



Wireless Components

PLL-Frequency Synthesizer PMB 2304R Version 2.1

Specification June 2002

preliminary

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Productinfo

Productinfo

General Description

The PMB 2304R PLL is a high speed CMOS IC, especially designed for use in battery powered radio equipment and mobile telephones and serves as a functional replacement of the PMB 2307R. The primary applications are in digital cellular and cordless systems e.g. GSM 900/1800/1900 and DECT systems. The wide range of dividing ratios also allows application in analog systems.

Package:



Features

- Low operating current consumption (typically 3.5 mA)
- High input sensitivity, high input frequencies (220 MHz)
- Extremely fast phase detector without dead zone
- Linearization of the phase detector output by current sources
- Synchronous programming of the counters (N-, N/A-, R-counters) and system parameters
- Fast modulus switchover for 65-MHz operation
- Switchable modulus trigger edge
- Serial control (3-wire bus: data, clock, enable) for fast programming (f_{max} ~ 10 MHz)

- Large dividing ratios for small channel spacing
 A counter 0 to 127
 N counter 3 to 16.383
 R counter 3 to 65.535
- Serial control (3-wire bus: data, clock, enable) for fast programming $(f_{\text{max}} \sim 10 \text{ MHz})$
- Switchable polarity and phase detector current programmable
- 2 Multifunction outputs f_m, f_m outputs of the R- and N/A- counters for test
- Output port (e.g. for standby of the prescaler)
- External current setting for PD output
- Lock detect output with quasidigital lock detect

Application

- GSM 900 / 1800 / 1900
- DECT

Analog systems

Ordering Information

Туре	Ordering Code	Package	
PMB 2304R	Q67106-H9100	P-TSSOP-16	

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Product Description

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Product Description

2.1 Overview

The PMB 2304R PLL is a high speed CMOS IC, especially designed for use in battery powered radio equipment and mobile telephones and serves as a functional replacement of the PMB 2307R. The primary applications are in digital cellular and cordless systems e.g. GSM 900/1800/1900 and DECT systems. The wide range of dividing ratios also allows application in analog systems.

2.2 Features

- Low operating current consumption (typically 3.5 mA)
- High input sensitivity, high input frequencies (220 MHz)
- Extremely fast phase detector without dead zone
- Linearization of the phase detector output by current sources
- Synchronous programming of the counters (N-, N/A-, R-counters) and system parameters
- Fast modulus switchover for 65-MHz operation
- Switchable modulus trigger edge
- Serial control (3-wire bus: data, clock, enable) for fast programming $(f_{max} \sim 10 \text{ MHz})$
- Large dividing ratios for small channel spacing A counter 0 to 127
 N counter 3 to 16.383
 R counter 3 to 65.535
- Serial control (3-wire bus: data, clock, enable) for fast programming $(f_{max} \sim 10 \text{ MHz})$
- Switchable polarity and phase detector current programmable
- **2** Multifunction outputs f_m , f_{vn} outputs of the R- and N/A- counters for test
- Output port (e.g. for standby of the prescaler)
- External current setting for PD output
- Lock detect output with quasidigital lock detect



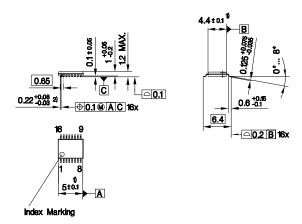
Product Description

2.3 Application

- GSM 900 / 1800 / 1900
- DECT
- Analog systems

2.4 Package Outlines

P-TSSOP-16

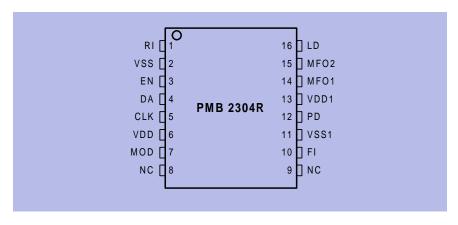


- 1) Does not include plastic or metal protrusion of 0.15 max. per side
- 2) Does not include dambar protrusion

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3.5	Programming
3.6	Data acquisition



