NOT RECOMMENDED FOR NEW DESIGNS
SEE EL5106, EL5108

Data Sheet | March 12, 2004 | FN7183.1

Single 400MHz Fixed Gain Amplifier with Enable

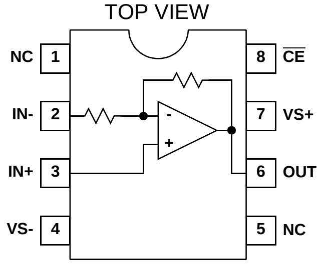
The EL5196 and the EL5196A are fixed gain amplifiers with a bandwidth of 400MHz, making these amplifiers ideal for today's high speed video and monitor applications. These amplifiers feature internal gain setting resistors and can be configured in a gain of +1, -1 or +2. The same bandwidth is seen in both gain-of-1 and gain-of-2 applications.

The EL5196A also incorporates an enable and disable function to reduce the supply current to 100µA typical per amplifier. Allowing the $\overline{CE}$ pin to float or applying a low logic level will enable the amplifier.

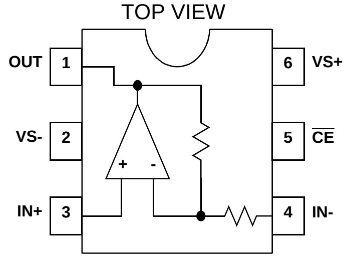
The EL5196 is offered in the 5-pin SOT-23 package and the EL5196A is available in the 6-pin SOT-23 as well as the industry-standard 8-pin SO packages. Both operate over the industrial temperature range of -40°C to +85°C.

Pinouts

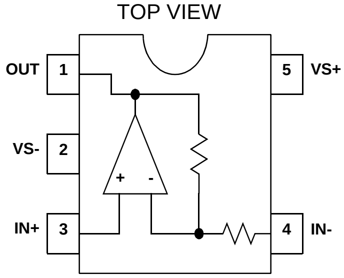
EL5196ACS
(8-PIN SO)
TOP VIEW



EL5196ACW
(6-PIN SOT-23)
TOP VIEW



EL5196CW
(5-PIN SOT-23)
TOP VIEW



Features

- Gain selectable (+1, -1, +2)
- 400MHz -3dB BW (A_V = 1, 2)
- 9mA supply current
- Fast enable/disable (EL5196A only)
- Single and dual supply operation, from 5V to 10V or ±2.5V to ±5V
- Available in SOT-23 packages
- Triple (EL5396) available
- 200MHz, 4mA products available (EL5197 & EL5397)

Applications

- Video amplifiers
- Cable drivers
- RGB amplifiers
- Test equipment
- Instrumentation
- Current to voltage converters

Ordering Information

PART NUMBER	PACKAGE	TAPE & REEL	PKG. DWG. #
EL5196CW-T7	5-Pin SOT-23	7" (3K pcs)	MDP0038
EL5196CW-T7A	5-Pin SOT-23	7" (250 pcs)	MDP0038
EL5196ACW-T7	6-Pin SOT-23	7"(3K pcs)	MDP0038
EL5196ACW-T7A	6-Pin SOT-23	7"(250 pcs)	MDP0038
EL5196ACS	8-Pin SO	-	MDP0027
EL5196ACS-T7	8-Pin SO	7"	MDP0027
EL5196ACS-T13	8-Pin SO	13"	MDP0027

Absolute Maximum Ratings (T_A = 25°C)

Supply Voltage between V_S+ and V_S- 11V
Maximum Continuous Output Current 50mA
Operating Junction Temperature . . . . 125°C
Power Dissipation . . . . See Curves
Pin Voltages. . . . . V_S- - 0.5V to V_S+ +0.5V
Storage Temperature. . . . . -65°C to +150°C
Operating Ambient Temperature . . . . -40°C to +85°C

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

Electrical Specifications

V_S+ = +5V, V_S- = -5V, R_L = 150Ω, T_A = 25°C unless otherwise specified.

PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
AC PERFORMANCE						
BW	-3dB Bandwidth	A_V = +1		400		MHz
		A_V = -1		400		MHz
		A_V = +2		400		MHz
BW1	0.1dB Bandwidth			35		MHz
SR	Slew Rate	V_O = -2.5V to +2.5V, A_V = +2	2400	2900		V/µs
t_S	0.1% Settling Time	V_{OUT} = -2.5V to +2.5V, AV = -1		9		ns
e_N	Input Voltage Noise			3.8		nV/√Hz
i_N-	IN- Input Current Noise			25		pA/√Hz
i_N+	IN+ Input Current Noise			55		pA/√Hz
dG	Differential Gain Error (Note 1)	A_V = +2		0.035		%
dP	Differential Phase Error (Note 2)	A_V = +2		0.04		°
DC PERFORMANCE						
V_{OS}	Offset Voltage		-15	1	15	mV
T_CV_{OS}	Input Offset Voltage Temperature Coefficient	Measured from T_{MIN} to T_{MAX}		5		µV/°C
A_E	Gain Error	V_O = -3V to +3V	-2	1.3	2	%
R_F, R_G	Internal R_F and R_G		320	400	480	Ω
INPUT CHARACTERISTICS						
CMIR	Common Mode Input Range		±3V	±3.3V		V
$+I_{IN}$	+ Input Current		-120	40	120	µA
$-I_{IN}$	- Input Current		-40	4	40	µA
R_{IN}	Input Resistance	at I_N+		27		kΩ
C_{IN}	Input Capacitance			0.5		pF
OUTPUT CHARACTERISTICS						
V_O	Output Voltage Swing	R_L = 150Ω to GND	±3.4V	±3.7V		V
		R_L = 1kΩ to GND	±3.8V	±4.0V		V
I_{OUT}	Output Current	R_L = 10Ω to GND	95	120		mA
SUPPLY						
I_{SON}	Supply Current - Enabled	No load, V_{IN} = 0V	8	9	11	mA
I_{SOFF}	Supply Current - Disabled	No load, V_{IN} = 0V		100	150	µA
PSRR	Power Supply Rejection Ratio	DC, V_S = ±4.75V to ±5.25V	55	75		dB

