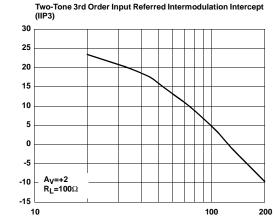






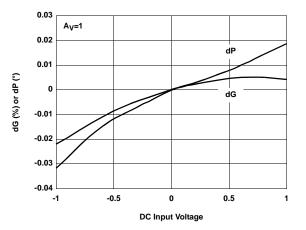
Supply Current vs Supply Voltage



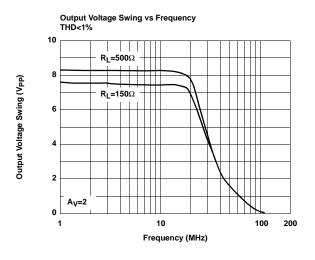


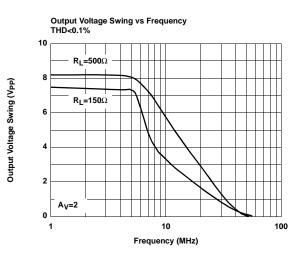
Frequency (MHz)

Differential Gain/Phase vs DC Input Voltage at 3.58MHz

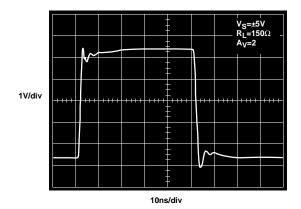


Small Signal Step Response

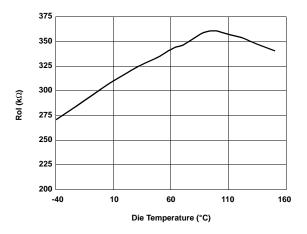




Large Signal Step Response



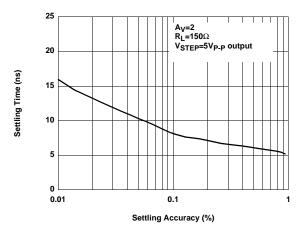
Transimpedance (Rol) vs Temperature



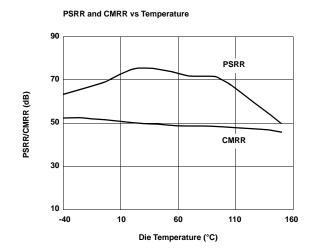
200mV/div

10ns/div

Settling Time vs Settling Accuracy



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Offset Voltage vs Temperature

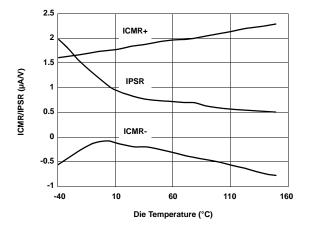
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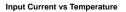
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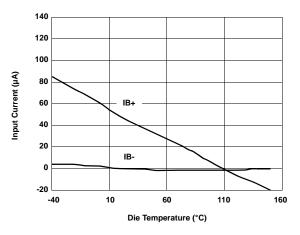
-1 -40

V_{OS} (mV)





ICMR and IPSR vs Temperature





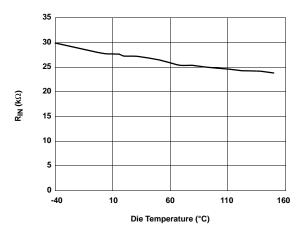
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Die Temperature (°C)

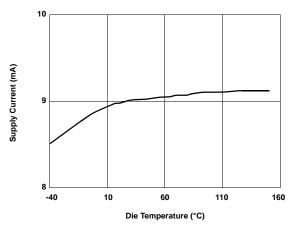
110

160

10



Supply Current vs Temperature



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Output Current vs Temperature

Sink

Source

10

140

135

130

125

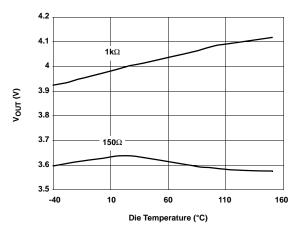
120

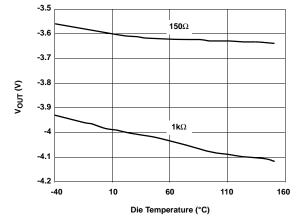
115

-40

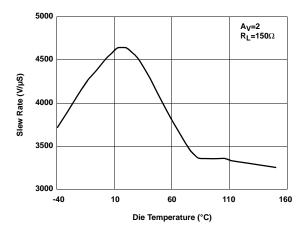
lout (mA)