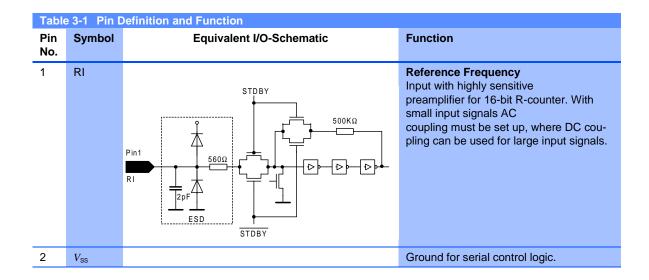
3.1 Pin Configuration



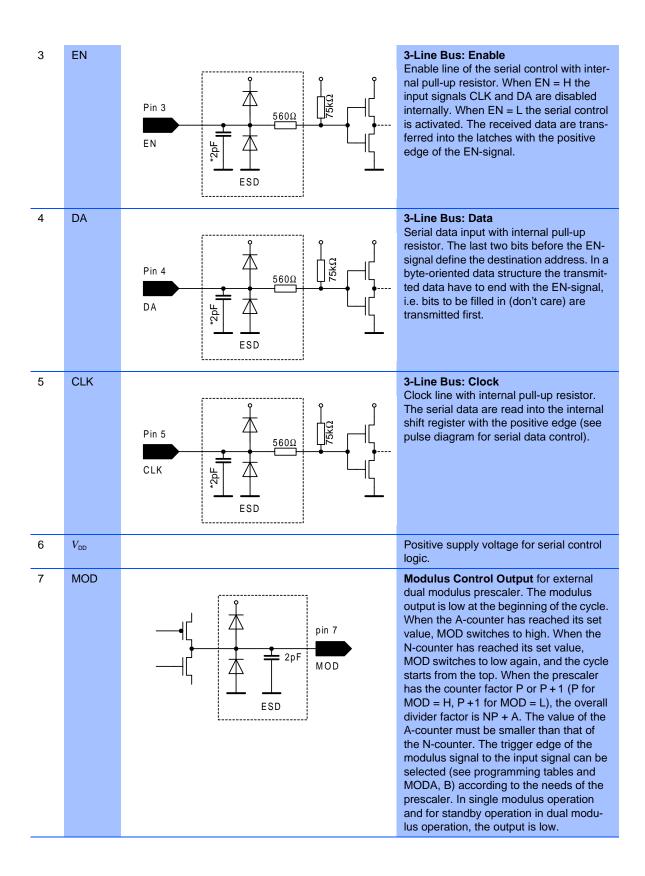
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Figure 3-1 Pin Configuration

3.2 Pin Definition and Function









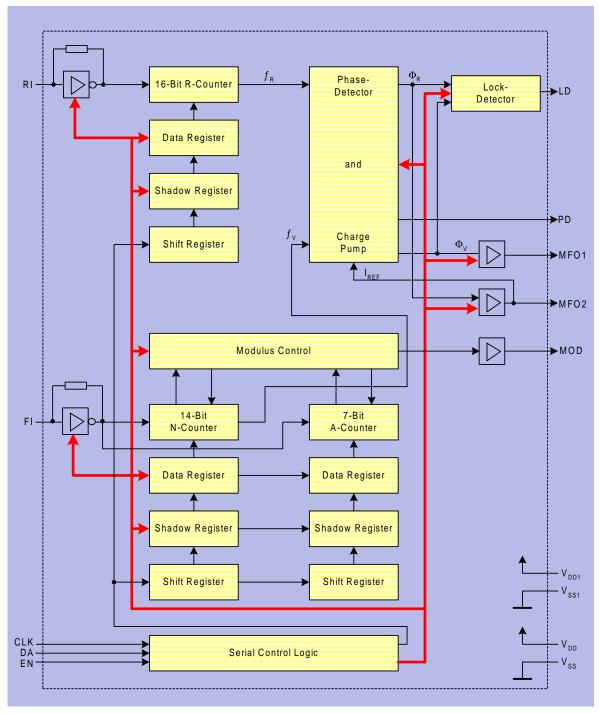
8	NC		not connected
9	NC		not connected
10	FI	Pin10 500KΩ 500KΩ FI STDBY	VCO-Frequency Input with highly sensitive preamplifier for 14-bit N-counter and 7-bit A-counter. With small input signals AC coupling must be set up, where DC coupling can be used for large input signals.
11	$V_{ m SS1}$		Ground for the preamplifiers, counters, phase detector and charge pump. (Note: The pins V _{DD} and V _{DD1} respec-
			tively $V_{\rm ss}$ and $V_{\rm ss1}$ have to have the same supply voltage.)
12	PD	* Only this pin has limited build-in ESD protection	Phase Detector Tristate charge pump output. The integrated, positive and negative current sources can be programmed with respect to their current density by means of the serial control. Activation and deactivation depend on the phase relationship of the scaled-down input signals FI:N, RI:R. (See phase detector output waveforms.) frequency $f_V < f_R$ or f_V lagging:p-channel current source active frequency $f_V > f_R$ or f_R leading:n-channel current source active frequency $f_V = f_R$ and PLL locked:current sources are switched off, PD-output is tristate In standby mode the PD-output is set to tristate. The assignment of the current sources to the output signals of the phase detector can be swapped in it's polarity, i.e. the sign of the phase detector constant can be controlled.
13	$V_{ extsf{DD1}}$		Positive supply voltage for the preamplifiers, counters, phase detector and charge pump.



14	MFO1	pin 14 pin 14 MFO1	Multifunction Output for the signals $f_{\rm RN}$, $\Phi_{\rm VN}$, and port1. The signal $f_{\rm RN}$ is the divided signal of the reference frequency. The L-time corresponds to $1/f_{\rm RI}$ respectively —In the port function the port 1 output signal is assigned to the information of the programmed status. The output switches with the rising edge of the EN-signal The standby mode does not affect the port function.
15	MFO2	Internal Charge Pump Mode & standby	Multifunction I/O-Pin for the external reference current setting $I_{\rm REF}$ and the signals $\Phi_{\rm RN}$ and $f_{\rm VN}$ (in testmode). -The signal $f_{\rm VN}$ is the divided signal of Flinput. The L-time corresponds to $1/f_{\rm Fl}$ respectively. Output levels are not specified, the signal should only be used for test purpose. -In the internal charge pump mode the reference current $I_{\rm REF}$ at MFO2 determines the value of the PD-output current.
16	LD	pin 16 ESD Divided to the point of the poi	Lock Detector Output (open drain). Unipolar output of the phase detector in the form of a pulse-width modulated signal. The LD-pulse width corresponds to the phase difference. In the locked state the LD-signal is at H-level. For standby mode see Standby Table. Only for ABL status 11 no gating of ABL impulse is performed.