Electrical Specifications $V_S+ = +5V$, $V_S- = -5V$, $R_L = 150\Omega$, $T_A = 25°C$ unless otherwise specified. **(Continued)**

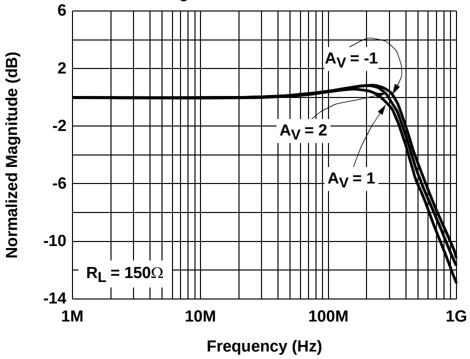
PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
-IPSR	- Input Current Power Supply Rejection	DC, $V_S = \pm4.75V$ to $\pm5.25V$	-2		2	µA/V
ENABLE (EL5196A ONLY)						
t_{EN}	Enable Time			40		ns
t_{DIS}	Disable Time			600		ns
I_{IHCE}	CE Pin Input High Current	$\overline{CE} = V_S+$		0.8	6	µA
I_{ILCE}	CE Pin Input Low Current	$\overline{CE} = V_S-$		0	-0.1	µA
V_{IHCE}	CE Input High Voltage for Disable		$V_S+ - 1$			V
V_{ILCE}	CE Input Low Voltage for Enable				$V_S+ - 3$	V

NOTES:

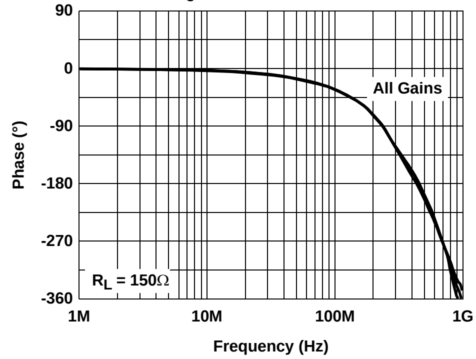
1. Standard NTSC test, AC signal amplitude = $286mV_{P-P}$, f = 3.58MHz.
2. Measured from the application of the $\overline{CE}$ logic signal until the output voltage is at the 50% point between initial and final values.

Typical Performance Curves

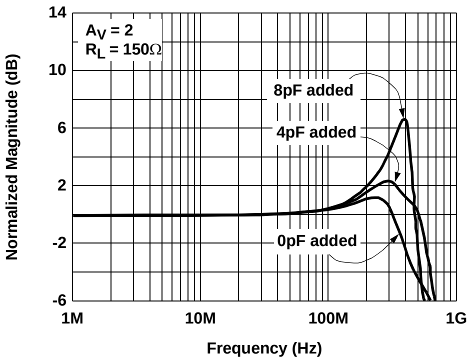
Frequency Response (Gain)
SOT-23 Package



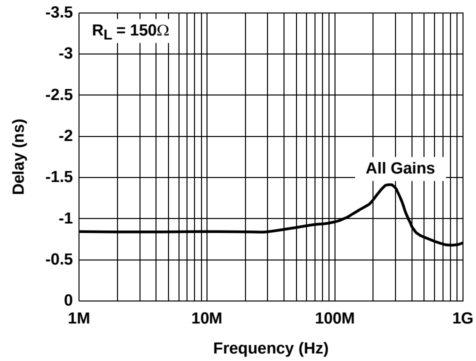
Frequency Response (Phase)
SOT-23 Package



