19-0461; Rev 2; 4/97

EVALUATION KIT AVAILABLE

400MHz, Ultra-Low-Distortion Op Amps

General Description

The MAX4108/MAX4109/MAX4308/MAX4309 op amps combine ultra-high-speed performance with ultra-low-distortion operation. The MAX4108 is compensated for unity-gain stability; the MAX4109, MAX4308, and MAX4309 are compensated for minimum closed-loop gains (A_{VCL}) of 2V/V, 5V/V, and 10V/V, respectively.

The MAX4108 delivers a 400MHz unity-gain bandwidth with a 1200V/µs slew rate. An ultra-low-distortion design provides an unprecedented spurious-free dynamic range of -93dBc (MAX4108) at 5MHz (V_{OUT} = 2Vp-p, R_L = 100Ω), making these amplifiers ideal for high-performance RF signal processing.

These high-speed op amps feature a wide output voltage swing and a high-current output-drive capability of 90mA.

Applications

High-Speed ADC/DAC Preamp

RGB and Composite Video

High-Performance Receivers

Pulse/RF Amplifier

Active Filters

Ultrasound

Broadcast and High-Definition TV

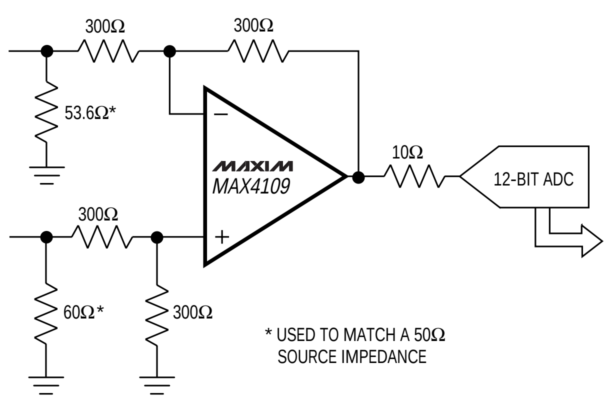
Features

- **High Speed:**
  - **400MHz Unity-Gain Bandwidth (MAX4108)**
  - **225MHz -3dB Bandwidth (A_{VCL} = +2, MAX4109)**
  - **220MHz -3dB Bandwidth (A_{VCL} = +5, MAX4308)**
  - **200MHz -3dB Bandwidth (A_{VCL} = +10, MAX4309)**
- **1200V/µs Slew Rate**
- **Excellent Spurious-Free Dynamic Range:**
  - **-93dBc at f_C = 5MHz (MAX4108)**
  - **-90dBc at f_C = 5MHz (MAX4109)**
- **100MHz 0.1dB Gain Flatness (MAX4108)**
- **High Full-Power Bandwidth: 300MHz (MAX4108, V_O = 2Vp-p)**
- **High Output Drive: 90mA**
- **Output Short-Circuit Protected**
- **Low Differential Gain/Phase: 0.004%/0.008°**

Ordering Information

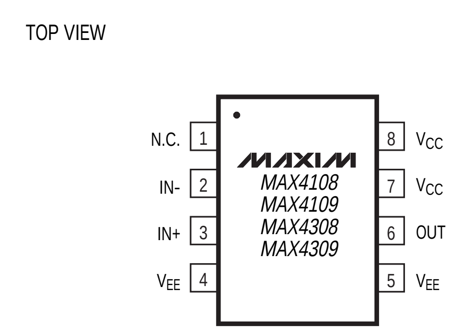
PART	TEMP. RANGE	PIN-PACKAGE
MAX4108ESA	-40°C to +85°C	8 SO
MAX4109ESA	-40°C to +85°C	8 SO
MAX4308ESA	-40°C to +85°C	8 SO
MAX4309ESA	-40°C to +85°C	8 SO

Typical Application Circuit



DIFFERENCE AMPLIFIER/ADC PREAMPLIFIER

Pin Configuration



SO

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V_{CC} to V_{EE}) 12V
Voltage on Any Pin to Ground or
Any Other Pin ....(V_{EE} - 0.3V) to (V_{CC} + 0.3V)
Short-Circuit Duration (OUT to GND) ....Continuous
Continuous Power Dissipation (T_A = +70°C)
SO (derate 5.88mW/°C above +70°C) 471mW
Operating Temperature Range ....-40°C to +85°C
Storage Temperature Range ....-65°C to +150°C
Junction Temperature ....+150°C
Lead Temperature (soldering, 10sec) ....+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{CC} = +5V, V_{EE} = -5V, T_A = T_{MIN} to T_{MAX}, typical values are at T_A = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
DC SPECIFICATIONS (R_L = ∞)							
Input Offset Voltage	V_{OS}	V_{OUT} = 0V			1	8	mV
Input Offset Voltage Drift	TCV_{OS}	V_{OUT} = 0V			13		µV/°C
Input Bias Current	I_B	V_{OUT} = 0V, V_{IN} = $-V_{OS}$			12	34	µA
Input Offset Current	I_{OS}	V_{OUT} = 0V, V_{IN} = $-V_{OS}$			0.05	2.5	µA
Common-Mode Input Resistance	R_{INCM}	Either input			1.5		MΩ
Common-Mode Input Capacitance	C_{INCM}	Either input			1		pF
Input Voltage Noise	e_n	f = 10kHz			6		nV/√Hz
Integrated Voltage Noise	E_{nRMS}	f_B = 1MHz to 100MHz			75		$µV_{RMS}$
Input Current Noise	i_n	f = 10kHz			2		pA/√Hz
Integrated Current Noise	I_n	f_B = 1MHz to 100MHz			25		nA_{RMS}
Common-Mode Input Voltage	V_{CM}			-2.5		2.5	V
Common-Mode Rejection	CMR	V_{CM} = ±2.5V		70	100		dB
Power-Supply Rejection	PSR	V_S = ±4.5V to ±5.5V		70	90		dB
Open-Loop Voltage Gain	A_{OL}	V_{OUT} = ±2.0V, V_{CM} = 0V, R_L = 100Ω		70	100		dB
Quiescent Supply Current	I_S	V_{IN} = 0V			20	27	mA
Output Voltage Swing	V_{OUT}	R_L = ∞		2.5 to -3.1	2.9 to -3.8		V
		R_L = 100Ω		2.5 to -3.1	2.7 to -3.7		
Output Current Drive	I_{OUT}	R_L = 33Ω, T_A = 0°C to +85°C		65	90		mA
Short-Circuit Output Current	I_{SC}	Short to ground			100		mA
AC SPECIFICATIONS (R_L = 100Ω)							
-3dB Bandwidth	BW_{-3dB}	V_{OUT} ≤ $0.1V_{RMS}$	MAX4108		400		MHz
			MAX4109		225		
			MAX4308		220		
			MAX4309		200		

ELECTRICAL CHARACTERISTICS (continued)