3.3 Functional Block Diagram



Funct_block.wmf

Figure 3-2 Functional Block Diagram



3.4 General Description

The circuit consists of a reference-, A- and N-counter, a dual modulus control logic, a phase detector with charge pump output and a serial control logic. The setting of the operating mode and the selection of the counter ratios is done serially at the ports CLK, DA and EN.

The operating modes allow the selection of single or dual operation, asynchronous or synchronous data acquisition, 4 different antibacklash-impulse times, 8 different PD-output current modes, polarity setting of the PD-output signal, adjustment of the trigger-edge of the MOD-output signal, 2 standby modes and the control of the multifunction outputs MFO1 and MFO2.

The reference frequency is applied at the RI-input and divided by the R-counter. It's maximum value is 100 MHz. The VCO-frequency is applied at the FI-input and divided by the N- or N/A-counter according to single or dual mode operation. The maximum value at FI is 220 MHz at single-, and 65 MHz at dual mode operation.

The phase and frequency sensitive phase detector produces an output signal with adjustable anti-backlash impulses in order to prevent a dead zone for very small phase deviations. Phase differences of less than 100 ps can be resolved. In general the shortest anti-backlash pulse gives the best system performance.

3.5 Programming

Programming of the IC is done by a serial data control. The contents of the message are assigned to the functional units according to the address. Single or dual mode operation as well as asynchronous or synchronous data acquisition is set by status 2 and should therefore precede the programming of the counters.

3.6 Data acquisition

The PMB 2304R offers the possibility of synchronous data acquisition to avoid error signals at the phase detector due to non-corresponding dividing factors in the counters produced by asynchronous loading.

Synchronous programming guarantees control during changes of frequency or channel. That means that the state of the phase detector or the phase difference is kept maintained, and in case of "lock in", the control process starts with the phase difference "zero".



This is done as follows:

- 1. Setting of synchronous data acquisition by status 2.
- 2. Programming of the R-counter, status 1 (optional)-data is being loaded into shadow registers.
- Programming of the N- or N/A-counter-data is being loaded into shadow registers, the EN-signal starts the synchronous loading procedure.
- 4. Synchronous programming which means data transfer of all data from the shadow registers to the data registers takes place at that point in time when the respective counter reaches "zero + 1", the maximum repetition rate for channel change is therefore f_F : N.
- 5. Transfer of status 1 information into the corresponding data register is tied to the N-counter loading, but follows the loading of the N-data register in the distance of one N-counter dividing ratio, this guarantees that for example a new PD-current value becomes valid at the same time when the counters are loaded with the new data.

Synchronous avoids additional phase error caused by programming. Synchronous data acquisition is of especial advantage, when large steps in frequency are to be made in a short time. For this purpose a high reference frequency can be programmed in order to achieve rapid – "rough" – transient response. This method increases the fundamental frequency nearly by the square route of the reference frequency relation. When rough lock is achieved, another synchronous data transfer is needed to switch back to the original channel spacing. A "fine" lock in will finish the total step response. It may not be necessary to change reference frequency, but it make sense to perform synchronous data acquisition in any case. Especially for GSM, PCN, PCS, DECT, DAMPS, PHP systems the synchronous mode should be used to get best performance of the PMB 2304R.

Standby Condition:

The PMB 2304R has two standby modes (standby 1, 2) to reduce the current consumption.

Standby 1 switches off the whole circuit, the current consumption is reduced below 1 μA .

Standby 2 switches off the counters, the charge pump and the outputs, only the preamplifiers stay active.

For the influence on the output signals see standby table (5-10).

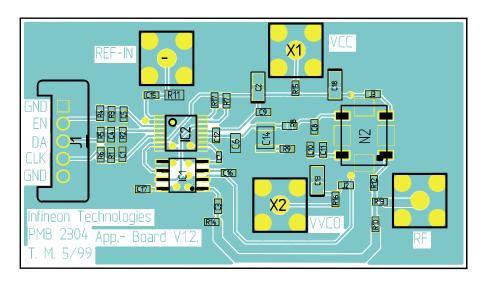
 f_{RN} , f_{VN} , Φ_{RN} , Φ_{VN} are the inverted signals of f_{R} , f_{V} , Φ_{R} , Φ_{V}

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4.1 PCB Layout

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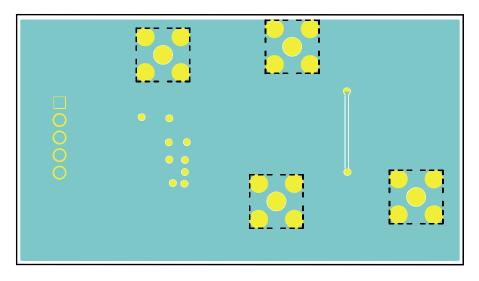