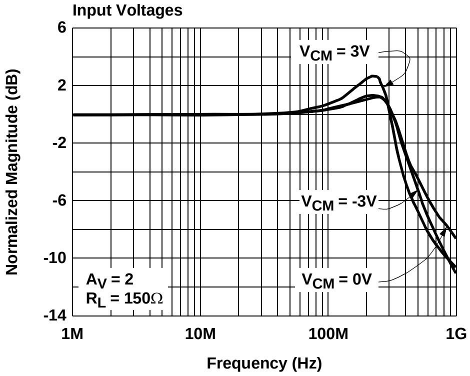
Frequency Response for Various C_L



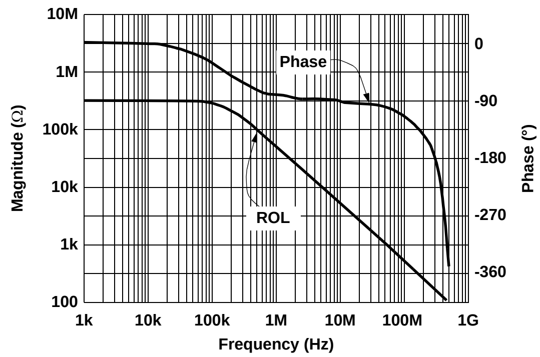
Group Delay vs Frequency, All Gains



Frequency Response for Various Common-Mode Input Voltages

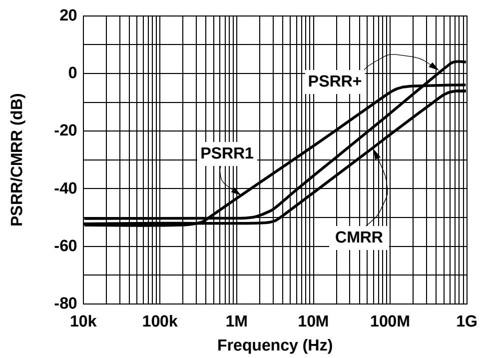


Transimpedance (ROL) vs Frequency

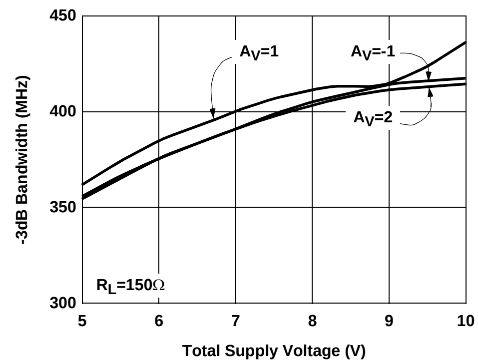


Typical Performance Curves (Continued)

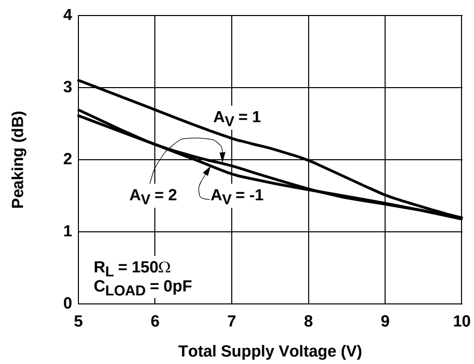
PSRR and CMRR vs Frequency



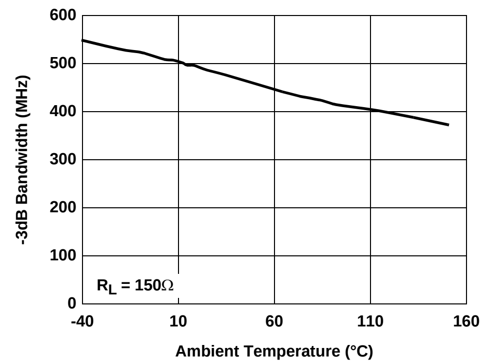
-3dB Bandwidth vs Supply Voltage



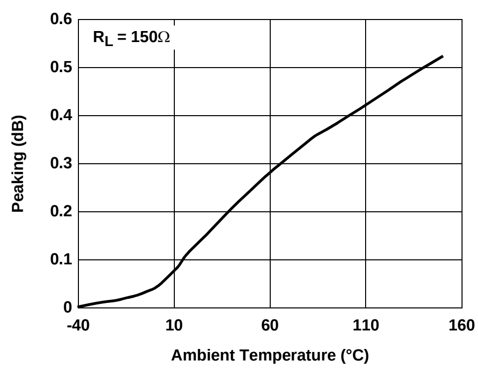
Peaking vs Supply Voltage



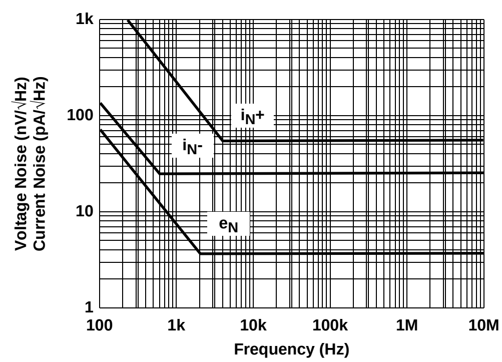
-3dB Bandwidth vs Temperature



Peaking vs Temperature

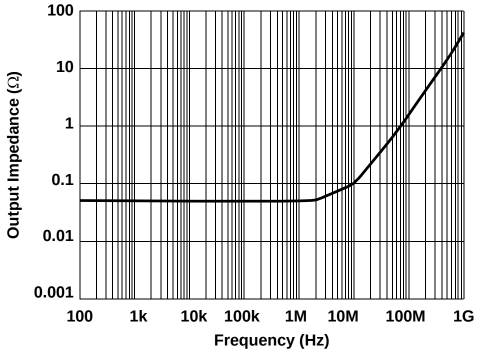


Voltage and Current Noise vs Frequency

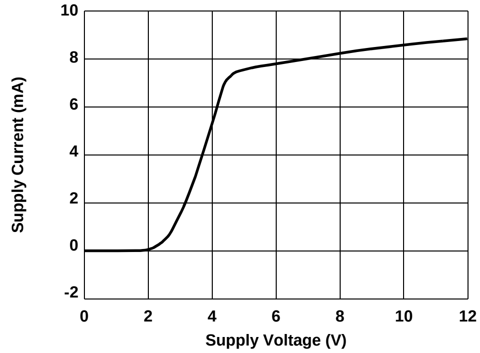


Typical Performance Curves (Continued)

