## 6-Channel SD/ED/HD Video Filter with Charge Pump

## FEATURES

3 SD channels; 18 MHz typical 1 dB bandwidth (BW) 3 ED/HD channels; 25 MHz or 34 MHz typical 1 dB BW
Fixed gain of 6.2 dB ( $\mathbf{2 . 0 4 2} \mathbf{~ V / V ) ~}$
On-board negative supply for output coupling without capacitors
Minimal dc offset at the output pins
Internal summation of $Y$ and C channels for CVBS output
Flexible input dc offset cancellation for luma channels
D-terminal (EIAJ RC-5237 D5) and S-terminal (S1/S2) support
Capable of driving 2 back-terminated $75 \Omega$ video loads simultaneously
Separate power-down pins for SD and ED/HD sections
38-lead TSSOP package
Sony Green Partner Environmental Quality Approval Program compliant

## APPLICATIONS

DVD players and recorders
Set-top boxes
Projectors
Personal video recorders

## GENERAL DESCRIPTION

The ADA4424-6 is a high performance video reconstruction filter specifically designed for consumer applications. It consists of a standard definition (SD) section with two fifth-order Butterworth filters and a high definition (HD) section with three fifth-order filters. The SD section contains an internal Y/C summer for CVBS output, whereas the HD section provides selectable corner frequencies for either extended definition (ED) or HD signals.
The ADA4424-6 filter/buffer section operates from a single 3.3 V supply. Full support for D-terminal (EIAJ RC-5237 D5) and S1/S2 signaling is provided, along with a dedicated 5 V supply pin. Separate enable pins are provided for the SD and HD sections.


Figure 1.

The luma channels (Y_IN, HY_IN) of the ADA4424-6 are capable of detecting and cancelling dc input offsets of up to 1.1 V. Four distinct modes of detection/cancellation are available.
The output drivers on the ADA4424-6 feature rail-to-rail outputs with 6.2 dB gain. An on-board charge pump allows the outputs to swing up to 1.4 V below ground, eliminating the need for large coupling capacitors. Each output is capable of driving two $75 \Omega$ doubly terminated cables.

The ADA4424-6 is available in a 38 -lead TSSOP and operates in the industrial temperature range of $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$.

Rev. C
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## SPECIFICATIONS

$\mathrm{V}_{\mathrm{DD} 3}=3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{O}}=2.042 \mathrm{~V}$ p-p, $\mathrm{R}_{\mathrm{L}}=150 \Omega$, dc-coupled outputs, unless otherwise noted. Charge pump configured as shown in Figure 18.

Table 1.