(3.81 07/15/99 pmb23045.tc)

Figure 4-1 Top Side

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(3.81 07/15/99 pmb23045.tc)

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Figure 4-2 Bottom Side



4.2 Application Board

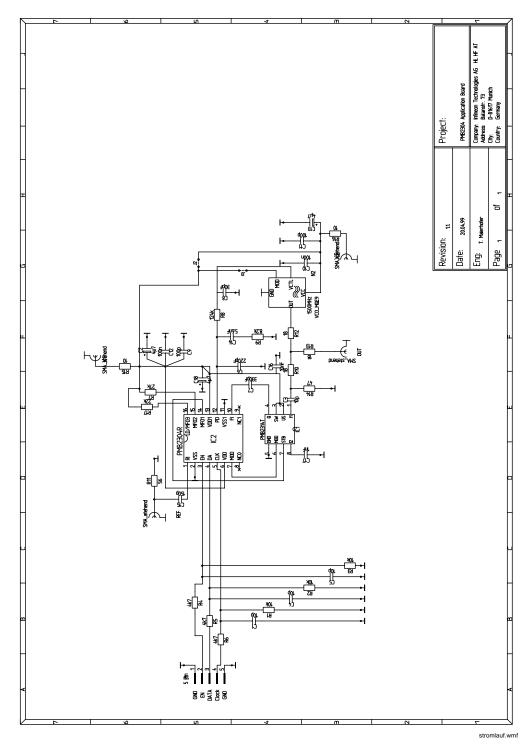


Figure 4-3 Application board



4.3 Bill of material

Table 4-1					
Nr	Reference	Symbol name	Technology		
1	C1	CAP	10p		
2	C2	CAPELK	4μ7		
3	C3	CAP	10p		
4	C4	CAP	10p		
5	C5	CAP	10p		
6	C6	CAP	220pF		
7	C7	CAP	330pF		
8	C8	CAP	30pF		
9	C9	CAP	100p		
10	C10	CAP	100n		
11	C11	CAP	100p		
12	C12	CAP	100n		
13	C13	CAPELK	4µ7		
14	C14	CAP	5.6nF		
15	C15	CAP	150p		
16	C16	CAP	10nF		
17	C17	CAP	1nF		
18	C18	CAPELK	4µ7		
19	IC1	PMB2314T	PMB2314T		
20	IC2	PMB2305	PMB2304R		
21	J1	CON-5	5 Pin		
22	J2	JUMP-2SMD0603	JUMPER_2SMD06031		
23	J3	JUMP-2SMD0603	JUMPER_2SMD06031		
24	N2	VCO2	1500MHz		
25	R1	RES	10k		
26	R2	RES	10k		
27	R3	RES	10k		
28	R4	RES	4k7		
29	R5	RES	4k7		
30	R6	RES	4k7		
31	R7	RES	27k		
32	R8	RES	124k		



33	R9	RES	8.2k		
34	R10	RES	18		
35	R11	RES	56		
36	R12	RES	18		
37	R13	RES	18		
38	R14	RES	47		
39	R15	RES	10		
40	R16	RES	10		
41	R17	RES	22k		
42		SMA	SMA_stehend		
43	-	SMA	SMA_stehend		
44	X1	SMA	SMA_stehend		
45	X2	SMA	SMA_stehend		

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5.1 Electrical Data

5.1.1 Absolute Maximum Range



WARNING

The maximum ratings may not be exceeded under any circumstances, not even momentarily and individually, as permanent damage to the IC will result.

Table 5-1 Absolute Maximum Range						
Parameter	Symbol	Limit	Limit Values		Remarks	
		min	max			
Supply Voltage	V_{DD}	-0.3	6	V		
Input Voltage	V_1	-0.3	V _{DD} + 0.3	V		
Output Voltage	V_{Q}	GND	V_{DD}	V		
Power dissipation per output	P_{Q}		10	mW		
Total power dissipation	P_{tot}		300	mW		
Ambient temperature	T_{A}	-40	85	°C	in operation	
Storage temperature	T_{stg}	-50	125	°C		
Thermal Resistance	R _{thJA}		180	K/W		
ESD Integrity except @Pin 12 (PD) (according to MIL833 Method 3015.7)	V _{ESD}		1	KV		
ESD Integrity except @Pin 12 (PD) (according to MIL833 Method 3015.7)	V _{ESD}		400	V		



5.1.2 Operating Ratings

Within the operating ratings the IC operates as described in the circuit description. The AC/DC characteristic limits are not guaranteed.

