DAC1403D160

Dual 14 bits DAC, up to 160 MHz, 2 x interpolating

Rev. 02 — 14 August 2008

Product data sheet

1. General description

The DAC1403D160 is optimized to reduce architecture complexity and overall system cost. It leads to dynamic performance in multi-carrier support, because of its direct IF conversion capabilities. With an internal sampling rate of up to 160 MHz, DAC1403D160 is an extremely competitive solution for broadband wireless systems transmitters, as well as a wide range of applications.

2. Features

- Dual 14-bit resolution
- Spurious-Free Dynamic Range (SFDR) = 80 dBc at 2.5 MHz
- Input data rate up to 80 MHz
- 2 × interpolation filter
- Output data rate up to 160 MHz
- Single 3.3 V power supply
- Low noise capacitor-free integrated Phase-Locked Loup (PLL)
- Low power dissipation
- HTQFP80 package
- Ambient temperature from -40 °C to +85 °C

3. Applications

- Broadband wireless systems
- Digital radio links
- Cellular base stations
- Instrumentation
- Cable modem
- Cable Modem Termination System (CMTS)/Data Over Cable Service Interface Specification (DOCSIS)



Dual 14 bits DAC, up to 160 MHz, 2 x interpolating

Ordering information

Table 1. **Ordering information**

Type number	Package			
	Name	Description	Version	
DAC1403D160HW	HTQFP80	plastic thermal enhanced thin quad flat package; 80 leads; body $12 \times 12 \times 1$ mm; exposed die pad	SOT841-1	

Block diagram

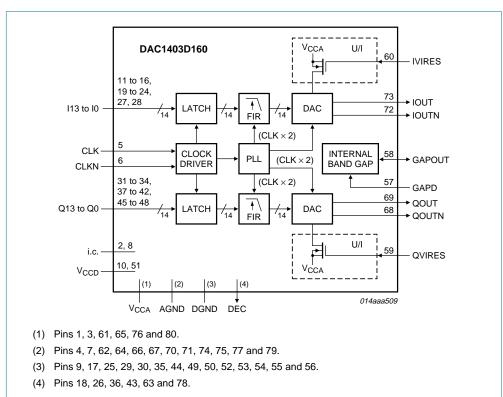
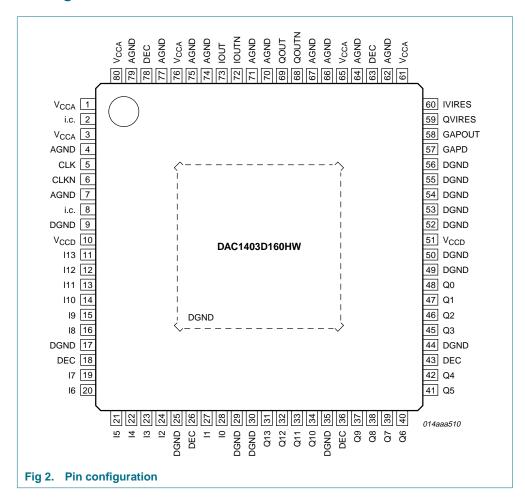


Fig 1. Block diagram

6. Pinning information

6.1 Pinning



6.2 Pin description

Table 2. Pin description

Tubic 2.	i ili descripti	OII	
Symbol	Pin	Type ^[1]	Description
V_{CCA}	1	S	analog supply voltage
i.c.	2	I/O	internally connected; leave open
V_{CCA}	3	S	analog supply voltage
AGND	4	G	analog ground
CLK	5	I	clock input
CLKN	6	I	complementary clock input
AGND	7	G	analog ground
i.c.	8	0	internally connected; leave open
DGND	9	G	digital ground