| Parameter | Test Conditions/Comments | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| OVERALL PERFORMANCE <br> DC Voltage Gain Input Voltage Range, All Inputs Output Voltage Range, All Outputs Input Bias Current Input Impedance Output Resistance | All channels <br> Not including dc offset <br> Y_IN, HY_IN, dc-coupled <br> C_IN, HPb_IN, HPr_IN, ac-coupled <br> Y_OUT, C_OUT, CVBS_OUT, HY_OUT, HPb_OUT, <br> HPr_OUT, dc-coupled <br> L1_OUT, L2_OUT, L3_OUT, S1/S2_OUT, dc-coupled | 6.0 | $\begin{aligned} & 6.2 \\ & -0.6 \text { to }+1.4 \\ & -1.6 \text { to }+3.0 \\ & 30 \\ & 800 \\ & 0.5 \\ & 10.5 \end{aligned}$ | 6.4 | dB <br> V <br> V <br> pA <br> $\mathrm{k} \Omega$ <br> $\Omega$ <br> $\mathrm{k} \Omega$ |
| SD CHANNEL DYNAMIC PERFORMANCE <br> In-Band Peaking <br> 1 dB Bandwidth <br> Out-of-Band Rejection <br> Crosstalk <br> Total Harmonic Distortion <br> Signal-to-Noise Ratio <br> Group Delay Variation <br> Differential Gain <br> Differential Phase | $\begin{aligned} & f=100 \mathrm{kHz} \text { to } 6.75 \mathrm{MHz} \\ & \mathrm{f}=148.5 \mathrm{MHz} \\ & \mathrm{f}=1 \mathrm{MHz} \\ & \mathrm{f}=1 \mathrm{MHz}, \mathrm{~V}_{0}=1.4 \mathrm{~V} \text { p-p } \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 6 \mathrm{MHz} \text {, unweighted } \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 5 \mathrm{MHz} \\ & \text { NTSC } \\ & \text { NTSC } \end{aligned}$ |  | 0.00 18 42 67 0.07 68 1 0.2 0.5 | 0.01 | dB <br> MHz <br> dB <br> dB <br> \% <br> dB <br> ns <br> \% <br> Degrees |
| ED CHANNEL DYNAMIC PERFORMANCE <br> In-Band Peaking <br> 1 dB Bandwidth <br> Out-of-Band Rejection <br> Crosstalk <br> Total Harmonic Distortion <br> Signal-to-Noise Ratio <br> Group Delay Variation | $\begin{aligned} & \text { FC_SEL = low (0) } \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 13.5 \mathrm{MHz} \\ & \mathrm{f}=148.5 \mathrm{MHz} \\ & \mathrm{f}=1 \mathrm{MHz} \\ & \mathrm{f}=5 \mathrm{MHz}, \mathrm{~V}_{\mathrm{o}}=1.4 \mathrm{~V} \mathrm{p-p} \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 13.5 \mathrm{MHz} \text {, unweighted } \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 13.5 \mathrm{MHz} \end{aligned}$ |  | 0.02 25 42 65 0.45 66 1.5 | 0.1 | dB <br> MHz <br> dB <br> dB <br> \% <br> dB <br> ns |
| HD CHANNEL DYNAMIC PERFORMANCE <br> In-Band Peaking <br> 1 dB Bandwidth <br> Out-of-Band Rejection <br> Crosstalk <br> Total Harmonic Distortion <br> Signal-to-Noise Ratio Group Delay Variation | $\begin{aligned} & \hline \text { FC_SEL = high (1) } \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 30 \mathrm{MHz} \\ & \text { Y Channel (HY_OUT) } \\ & \text { P Channels (HPb_OUT, HPr_OUT) } \\ & \mathrm{f}=148.5 \mathrm{MHz} \\ & \mathrm{f}=1 \mathrm{MHz} \\ & \mathrm{f}=10 \mathrm{MHz}, \mathrm{~V}_{\mathrm{o}}=1.4 \mathrm{~V} \text { p-p } \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 30 \mathrm{MHz}, \text { unweighted } \\ & \mathrm{f}=100 \mathrm{kHz} \text { to } 30 \mathrm{MHz} \\ & \hline \end{aligned}$ | 30 25 33 | $\begin{aligned} & 0.1 \\ & 39 \\ & 34 \\ & 37 \\ & 65 \\ & 1.2 \\ & 65 \\ & 2.2 \end{aligned}$ | 0.2 | dB <br> MHz <br> MHz <br> dB <br> dB <br> \% <br> dB <br> ns |
| DC CHARACTERISTICS <br> Operating Voltage, 3.3 V Supply Quiescent Supply Current, 3.3 V Supply Operating Voltage, 5 V Supply | Both active, SD_ENABLE = high, HD_ENABLE = high, no load, no signal, not including D/S terminal outputs SD disabled, SD_ENABLE = low, HD_ENABLE = high HD disabled, SD_ENABLE = high, HD_ENABLE = low Both disabled, SD_ENABLE = low, HD_ENABLE = low |  | $\begin{aligned} & 3.14 \text { to } 3.46 \\ & 93 \\ & 54 \\ & 45 \\ & 6.1 \\ & 4.75 \text { to } 5.25 \end{aligned}$ | 133 10 | V <br> mA <br> mA <br> mA <br> mA <br> V |