Table 5-2 Operating Ratings, S	upply Volta	ge V _{VCC} = 2.7	' V 4.5 V, Ar	nbient te	emp. T _{AMB} = -30°C + 85	°C
Parameter	Symbol	Limit Values		Unit	Test Conditions L	Item
		min	max			
Supply Voltage	V_{DD}	2.7	5.5	V		
Input frequency dual	$f_{\sf FI}$	0.1	65	MHz	V _{DD} = 4.55.5V	
Input frequency single HF-mode	$f_{\sf FI}$	0.1	220	MHz	V _{DD} = 4.55.5V	
Input frequency single LF-mode	$f_{\sf FI}$	0.1	90	MHz	V _{DD} = 4.55.5V	
Input reference frequency	f_{RI}		100	MHz	$V_{\rm DD} = 4.55.5 \text{V}$	
Input frequency dual mode	f_{FI}	0.1	30	MHz	$V_{\rm DD}$ = 2.7V	
Input frequency single HF-mode	f_{FI}	0.1	90	MHz	$V_{\rm DD}$ = 2.7V	
Input frequency single LF-mode	f_{FI}	0.1	35	MHz	$V_{\rm DD}$ = 2.7V	
Input reference frequency	f_{RI}		20	MHz	$V_{\rm DD}$ = 2.7V	
PD-output current	/ I _{PD}		4	mA		
PD-output voltage	V_{PD}	0.5	V _{DD} - 0.5	V	V _{DD} = 4.5- 5.5V	
PD-output voltage	<i>V</i> PD	0.5	V _{DD} - 0.5	С	$V_{\rm DD}$ = 2.7V	
Ambient temperature	T_{A}	-40	85	°C		

5.1.3 Typical Supply Current I_{DD}

All pins are protected against ESD. Unused inputs without pullup resistors must be connected to either $V_{\rm DD}$ or $V_{\rm SS}$.

Table 5-3 Typical Supply Current I _{DD}										
Parameter	Symbol	Limit Values		Unit	Test Conditions L	Item				
		min		max						
Supply voltage	V_{DD}	2.7	5	5.5	V					
Supply current:						f _{FI} = 50MHz				
single mode HF	I_{DD}	1.63	2.6	2.94	mA	$V_{\rm FI}$ = 150mVrms				
dual mode	I_{DD}	1.76	2.80	3.17	mA	f_{RI} = 10MHz				
standby 2	I_{DD}	0.11	0.62	0.75	mA	V_{RI} = 150mVrms				
standby 1	I_{DD}			1	μA	$I_{PD} = 0.25 \text{mA}$				
						$I_{\text{ref}} = 100 \mu\text{A}$				



5.1.4 AC/DC Characteristics

AC/DC characteristics involve the spread of values guaranteed within the specified supply voltage and ambient temperature range. Typical characteristics are the median of the production.

Table 5-4 AC/DC Characteristics with		Ambient	temp. T _{amb} = -20 .	85 °C, Supply Voltage V _{VCC} =		.7 4	4.5V
	Symbol	Li	mit Values	Unit	Test Conditions	L	Item
		min	typ max				
Input Signals DA, CLK, EN	(with inter	nal pull-up	resistors)				
H-input voltage	V_{IH}	0.7- V _{DD}	V_{DD}	V			1.1
L-input voltage	V_{IL}		0.3- V _{DD}	V			1.2
Input capacity	C_{I}		5	pF			
H-input current	I_{H}		10	μΑ	$V_{I} = V_{DD} = 5.5V$		1.3
L-input current	I_{L}	-60		μΑ	$V_{I} = GND$		1.4
Input Signal RI							
Input voltage	V_{l}	100		mVrms	$f = 4100MHz, V_{DD} = 4.5V$		2.1
Input voltage	V_{l}	100		mVrms	$f = 430MHz, V_{DD} = 2.7V$		2.2
Slew rate		4		V/μs	V _{DD} = 2.75.5V		2.3
Input capacity	C_{I}		3	pF			
H-input current	I_{H}		30	μΑ	$V_{I} = V_{DD} = 5.5 V$		2.4
L-input current	I_{L}	-30		μΑ	$V_{I} = GND$		2.5
Input Signal FI (dual mode)							
Input voltage	V_{l}	180		mVrms	$f = 465MHz, V_{DD} = 4.5V$		3.1
Input voltage	V_{I}	50		mVrms	$f = 1025MHz, V_{DD} = 2.7V$		3.2
Slew rate		4		V/μs	V _{DD} = 2.75.5V		3.3
Input capacity	C_{I}		3	pF			
H-input current	I_{H}		30	μΑ	$V_{\rm I} = V_{\rm DD} = 5.5 \text{V}$		3.4
L-input current	I_{L}	-30		μΑ	$V_{I} = GND$		3.5
Input Signal FI (single HF-n	node)						
Input voltage	V_{I}	200		mVrms	$f = 4200$ MHz, $V_{DD} = 4.5$ V		4.1
Input voltage	V_{I}	200		mVrms	f = 490MHz, V _{DD} = 2.7V		4.2
Input voltage	V_{I}	50		mVrms	f = 1040MHz, V _{DD} = 4.5V		4.3
Slew rate		4		V/µs	V _{DD} = 2.75.5V		4.4
Input capacity	C_{I}		3	pF			
H-input current	I_{H}		30	μΑ	$V_{\rm I} = V_{\rm DD} = 5.5 \text{V}$		4.5
L-input current	I_{L}	-30		μΑ	$V_{\rm I} = {\sf GND}$		4.6