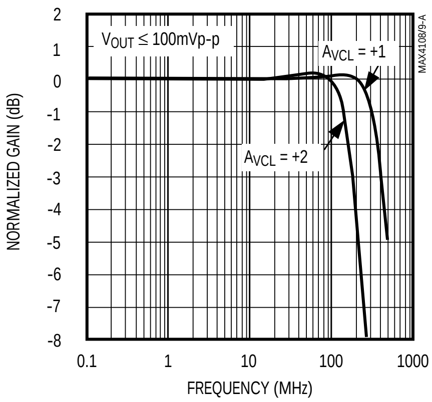
(V_{CC} = +5V, V_{EE} = -5V, T_A = T_{MIN} to T_{MAX}, typical values are at T_A = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
AC SPECIFICATIONS (R_L = 100Ω) (continued)							
Full-Power Bandwidth	FPBW	V_{OUT} = 2Vp-p	MAX4108		300		MHz
			MAX4109		200		
			MAX4308		190		
			MAX4309		130		
0.1dB Gain Flatness	$BW_{0.1dB}$	MAX4108, A_{VCL} = +1			100		MHz
		MAX4109, A_{VCL} = +2			25		
		MAX4308, A_{VCL} = +5			100		
		MAX4309, A_{VCL} = +10			30		
Slew Rate	SR	-2V ≤ V_{OUT} ≤ 2V			1200		V/µs
Settling Time	t_S	-1V ≤ V_{OUT} ≤ 1V	To 0.1%		8		ns
			To 0.01%		12		
Rise/Fall Times	t_R, t_F	10% to 90%	-2V ≤ V_{OUT} ≤ 2V		3		ns
			-50mV ≤ V_{OUT} ≤ 50mV		2		
Differential Gain	DG	f = 3.58MHz, R_L = 150Ω			0.004		%
Differential Phase	DP	f = 3.58MHz, R_L = 150Ω			0.008		degrees
Input Capacitance	C_{IN}				2		pF
Output Resistance	R_{OUT}	f = 10MHz			1		Ω
Spurious-Free Dynamic Range	SFDR	MAX4108, V_{OUT} = 2Vp-p, A_{VCL} = +1	f_C = 5MHz, R_L = 100Ω		-93		dBc
			f_C = 20MHz, R_L = 100Ω		-81		
		MAX4109, V_{OUT} = 2Vp-p, A_{VCL} = +2	f_C = 5MHz, R_L = 100Ω		-90		
			f_C = 20MHz, R_L = 100Ω		-80		
		MAX4308, V_{OUT} = 2Vp-p, A_{VCL} = +5	f_C = 5MHz, R_L = 100Ω		-83		
			f_C = 20MHz, R_L = 100Ω		-80		
		MAX4309, V_{OUT} = 2Vp-p, A_{VCL} = +10	f_C = 5MHz, R_L = 100Ω		-83		
			f_C = 20MHz, R_L = 100Ω		-80		
Third-Order Intercept	IP3	f_C = 10MHz	MAX4108		39		dBm
			MAX4109		36		
			MAX4308		46		
			MAX4309		43		

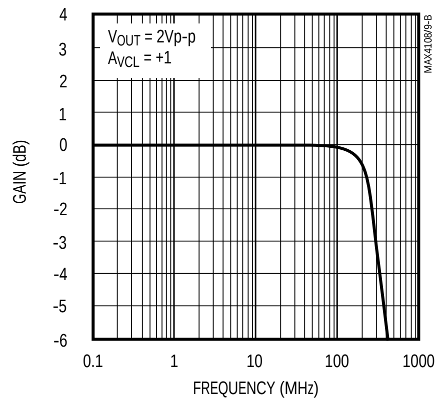
Typical Operating Characteristics

(V_{CC} = +5V, V_{EE} = -5V, R_L = 100Ω, T_A = +25°C, unless otherwise noted.)