| Parameter | Test Conditions/Comments | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Quiescent Supply Current, 5 V Supply | SD_ENABLE = high, HD_ENABLE = high, $\mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega, \mathrm{D} 1, \mathrm{D} 2, \mathrm{D} 3=$ high, $\mathrm{S}=$ high |  | 190 | 200 | $\mu \mathrm{A}$ |
|  | SD_ENABLE = low, HD_ENABLE = low |  | 5 | 15 | $\mu \mathrm{A}$ |
| PSRR | ED/HD channels, output referred |  | -42 |  | dB |
|  | SD channels, output referred |  | -41 |  | $d B$ |
| DC Offset | See Table 6 and Table 7 |  |  |  |  |
| Input Referred, Offset Cancellation Disabled Mode | OFFSET_ENB = low |  |  |  |  |
| SD Channels | Y - $\mathrm{IN}=0 \mathrm{~V}$ dc | -60 | -20 | +60 | mV |
| CVBS Channel | Y - $\mathrm{IN}=0 \mathrm{~V}$ dc | -100 | -40 | +100 | mV |
| ED/HD Channels | HY_IN $=0 \mathrm{~V}$ dc | -60 | -20 | +60 | mV |
| Input Referred, Fixed Offset Cancellation Mode | OFFSET_ENB = high, MODE1 = high |  |  |  |  |
| SD Fixed High Offset Mode | Y_IN $=1.0 \mathrm{~V} \mathrm{dc}, \mathrm{MODE}$ = low | -100 | -30 | +100 | mV |
| ED/HD Fixed High Offset Mode | HY_IN = 1.1 V dc, MODE0 = low | -100 | -38 | +100 | mV |
| SD Fixed Low Offset Mode | Y_IN $=0.33 \mathrm{~V}$ dc, MODE0 $=$ high | -90 | -17 | +90 | mV |
| ED/HD Fixed Low Offset Mode | HY_IN $=0.33 \mathrm{Vdc}, \mathrm{MODE0}=$ high | -100 | -25 | +100 | mV |
| Input Referred, Auto Offset Cancellation Mode | OFFSET_ENB = high, MODE1 = low |  |  |  |  |
| SD Auto Offset Mode Sync Tip Sampling | Y _ $\mathrm{IN}=0 \mathrm{~V}$ to 1.0 V dc, MODE0 = low | -70 | -36 | +70 | mV |
| ED/HD Auto Offset Mode Sync Tip Sampling | HY_IN $=0 \mathrm{~V}$ to 1.1 V dc, MODE0 = low | -95 | -46 | +95 | mV |
| SD Auto Offset Mode Back Porch Sampling | Y _IN $=0 \mathrm{~V}$ to 1.0 V dc, MODE0 $=$ high | -25 | -6 | +25 | mV |
| ED/HD Auto Offset Mode Back Porch Sampling | HY_IN = 0 V to 1.1 V dc, MODE0 = high | -25 | -5 | +25 | mV |
| FC_SEL Input Logic Low Level |  | 0 |  | 0.6 | V |
| FC_SEL Input Logic High Level |  | 1.2 |  | $V_{\text {DD3 }}$ | V |
| xD_ENABLE, OFFSET_ENB, MODEx Input Logic Low Level |  | 0 |  | 0.8 | V |
| xD_ENABLE, OFFSET_ENB, MODEx Input Logic High Level |  | 2.0 |  | $V_{\text {DD3 }}$ | V |
| xD_ENABLE Assert Time | $x D_{\text {d }}$ ENABLE $=$ low to high |  | 95 |  | ns |
| xD_ENABLE Deassert Time | xD_ENABLE = high to low |  | 20 |  | ns |
| xD_ENABLE Input Bias Current | Disabled, xD_ENABLE = low |  | 6.1 |  | $\mu \mathrm{A}$ |
| Input-to-Output Isolation | Disabled, $x$ D_ENABLE $=$ low, $\mathrm{f}=5 \mathrm{MHz}$ |  | -100 |  | dB |
| D- and S-Terminal Input Logic Low Level | $\mathrm{RL}=100 \mathrm{k} \Omega$ | 0 |  | 0.6 | V |
| D- and S-Terminal Input Logic Mid Level | $\mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega$ | 0.9 |  | 1.9 | V |
| D- and S-Terminal Input Logic High Level | $\mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega$ | 2.7 |  | $V_{\text {DD3 }}$ | V |
| D- and S-Terminal Input Logic Open (Hi-Z) Resistance Value | $\mathrm{RL}=100 \mathrm{k} \Omega$ | 200 |  |  | $k \Omega$ |
| D-Terminal (L1_OUT, L2_OUT, L3_OUT) Low Level Output | $\mathrm{V}_{\mathrm{DD} 5}=5.0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega, \mathrm{D} 1, \mathrm{D} 2, \mathrm{D} 3=$ low |  | 0.0 |  | V |
| D-Terminal (L1_OUT, L3_OUT) Mid Level Output | $\mathrm{V}_{\mathrm{DD5}}=5.0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega, \mathrm{D} 1, \mathrm{D} 3=$ mid or open |  | 2.1 |  | V |
| D-Terminal (L1_OUT, L2_OUT, L3_OUT) High Level Output | $V_{D D 5}=5.0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega, \mathrm{D} 1, \mathrm{D} 2, \mathrm{D} 3=\mathrm{high}$ |  | 4.5 |  | V |
| S-Terminal (S1/S2_OUT) Low Level Output | $\mathrm{V}_{\mathrm{DD} 5}=5.0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega, \mathrm{S}=$ low |  | 0.0 |  | V |