Table 5-4 AC/DC Characte	ristics with	Ambien	t temp. T	amb= -20	85 °C,	Supply Voltage V _{VCC} = 2	.7	4.5V
	Symbol	L	imit Value	es	Unit	Test Conditions	L	Item
		min	typ	max				
Input Signal FI (single LF-n	node)							
Input voltage	V_{I}	100			mVrms	$f = 490MHz, V_{DD} = 4.5V$		5.1
Input voltage	V_{I}	100			mVrms	$f = 435$ MHz, $V_{DD} = 2.7$ V		5.2
Slew rate		4			V/μs	V _{DD} = 2.75.5V		5.3
Input capacity	C_{I}			3	pF			
H-input current	I_{H}			30	μΑ	$V_{\rm I} = V_{\rm DD} = 5.5 \text{V}$		5.4
L-input current	I_{L}	-30			μΑ	$V_{I} = GND$		5.5
Output Current /IPD/								
Current mode:								6.1
"0.175 mA"	I_{PROG}	-20%	0.175	+20%	mA			6.2
"0.25 mA"	I_{PROG}	-20%	0.25	+20%	mA	V _{DD} = 4.55.5V		6.3
"0.35 mA"	I_{PROG}	-20%	0.35	+20%	mA	$V_{\rm PD} = V_{\rm DD}/2$		6.4
"0.5 mA"	I_{PROG}	-20%	0.5	+20%	mA	$I_{REF} = 100 \mu A$		6.5
"0.7 mA"	I_{PROG}	-20%	0.7	+20%	mA			6.6
"1.0 mA"	I_{PROG}	-15%	1.0	+15%	mA	- V _{DD} = 4.5V		6.7
"1.4 mA"	I_{PROG}	-15%	1.0	+15%	mA			6.8
"2.0 mA"	I_{PROG}	-10%	2.0	+10%	mA			6.9
Standby"	I_{PD}		0.1	1	nA			6.10
Output Tolerance I _{PD}								
$\Delta I_{ extsf{PD}}$ / $I_{ extsf{PROG}}$		-10%	-5%	+0%		$V_{PD} = V_{DD}/2, V_{DD} = 2.7V$		7.1
$\Delta I_{ extsf{PD}}$ / $I_{ extsf{PROG}}$			±2.5%			$V_{PD} = 0.52.2V, V_{DD} = 2.7V$		7.2
Input Voltage MFO2 (Intern	al charge p	oump mo	de)					
Reference voltage	V_{REF}	0.9	1.1	1.3	V	$V_{\rm DD} = 2.75.5 \text{V}$ $I_{\rm REF} = 100 \mu \text{A}$		8.1
Output Signal MFO1 (push	pull)							
H-output voltage	V_{QH}	$V_{ m DD}$ - 1			V	$V_{\rm DD} = 4.55.5 \text{V}, I_{\rm QH} = -2 \text{mA}$		9.1
L-output voltage	V_{QL}			1	V	$V_{\rm DD} = 4.55.5 \text{V}, I_{\rm QL} = 2\text{mA}$		9.2
H-output voltage	V_{QH}	V _{DD} - 1			V	$V_{\rm DD} = 2.7 \text{V}, I_{\rm QH} = -1.2 \text{mA}$		9.3
L-output voltage	V_{QL}			1	V	$V_{\rm DD} = 2.7 \text{V}, I_{\rm QL} = 1.2 \text{mA}$		9.4
Rise time	t_{R}		2.5	10	ns	V _{DD} = 4.55.5V,C _I = 10pF		9.5
Fall time	t_{F}		2.0	10	ns	$V_{\rm DD} = 4.55.5 \text{V}, C_{\rm l} = 10 \text{p}$		9.6
Rise time	t_{R}		5	12	ns	$V_{\rm DD} = 2.7 \text{V}, C_{\rm l} = 10 \text{pF}$		9.7
Fall time	t _F		4	12	ns	$V_{\rm DD} = 2.7 \text{V}, C_{\rm l} = 10 \text{pF}$		9.8