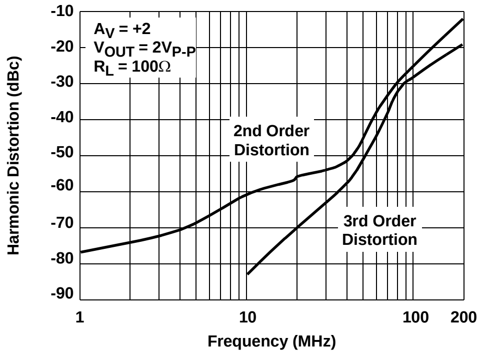
Closed Loop Output Impedance vs Frequency



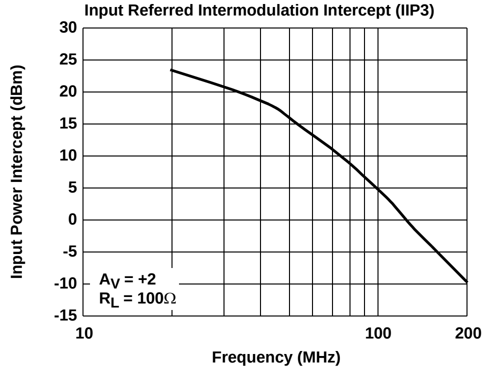
Supply Current vs Supply Voltage



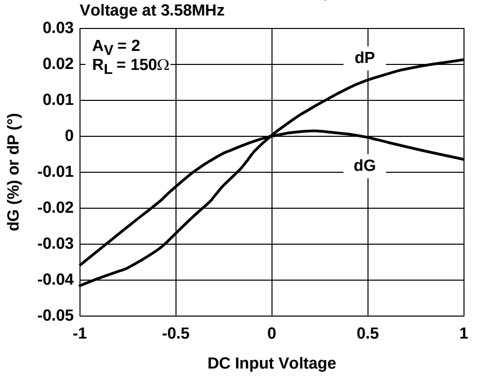
2nd and 3rd Harmonic Distortion vs Frequency



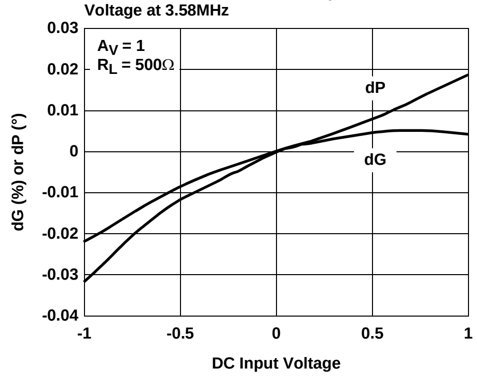
Two-Tone 3rd Order
Input Referred Intermodulation Intercept (IIP3)



Differential Gain/Phase vs DC Input Voltage at 3.58MHz

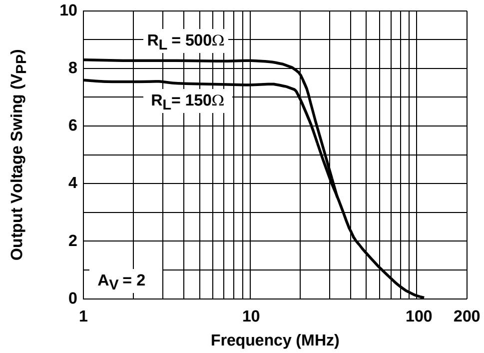


Differential Gain/Phase vs DC Input Voltage at 3.58MHz

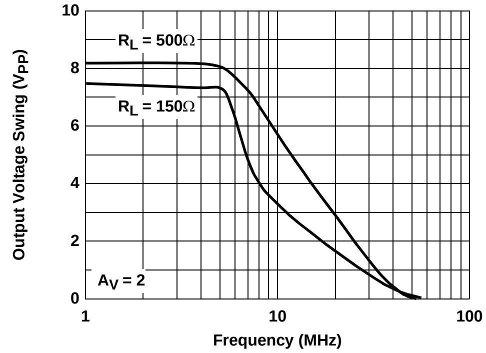


Typical Performance Curves (Continued)

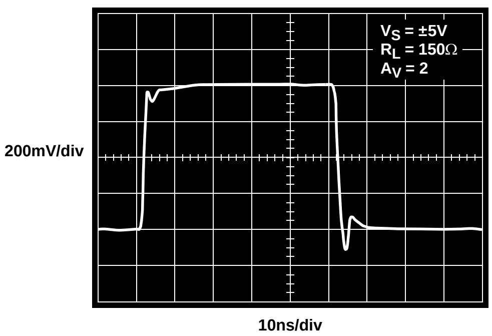
Output Voltage Swing vs Frequency
THD < 1%



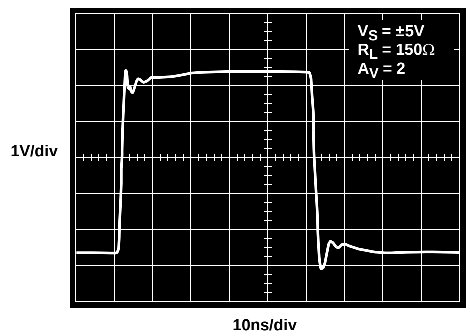
Output Voltage Swing vs Frequency
THD < 0.1%