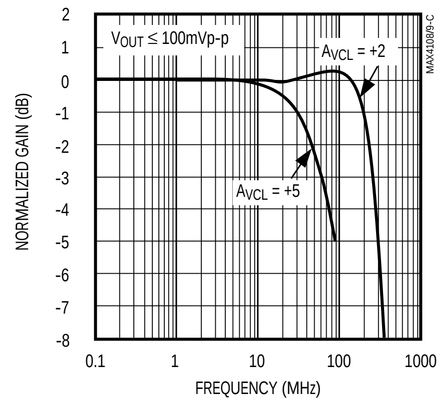
MAX4108
SMALL-SIGNAL GAIN vs. FREQUENCY



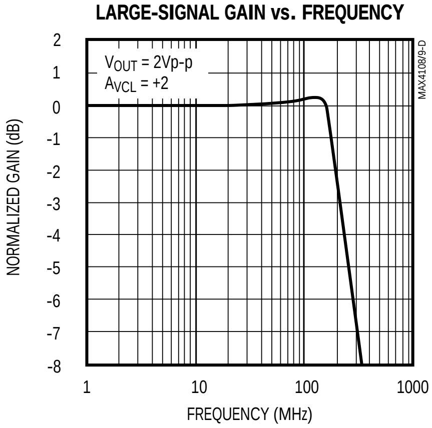
MAX4108
LARGE-SIGNAL GAIN vs. FREQUENCY



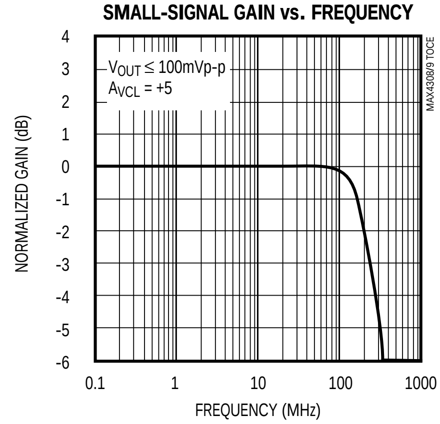
MAX4109
SMALL-SIGNAL GAIN vs. FREQUENCY



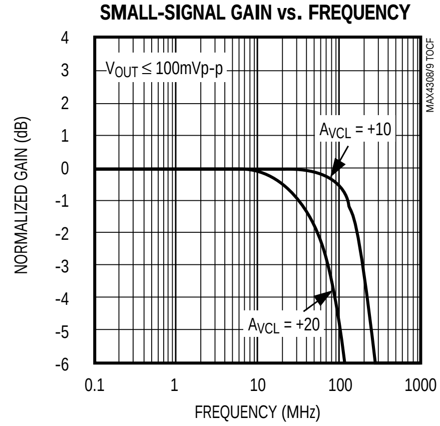
MAX4109
LARGE-SIGNAL GAIN vs. FREQUENCY



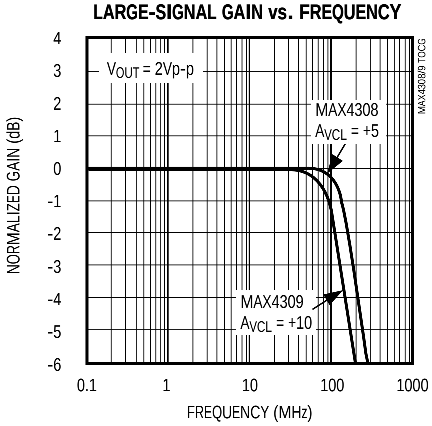
MAX4308
SMALL-SIGNAL GAIN vs. FREQUENCY



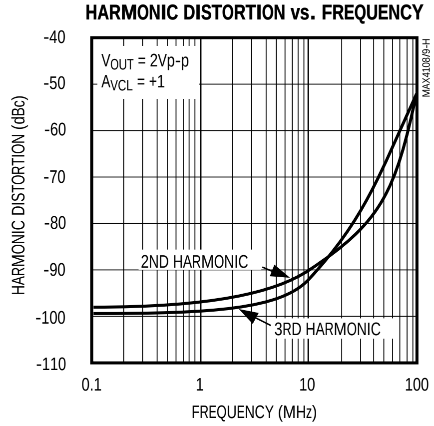
MAX4309
SMALL-SIGNAL GAIN vs. FREQUENCY



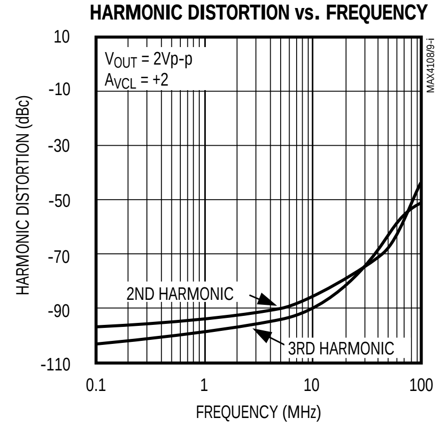
MAX4308/MAX4309
LARGE-SIGNAL GAIN vs. FREQUENCY



MAX4108
HARMONIC DISTORTION vs. FREQUENCY



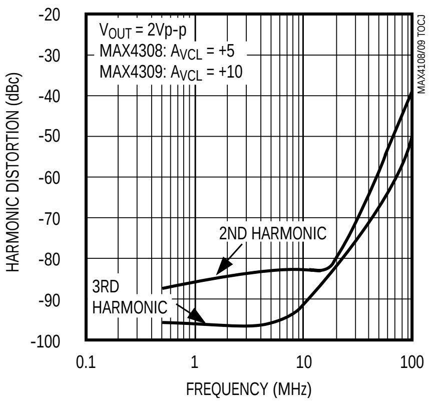
MAX4109
HARMONIC DISTORTION vs. FREQUENCY



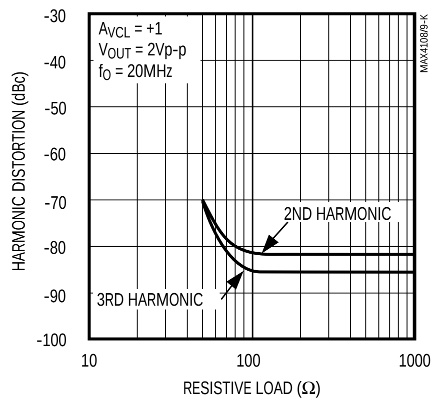
Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = -5V, R_L = 100Ω, T_A = +25°C, unless otherwise noted.)

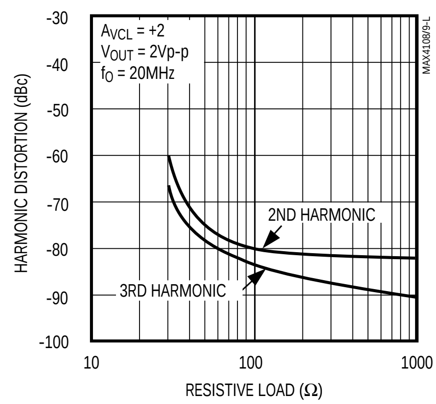
MAX4308/MAX4309
HARMONIC DISTORTION vs.FREQUENCY



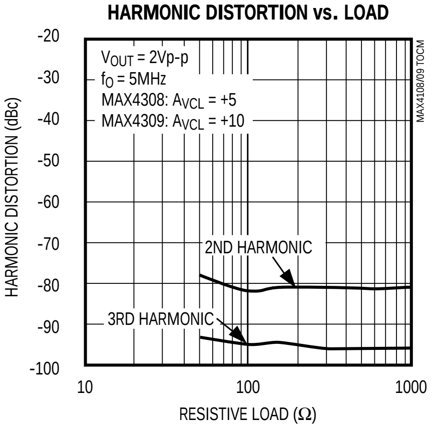
MAX4108
HARMONIC DISTORTION vs. LOAD



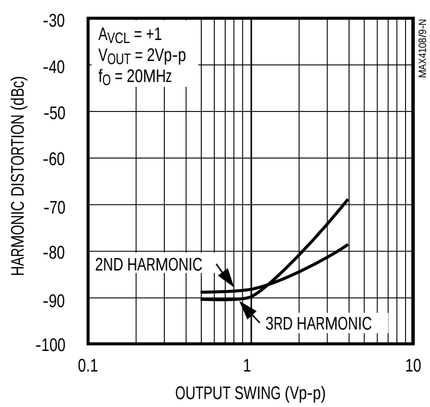
MAX4109
HARMONIC DISTORTION vs. LOAD



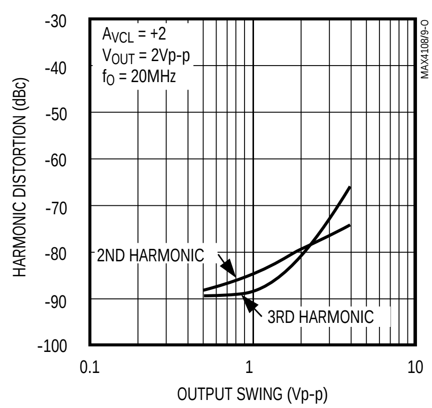
MAX4308/MAX4309
HARMONIC DISTORTION vs. LOAD



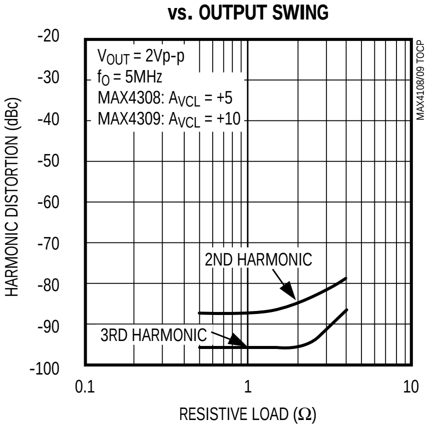
MAX4108
HARMONIC DISTORTION vs. OUTPUT SWING



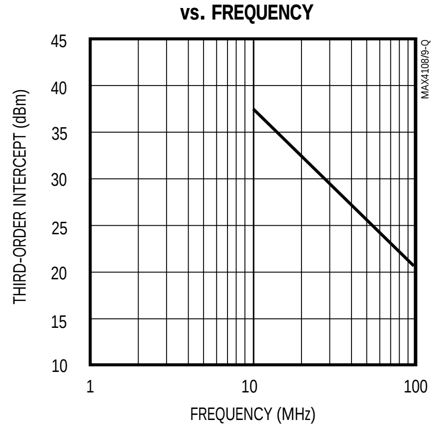
MAX4109
HARMONIC DISTORTION vs. OUTPUT SWING



