



General Description

The MAX5391/MAX5393 dual 256-tap, volatile, low-voltage linear taper digital potentiometers offer three end-to-end resistance values of $10k\Omega$, $50k\Omega$, and $100k\Omega$. Operating from a single +1.7V to +5.5V power supply, these devices provide a low 35ppm/°C end-to-end temperature coefficient. The devices feature an SPITM interface.

The small package size, low supply voltage, low supply current, and automotive temperature range of the MAX5391/MAX5393 make the devices uniquely suitable for the portable consumer market, battery backup industrial applications, and the automotive market.

The MAX5391/MAX5393 include two digital potentiometers in a voltage-divider configuration. The MAX5391/MAX5393 are specified over the -40°C to +125°C automotive temperature range and are available in a 16-pin, 3mm x 3mm TQFN and a 14-pin TSSOP package, respectively.

Applications

Low-Voltage Battery Applications
Portable Electronics
Mechanical Potentiometer Replacement
Offset and Gain Control
Adjustable Voltage References/Linear Regulators
Automotive Electronics

Features

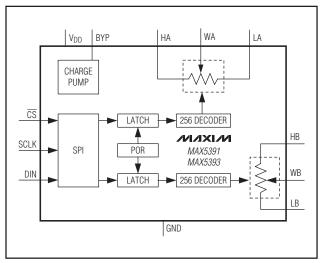
- ♦ Dual 256-Tap Linear Taper Positions
- ♦ Single +1.7V to +5.5V Supply Operation
- ♦ Low 12μA Quiescent Supply Current
- 10kΩ, 50kΩ, and 100kΩ End-to-End Resistance Values
- ◆ SPI-Compatible Interface
- ♦ Wiper Set to Midscale on Power-Up
- ◆ -40°C to +125°C Operating Temperature Range

Ordering Information

PART	PIN-PACKAGE	$\begin{array}{c} \textbf{END-TO-END} \\ \textbf{RESISTANCE} \ (\textbf{k}\Omega) \end{array}$				
MAX5391LATE+	16 TQFN-EP*	10				
MAX5391MATE+	16 TQFN-EP*	50				
MAX5391NATE+	16 TQFN-EP*	100				
MAX5393LAUD+	14 TSSOP	10				
MAX5393MAUD+	14 TSSOP	50				
MAX5393NAUD+	14 TSSOP	100				

Note: All devices are specified in the -40°C to +125°C temperature range.

Functional Diagram



SPI is a trademark of Motorola, Inc.

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For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

⁺Denotes a lead(Pb)-free/RoHS-compliant package

^{*}EP = Exposed pad.

ABSOLUTE MAXIMUM RATINGS

V _{DD} to GND0.3V to +6V	Continuous Power Dissipation (T _A = +70°C)
H_, W_, L_ to GND0.3V to the lower of	14-Pin TSSOP (derate 10mW/°C above +70°C)796.8mW
(V _{DD} + 0.3V) or +6V	16-Pin TQFN (derate 14.7mW/°C above +70°C)1176.5mW
All Other Pins to GND0.3V to +6V Continuous Current into H_, W_, and L_ MAX5391L/MAX5393L±5mA MAX5391M/MAX5393M±2mA MAX5391N/MAX5393N±1mA	Operating Temperature Range40°C to +125°C Junction Temperature+150°C Storage Temperature Range65°C to +150°C Lead Temperature (soldering, 10s)+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{DD} = +1.7V \text{ to } +5.5V, V_{H_} = V_{DD}, V_{L_} = GND, T_{A} = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{DD} = +1.8V, T_{A} = +25^{\circ}C.) \text{ (Note 1)}$