Positive Output Swing vs Temperature for Various Loads

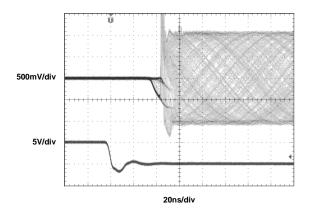












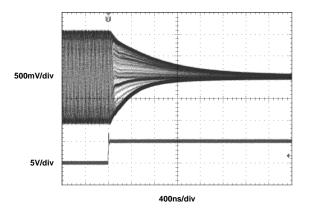
60

Die Temperature (°C)

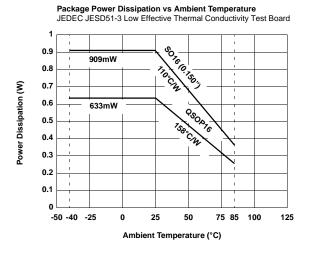
110

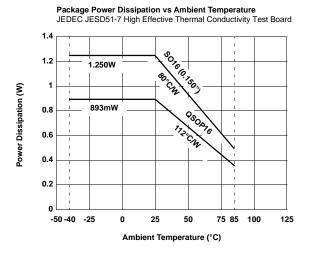
160





Negative Output Swing vs Temperature for Various Loads





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Pin Descriptions

16-PIN SO (0.150")	16-PIN QSOP	PIN NAME	FUNCTION	EQUIVALENT CIRCUIT
1	1	INA+	Non-inverting input, channel A	
2	2	CEA	Chip enable, channel A	CE CE Circuit 2
3	3	VS-	Negative supply	
4	4	CEB	Chip enable, channel B	(See circuit 2)
5	5	INB+	Non-inverting input, channel B	(See circuit 1)
6, 11	6, 11	NC	Not connected	
7	7	CEC	Chip enable, channel C	(See circuit 2)
8	8	INC+	Non-inverting input, channel C	(See circuit 1)
9	9	INC-	Inverting input, channel C	(See circuit 1)
10	10	OUTC	Output, channel C	
12	12	INB-	Inverting input, channel B	(See circuit 1)
13	13	OUTB	Output, channel B	(See circuit 3)
14	14	VS+	Positive supply	
15	15	OUTA	Output, channel A	(See circuit 3)
16	16	INA-	Inverting input, channel A	(See circuit 1)

Applications Information

Product Description

The EL5396A is a triple channel fixed gain amplifier that offers a wide -3dB bandwidth of 400MHz and a low supply current of 9mA per amplifier. The EL5396A works with supply voltages ranging from a single 5V to 10V and they are also capable of swinging to within 1V of either supply on the output. This combination of high bandwidth and low power, together with aggressive pricing make the EL5396A the ideal choice for many low-power/high-bandwidth applications such as portable, handheld, or battery-powered equipment.

For varying bandwidth and higher gains, consider the EL5191 with 1GHz on a 9mA supply current or the EL5193 with 300MHz on a 4mA supply current. Versions include single, dual, and triple amp packages with 5-pin SOT23, 16-pin QSOP, and 8-pin or 16-pin SO outlines.

Power Supply Bypassing and Printed Circuit Board Layout

As with any high frequency device, good printed circuit board layout is necessary for optimum performance. Low impedance ground plane construction is essential. Surface mount components are recommended, but if leaded components are used, lead lengths should be as short as possible. The power supply pins must be well bypassed to reduce the risk of oscillation. The combination of a 4.7μ F tantalum capacitor in parallel with a 0.01μ F capacitor has been shown to work well when placed at each supply pin.

Disable/Power-Down

The EL5396A amplifier can be disabled placing its output in a high impedance state. When disabled, the amplifier supply current is reduced to < 150µA. The EL5396A is disabled when its \overline{CE} pin is pulled up to within 1V of the positive supply. Similarly, the amplifier is enabled by floating or pulling its \overline{CE} pin to at least 3V below the positive supply. For ±5V supply, this means that an EL5396A amplifier will be enabled when \overline{CE} is 2V or less, and disabled when \overline{CE} is above 4V. Although the logic levels are not standard TTL, this choice of logic voltages allows the EL5396A to be enabled by tying \overline{CE} to ground, even in 5V single supply applications. The \overline{CE} pin can be driven from CMOS outputs.

Gain Setting

The EL5396A is built with internal feedback and gain resistors. The internal feedback resistors have equal value; as a result, the amplifier can be configured into gain of +1, -1, and +2 without any external resistors. Figure 1 shows the amplifier in gain of +2 configuration. The gain error is $\pm 2\%$ maximum. Figure 2 shows the amplifier in gain of -1 configuration. For gain of +1, IN+ and IN- should be connected together as shown in Figure 3. This configuration avoids the effects of any parasitic capacitance on the IN- pin. Since the internal feedback and gain resistors change with

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temperature and process, external resistor should not be used to adjust the gain settings.