 Table 2.
 Pin description ...continued

Pin descriptioncontinued			
Pin	Type ^[1]	Description	
10	S	digital supply voltage	
11	I	I data input bit 13 (Most Significant Bit (MSB))	
12	I	I data input bit 12	
13	I	I data input bit 11	
14	I	I data input bit 10	
15	I	I data input bit 9	
16	I	I data input bit 8	
17	G	digital ground	
18	О	decoupling node	
19	I	I data input bit 7	
20	I	I data input bit 6	
21	I	I data input bit 5	
22	I	I data input bit 4	
23	I	I data input bit 3	
24	I	I data input bit 2	
25	G	digital ground	
26	0	decoupling node	
27	I	I data input bit 1	
28	I	I data input bit 0 (Least Significant Bit (LSB))	
29	G	digital ground	
30	G	digital ground	
31	I	Q data input bit 13 (MSB)	
32	I	Q data input bit 12	
33	I	Q data input bit 11	
34	I	Q data input bit 10	
35	G	digital ground	
36	0	decoupling node	
37	I	Q data input bit 9	
38	I	Q data input bit 8	
39	I	Q data input bit 7	
40	I	Q data input bit 6	
41	ı	Q data input bit 5	
42	I	Q data input bit 4	
43	0	decoupling node	
44	G	digital ground	
45	I	Q data input bit 3	
46	I	Q data input bit 2	
47	ı	Q data input bit 1	
48	ı	Q data input bit 0 (LSB)	
49	G	digital ground	
50	G	digital ground	
	Pin 10 11 12 13 14 15 16 17 18 19 20 21 22 23 24 25 26 27 28 29 30 31 32 29 30 31 32 33 34 35 36 37 38 39 40 41 42 43 44 45 46 47 48 49	Pin Type[1] 10 S 11 I 12 I 13 I 14 I 15 I 16 I 17 G 18 O 19 I 20 I 21 I 22 I 23 I 24 I 25 G 26 O 27 I 28 I 29 G 30 G 31 I 32 I 33 I 34 I 35 G 36 O 37 I 38 I 39 I 40 I 41 I 42 I 43 O <t< td=""></t<>	

Dual 14 bits DAC, up to 160 MHz, 2 x interpolating

Table 2. Pin description ... continued

I GOIG Z.	i ili ucaciipi		uou
Symbol	Pin	Type ^[1]	Description
V_{CCD}	51	S	digital supply voltage
DGND	52	G	digital ground
DGND	53	G	digital ground
DGND	54	G	digital ground
DGND	55	G	digital ground
DGND	56	G	digital ground
GAPD	57	I	internal band gap power disable input
GAPOUT	58	I/O	band gap output voltage
QVIRES	59	I	Q DAC biasing resistor
IVIRES	60	I	I DAC biasing resistor
V_{CCA}	61	S	analog supply voltage
AGND	62	G	analog ground
DEC	63	0	decoupling node
AGND	64	G	analog ground
V_{CCA}	65	S	analog supply voltage
AGND	66	G	analog ground
AGND	67	G	analog ground
QOUTN	68	0	complementary Q DAC output current
QOUT	69	0	Q DAC output current
AGND	70	G	analog ground
AGND	71	G	analog ground
IOUTN	72	0	complementary I DAC output current
IOUT	73	0	I DAC output current
AGND	74	G	analog ground
AGND	75	G	analog ground
V_{CCA}	76	S	analog supply voltage
AGND	77	G	analog ground
DEC	78	0	decoupling node
AGND	79	G	analog ground
V_{CCA}	80	S	analog supply voltage

^[1] Type description: S: Supply; G: Ground; I: Input; O: Output.

Functional description

The Digital-to-Analog Converter (DAC) is a segmented architecture composed of a 7-bit thermometer sub-DAC and the remaining 7-bit in a binary weighted sub-DAC.

The device produces two complementary current outputs on both channels, respectively pins IOUT/IOUTN and QOUT/QOUTN which need to be connected via a load resistor to the ground.

Figure 3 shows the equivalent analog output circuit of one DAC, which consists of a parallel combination of PMOS current sources and associated switches for each segment.

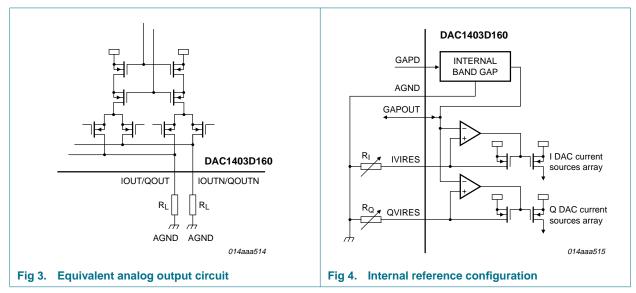
The cascade source configuration enables the increasing of the output impedance of the source and set to improve the dynamic performance of the DAC by introducing less distortion.