PARAMETER	SYMBOL		MIN	TYP	MAX	UNITS		
Resolution	N			256			Тар	
DC PERFORMANCE (Voltage-D	ivider Mode)							
Integral Nonlinearity	INL	(Note 2)		-0.5		+0.5	LSB	
Differential Nonlinearity	DNL	(Note 2)		-0.5		+0.5	LSB	
Dual-Code Matching		Register A = R	legister B	-0.5		+0.5	LSB	
Ratiometric Resistor Tempco		$(\Delta V_W/V_W)/\Delta T$,	no load		5		ppm/°C	
			MAX5391L/MAX5393L	-3	-2.2			
Full-Scale Error		Code = FFh	MAX5391M/MAX5393M	-1	-0.6		LSB	
			MAX5391N/MAX5393N	-0.5	-0.3			
			MAX5391L/MAX5393L		2.2	3		
Zero-Scale Error		Code = 00h	MAX5391M/MAX5393M		0.6	1	LSB	
			MAX5391N/MAX5393N		0.3	0.5		
DC PERFORMANCE (Variable F	Resistor Mode	e)						
Integral Nonlinearity	R-INL	(Note 3)		-1.0		+1.5	LSB	
Differential Nonlinearity	R-DNL	(Note 3)		-0.5		+0.5	LSB	
DC PERFORMANCE (Resistor C	haracteristic	s)						
Wiper Resistance	RwL	(Note 4)				200	Ω	
Terminal Capacitance	CH_, CL_	Measured to GND			10		pF	
Wiper Capacitance	Cw_	Measured to GND			50		pF	
End-to-End Resistor Tempco	TCR	No load			35		ppm/°C	
End-to-End Resistor Tolerance	ΔRHL	Wiper not connected		-25		+25	%	
AC PERFORMANCE								
Crosstalk		(Note 5)			-90		dB	
-3dB Bandwidth		Code = 08H,	MAX5391L/MAX5393L		600			
	BW	10pF load,	MAX5391M/MAX5393M		100		kHz	
		$V_{DD} = 1.8V$	MAX5391N/MAX5393N	50				
Total Harmonic Distortion Plus Noise	THD+N	Measured at W, VH_ = 1VRMS at 1kHz			0.02		%	
		MAX5391L/MA	MAX5391L/MAX5393L		400			
Wiper Settling Time (Note 6)	ts	MAX5391M/M/		1200		ns		
		MAX5391N/MA	MAX5391N/MAX5393N					

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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +1.7V \text{ to } +5.5V, V_{H_{-}} = V_{DD}, V_{L_{-}} = GND, T_{A} = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{DD} = +1.8V, T_{A} = +25^{\circ}C.) \text{ (Note 1)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Charge-Pump Feedthrough at W_	VRW	fCLK = 600kHz, COUT = 0nF		200		nV _{P-P}	
POWER SUPPLIES							
Supply Voltage Range	V _{DD}		1.7		5.5	V	
Chan allow Course at		V _{DD} = 5.5V	27				
Standby Current		$V_{DD} = 1.7V$	12			μΑ	
DIGITAL INPUTS							
Minimum Instant High Valle on	\/	$V_{DD} = 2.6V \text{ to } 5.5V$	70			0/ \/	
Minimum Input High Voltage	VIH	V _{DD} = 1.7V to 2.6V	75			% x V _{DD}	
Maximum langut lang Valtage	\ /	V _{DD} = 2.6V to 5.5V			30	% x V _{DD}	
Maximum Input Low Voltage	VIL	$V_{DD} = 1.7V \text{ to } 2.6V$			25		
Input Leakage Current			-1		+1	μΑ	
Input Capacitance				5		pF	
TIMING CHARACTERISTICS—SF	I (Note 7)						
SCLK Frequency	fMAX				10	MHz	
SCLK Clock Period	tCP		100			ns	
SCLK Pulse-Width High	tCH		40			ns	
SCLK Pulse-Width Low	tCL		40			ns	
CS Fall to SCK Rise Setup Time	tcss		40			ns	
SCLK Rise to CS Rise Hold Time	tcsh		0			ns	
DIN Setup Time	tDS		40			ns	
DIN Hold Time	tDH		0			ns	
SCLK Rise to CS Fall Delay	tcso		10			ns	
SCLK Rise to SCLK Rise Hold Time	tCS1		40			ns	
CS Pulse-Width High	tcsw		100			ns	

