



MOTOROLA

MPQ6001,N MPQ6002,N MPQ6501,N MPQ6502,N

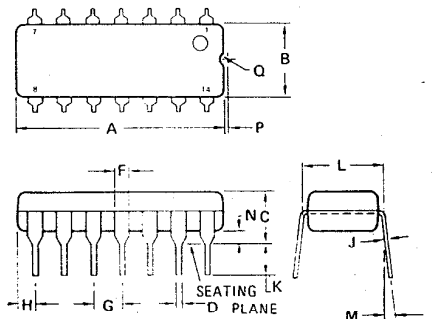
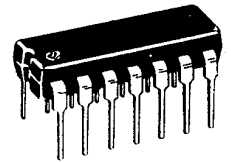
QUAD DUAL-IN-LINE SILICON ANNULAR[♦] COMPLEMENTARY PAIR TRANSISTORS

... designed for high-speed switching circuits, DC to VHF amplifier applications and complementary circuitry.

- Maximum Power Dissipation @ $T_A = 25^\circ\text{C}$
 $P_D = 1.25$ Watts – MPQ6001,MPQ6002,MPQ6501,MPQ6502
 $= 0.9$ Watts – MPQ6001N,MPQ6002N,MPQ6501N,MPQ6502N
- DC Current Gain Specified – 1.0 to 300 mAdc
- High Current-Gain-Bandwidth Product –
 $f_T = 400$ MHz (Typ) @ $I_C = 50$ mAdc
- NPN Transistor Similar to 2N2218 or 2N2219
- PNP Transistor Similar to 2N2904 or 2N2905
- Compact Size Compatible with IC Automatic Insertion Equipment

QUAD DUAL-IN-LINE SILICON COMPLEMENTARY PAIR TRANSISTORS

CASE 646
PLASTIC PACKAGE



DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	18.16	18.80	0.715	0.740
B	6.10	6.60	0.240	0.260
C	4.06	4.57	0.160	0.180
D	0.38	0.51	0.015	0.020
F	1.02	1.52	0.040	0.060
G	2.54 BSC		0.100 BSC	
H	1.32	1.83	0.052	0.072
J	0.20	0.30	0.008	0.012
K	2.92	3.43	0.115	0.135
L	7.37	7.87	0.290	0.310
M	— 10 ⁰		— 10 ⁰	
N	0.51	1.02	0.020	0.040
P	0.13	0.38	0.005	0.015
Q	0.51	0.76	0.020	0.030

NOTES:

- LEADS WITHIN 0.13 mm (0.005) RADIUS OF TRUE POSITION AT SEATING PLANE AT MAXIMUM MATERIAL CONDITION.
- DIMENSION "L" TO CENTER OF LEADS WHEN FORMED PARALLEL

MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Collector-Emitter Voltage	V_{CEO}	30	Vdc
Collector-Base Voltage	V_{CB}	60	Vdc
Emitter-Base Voltage	V_{EB}	5.0	Vdc
Collector Current – Continuous	I_C	500	mAdc
		Each Transistor	Four Transistors Equal Power
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ (1) MPQ6001, MPQ6002, MPQ6501, MPQ6502 MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N Derate above 25°C	P_D	0.65 0.5	1.25 0.9 mW/ $^\circ\text{C}$
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ MPQ6001, MPQ6002, MPQ6501, MPQ6502 MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N Derate above 25°C	P_D	5.18 4.0	10 7.2 mW/ $^\circ\text{C}$
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ MPQ6001, MPQ6002, MPQ6501, MPQ6502 MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N Derate above 25°C	P_D	1.0 0.825	3.0 2.4 mW/ $^\circ\text{C}$
MPQ6001, MPQ6002, MPQ6501, MPQ6502 MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N		8.0 6.7	24 19.2 mW/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	T_J, T_{stg}	-55 to +150	$^\circ\text{C}$

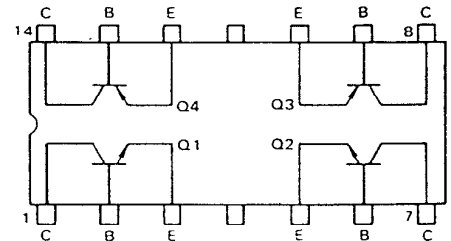
(1) Second Breakdown occurs at power levels greater than 3 times the power dissipation rating.