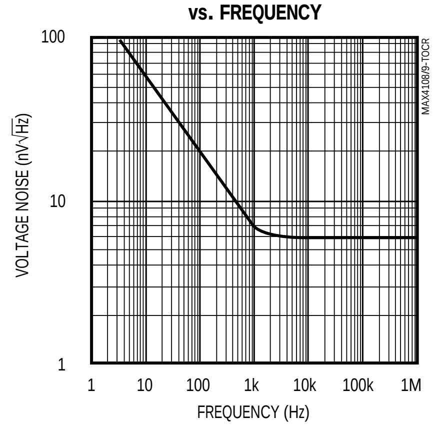
MAX4308/MAX4309
HARMONIC DISTORTION
vs. OUTPUT SWING



MAX4109
TWO-TONE THIRD-ORDER INTERCEPT
vs. FREQUENCY



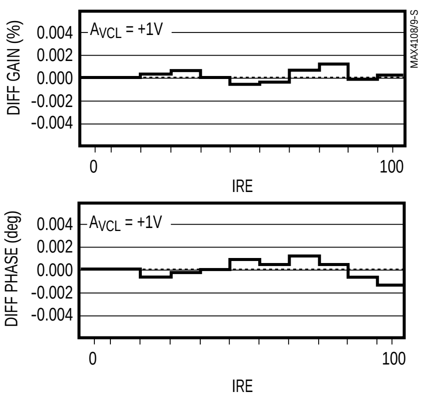
INPUT VOLTAGE NOISE
vs. FREQUENCY



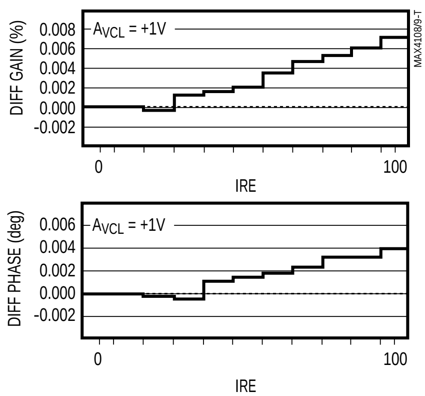
Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = -5V, R_L = 100Ω, T_A = +25°C, unless otherwise noted.)

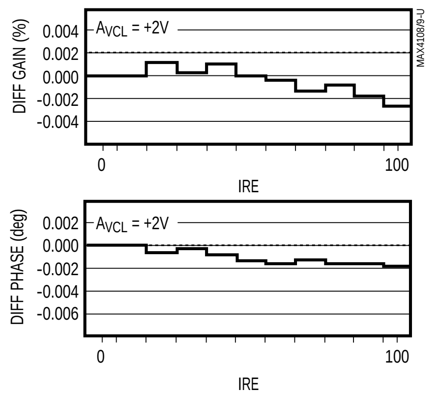
MAX4108 DIFFERENTIAL GAIN AND PHASE (R_L = 150Ω)



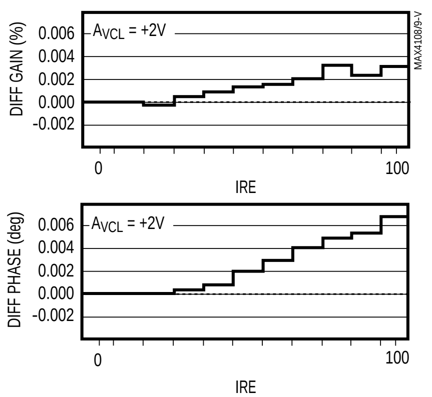
MAX4108 DIFFERENTIAL GAIN AND PHASE (R_L = 75Ω)



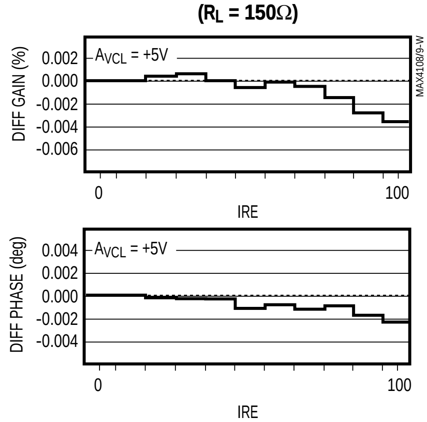
MAX4109 DIFFERENTIAL GAIN AND PHASE (R_L = 150Ω)



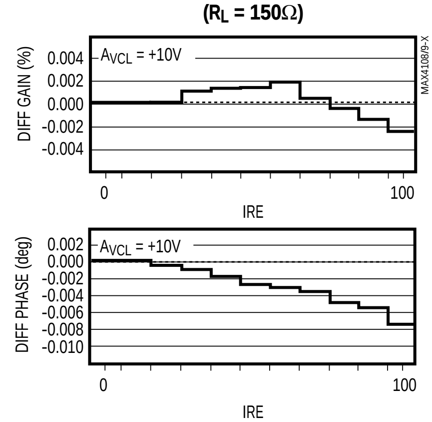
MAX4109 DIFFERENTIAL GAIN AND PHASE (R_L = 75Ω)



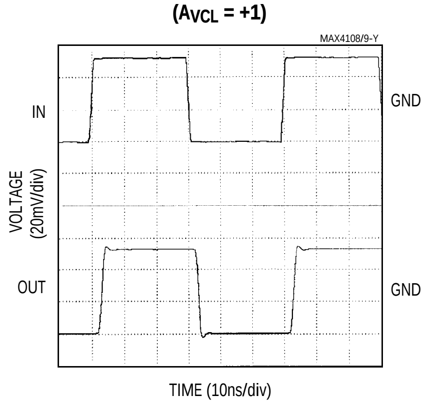
MAX4308 DIFFERENTIAL GAIN AND PHASE (R_L = 150Ω)



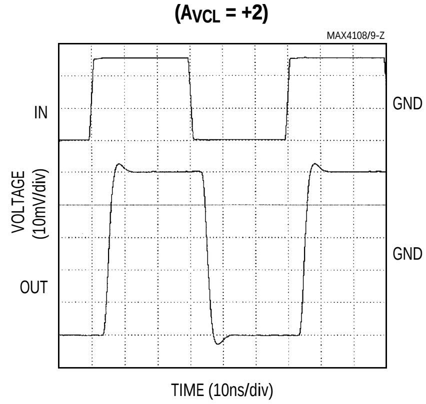
MAX4309 DIFFERENTIAL GAIN AND PHASE (R_L = 150Ω)



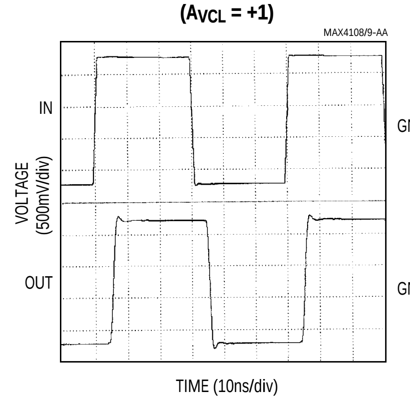
MAX4108 SMALL-SIGNAL PULSE RESPONSE (A_{VCL} = +1)



MAX4108 SMALL-SIGNAL PULSE RESPONSE (A_{VCL} = +2)