- **Note 1:** All devices are 100% production tested at T_A = +25°C. Specifications over temperature limits are guaranteed by design and characterization.
- Note 2: DNL and INL are measured with the potentiometer configured as a voltage-divider (Figure 1) with H_ = V_{DD} and L_ = GND. The wiper terminal is unloaded and measured with a high-input-impedance voltmeter.
- Note 3: R-DNL and R-INL are measured with the potentiometer configured as a variable resistor (Figure 1). DNL and INL are measured with the potentiometer configured as a variable resistor. H_ is unconnected and L_ = GND. For V_{DD} = +5V, the wiper terminal is driven with a source current of 400μA for the 10kΩ configuration, 80μA for the 50kΩ configuration, and 40μA for the 100kΩ configuration. For V_{DD} = +1.7V, the wiper terminal is driven with a source current of 150μA for the 10kΩ configuration, 30μA for the 50kΩ configuration, and 15μA for the 100kΩ configuration.
- Note 4: The wiper resistance is the value measured by injecting the currents given in Note 3 into W_ with L_ = GND.

 RW = (VW VH)/IW.
- Note 5: Drive HA with a 1kHz GND to V_{DD} amplitude tone. LA = LB = GND. No load. WB is at midscale with a 10pF load. Measure WB.
- **Note 6:** The wiper-settling time is the worst-case 0 to 50% rise time, measured between tap 0 and tap 127. H_ = V_{DD}, L_ = GND, and the wiper terminal is loaded with 10pF capacitance to ground.
- Note 7: Digital timing is guaranteed by design and characterization, not production tested.



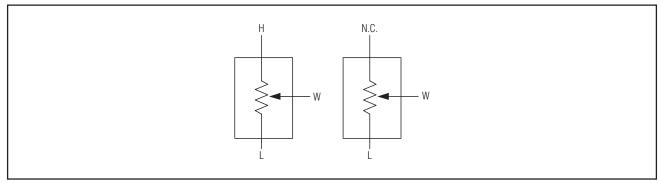
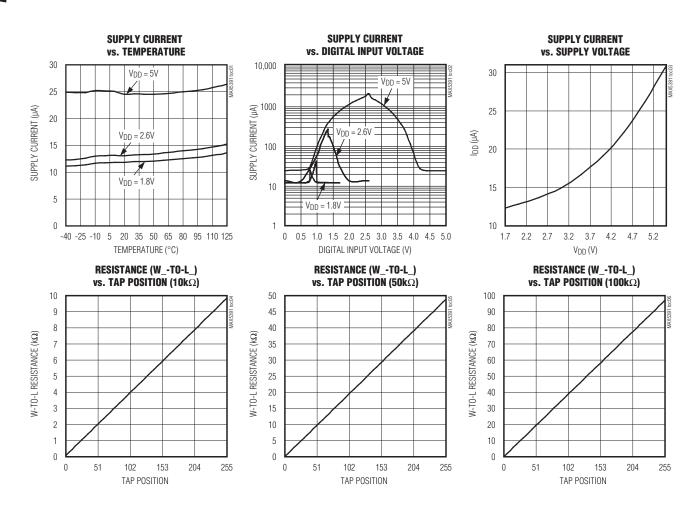