| Parameter | Test Conditions/Comments | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| S-Terminal (S1/S2_OUT) Mid Level Output | $V_{D D 5}=5.0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega, \mathrm{S}=$ mid or open |  | 2.1 |  |  |
| S-Terminal (S1/S2_OUT) High Level Output | $V_{\text {DD } 5}=5.0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \mathrm{k} \Omega, \mathrm{S}=$ high |  | 4.5 |  | V |
| CHARGE PUMP CHARACTERISTICS | All channels operating; C1 $=\mathrm{C} 2=4.7 \mu \mathrm{~F}, \mathrm{C} 3=\mathrm{C} 4=$ $1.0 \mu \mathrm{~F}, \mathrm{R} 1=1 \Omega$ (see Figure 18) |  |  |  |  |
| Output Voltage |  |  | -1.66 |  |  |
| Output Voltage Ripple |  |  | 180 |  | mV p-p |
| Output Ripple Frequency |  |  | 100 |  |  |

## ABSOLUTE MAXIMUM RATINGS

Table 2.

| Parameter | Rating |
| :--- | :--- |
| 3.3 V Supply Voltage | 3.6 V |
| 5 V Supply Voltage | 5.5 V |
| Digital Input Voltage (Pin 2 to Pin 5, Pin 8, | 3.6 V |
| Pin 12, Pin 15, Pin 16, Pin 23) |  |
| Power Dissipation | See Figure 2 |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Operating Temperature Range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Lead Temperature (Soldering, 10 sec) | $300^{\circ} \mathrm{C}$ |
| Junction Temperature | $150^{\circ} \mathrm{C}$ |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## THERMAL RESISTANCE

$\theta_{\mathrm{JA}}$ is specified for the device soldered to a high thermal conductivity 4-layer ( 2 s 2 p ) circuit board, as described in EIA/JESD 51-7.

Table 3.

| Package Type | $\boldsymbol{\theta}_{\mathrm{JA}}$ | $\boldsymbol{\theta}_{\text {Jc }}$ | Unit |
| :--- | :--- | :--- | :--- |
| 38-Lead TSSOP | 67.6 | 14.0 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## MAXIMUM POWER DISSIPATION

The maximum safe power dissipation in the ADA4424-6 package is limited by the associated rise in junction temperature $\left(\mathrm{T}_{\mathrm{J}}\right)$ on the die. At approximately $150^{\circ} \mathrm{C}$, which is the glass transition temperature, the plastic changes its properties. Even temporarily exceeding this temperature limit can change the stresses that the package exerts on the die, permanently shifting the parametric performance of the ADA4424-6. Exceeding a junction temperature of $150^{\circ} \mathrm{C}$ for an extended time can result in changes in the silicon devices, potentially causing failure.