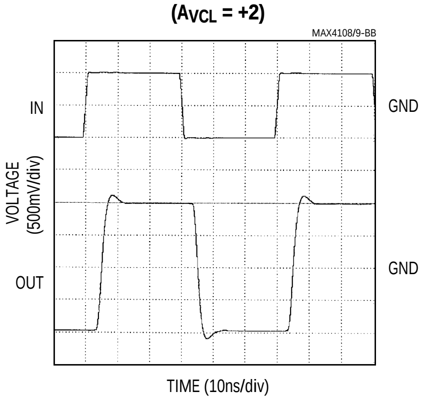
MAX4108 LARGE-SIGNAL PULSE RESPONSE (A_{VCL} = +1)



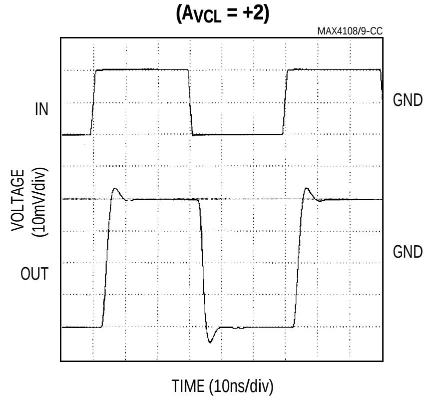
Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = -5V, R_L = 100Ω, T_A = +25°C, unless otherwise noted.)

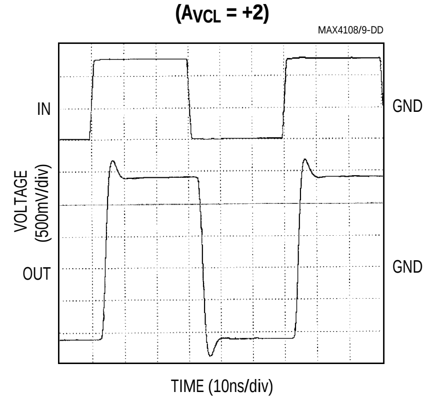
MAX4108
LARGE-SIGNAL PULSE RESPONSE
(A_{VCL} = +2)



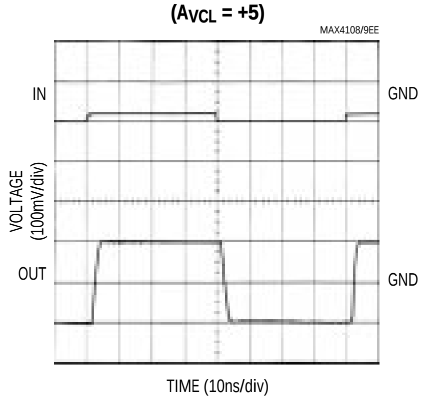
MAX4109
SMALL-SIGNAL PULSE RESPONSE
(A_{VCL} = +2)



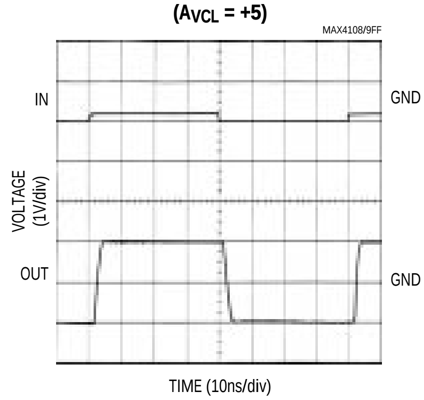
MAX4109
LARGE-SIGNAL PULSE RESPONSE
(A_{VCL} = +2)



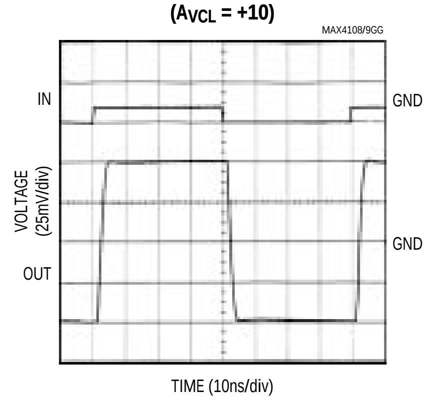
MAX4308
SMALL-SIGNAL PULSE RESPONSE
(A_{VCL} = +5)



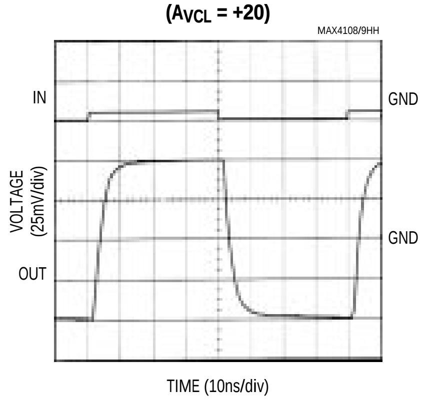
MAX4308
LARGE-SIGNAL PULSE RESPONSE
(A_{VCL} = +5)



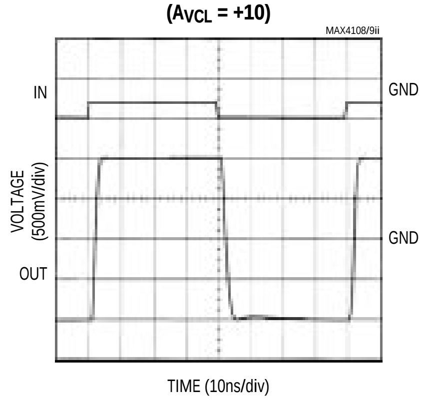
MAX4309
SMALL-SIGNAL PULSE RESPONSE
(A_{VCL} = +10)



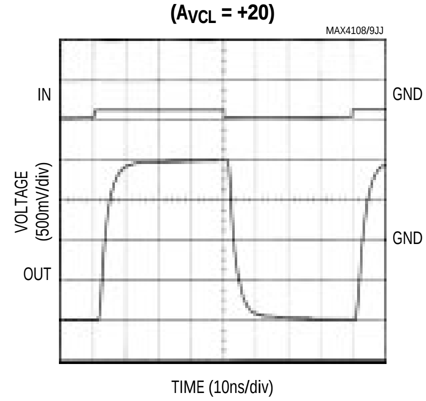
MAX4309
SMALL-SIGNAL PULSE RESPONSE
(A_{VCL} = +20)



MAX4309
LARGE-SIGNAL PULSE RESPONSE
(A_{VCL} = +10)



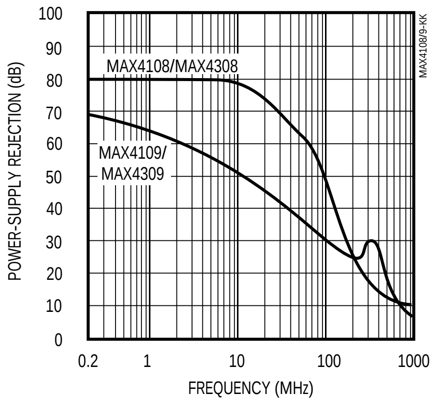
MAX4309
LARGE-SIGNAL PULSE RESPONSE
(A_{VCL} = +20)



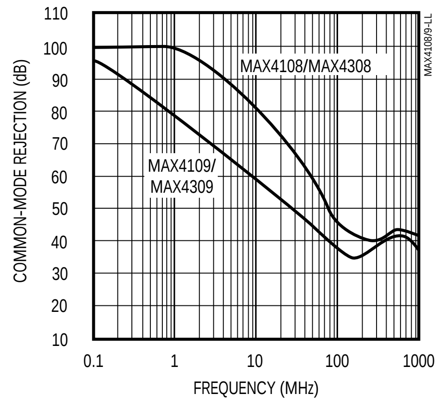
Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = -5V, R_L = 100Ω, T_A = +25°C, unless otherwise noted.)