Figure 4 shows the internal reference configuration. In this case the bias current is given by the output of the internal regulator connected to the inverting input of the internal operational amplifiers, while external resistors R_I and R_Q are connected respectively to pins IVIRES and QVIRES. Thus the output current of the two DACs is typically fixed to 20 mA with an appropriate choice of these resistors. This configuration is optimal for temperature drift compensation because the band gap can be matched with the voltage on the feedback resistors.

The relation between full-scale output current $I_{O(fs)}$ and the R_I (R_Q) is:

$$R_{I} = \frac{2048 \times V_{GAPOUT}}{82 \times I_{O(fs)}} \Omega$$

The output current can also be adjusted by imposing an external reference voltage to the inverting input pin GAPOUT and disabling the internal band gap with pin GAPD set to HIGH. At a voltage lower than 1.2 V the current can be set at values lower than 20 mA. The input references at pins IVIRES and QVIRES may also be driven by separate reference voltages to adjust independently the two DAC currents.



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8. Limiting values

Table 3. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions		Min	Max	Unit
V _{CCD}	digital supply voltage		[1]	-0.3	+3.9	٧
V _{CCA}	analog supply voltage		[1]	-0.3	+3.9	V
ΔV_{CC}	supply voltage difference	V _{CCA} – V _{CCD}		-150	+150	mV
VI	input voltage	pins Qn and In referenced to DGND		-0.3	V _{CCD} + 0.3	V
		pins IVIRES, QVIRES and GAPD referenced to AGND		-0.3	V _{CCA} + 0.3	V
		pins CLK and CLKN referenced to AGND		-0.3	V _{CCA} + 0.3	V
Vo	output voltage	pins IOUT, IOUTN, QOUT and QOUTN referenced to AGND		-0.3	V _{CCA} + 0.3	V
T _{stg}	storage temperature			-55	+150	°C
T _{amb}	ambient temperature			-40	+85	°C
Tj	junction temperature			-	125	°C

^[1] All supplies are connected together.

9. Thermal characteristics

Table 4. Thermal characteristics

Symbol	Parameter	Conditions	Тур	Unit
$R_{th(j-a)}$	thermal resistance from junction to ambient	in free air	27.1	K/W
$R_{th(c-a)}$	thermal resistance from case to ambient	in free air	11.8	K/W

10. Characteristics

Table 5. Characteristics

 $V_{CCD} = V_{CCA} = 3.0 \text{ V}$ to 3.6 V; AGND and DGND connected together; $T_{amb} = -40 \,^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$; typical values measured at $V_{CCD} = V_{CCA} = 3.3 \text{ V}$, $I_{O(fs)} = 20 \text{ mA}$ and $T_{amb} = 25 \,^{\circ}\text{C}$; dynamic parameters measured using output schematic given in Figure 10; unless otherwise specified.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
Supplies							
V_{CCD}	digital supply voltage			3.0	3.3	3.6	V
V_{CCA}	analog supply voltage			3.0	3.3	3.6	V
I _{CCD}	digital supply current			-	55	65	mA
I _{CCA}	analog supply current			-	73	85	mA
P _{tot}	total power dissipation	$f_{Clk} = 80 \text{ MHz};$ $f_{IOUT} = f_{QOUT} = 5 \text{ MHz}$		-	422	540	mW
Clock inpu	ts (CLK and CLKN)						
$V_{I(cm)}$	common-mode input voltage			-	1.65	-	V
$V_{i(dif)(p\text{-}p)}$	peak-to-peak differential input voltage			-	1.0	-	V
Analog ou	tputs (IOUT, IOUTN, QOUT a	and QOUTN)					
I _{O(fs)}	full-scale output current	differential outputs		4	-	20	mA
R _o	output resistance		[1]	-	150	-	kΩ
Co	output capacitance		<u>[1]</u>	-	3	-	pF
Digital inp	uts (I0 to I13, Q0 to Q13 and	GAPD)					
V_{IL}	LOW-level input voltage			DGND	-	$0.3 V_{CCD}$	V
V_{IH}	HIGH-level input voltage			$0.7 V_{CCD}$	-	V_{CCD}	V
I _{IL}	LOW-level input current	$V_{IL} = 0.3 V_{CCD}$		-	5	-	μΑ
I_{IH}	HIGH-level input current	$V_{IH} = 0.7 V_{CCD}$		-	5	-	μΑ
Reference	voltage output (GAPOUT)						
V_{GAPOUT}	voltage on pin GAPOUT			-	1.31	-	V
I_{GAPOUT}	current on pin GAPOUT	external voltage		-	1	-	μΑ
ΔV_{GAPOUT}	voltage variation on pin GAPOUT			-	±133	-	ppm/°C
Clock timi	ng inputs (CLK and CLKN)						
f _{clk}	clock frequency			-	-	80	MHz
t _{w(clk)H}	HIGH clock pulse width			5	-	-	ns
t _{w(clk)L}	LOW clock pulse width			5	-	-	ns
Input timin	g (I0 to I13 and Q0 to Q13);	see Figure 5					
t _{h(i)}	input hold time			1.1	-	3.4	ns
$t_{su(i)}$	input set-up time			-1.5	-	+0.7	ns
Output tim	ing (IOUT, IOUTN, QOUT, Q	OutN)					
ts	settling time	$t_o = \pm 0.5 \text{ LSB}$	[1]	-	43	-	ns