The power dissipated in the package $\left(\mathrm{P}_{\mathrm{D}}\right)$ is the sum of the quiescent power dissipation and the power dissipated in the package due to the load drive for all outputs. The quiescent power is the voltage between the supply pins ( $\mathrm{V}_{\mathrm{s}}$ ) times the quiescent current ( $\mathrm{I}_{\mathrm{s}}$ ). The power dissipated due to load drive depends on the particular application. For each output, the power due to load drive is calculated by multiplying the load current by the associated voltage drop across the device. The power dissipated due to the loads is equal to the sum of the power dissipations due to each individual load. RMS voltages and currents must be used in these calculations.

Airflow increases heat dissipation, effectively reducing $\theta_{J A}$. Figure 2 shows the maximum power dissipation in the package vs. the ambient temperature for the 38 -lead TSSOP $\left(67.6^{\circ} \mathrm{C} / \mathrm{W}\right)$ on a JEDEC standard 4-layer board. $\theta_{\mathrm{JA}}$ values are approximate.


Figure 2. Maximum Power Dissipation vs. Ambient Temperature for a 4-Layer Board

## ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Figure 3. Pin Configuration, Top View

Table 4. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | GND5 | Ground Pin for 5 V Supply. |
| 2 | D1 | D-Terminal Vertical Resolution Selection Input. |
| 3 | D2 | D-Terminal Scan Selection Input. |
| 4 | D3 | D-Terminal Aspect Ratio Selection Input. |
| 5 | S | S-Terminal Aspect Ratio Selection Input. |
| 6 | Y_IN | SD Luma (Y) Input. |
| 7 | VDD3_SD | 3.3 V Supply Pin for SD Filter Section. |
| 8 | SD_ENABLE | Output Enable Pin for SD (Y, C, CVBS). |
| 9 | C_IN | SD Chroma (C) Input. |
| 10, 29 | GND3 | Ground Pins for 3.3 V Supply. |
| 11 | HY_IN | ED/HD Y Component Input. |
| 12 | FC_SEL | Filter Corner Frequency Selection Pin for HY, HPb, HPr Channels. |
| 13 | HPb_IN | ED/HD Pb Component Input. |
| 14 | HPr_IN | ED/HD Pr Component Input. |
| 15 | HD_ENABLE | Output Enable Pin for ED/HD (HY, HPb, HPr). |
| 16 | MODE0 | This pin selects sync tip or back porch sampling when MODE1 $=0$ and selects high or low fixed offset subtraction when MODE1 $=1$. |
| 17 | OFFSET_ENB | Offset Cancellation Enable Pin. |
| 18 | VDD3_CP | 3.3V Supply Pins for Charge Pump Section. |
| 19, 21 | C1a, C1b | Charge Pump Capacitor C1 Connection Pin. |
| 20 | GND_CP | Ground Pin for 3.3 V Charge Pump Supply. |
| 22 | C2/CP_OUT | Charge Pump Output Pin. Connect Capacitor C2 from this pin to ground. |
| 23 | MODE1 | Selects Automatic or Fixed Offset Subtraction Mode. |
| 24 | VDD3_HD | 3.3 V Supply Pin for ED/HD Filter Section. |
| 25 | HPr_OUT | ED/HD Pr Component Output. |
| 26 | HPb_OUT | ED/HD Pb Component Output. |
| 27 | VSS_HD | Negative Supply Pin for ED/HD Filter Section. This pin should be connected to the charge pump output (Pin 22), as shown in Figure 18. |
| 28 | HY_OUT | ED/HD Y Component Output. |
| 30 | C_OUT | SD Chroma (C) Output. |
| 31 | CVBS_OUT | SD Composite Video (CVBS) Output. |


| Pin No. | Mnemonic | Description |
| :--- | :--- | :--- |
| 32 | VSS_SD | Negative Supply Pin for SD Filter Section. This pin should be connected to the charge pump output (Pin 22), as <br> shown in Figure 18. |
| 33 | Y_OUT | SD Luma (Y) Output. |
| 34 | S1/S2_OUT | S-Terminal Aspect Ratio Selection Output. |
| 35 | L3_OUT | D-Terminal Aspect Ratio Selection Output. |
| 36 | L2_OUT | D-Terminal Scan Selection Output. |
| 37 | L1_OUT | D-Terminal Vertical Resolution Selection Output. |
| 38 | VDD5 | 5 V Supply Pin for D-Terminal and S-Terminal Signaling. |