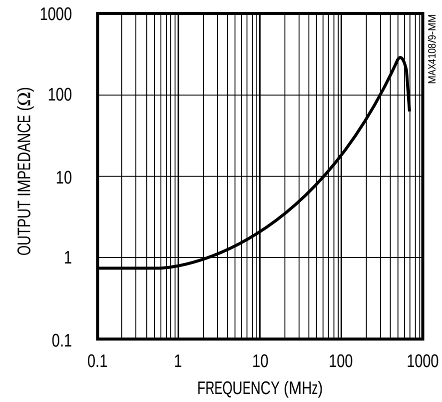
POWER-SUPPLY REJECTION vs. FREQUENCY



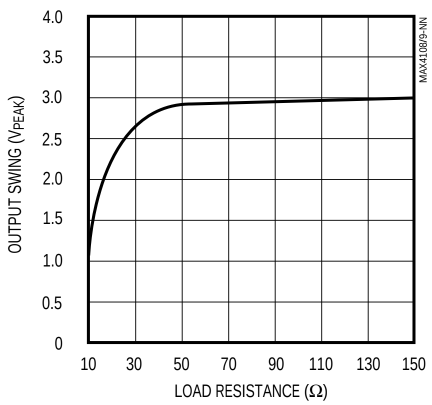
COMMON-MODE REJECTION vs. FREQUENCY



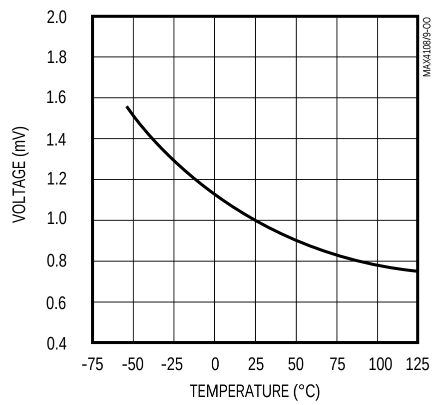
CLOSED-LOOP OUTPUT IMPEDANCE vs. FREQUENCY



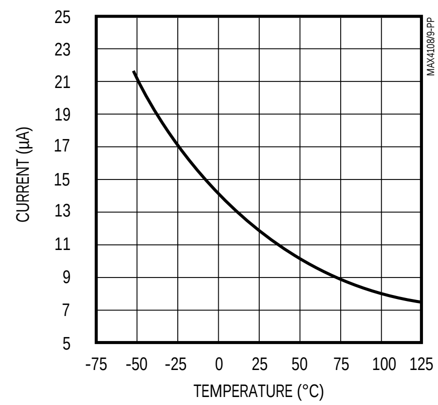
OUTPUT SWING vs. LOAD RESISTANCE



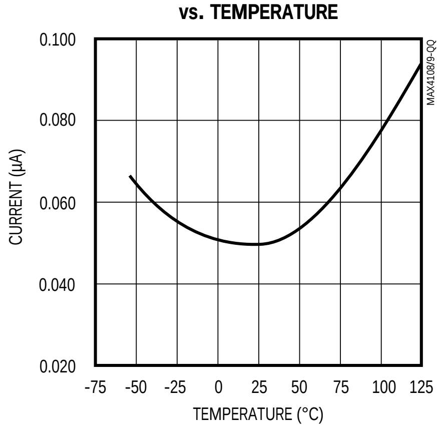
INPUT OFFSET VOLTAGE vs. TEMPERATURE



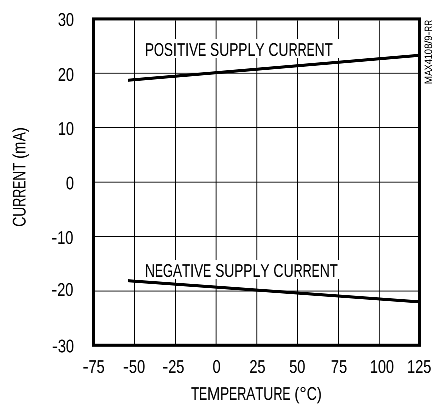
INPUT BIAS CURRENT vs. TEMPERATURE



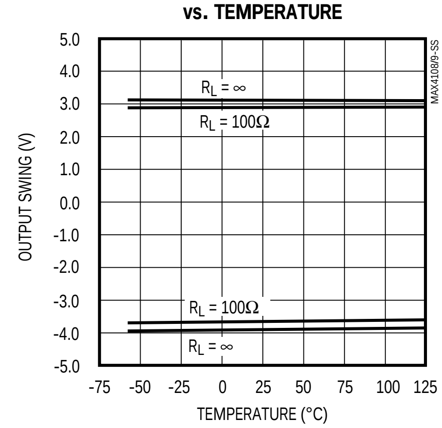
INPUT OFFSET CURRENT vs. TEMPERATURE



POWER-SUPPLY CURRENT vs. TEMPERATURE



OUTPUT SWING vs. TEMPERATURE



Pin Description

PIN	NAME	FUNCTION
1	N.C.	No Connection. Not internally connected.
2	IN-	Inverting Input
3	IN+	Noninverting Input
4, 5	V_{EE}	Negative Power Supply, connect to $-5V_{DC}$.
6	OUT	Amplifier Output
7, 8	V_{CC}	Positive Power Supply, connect to $+5V_{DC}$.

Detailed Description

Choosing Resistor Values

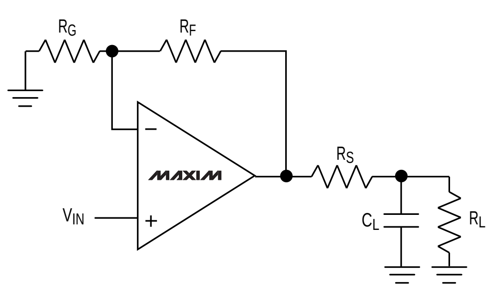
Unity-Gain Configuration

The MAX4108 is internally compensated for unity gain. When configured for unity gain, the device requires a small resistor in series with the feedback path. This resistor improves the AC response by reducing the Q of the tank circuit, which is formed by parasitic feedback inductance and capacitance.

Inverting and Noninverting Configurations

The values of the gain-setting feedback and input resistors are important design considerations. Large resistor values will increase voltage noise, and will interact with the amplifier's input and PC board capacitance to generate undesirable poles and zeros, which can decrease bandwidth or cause oscillations. For example, a noninverting gain of +2, using 1kΩ resistors combined with 2pF of input capacitance and 0.5pF of board capacitance, will cause a feedback pole at 128MHz. If this pole is within the anticipated amplifier bandwidth, it will jeopardize stability. Reducing these 1kΩ resistors to 100Ω will extend the pole frequency to 1.28GHz, but could limit output swing by adding 200Ω in parallel with the amplifier's load. Clearly, the selection of resistor values must be tailored to the specific application.