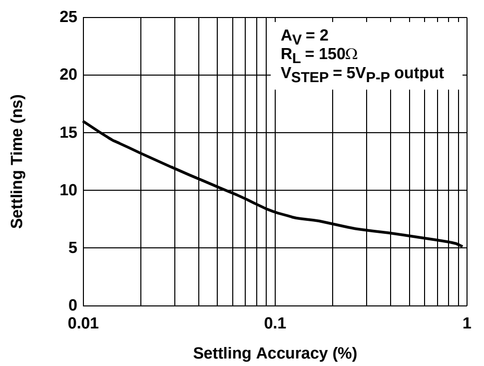
Small Signal Step Response



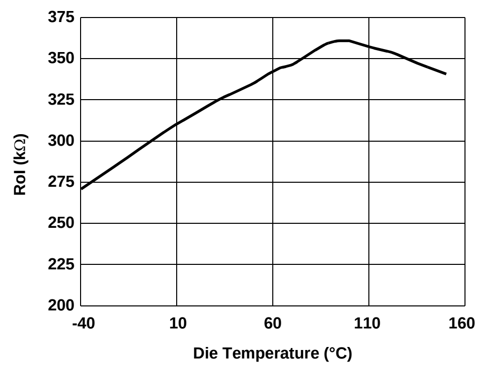
Large Signal Step Response



Settling Time vs Settling Accuracy

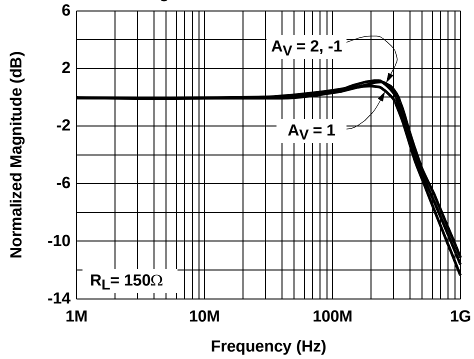


Transimpedance (RoI) vs Temperature

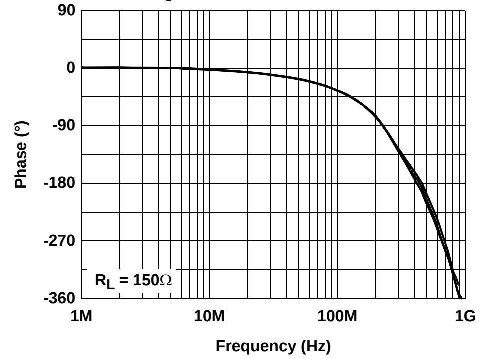


Typical Performance Curves (Continued)