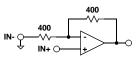


FIGURE 1. $A_V = +2$

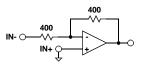


FIGURE 2. A_V = -1

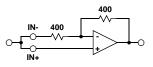
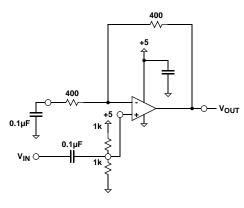


FIGURE 3. A_V = +1

Supply Voltage Range and Single-Supply Operation

The EL5396A has been designed to operate with supply voltages having a span of greater than or equal to 5V and less than 11V. In practical terms, this means that the EL5396A will operate on dual supplies ranging from $\pm 2.5V$ to $\pm 5V$. With single-supply, the EL5396A will operate from 5V to 10V.

As supply voltages continue to decrease, it becomes necessary to provide input and output voltage ranges that can get as close as possible to the supply voltages. The EL5396A has an input range which extends to within 2V of either supply. So, for example, on $\pm 5V$ supplies, the EL5396A has an input range which spans $\pm 3V$. The output range of the EL5396A is also quite large, extending to within 1V of the supply rail. On a $\pm 5V$ supply, the output is therefore capable of swinging from -4V to +4V. Single-supply output range is larger because of the increased negative swing due to the external pull-down resistor to ground. Figure 4 shows an AC-coupled, gain of +2, +5V single supply circuit configuration.





Video Performance

For good video performance, an amplifier is required to maintain the same output impedance and the same frequency response as DC levels are changed at the output. This is especially difficult when driving a standard video load of 150Ω , because of the change in output current with DC level. Previously, good differential gain could only be achieved by running high idle currents through the output transistors (to reduce variations in output impedance.) These currents were typically comparable to the entire 9mA supply current of each EL5396A amplifier. Special circuitry has been incorporated in the EL5396A to reduce the variation of output impedance with current output. This results in dG and dP specifications of 0.0035% and 0.04°, while driving 150Ω at a gain of 2.

Video performance has also been measured with a 500Ω load at a gain of +1. Under these conditions, the EL5396A has dG and dP specifications of 0.03% and 0.05°, respectively.

Output Drive Capability

In spite of its low 9mA of supply current, the EL5396A is capable of providing a minimum of \pm 95mA of output current. With a minimum of \pm 95mA of output drive.

Driving Cables and Capacitive Loads

When used as a cable driver, double termination is always recommended for reflection-free performance. For those applications, the back-termination series resistor will decouple the EL5396A from the cable and allow extensive capacitive drive. However, other applications may have high capacitive loads without a back-termination resistor. In these applications, a small series resistor (usually between 5Ω and 50Ω) can be placed in series with the output to eliminate most peaking.

Current Limiting

The EL5396A has no internal current-limiting circuitry. If the output is shorted, it is possible to exceed the Absolute Maximum Rating for output current or power dissipation, potentially resulting in the destruction of the device.

Power Dissipation

With the high output drive capability of the EL5396A, it is possible to exceed the 125°C Absolute Maximum junction temperature under certain very high load current conditions. Generally speaking when R_L falls below about 25 Ω , it is important to calculate the maximum junction temperature (T_{JMAX}) for the application to determine if power supply voltages, load conditions, or package type need to be modified for the EL5396A to remain in the safe operating area. These parameters are calculated as follows:

$$\mathsf{T}_{\mathsf{JMAX}} = \mathsf{T}_{\mathsf{MAX}} + (\theta_{\mathsf{JA}} \times \mathsf{n} \times \mathsf{PD}_{\mathsf{MAX}})$$

where:

T_{MAX} = Maximum ambient temperature

 θ_{JA} = Thermal resistance of the package

n = Number of amplifiers in the package

 PD_{MAX} = Maximum power dissipation of each amplifier in the package

PD_{MAX} for each amplifier can be calculated as follows:

$$\mathsf{PD}_{\mathsf{MAX}} = (2 \times \mathsf{V}_{\mathsf{S}} \times \mathsf{I}_{\mathsf{SMAX}}) + \left[(\mathsf{V}_{\mathsf{S}} - \mathsf{V}_{\mathsf{OUTMAX}}) \times \frac{\mathsf{V}_{\mathsf{OUTMAX}}}{\mathsf{R}_{\mathsf{L}}} \right]$$

where:

V_S = Supply voltage

I_{SMAX} = Maximum supply current

V_{OUTMAX} = Maximum output voltage (required)

R_L = Load resistance

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