The MAX4108/MAX4109/MAX4308/MAX4309 are ultra-low-distortion, high-bandwidth op amps. The output distortion will be degraded as the total load resistance seen by the amplifier decreases. To minimize distortion products, keep the input and gain-setting resistors relatively large. A 500Ω feedback resistor combined with an appropriate input resistor to set the gain will provide excellent AC performance without significantly increasing distortion.



PART	R_F (Ω)	R_G (Ω)	GAIN (V/V)
MAX4108	24	∞	1
MAX4109	500	500	2
MAX4308	500	125	5
MAX4309	500	56	10

Figure 1a. Using an Isolation Resistor for High Capacitive Loads

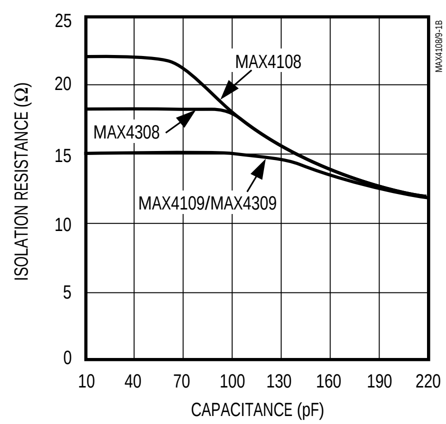


Figure 1b. Optimal Isolation Resistor (R_S) vs. Capacitive Load

Driving Capacitive Loads

The MAX4108/MAX4109/MAX4308/MAX4309 are optimized for AC performance. They are not designed to drive highly reactive loads. Reactive loads will decrease phase margin and may produce excessive ringing and oscillation. Figure 1a shows a circuit that eliminates this problem, and Figure 1b is a graph of the optimal isolation resistor (R_S) vs. capacitive load. Figures 2a–2d show how a capacitive load causes excessive peaking of the amplifier's bandwidth if the capacitive load is not isolated (R_S) from the amplifier. A small isolation resistor (usually 15Ω to 22Ω) placed

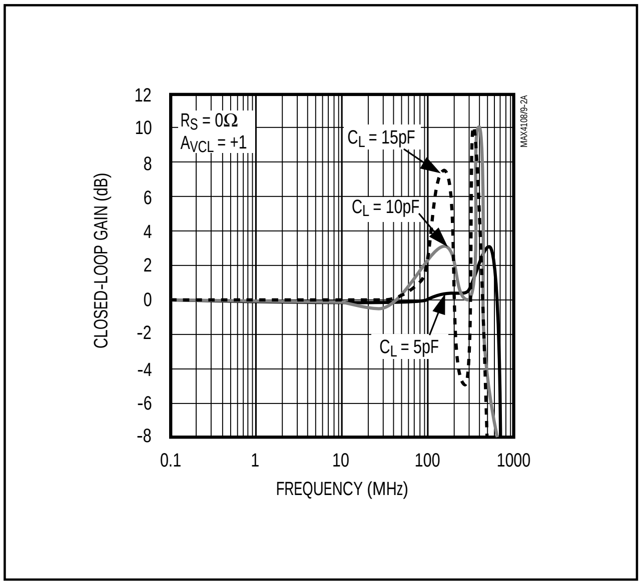


Figure 2a. MAX4108 Response vs. Capacitive Load—No Resistive (R_S) Isolation (circuit shown in Figure 1a)

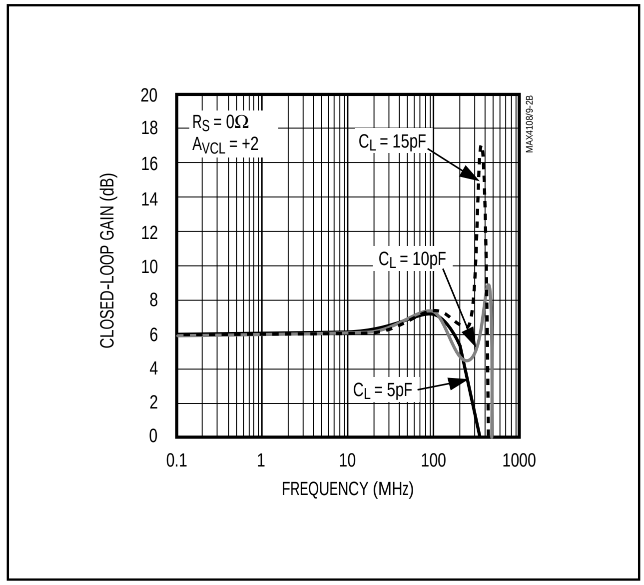


Figure 2b. MAX4109 Response vs. Capacitive Load—No Resistive (R_S) Isolation (circuit shown in Figure 1a)

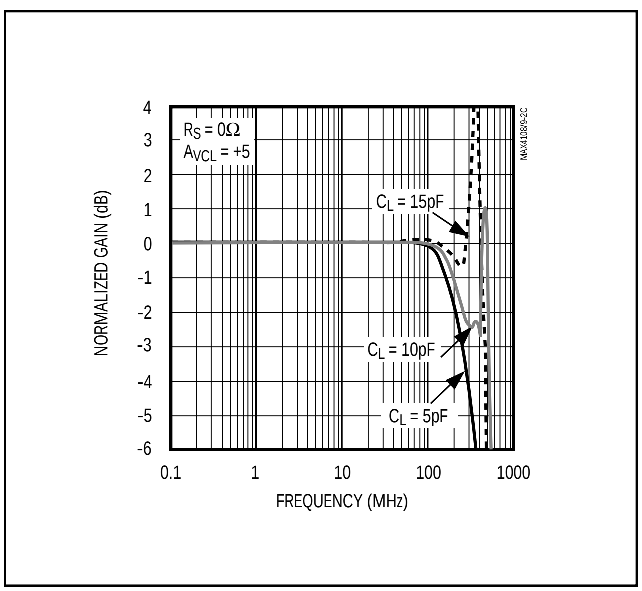


Figure 2c. MAX4308 Response vs. Capacitive Load—No Resistive (R_S) Isolation (circuit shown in Figure 1a)

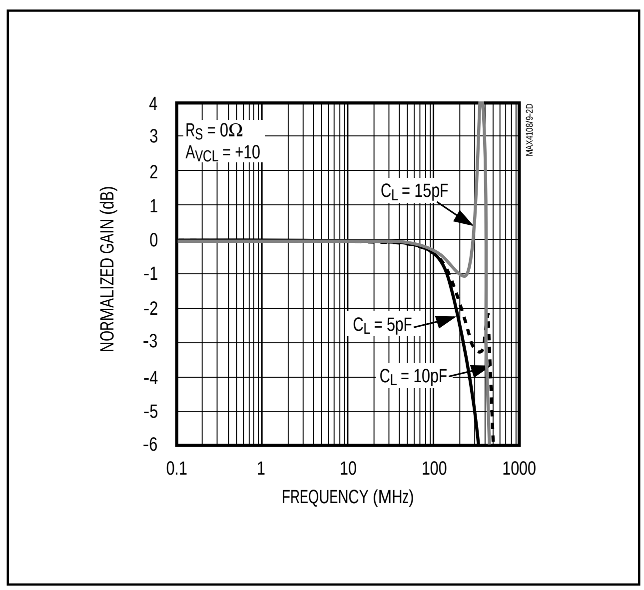


Figure 2d. MAX4309 Response vs. Capacitive Load—No Resistive (R_S) Isolation (circuit shown in Figure 1a)

before the reactive load prevents ringing and oscillation. At higher capacitive loads, AC performance will be controlled by the interaction of the load capacitance and isolation resistor. Figures 3a–3c show the effect of an isolation resistor on the MAX4108/MAX4109/MAX4308/MAX4309 closed-loop response.