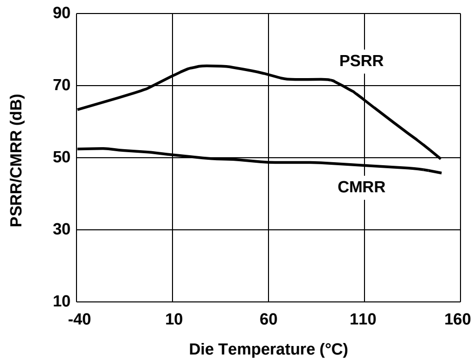
Frequency Response (Gain)
SO8 Package



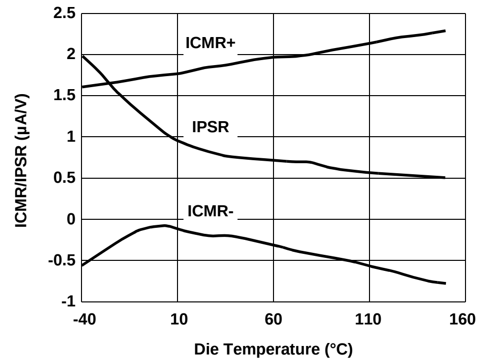
Frequency Response (Phase)
SO8 Package



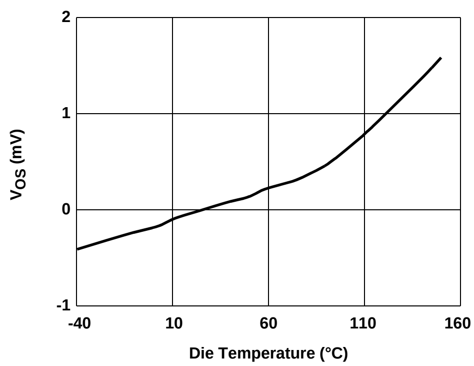
PSRR and CMRR vs Temperature



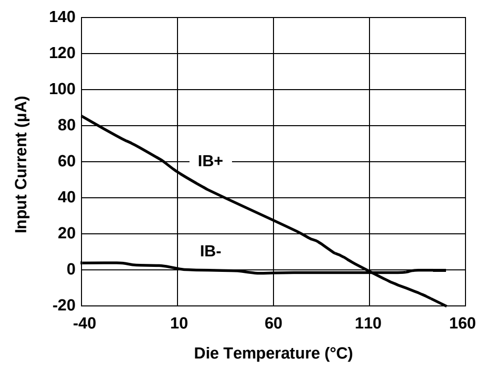
ICMR and IPSR vs Temperature



Offset Voltage vs Temperature

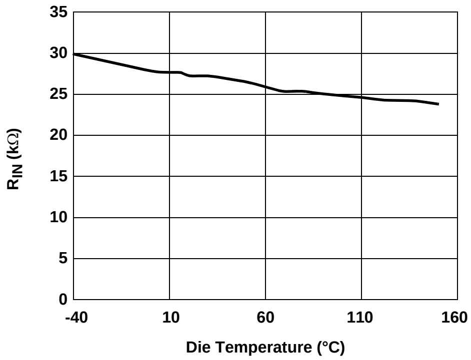


Input Current vs Temperature

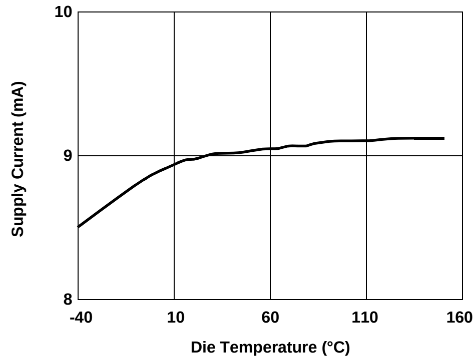


Typical Performance Curves (Continued)

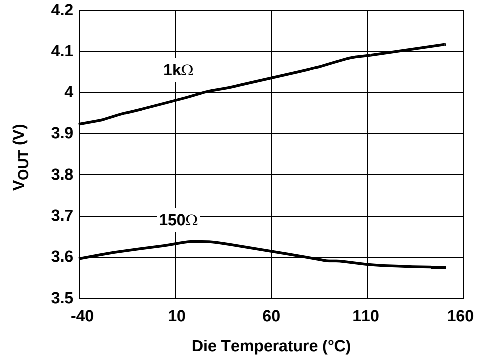
Positive Input Resistance vs Temperature



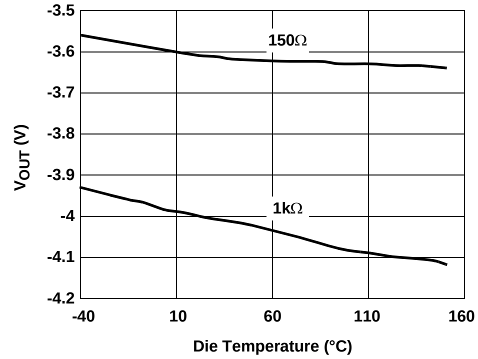
Supply Current vs Temperature



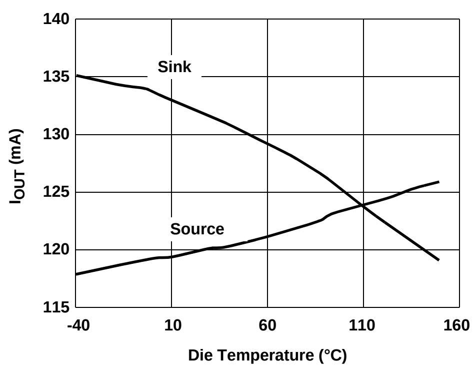
Positive Output Swing vs Temperature for Various Loads



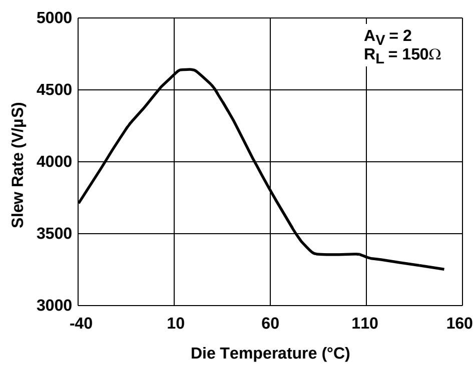
Negative Output Swing vs Temperature for Various Loads



Output Current vs Temperature

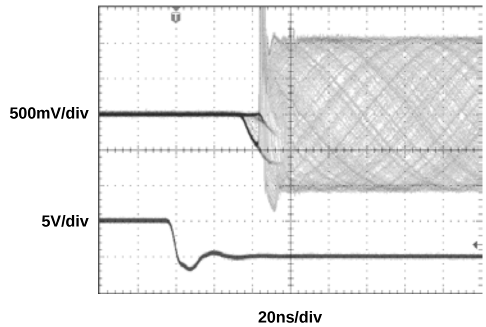


Slew Rate vs Temperature

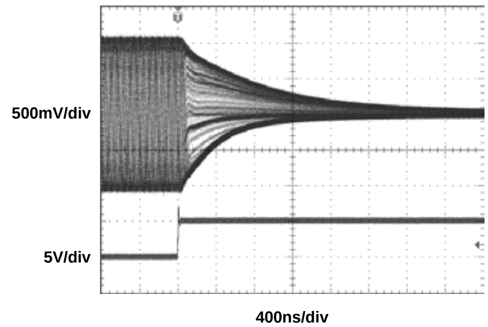


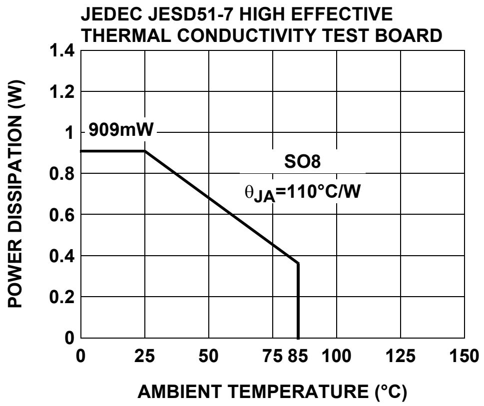
Typical Performance Curves (Continued)