## TYPICAL PERFORMANCE CHARACTERISTICS

$\mathrm{V}_{\mathrm{DD} 3}=3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{O}}=2.042 \mathrm{~V}$ p-p, $\mathrm{R}_{\mathrm{L}}=150 \Omega$, dc-coupled outputs, unless otherwise noted. Charge pump configured as shown in Figure 18.


Figure 4. Frequency Response—SD Channels


Figure 5. Frequency Response-ED Channels


Figure 6. Frequency Response-HD Channels


Figure 7. Frequency Response Peaking-SD Channels


Figure 8. Frequency Response Peaking-ED Channels


Figure 9. Frequency Response Peaking—HD Channels


Figure 10. Crosstalk—SD Channels


Figure 11. Group Delay-SD Channels


Figure 12. Group Delay—HD Channels


Figure 13. Crosstalk-ED and HD Channels


Figure 14. Group Delay—ED Channels


Figure 15. Input-to-Output Isolation-All Channels

## APPLICATIONS INFORMATION CVBS OUTPUT

The composite video signal (CVBS_OUT) is produced by passively summing the Y and C channels (see Figure 1), after amplification by their respective gain stages. Each signal experiences a 6.2 dB loss from the passive summation and is subsequently amplified by 6.2 dB in the fixed stage following the summer. The net signal gain through the composite video path is, therefore, 0 dB , and the resulting composite signal present at the ADA4424-6 output is the sum of $Y$ and $C$ with unity gain.

## CORNER FREQUENCY SELECTION

The component channels of the ADA4424-6 allow for a 1 dB filter corner frequency of either 25 MHz or $39 \mathrm{MHz} / 34 \mathrm{MHz}$. The FC_SEL pin operates as described in Table 5.

Table 5. ED/HD Bandwidth Selection

| FC_SEL (Pin 12) | HD/ED | $\mathbf{- 1 ~ d B ~ C o r n e r ~ F r e q u e n c y ~ ( T y p ) ~}$ |
| :--- | :--- | :--- |
| Low (0) | ED | 25 MHz |
| High (1) | HD | $39 \mathrm{MHz}(\mathrm{HY}) ; 34 \mathrm{MHz}(\mathrm{Pb}, \mathrm{Pr})$ |

## INPUT DC OFFSET CANCELLATION

The luma channels (Y_IN, HY_IN) of the ADA4424-6 are capable of detecting and cancelling dc input offsets of up to 1.1 V . Four distinct modes of detection/cancellation are available. These are selected via the MODE1 and MODE0 pins. The chroma (C_IN) and color difference (HPb_IN, HPr_IN) inputs do not support offset cancellation. It is recommended that these inputs be accoupled.

## Automatic Detection/Cancellation Mode

There are two modes of automatic operation. The primary mode samples the input signal between the rising edge of the horizontal sync pulse and the start of active video (back porch), averages the value over the sampling interval, and subtracts it from the output signal. This is the more accurate method and is able to reduce the input-referred offsets to less than 25 mV .
An alternate method is available for copy-protected signals, where sampling the back porch may not provide a reliable dc average. This method detects the input negative sync tip, and clamps it to a fixed value ( -214 mV for SD , and -300 mV for $\mathrm{ED} / \mathrm{HD}$ ). Sample intervals for SD and ED are shown in Figure 16, and the HD sample intervals are shown in Figure 17.


Figure 16. Back Porch and Sync Tip Sample Intervals (SD/ED)


Figure 17. Back Porch and Sync Tip Sample Intervals (HD)

## ADA4424-6

## Fixed Offset Cancellation Mode

In addition to the automatic mode, there are two levels of fixed offset correction available. In high offset mode, fixed voltages of 1.0 V and 1.1 V are subtracted from the Y_IN and HY_IN inputs, respectively. In low offset mode, a fixed voltage of 0.33 V is subtracted from both Y_IN and HY_IN. The various modes of offset cancellation are outlined in Table 6.