 Table 5.
 Characteristics ...continued

 $V_{CCD} = V_{CCA} = 3.0 \text{ V}$ to 3.6 V; AGND and DGND connected together; $T_{amb} = -40 \,^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$; typical values measured at $V_{CCD} = V_{CCA} = 3.3 \text{ V}$, $I_{O(fs)} = 20 \text{ mA}$ and $T_{amb} = 25 \,^{\circ}\text{C}$; dynamic parameters measured using output schematic given in Figure 10; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Digital filt	er specification (FIR); order	N = 42; see <u>Figure 6</u> and <u>7</u> and <u>Ta</u>	ble 7			
f _{data}	data rate		-	-	80	MHz
$\alpha_{\text{ripple(pb)}}$	pass-band ripple	f_{data}/f_{clk} ; 0.005 dB attenuation	-	0.405	-	
Bp	power bandwidth	f _{data} /f _{clk} ; 3 dB attenuation	-	0.479	-	
α_{stpb}	stop-band attenuation	$f_{data}/f_{clk} = 0.6 \text{ to } 1$	-	69	-	dB
t _{d(grp)}	group delay time		-	11T _{CLK}	-	ns
Analog sig	gnal processing					
Linearity						
INL	integral non-linearity		-	±2.9	-	LSB
DNL	differential non-linearity		-	±1.5	-	LSB
$I_{n(o)}$	output noise current		-	120	-	pA/√Hz
E _{offset}	offset error	relative to full-scale	-	-0.3	-	%
E_G	gain error	relative to full-scale	-5.4	-	+5.4	%
ΔG_{IQ}	IQ gain mismatch	between I and Q, relative to full-scale	-	±0.2	-	%
Spurious f	ree dynamic range					
SFDR	spurious free dynamic	f _{clk} = 80 MHz; BW = Nyquist				
	range	f _o = 2.5 MHz at 0 dBFS	-	80	-	dBc
		f _o = 5 MHz at 0 dBFS	-	72	-	dBc
		f _o = 13 MHz at 0 dBFS	-	64	-	dBc
Harmonics	3					
$\alpha_{2\text{H}}$	second harmonic level	$f_0 = 5 \text{ MHz}$	-	73	-	dBc
		f _o = 13 MHz	-	65	-	dBc
α_{3H}	third harmonic level	f _o = 5 MHz	-	88	-	dBc
		f _o = 13 MHz	-	86	-	dBc
Two-tone i	ntermodulation					
IMD2	second-order intermodulation distortion	f_{clk} = 80 MHZ; f_o 1 = 10 MHz; f_o 2 = 12 MHz; BW = Nyquist	-	65	-	dBc
IMD3	third-order intermodulation distortion	$f_{clk} = 80 \text{ MHz}; f_0 1 = 10 \text{ MHz};$ $f_0 2 = 12 \text{ MHz}$	-	84	-	dBc
THD	total harmonic distortion	f _{clk} = 80 MHz; BW = Nyquist (T _{amb}	, = 25 °C)			
		f _o = 2.5 MHz	-	75	-	dBc
		f _o = 5 MHz	68	71	-	dBc
NSD	noise spectral density	f _{clk} = 80 MHz				
		f _o = 2.5 MHz	-	-155	-	dBm/Hz
		f _o = 5 MHz	-	-155	-	dBm/Hz
		f _o = 19 MHz	-	-153	-	dBm/Hz