Figure 1. Voltage-Divider and Variable Resistor Configurations

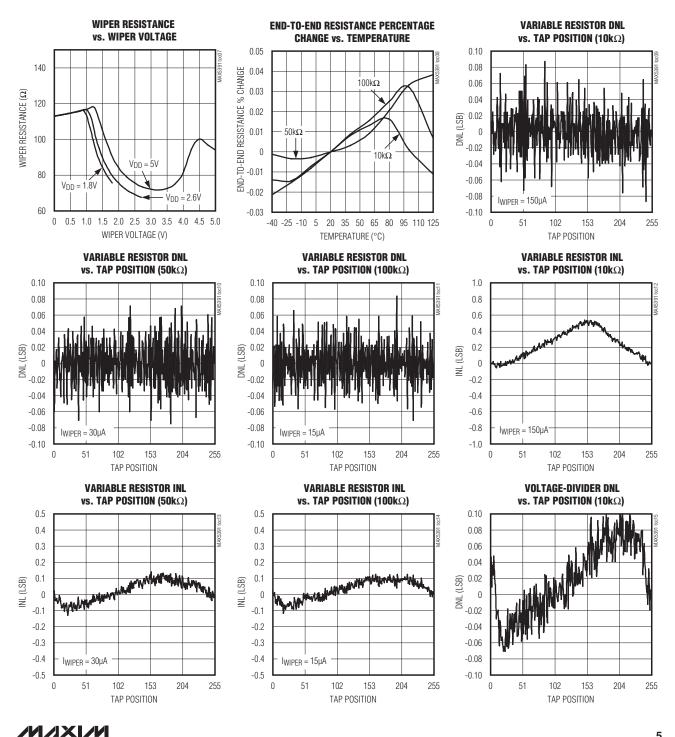
Typical Operating Characteristics

 $(V_{DD} = 1.8V, T_A = +25^{\circ}C, unless otherwise noted.)$



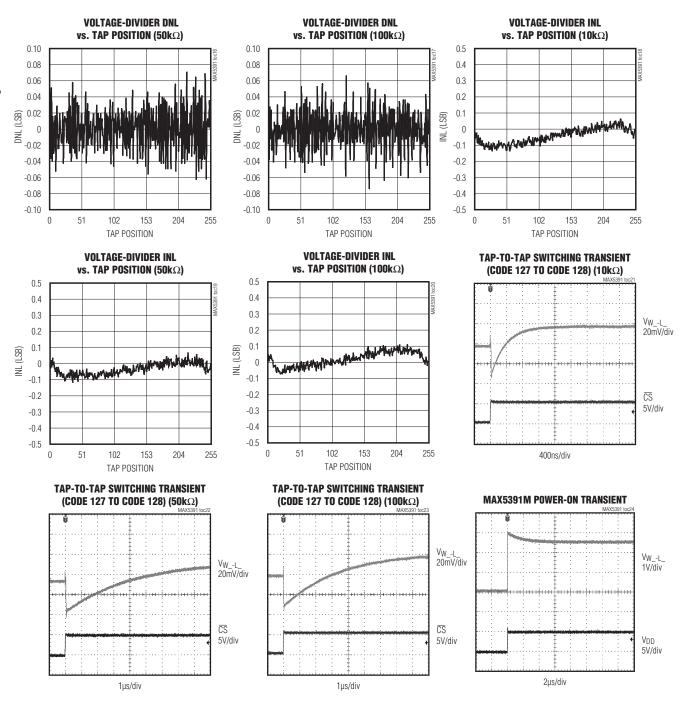
Typical Operating Characteristics (continued)

 $(V_{DD} = 1.8V, T_A = +25^{\circ}C, unless otherwise noted.)$



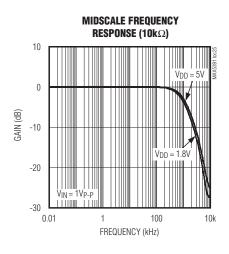
Typical Operating Characteristics (continued)

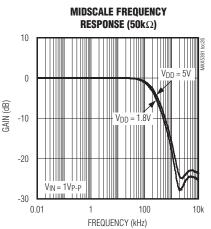
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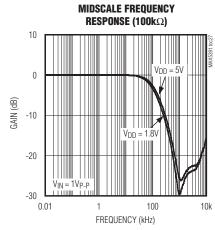