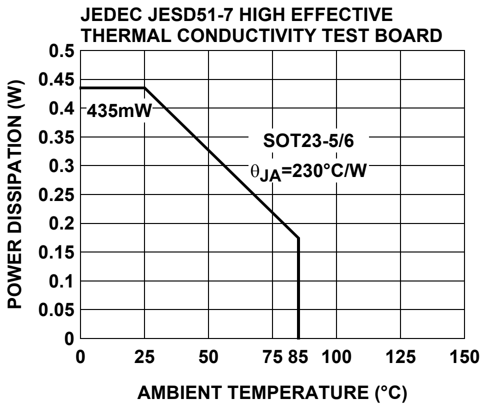
Enable Response

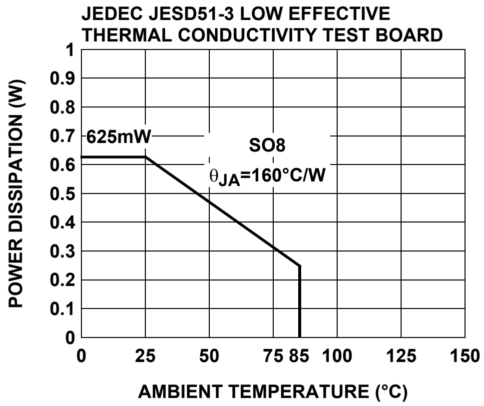


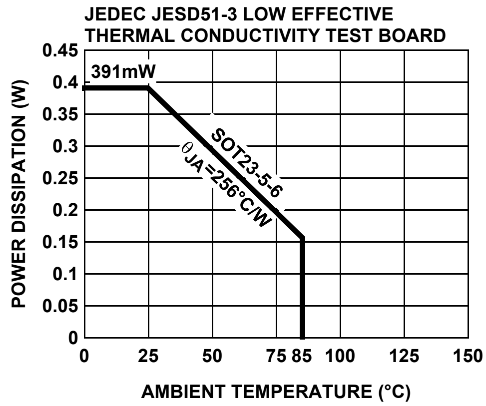
Disable Response



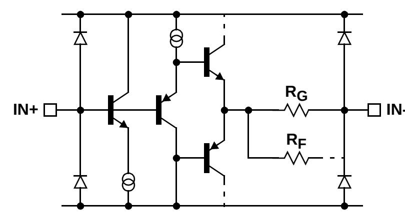
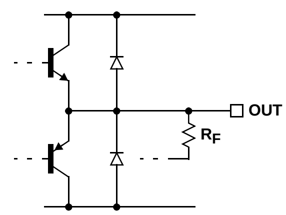
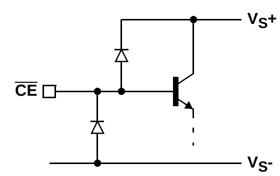








Pin Descriptions

8-PIN SO	5-PIN SOT-23	6-PIN SOT-23	PIN NAME	FUNCTION	EQUIVALENT CIRCUIT
1, 5			NC	Not connected	
2	4	4	IN-	Inverting input	 Circuit 1
3	3	3	IN+	Non-inverting input	(See circuit 1)
4	2	2	V_S-	Negative supply	
6	1	1	OUT	Output	 Circuit 2
7	5	6	V_S+	Positive supply	
8		5	$\overline{CE}$	Chip enable	 Circuit 3

Applications Information

Product Description

The EL5196 is a fixed gain amplifier that offers a wide -3dB bandwidth of 400MHz and a low supply current of 9mA per amplifier. The EL5196 works with supply voltages ranging from a single 5V to 10V and they are also capable of swinging to within 1V of either supply on the output. This combination of high bandwidth and low power, together with aggressive pricing make the EL5196 the ideal choice for many low-power/high-bandwidth applications such as portable, handheld, or battery-powered equipment.

For varying bandwidth and higher gains, consider the EL5191 with 1GHz on a 9mA supply current or the EL5193 with 300MHz on a 4mA supply current. Versions include single, dual, and triple amp packages with 5-pin SOT-23, 16-pin QSOP, and 8-pin or 16-pin SO outlines.

Power Supply Bypassing and Printed Circuit Board Layout

As with any high frequency device, good printed circuit board layout is necessary for optimum performance. Low impedance ground plane construction is essential. Surface mount components are recommended, but if leaded components are used, lead lengths should be as short as possible. The power supply pins must be well bypassed to reduce the risk of oscillation. The combination of a 4.7µF tantalum capacitor in parallel with a 0.01µF capacitor has been shown to work well when placed at each supply pin.

Disable/Power-Down

The EL5196A amplifier can be disabled placing its output in a high impedance state. When disabled, the amplifier supply current is reduced to < 150µA. The EL5196A is disabled when its $\overline{CE}$ pin is pulled up to within 1V of the positive supply. Similarly, the amplifier is enabled by floating or pulling its $\overline{CE}$ pin to at least 3V below the positive supply. For ±5V supply, this means that an EL5196A amplifier will be enabled when $\overline{CE}$ is 2V or less, and disabled when $\overline{CE}$ is above 4V. Although the logic levels are not standard TTL, this choice of logic voltages allows the EL5196A to be enabled by tying $\overline{CE}$ to ground, even in 5V single supply applications. The $\overline{CE}$ pin can be driven from CMOS outputs.