Table 5. Characteristics ...continued

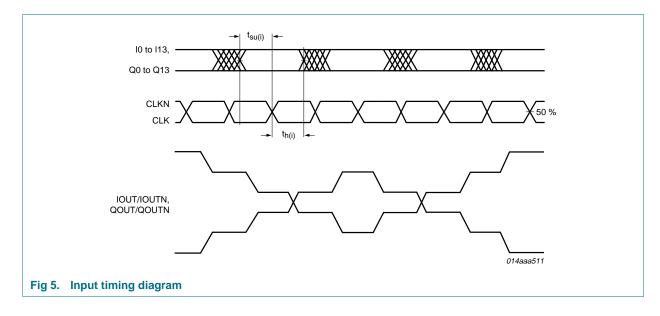
 $V_{CCD} = V_{CCA} = 3.0 \text{ V}$ to 3.6 V; AGND and DGND connected together; $T_{amb} = -40 \,^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$; typical values measured at $V_{CCD} = V_{CCA} = 3.3 \text{ V}$, $I_{O(fs)} = 20 \text{ mA}$ and $T_{amb} = 25 \,^{\circ}\text{C}$; dynamic parameters measured using output schematic given in Figure 10; unless otherwise specified.

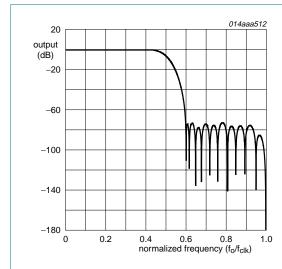
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Signal-to-n	oise ratio					
S/N	signal-to-noise ratio	f _{clk} = 80 MHz; BW = Nyquist				
		f _o = 2.5 MHz	-	80	-	dBc
		f _o = 5 MHz	70	80	-	dBc
		f _o = 19 MHz	-	78	-	dBc
ACPR	adjacent channel power	baseband; 5 MHz channel spacing	; BW = 3.84	MHz		
	ratio	f _o = 2.5 MHz	-	69	-	dBc
		f _o = 20 MHz	-	71	-	dBc

[1] Guaranteed by design.

Table 6. Band gap

Band gap disable (GAPD)	Band gap input/output (GAPOUT)	Internal band gap
LOW	output ($V_{GAPOUT} = 1.2 \text{ V}$)	enable
HIGH	input	disable