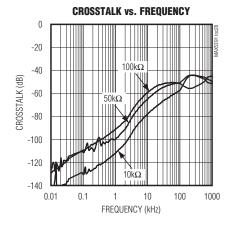
Typical Operating Characteristics (continued)

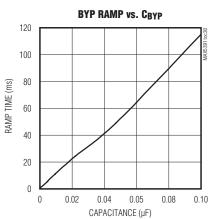
 $(V_{DD} = 1.8V, T_A = +25^{\circ}C, unless otherwise noted.)$

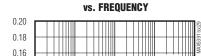




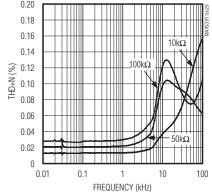


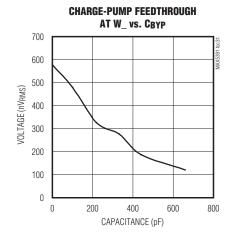




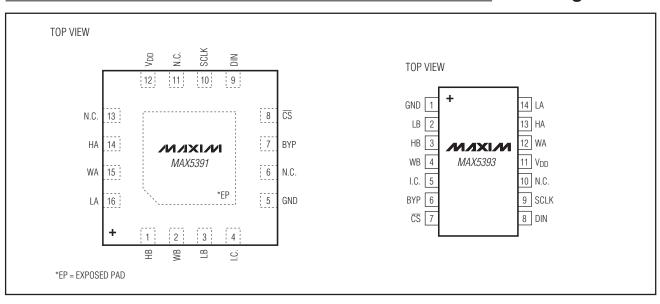


TOTAL HARMONIC DISTORTION PLUS NOISE





Pin Configurations



Pin Description

PII	N		
MAX5391 (TQFN-EP)	MAX5393 (TSSOP)	NAME	FUNCTION
1	3	НВ	Resistor B High Terminal. The voltage at HB can be higher or lower than the voltage at LB. Current can flow into or out of HB.
2	4	WB	Resistor B Wiper Terminal
3	2	LB	Resistor B Low Terminal. The voltage at LB can be higher or lower than the voltage at HB. Current can flow into or out of LB.
4	5	I.C.	Internally Connected. Connect to GND.
5	1	GND	Ground
6, 11, 13	10	N.C.	No Connection. Not internally connected.
7	6	BYP	Internal Power-Supply Bypass. For additional charge-pump filtering, bypass to GND with a capacitor close to the device.
8	7	CS	Active-Low Chip-Select Input
9	8	DIN	Serial-Interface Data Input
10	9	SCLK	Serial-Interface Clock Input
12	11	VDD	Power-Supply Input. Bypass VDD to GND with a 0.1µF capacitor close to the device.
14	13	НА	Resistor A High Terminal. The voltage at HA can be higher or lower than the voltage at LA. Current can flow into or out of HA.
15	12	WA	Resistor A Wiper Terminal
16	14	LA	Resistor A Low Terminal. The voltage at LA can be higher or lower than the voltage at HA. Current can flow into or out of LA.
_	_	EP	Exposed Pad (MAX5391 Only). Connect to GND.

Detailed Description

The MAX5391/MAX5393 dual 256-tap, volatile, low-voltage linear taper digital potentiometers offer three end-to-end resistance values of $10k\Omega,\,50k\Omega,\,$ and $100k\Omega.$ Each potentiometer consists of 255 fixed resistors in series between terminals H_ and L_. The potentiometer wiper, W_, is programmable to access any one of the 256 tap points on the resistor string.

The potentiometers in each device are programmable independently of each other. The MAX5391/MAX5393 feature an SPI interface.