Gain Setting

The EL5196A is built with internal feedback and gain resistors. The internal feedback resistors have equal value; as a result, the amplifier can be configured into gain of +1, -1, and +2 without any external resistors. Figure 1 shows the amplifier in gain of +2 configuration. The gain error is ±2% maximum. Figure 2 shows the amplifier in gain of -1 configuration. For gain of +1, IN+ and IN- should be connected together as shown in Figure 3. This configuration avoids the effects of any parasitic capacitance on the IN- pin. Since the internal feedback and gain resistors change with temperature and process, external resistor should not be used to adjust the gain settings.

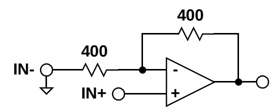


FIGURE 1. $A_V = +2$

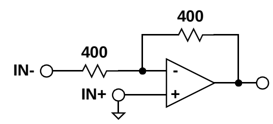


FIGURE 2. $A_V = -1$

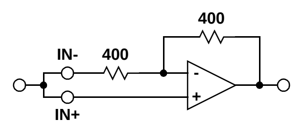


FIGURE 3. $A_V = +1$

Supply Voltage Range and Single-Supply Operation

The EL5196 has been designed to operate with supply voltages having a span of greater than or equal to 5V and less than 11V. In practical terms, this means that the EL5196 will operate on dual supplies ranging from ±2.5V to ±5V. With single-supply, the EL5196 will operate from 5V to 10V.

As supply voltages continue to decrease, it becomes necessary to provide input and output voltage ranges that can get as close as possible to the supply voltages. The EL5196 has an input range which extends to within 2V of either supply. So, for example, on ±5V supplies, the EL5196 has an input range which spans ±3V. The output range of the EL5196 is also quite large, extending to within 1V of the supply rail. On a ±5V supply, the output is therefore capable of swinging from -4V to +4V. Single-supply output range is larger because of the increased negative swing due to the external pull-down resistor to ground. Figure 4 shows an AC-coupled, gain of +2, +5V single supply circuit configuration.

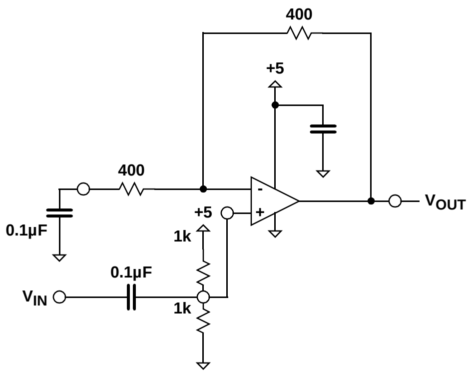


FIGURE 4.

Video Performance

For good video performance, an amplifier is required to maintain the same output impedance and the same frequency response as DC levels are changed at the output. This is especially difficult when driving a standard video load of 150Ω, because of the change in output current with DC level. Previously, good differential gain could only be achieved by running high idle currents through the output transistors (to reduce variations in output impedance.) These currents were typically comparable to the entire 9mA supply current of each EL5196 amplifier. Special circuitry has been incorporated in the EL5196 to reduce the variation of output impedance with current output. This results in dG and dP specifications of 0.0035% and 0.04°, while driving 150Ω at a gain of 2.

Video performance has also been measured with a 500Ω load at a gain of +1. Under these conditions, the EL5196 has dG and dP specifications of 0.03% and 0.05°, respectively.

Output Drive Capability

In spite of its low 9mA of supply current, the EL5196 is capable of providing a minimum of ±95mA of output current with a minimum of ±95mA of output drive.

Driving Cables and Capacitive Loads

When used as a cable driver, double termination is always recommended for reflection-free performance. For those applications, the back-termination series resistor will decouple the EL5196 from the cable and allow extensive capacitive drive. However, other applications may have high capacitive loads without a back-termination resistor. In these applications, a small series resistor (usually between 5Ω and 50Ω) can be placed in series with the output to eliminate most peaking.

Current Limiting

The EL5196 has no internal current-limiting circuitry. If the output is shorted, it is possible to exceed the Absolute Maximum Rating for output current or power dissipation, potentially resulting in the destruction of the device.

Power Dissipation

With the high output drive capability of the EL5196, it is possible to exceed the 125°C Absolute Maximum junction temperature under certain very high load current conditions. Generally speaking when R_L falls below about 25Ω, it is important to calculate the maximum junction temperature (T_{JMAX}) for the application to determine if power supply voltages, load conditions, or package type need to be modified for the EL5196 to remain in the safe operating area. These parameters are calculated as follows:

$$T_{JMAX} = T_{MAX} + (\theta_{JA} \times n \times PD_{MAX})$$

where:

T_{MAX} = Maximum ambient temperature

θ_{JA} = Thermal resistance of the package

n = Number of amplifiers in the package

PD_{MAX} = Maximum power dissipation of each amplifier in the package

PD_{MAX} for each amplifier can be calculated as follows:

$$PD_{MAX} = (2 \times V_S \times I_{SMAX}) + \left[(V_S - V_{OUTMAX}) \times \frac{V_{OUTMAX}}{R_L}\right]$$

where:

V_S = Supply voltage

I_{SMAX} = Maximum supply current of 1A

V_{OUTMAX} = Maximum output voltage (required)

R_L = Load resistance