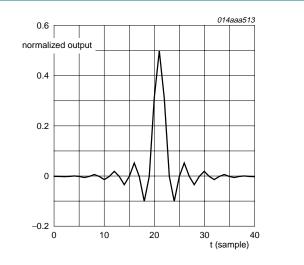


Fig 6. FIR filter frequency response

Fig 7. FIR filter impulse response

Table 7. Interpolation FIR filter coefficient

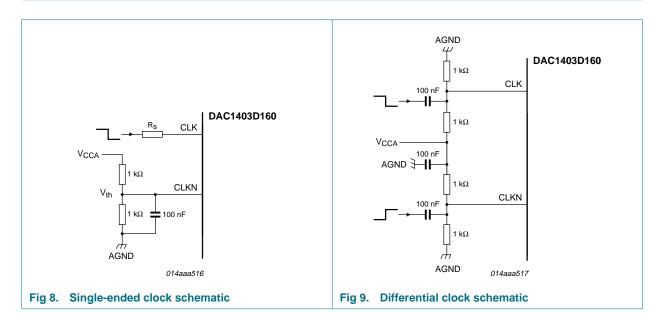
H(1) H(43) 10 H(2) H(42) 0 H(3) H(41) -31 H(4) H(40) 0 H(5) H(39) 69 H(6) H(38) 0 H(7) H(37) -138 H(8) H(36) 0 H(9) H(35) 248 H(10) H(34) 0 H(11) H(33) -419 H(12) H(32) 0 H(13) H(31) 678 H(14) H(30) 0 H(14) H(30) 0 H(15) H(29) -1083 H(16) H(28) 0 H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364 H(22) - 16384	Coefficient	Coefficient	Value
H(3) H(41) -31 H(4) H(40) 0 H(5) H(39) 69 H(6) H(38) 0 H(7) H(37) -138 H(8) H(36) 0 H(9) H(35) 248 H(10) H(34) 0 H(11) H(33) -419 H(12) H(32) 0 H(13) H(31) 678 H(14) H(30) 0 H(14) H(30) 0 H(15) H(29) -1083 H(16) H(28) 0 H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(1)	H(43)	10
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H(5) H(39) 69 H(6) H(38) 0 H(7) H(37) -138 H(8) H(36) 0 H(9) H(35) 248 H(10) H(34) 0 H(11) H(33) -419 H(12) H(32) 0 H(13) H(31) 678 H(14) H(30) 0 H(15) H(29) -1083 H(16) H(28) 0 H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(3)	H(41)	-31
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H(8) H(36) 0 H(9) H(35) 248 H(10) H(34) 0 H(11) H(33) -419 H(12) H(32) 0 H(13) H(31) 678 H(14) H(30) 0 H(15) H(29) -1083 H(16) H(28) 0 H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(6)	H(38)	0
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H(12) H(32) 0 H(13) H(31) 678 H(14) H(30) 0 H(15) H(29) -1083 H(16) H(28) 0 H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(10)	H(34)	0
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H(15) H(29) -1083 H(16) H(28) 0 H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(13)	H(31)	678
H(16) H(28) 0 H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(14)	H(30)	0
H(17) H(27) 1776 H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(15)	H(29)	-1083
H(18) H(26) 0 H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(16)	H(28)	0
H(19) H(25) -3282 H(20) H(24) 0 H(21) H(23) 10364	H(17)	H(27)	1776
H(20) H(24) 0 H(21) H(23) 10364	H(18)	H(26)	0
H(21) H(23) 10364	H(19)	H(25)	-3282
	H(20)	H(24)	0
H(22) - 16384	H(21)	H(23)	10364
	H(22)	-	16384

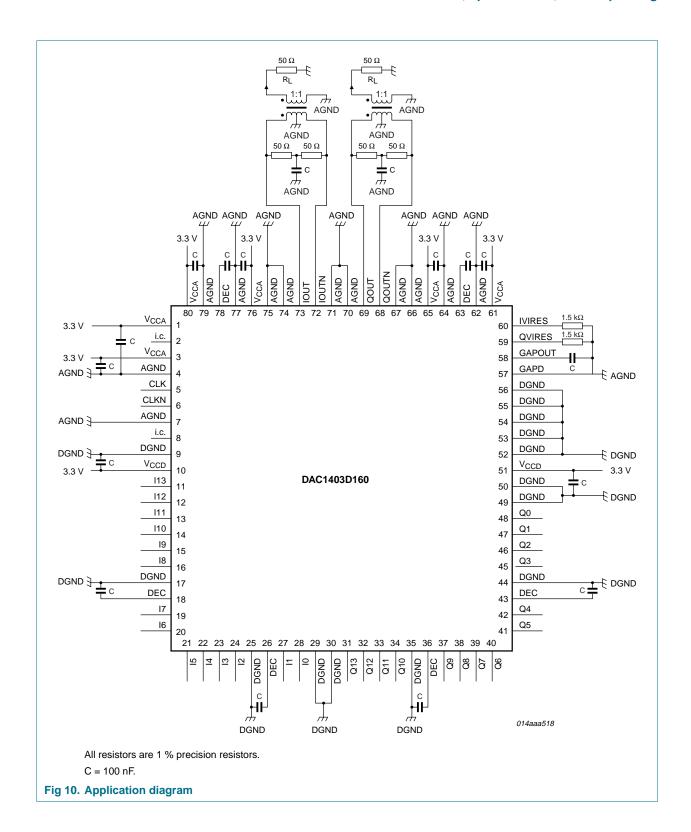
Product data sheet

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Dual 14 bits DAC, up to 160 MHz, 2 x interpolating

11. Application information





Dual 14 bits DAC, up to 160 MHz, 2 x interpolating

11.1 Alternative parts

The following alternative parts are also available:

Table 8. **Alternative parts**

Type number	Description		Sampling frequency
DAC1203D160	Dual 12 bits DAC, with 2 × interpolating	[1]	160 MHz
DAC1003D160	Dual 10 bits DAC, with 2 × interpolating	[1]	160 MHz

^[1] Pin to pin compatible

12. Package outline

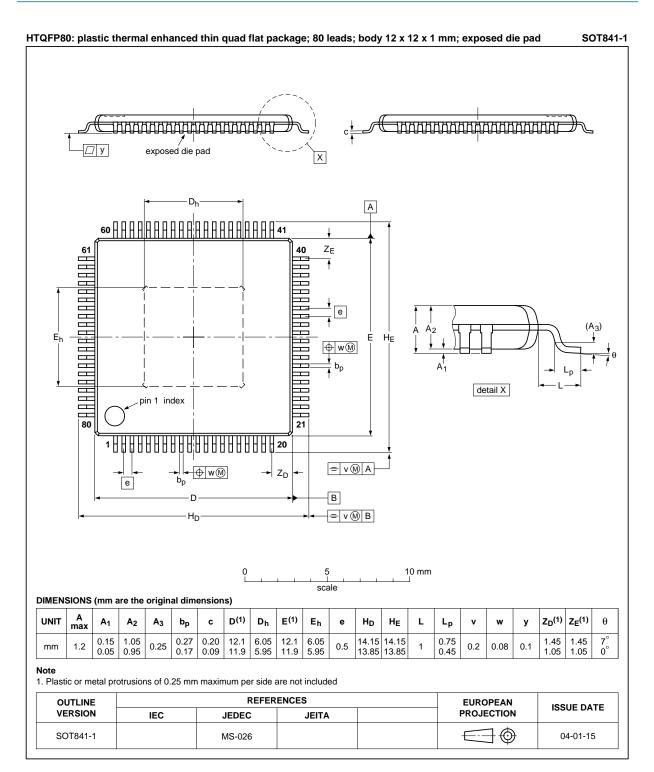


Fig 11. Package outline SOT841-1 (HTQFP80)

13. Abbreviations

Table 9. Abbreviations

Acronym	Description
BW	BandWidth
FIR	Finite Impulse Response
IF	Intermediate Frequency
LSB	Least Significant Bit
MSB	Most Significant Bit
PLL	Phase-Locked Loop
PMOS	Positive-Metal Oxide Semiconductor

14. Glossary

14.1 Static parameters

DNL — Differential Non-Linearity. The difference between the ideal and the measured output value between successive DAC codes.

INL — Integral Non-Linearity. The deviation of the transfer function from a best-fit straight line (linear regression computation).

14.2 Dynamic parameters

IMD2 — Second-order intermodulation distortion. From a dual-tone digital input sine wave (these two frequencies are close together), the intermodulation distortion product IMD2 is the ratio of the RMS value of either tone and the RMS value of the worst 2nd-order intermodulation product.

IMD3 — Third-order intermodulation distortion. From a dual-tone digital input sine wave (these two frequencies are close together), the intermodulation distortion product IMD3 is the ratio of the RMS value of either tone and the RMS value of the worst 3rd-order intermodulation product.

S/N — Signal-to-Noise ratio. The ratio of the RMS value of the reconstructed output sine wave to the RMS value of the noise excluding the harmonics and the DC component.

THD — Total Harmonic Distortion. The ratio of the RMS value of the harmonics of the output frequency to the RMS value of the output sine wave. Usually, the calculation of THD is done on the first 5 harmonics.

SFDR — Spurious-Free Dynamic Range. The ratio of the RMS value of the reconstructed output sine wave and the RMS value of the largest spurious (harmonic and non-harmonic, excluding DC component) observed in the frequency domain.

15. Revision history

Table 10. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes	
DAC1403D160_2	20080814	Product data sheet	-	DAC1403D160_1	
Modifications:	 Corrections made to row Q0 in <u>Table 2</u>. Added condition & correction of value to t_s in <u>Table 5</u>. Additional definition in <u>Section 14</u>. 				
DAC1403D160_1	20080616	Product data sheet	-	-	

16. Legal information

16.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
- [3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL http://www.nxp.com.

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