Charge Pump

The MAX5391/MAX5393 contain an internal charge pump that guarantees the maximum wiper resistance, RWL, to be less then 200Ω for supply voltages down to 1.7V. Pins H_, W_, and L_ are still required to be less than VDD + 0.3V. A bypass input, BYP, is provided to allow additional filtering of the charge-pump output, further reducing clock feed through that may occur on H_, W_, or L_. The nominal clock rate of the charge pump is 600kHz. BYP should remain resistively unloaded as any additional load would produce a ripple of approximately IBYP/(600kHz x CBYP) volts. See the Charge-Pump Feedthrough at W_ vs. CBYP graph in the *Typical Operating Characteristics* for CBYP sizing guidelines with respect to clock feedthrough to the wiper. The value of

CBYP does affect the startup time of the charge pump; however, CBYP does not impact the ability to communicate with the device, nor is there a minimum CBYP requirement. The maximum wiper impedance specification is not guaranteed until the charge pump is fully settled. See the BYP Ramp vs. CBYP graph in the *Typical Operating Characteristics* for CBYP impact on charge-pump settling time.

SPI Digital Interface

The MAX5391/MAX5393 include a SPI interface that provides a 3-wire write-only serial-data interface to control the wiper tap position through inputs chip select $(\overline{\text{CS}})$, data in (DIN), and data clock (SCLK). Drive $\overline{\text{CS}}$ low to load data from DIN synchronously into the serial shift register on the rising edge of each SCLK pulse. The MAX5391/MAX5393 load the last 10 bits of clocked data into the appropriate potentiometer control register once $\overline{\text{CS}}$ transitions high. See Figures 2 and 3. Data written to a memory register immediately updates the wiper position. Keep $\overline{\text{CS}}$ low during the entire data stream to prevent the data from being terminated.

The first two bits A1:A0 (address bits) address one of the two potentiometers. See Table 1. The power-on reset (POR) circuitry sets the wiper to midscale.

Table 1. SPI Register Map

Bit Number	1	2	3	4	5	6	7	8	9	10
Bit Name	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write Wiper Register A	0	0	D7	D6	D5	D4	D3	D2	D1	D0
Write Wiper Register B	0	1	D7	D6	D5	D4	D3	D2	D1	D0
Write to Both A and B	1	1	D7	D6	D5	D4	D3	D2	D1	D0

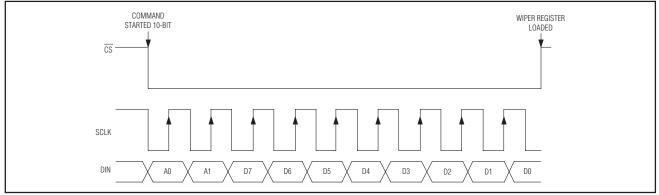


Figure 2. SPI Digital Interface Format

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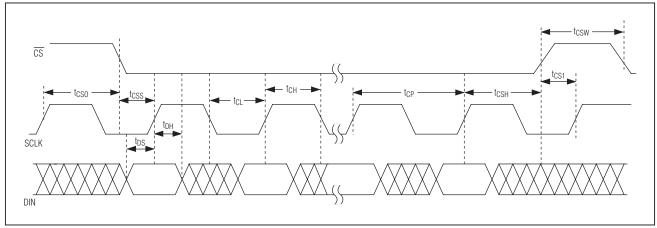


Figure 3. SPI Timing Diagram

REG A: The data byte writes to register A, and the wiper of potentiometer A moves to the appropriate position at the rising edge of \overline{CS} . D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wiper to the position closest to LA. D[7:0] = FFh moves the wiper closest to HA. D[7:0] is 80h following power-on.

REG B: The data byte writes to register B, and the wiper of potentiometer B moves to the appropriate position at the rising edge of \overline{CS} . D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wiper to the position closest to LB. D[7:0] = FFh moves the wiper to the position closest to HB. D[7:0] is 80h following power-on.