Coaxial cable and other transmission lines are easily driven when terminated at both ends with their characteristic impedance. When driving back-terminated transmission lines, the capacitance of the transmission line is essentially eliminated.

ADC Input Buffers

Input buffer amplifiers can be a source of significant errors in high-speed ADC applications. The input buffer is usually required to rapidly charge and discharge the ADC's input, which is often capacitive (see the section *Driving Capacitive Loads*). In addition, a high-speed ADC's input impedance often changes very rapidly during the conversion cycle, requiring an amplifier with very low output impedance at high frequencies to maintain measurement accuracy. The combination of high speed, fast slew rate, low noise, and a low and stable distortion over load makes the MAX4108/MAX4109/MAX4308/MAX4309 ideally suited for use as buffer amplifiers in high-speed ADC applications.

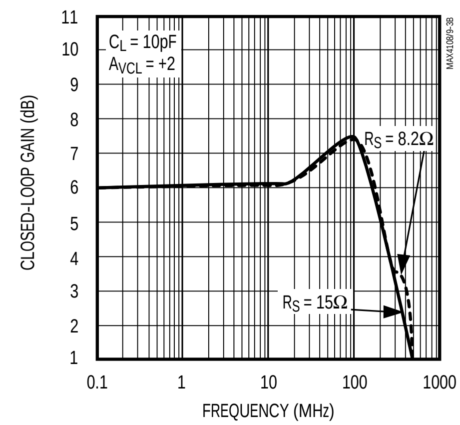


Figure 3b. MAX4308 Response vs. Capacitive Load with Resistive (R_S) Isolation (circuit shown in Figure 1a)

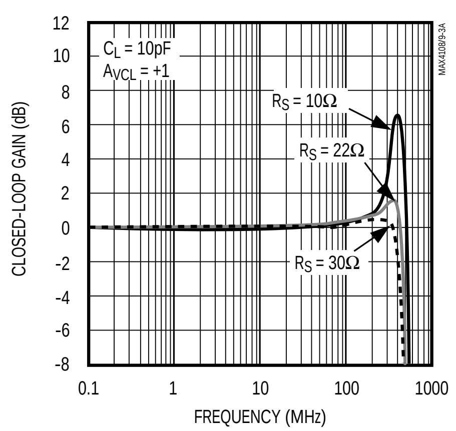


Figure 3a. MAX4108 Response vs. Capacitive Load with Resistive (R_S) Isolation (circuit shown in Figure 1a)

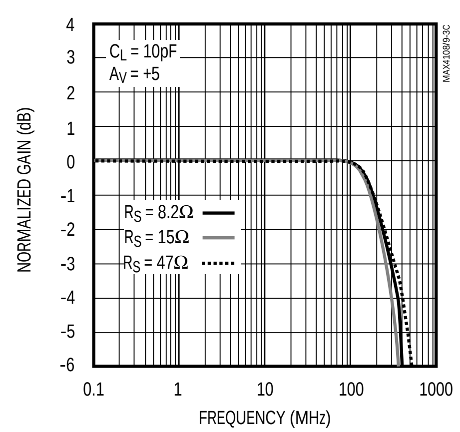


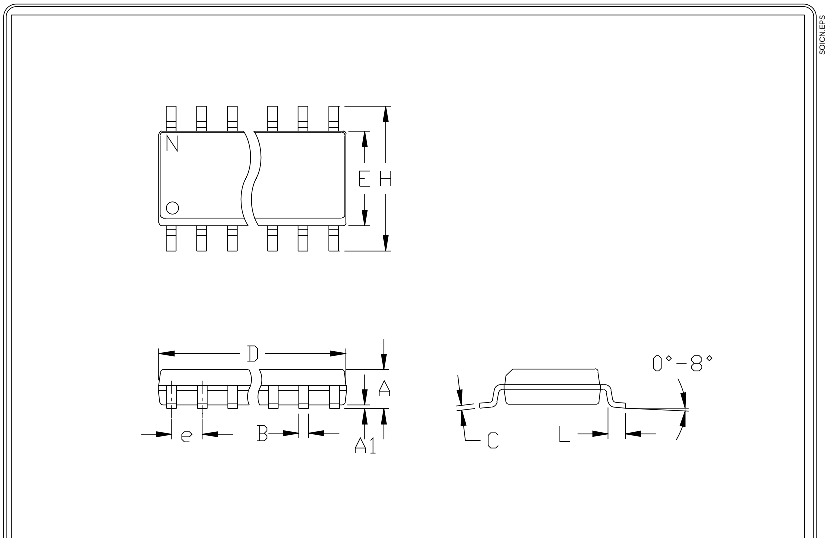
Figure 3c. MAX4108/MAX4309 Response vs. Capacitive Load with Resistive (R_S) Isolation (circuit shown in Figure 1a)

Chip Information

TRANSISTOR COUNT: 57

SUBSTRATE CONNECTED TO V_{EE}

Package Information



	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.053	0.069	1.35	1.75
A1	0.004	0.010	0.10	0.25
B	0.014	0.019	0.35	0.49
C	0.007	0.010	0.19	0.25
e	0.050		1.27	
E	0.150	0.157	3.80	4.00
H	0.228	0.244	5.80	6.20
h	0.010	0.020	0.25	0.50
L	0.016	0.050	0.40	1.27

	INCHES		MILLIMETERS			
	MIN	MAX	MIN	MAX	N	MS012
D	0.189	0.197	4.80	5.00	8	A
D	0.337	0.344	8.55	8.75	14	B
D	0.386	0.394	9.80	10.00	16	C

NOTES:
1. D&E DO NOT INCLUDE MOLD FLASH
2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED .15mm (.006")
3. LEADS TO BE COPLANAR WITHIN .102mm (.004")
4. CONTROLLING DIMENSION: MILLIMETER
5. MEETS JEDEC MS012-XX AS SHOWN IN ABOVE TABLE
6. N = NUMBER OF PINS

MAXIM 120 SAN GABRIEL DR. SUNNYVALE CA 94086 FAX (408) 737 7194 PROPRIETARY INFORMATION | TITLE: PACKAGE FAMILY OUTLINE: SOIC .150" | 1/1 | DOCUMENT CONTROL NUMBER: 21-0041 | REV: A

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