Table 6. Offset Cancellation Mode Selection

| MODE1 <br> (Pin 23) | MODEO <br> (Pin 16) | Output Offset Cancellation |
| :---: | :---: | :---: |
| Low (0) | Low (0) | Auto-cancel, sync-tip sampling mode. Clamps the input referred SD sync tip to -214 mV , and the input referred ED/HD sync tip to -300 mV . |
| Low (0) | High (1) | Auto-cancel, back porch sampling mode. Sets the output blanking level to 0 V , independent of sync depth. |
| High (1) | Low (0) | Fixed cancellation mode, high dc offset. <br> Subtracts 1.0 V from the Y _IN signal; subtracts 1.1 V from the HY _IN signal. |
| High (1) | High (1) | Fixed cancellation mode, low dc offset. Subtracts 0.33 V from both the Y _IN and HY_IN signals. |

## Offset Cancellation Disable

The offset cancellation function can be enabled or disabled via the OFFSET_ENB pin, as described in Table 7.

Table 7. Offset Cancellation Enable/Disable

| OFFSET_ENB <br> (Pin 17) | Offset Cancellation State |
| :--- | :--- |
| Low (0) | Offset cancellation is disabled. <br> Offset cancellation is enabled. Function is <br> determined by the MODE1 and MODE0 pins <br> (see Table 6). |

## D-TERMINAL AND S-TERMINAL SUPPORT

Full D-terminal support (EIAJ RC-5237 D5) is provided for the component channels (HY_OUT, HPb_OUT, HPr_OUT). Level D1 through Level D5 are supported for vertical resolution, scan type, and aspect ratio selection. Details are shown in Table 8, Table 9, and Table 10.

S-terminal (also known as S_DC or S1/S2) support for S-video aspect ratio selection is also provided, as described in Table 11.
The VDD5 pin (Pin 38) provides 5 V power for these outputs. If D - or S -terminal support is not required, it is recommended that Pin 2 to Pin 5 and Pin 34 to Pin 38 remain unconnected.

Table 8. D-Terminal Control for Vertical Resolution Selection

| Input Logic <br> Level D1 (Pin 2) | Nominal Output (V) <br> L1_OUT (Pin 37) <br> $\mathbf{R}_{\mathbf{L}}=\mathbf{1 0 0} \mathbf{~ k ~} \boldsymbol{\Omega}$ | Vertical Resolution <br> (Number of Lines) |
| :--- | :--- | :--- |
| Low (0) | 0.0 | 480 |
| Mid or open | 2.1 | 720 |
| High (1) | 4.5 | 1080 |

Table 9. D-Terminal Control for Scan Selection

| Input Logic Level <br> D2 (Pin 3) | Nominal Output (V) <br> L2_0UT (Pin 36) <br> $\mathbf{R}_{\mathrm{L}}=\mathbf{1 0 0} \mathbf{~ k \Omega}$ | Scan Type |
| :--- | :--- | :--- |
| Low (0) | 0.0 | Interlaced |
| Mid or open | 2.1 | N/A |
| High (1) | 4.5 | Progressive |

Table 10. D-Terminal Control for Aspect Ratio Selection

| Input Logic Level <br> D3 (Pin 4) | Nominal Output (V) <br> L3_OUT (Pin 35) |  |
| :--- | :--- | :--- |
| Row (0) | 0.0 | Aspect Ratio |
| Low | $4: 3$ |  |
| Mid or open | 2.1 | $4: 3$ letterbox |
| High (1) | 4.5 | $16: 9$ |

Table 11. S-Terminal Control for Aspect Ratio Selection

| Input Logic Level <br> S (Pin 5) | Nominal Output (V) <br> S1/S2_OUT (Pin 34) |  |
| :--- | :--- | :--- |
| Row (0) | 0.0 | Aspect Ratio |
| Mid or open | 2.1 | $4: 3$ |
| High (1) | 4.5 | $4: 3$ letterbox |

## POWER-DOWN

The ADA4424-6 provides separate output enable pins for the SD and ED/HD sections. In addition to powering down the Y , C, and CVBS outputs, the SD_ENABLE pin, when driven low, also places the S1/S2 output (S1/S2_OUT, Pin 34) in a high impedance state. Likewise, driving the HD_ENABLE pin low disables the component outputs (HY_OUT, HPb_OUT, and HPr_OUT) and changes the L1, L2, and L3 outputs (Lx_OUT, Pin 35 to Pin 37) to a high impedance state. Control details are shown in Table 12 and Table 13.