Table 5-4 AC/DC Characte								
	Symbol		mit Value		Unit	Test Conditions	L	Item
Output Signal MFO2 (push	(llua	min	typ	max				
H-output voltage	V_{QH}	V _{DD} - 1			V	$V_{\rm DD} = 4.55.5 \text{V}, I_{\rm QH} = 2\text{mA}$		10.1
L-output voltage	V_{QL}	DD		1	V	$V_{\rm DD} = 4.55.5 \text{V}, I_{\rm QL} = 2\text{mA}$		10.2
H-output voltage	V_{QH}	V _{DD} - 1			V	$V_{\rm DD} = 2.7 \text{V}, I_{\rm QH} = 1.2 \text{mA}$		10.3
L-output voltage	V_{QL}	55		1	V	$V_{\rm DD} = 2.7 \text{V}, I_{\rm QL} = 1.2 \text{mA}$		10.4
Rise time	t_{R}		2	10	ns	$V_{\rm DD} = 4.55.5 \text{V}, C_{\rm I} = 10 \text{pF}$		10.5
Fall time	t_{F}		2	10	ns	$V_{\rm DD} = 4.55.5 \text{V}, C_{\rm l} = 10 \text{p}$		10.6
Rise time	t_{R}		3	10	ns	$V_{\rm DD} = 2.7 \text{V}, C_{\rm l} = 10 \text{pF}$		10.7
Fall time	t _F		3	10	ns	$V_{\rm DD} = 2.7 \text{V}, C_{\rm l} = 10 \text{pF}$		10.8
Output Signal LD (n-chanr	nel open dra	ain)						
L-output voltage	V_{QL}			0.4	V	$V_{\rm DD}$ = 2.75.5V, $I_{\rm QL}$ = 0.3mA		11.1
H-output current	I_{QH}			5	μΑ	V _{DD} = 2.75.5V		11.2
Fall time	t_{F}		3	10	ns	V _{DD} = 4.55.5V,C _I =10pF		11.3
Fall time	t_{F}		5	12	ns	$V_{\rm DD} = 2.7 \text{V}, C_{\rm I} = 10 \text{pF}$		11.4
Output Signal MOD (push	pull)							
H-output voltage	V_{QH}	V _{DD} -0.4			V	$V_{\rm DD}$ = 4.55.5V, $I_{\rm QH}$ = -0.5mA		12.1
L-output voltage	V_{QL}			0.4	V	$V_{\rm DD}$ = 4.55.5V $I_{\rm QL}$ = 0.5mA		12.2
H-output voltage	V_{QH}	V _{DD} -0.4			V	$V_{\rm DD}$ = 2.7V, $I_{\rm QH}$ = - 0.3mA		12.3
L-output voltage	V_{QL}			0.4	V	$V_{\rm DD}$ = 2.7V, $I_{\rm QL}$ = 0.3mA		12.4
Rise time	t_{R}		1.5	3	ns	$V_{\rm DD}$ = 4.55.5V, $C_{\rm I}$ = 5pF		12.5
Fall time	t_{F}		1.3	3	ns	V _{DD} = 4.55.5V, C _I = 5pF		12.6
Propagation delay time H-L to FI	t _{DQHL}		8	12	ns	$V_{\rm DD}$ = 4.55.5V, $C_{\rm I}$ = 5pF		12.7
Propagation delay time L-H to FI	t _{DQHL}		8	12	ns	$V_{\rm DD}$ = 4.55.5V, $C_{\rm I}$ = 5pF		12.8
Rise time	t_{R}		3.2	5	ns	$V_{\rm DD} = 2.7 \text{V}, \ C_{\rm I} = 5 \text{pF}$		12.9
Fall time	t_{F}		2	5	ns	$V_{\rm DD} = 2.7 \text{V}, \ C_{\rm I} = 5 \text{pF}$		12.10
Propagation delay time H-L to FI	t_{DQHL}		15		ns	$V_{\rm DD} = 2.7 \text{V}, C_{\rm I} = 5 \text{pF}$		12.11
Propagation delay time L-H to FI	t_{DQHL}		15		ns	$V_{\rm DD} = 2.7 \text{V}, C_{\rm I} = 5 \text{pF}$		12.12

This value is only guaranteed in lab.



5.2 Phase detector outputs

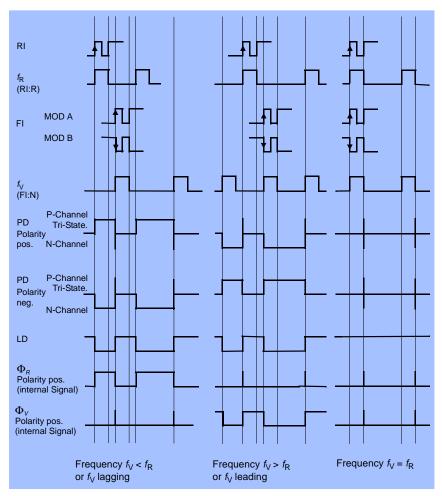


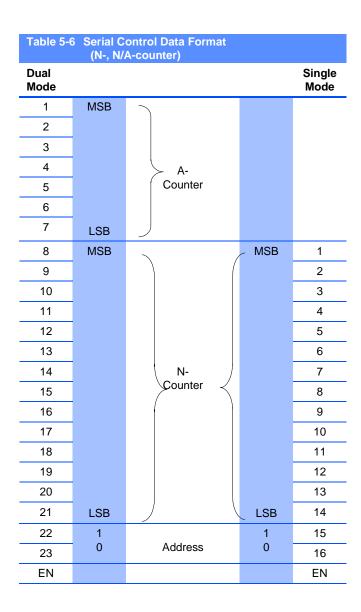
Figure 5-1 Phase detector output signals