REG A and B: The data byte writes to registers A and B, and the wipers of potentiometers A and B move to the appropriate position. D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wiper to the position closest to L_. D[7:0] = FFh moves the wiper to the position closest to H_. D[7:0] is 80h following power-on.

Applications Information

Variable Gain Amplifier

Figure 4 shows a potentiometer adjusting the gain of a noninverting amplifier. Figure 5 shows a potentiometer adjusting the gain of an inverting amplifier.

Adjustable Dual Regulator

Figure 6 shows an adjustable dual linear regulator using a dual potentiometer as two variable resistors.

Adjustable Voltage Reference

Figure 7 shows an adjustable voltage reference circuit using a potentiometer as a voltage divider.

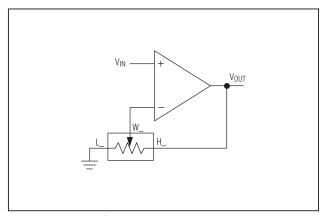


Figure 4. Variable-Gain Noninverting Amplifier

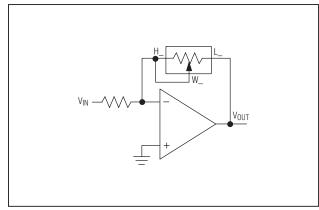


Figure 5. Variable-Gain Inverting Amplifier

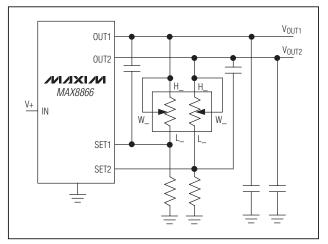


Figure 6. Adjustable Dual Linear Regulator

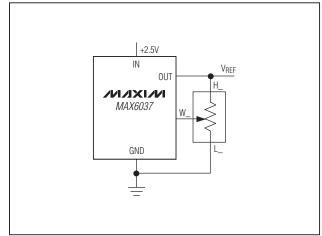


Figure 7. Adjustable Voltage Reference

Variable-Gain Current-to-Voltage Converter

Figure 8 shows a variable-gain current-to-voltage converter using a potentiometer as a variable resistor.

LCD Bias Control

Figure 9 shows a positive LCD bias control circuit using a potentiometer as a voltage-divider.

Programmable Filter

Figure 10 shows a programmable filter using a dual potentiometer.

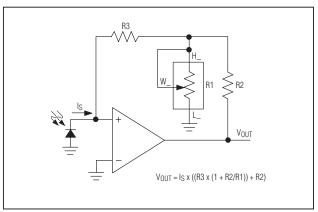


Figure 8. Variable Gain I-to-V Converter

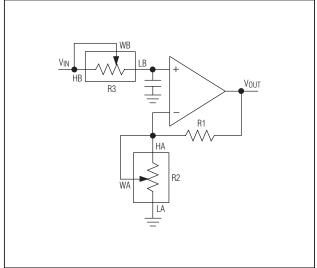


Figure 10. Programmable Filter

Offset Voltage Adjustment Circuit

Figure 11 shows an offset voltage adjustment circuit using a dual potentiometer.

Process Information

PROCESS: BICMOS

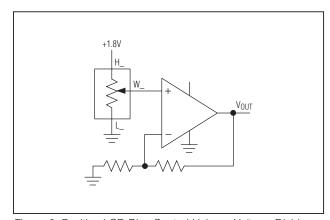


Figure 9. Positive LCD Bias Control Using a Voltage-Divider

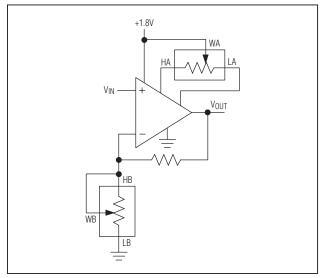


Figure 11. Offset Voltage Adjustment Circuit

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Package Information

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
14 TSSOP	U14+1	21-0066
16 TQFN-EP	T1633+5	21-0136

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