Table 12. Power-Down Control for SD Channels

| SD_ENABLE <br> (Pin 8) | SD Outputs <br> (Y, C, CVBS) | S1/S2_OUT <br> (Pin 34) |
| :--- | :--- | :--- |
| Low (0) | Disabled | High-Z (Open) |
| High (1) | Enabled | Active |

Table 13. Power-Down Control for ED/HD Channels

| HD_ENABLE <br> (Pin 15) | ED/HD Outputs <br> (HY, HPb, HPr) | Lx_OUT (Pin 35, <br> Pin 36, Pin 37) |
| :--- | :--- | :--- |
| Low (0) | Disabled | High-Z (Open) |
| High (1) | Enabled | Active |

## CHARGE PUMP

The ADA4424-6 features an on-chip charge pump that supplies a negative rail voltage for the output stages. To minimize internal noise coupling, the charge pump uses an external connection to the negative supply pins (VSS_SD and VSS_HD). These pins should be connected to the C2/CP_OUT pin, each decoupled with a $1.0 \mu \mathrm{~F}$ capacitor. It is also recommended to place a small ( $1 \Omega$ ) series resistor in this connection. This forms a low-pass filter with the VSS decoupling capacitors and further reduces coupled noise. The charge pump also requires two $4.7 \mu \mathrm{~F}$ ceramic capacitors, one connected across the C1a and C1b pins, and one connected from the C2/CP_OUT pin to ground. The recommended charge pump configuration is shown in the application diagram (Figure 18).
With the black or zero level of the outputs placed at approximately ground potential, the outputs can swing up to 1.6 V in the negative direction. This eliminates the need for large output coupling capacitors because the input-referred dc offsets does not exceed $\pm 100 \mathrm{mV}$ (depending on the selected cancellation mode).

## PRINTED CIRCUIT BOARD (PCB) LAYOUT

As with all high speed applications, attention to the PCB layout is of paramount importance. When designing with the ADA4424-6, adhere to standard high speed layout practices. A solid ground plane is recommended, and surface-mount, ceramic power supply decoupling capacitors should be placed as close as possible to the supply pins. Connect all of the ADA4424-6 GND pins to the ground plane with traces that are as short as possible. Controlled impedance traces of the shortest length possible should be used to connect to the signal I/O pins and should not pass over any voids in the ground plane. A $75 \Omega$ impedance level is typically used in video applications. When driving transmission lines, include series termination resistors on the signal outputs of the ADA4424-6.

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Figure 18. Typical Application Diagram for Using the ADA4424-6 in Auto Offset Cancellation Mode ( $D$ and S Terminal Connections Not Shown)

## OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-153-BD-1
Figure 19. 38-Lead Thin Shrink Small Outline Package [TSSOP] (RU-38)
Dimensions shown in millimeters

ORDERING GUIDE

| Model $^{1}$ | Temperature Range | Package Description | Package <br> Option | Ordering <br> Quantity |
| :--- | :--- | :--- | :--- | :--- |
| ADA4424-6ARUZ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 38 -Lead Thin Shrink Small Outline Package (TSSOP) | RU-38 | 50 |
| ADA4424-6ARUZ-R7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 38 -Lead Thin Shrink Small Outline Package (TSSOP) | RU-38 | 1,000 |
| ADA4424-6ARUZ-RL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 38 -Lead Thin Shrink Small Outline Package (TSSOP) | RU-38 | 2,500 |

[^0]
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NOTES


[^0]:    ${ }^{1} Z=$ RoHS Compliant Part.