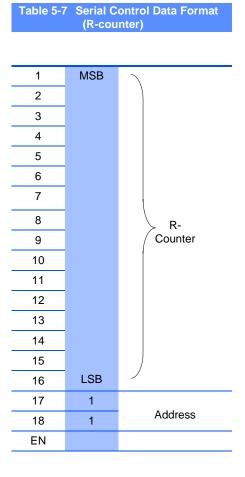


5.3 Serial Control Data Format

Table 5-	5 Serial Control Data Forma	t (status 1	,2)		
Status 1		Status 2	1		
			0	1	
	Data acquisition mode	1	asynchronous	synchronous	
	Mode 1	2	see t	table	
	Mode 2	3	see t	table	
	PD-polarity	4	negative	positive	
	Standby 1	5	standby	active	
	Standby 2	6	standby	active	
	Anti-backlash pulse width 1	7	see table		
	Anti-backlash pulse width 2	8	see table		
	Preamplifier select	9	see t	table	
	Single / dual mode	10	single	dual	
1	Port 1	11	low	high	
2	PD-current 1	12	see t	table	
3	PD-current 2	13	see t	table	
4	PD-current 3	14	see t	table	
5	0 Address 0	15			
6	0 1	16			
EN		EN			









5.4 Programming Tables

Table 5-8			
Status Bits			
Anti- Backlash Pulse Width 2	Anti- Backlash Pulse Width 1	t _₩ (typ.) [ns]	
0	0	1.3	$V_{\rm DD} = 5 \text{V}$
0	1	5	
1	0	10	not recommended
1	1	13*	any application where continuous lock detect is required

Table 5-9					
Status Bits		Preamplifier Function Mode			
Single/Dual Mode	Preamplifier Select				
0	0	FI-input frequency,single HF-mode			
0	1	FI-input frequency, single LF-mode			
1	0	FI-input frequency, dual-mode, FI- trigger edge LH, MOD A			
1	1	FI-input frequency, dual-mode, FI- trigger edge HL, MOD B			

Table 5-10 Standby Table										
		Output Pins								
Status	MF	·O1	MFO2	LD	PD	MOD				
	$\Phi_{f V}$	Φ_{VN}								
Standby 1	low	high	high	resistive	tristate	low				
Standby 2	low	high	high	resistive	tristate	low				

^{*} No ABL gating performed * In general the shortest anti-backlash pulse gives the best system performance



Table 5-11			
Status Bits		PD-Current Mode	
PD-Current 3	PD-Current 2	PD-Current 1	lpd/mA
0	0	0	0.175
0	0	1	0.25
0	1	0	0.35
0	1	1	0.5
1	0	0	0.7
1	0	1	1
1	1	0	1.4
1	1	1	2

Table 5-12									
Status Bits	i	Multifunction Output							
Mode 2	Mode 1	MFO 1	MFO 2	Remarks					
0	0	f _{RN}	f_{VN}	test mode					
0	1	Φ_{V}	Φ_{RN}	external charge pump mode 1					
1	0	Φ_{VN}	Φ_{RN}	external charge pump mode 2					
1	1	Port 1	I _{ref}	internal charge pump mode					

5.5 Pulse Diagram

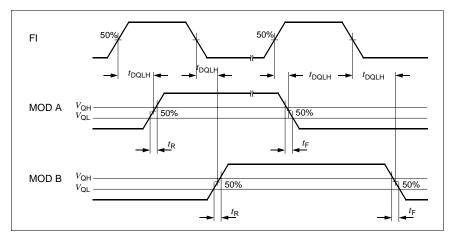


Figure 5-2 Pulse diagram



5.6 Serial Control Data Input Timing

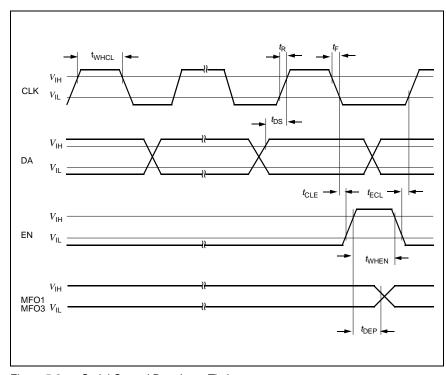


Figure 5-3 Serial Control Data Input Timing

Table 5-13								
Parameter	Symbol	Limit V	Limit Values					
		min	max					
Clock frequency	f_{CL}		12	MHz				
H-pulsewidth (CL)	t _{WHCL}	40		ns				
Data setup	t_{DS}	20		ns				
Setup time-clock enable	t _{CLE}	20		ns				
Setup time enable-clock	t _{ECL}	20		ns				
H-pulsewidth (enable)	t _{WHEN}	40		ns				
Rise, fall time	t _R , t _F		10	μs				
Propagation delay time EN-PORT	t _{DEP}		1	μs				



5.7 Diagram Input Sensitivity FI

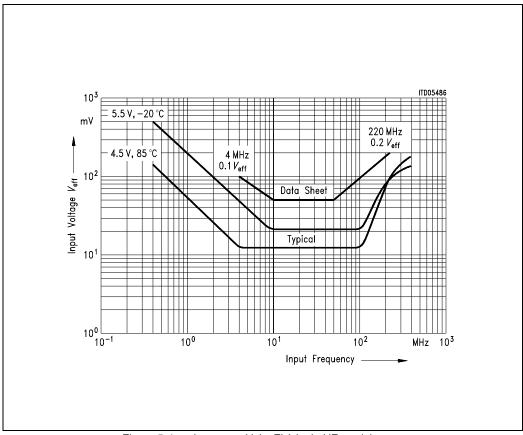


Figure 5-4 Input sensitivity FI (single HF-mode)