THERMAL CHARACTERISTICS

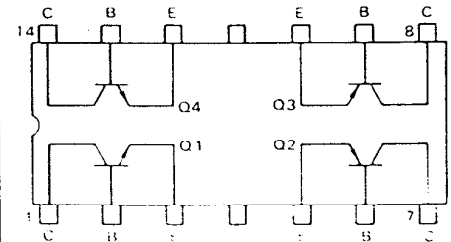
Characteristic	Junction to Case	Junction to Ambient	Unit
Thermal Resistance			$^\circ\text{C}/\text{W}$
Each Die			
MPQ6001, MPQ6002, MPQ6501, MPQ6502	125	193	
MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N	151	250	
Effective, 4 Die			
MPQ6001, MPQ6002, MPQ6501, MPQ6502	41.6	100	
MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N	52	134	
Coupling Factors			%
Q1-Q4 or Q2-Q3	MPQ6001, MPQ6002 MPQ6501, MPQ6502	30	60
	MPQ6001N, MPQ6002N MPQ6501N, MPQ6502N	34	70
Q1-Q2 or Q3-Q4	MPQ6001, MPQ6002 MPQ6501, MPQ6502	2.0	24
	MPQ6001N, MPQ6002N MPQ6501N, MPQ6502N	2.0	26

CONNECTION DIAGRAM

MPQ6001,N, MPQ6002,N,



MPQ6501,N, MPQ6502,N



[♦] Annular Semiconductors Patented by Motorola Inc.

THERMAL COUPLING AND EFFECTIVE THERMAL RESISTANCE

In multiple chip devices, coupling of heat between die occurs. The junction temperature can be calculated as follows:

$$(1) \Delta T_{J1} = R_{\theta 1} P_{D1} + R_{\theta 2} K_{\theta 2} P_{D2} + R_{\theta 3} K_{\theta 3} P_{D3} + R_{\theta 4} K_{\theta 4} P_{D4}$$

Where ΔT_{J1} is the change in junction temperature of die 1
 $R_{\theta 1}$ thru 4 is the thermal resistance of die 1 through 4
 P_{D1} thru 4 is the power dissipated in die 1 through 4
 $K_{\theta 2}$ thru 4 is the thermal coupling between die 1 and die 2 through 4.

An effective package thermal resistance can be defined as follows:

$$(2) R_{\theta(EFF)} = \Delta T_{J1} / P_{DT}$$

where: P_{DT} is the total package power dissipation.

Assuming equal thermal resistance for each die, equation (1) simplifies to

$$(3) \Delta T_{J1} = R_{\theta 1} (P_{D1} + K_{\theta 2} P_{D2} + K_{\theta 3} P_{D3} + K_{\theta 4} P_{D4})$$

For the conditions where $P_{D1} = P_{D2} = P_{D3} = P_{D4}$, $P_{DT} = 4 P_D$, equation (3) can be further simplified and by substituting into equation (2) results in

$$(4) R_{\theta(EFF)} = R_{\theta 1} (1 + K_{\theta 2} + K_{\theta 3} + K_{\theta 4}) / 4$$

Values for the coupling factors when either the case or the ambient is used as a reference are given in the table on page 1. If significant power is to be dissipated in two die, die at the opposite ends of the package should be used so that lowest possible junction temperatures will result.

ELECTRICAL CHARACTERISTICS ($T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
OFF CHARACTERISTICS					
Collector-Emitter Breakdown Voltage(1) ($I_C = 10 \text{ mA}$, $I_B = 0$)	BV_{CEO}	30	—	—	Vdc
Collector-Base Breakdown Voltage ($I_C = 10 \mu\text{A}$, $I_E = 0$)	BV_{CBO}	60	—	—	Vdc
Emitter-Base Breakdown Voltage ($I_E = 10 \mu\text{A}$, $I_C = 0$)	BV_{EBO}	5.0	—	—	Vdc
Collector Cutoff Current ($V_{CB} = 50 \text{ Vdc}$, $I_E = 0$)	I_{CBO}	—	—	30	nA
Emitter Cutoff Current ($V_{EB} = 3.0 \text{ Vdc}$, $I_C = 0$)	I_{EBO}	—	—	30	nA
ON CHARACTERISTICS					
DC Current Gain(1) ($I_C = 1.0 \text{ mA}$, $V_{CE} = 10 \text{ Vdc}$)	h_{FE}	25 50	—	—	—
($I_C = 10 \text{ mA}$, $V_{CE} = 10 \text{ Vdc}$)		35 75	—	—	—
($I_C = 150 \text{ mA}$, $V_{CE} = 10 \text{ Vdc}$)		40 100	—	—	—
($I_C = 300 \text{ mA}$, $V_{CE} = 10 \text{ Vdc}$)		20 30	—	—	—
Collector-Emitter Saturation Voltage (1) ($I_C = 150 \text{ mA}$, $I_B = 15 \text{ mA}$)	$V_{CE(sat)}$	—	—	0.4	Vdc
($I_C = 300 \text{ mA}$, $I_B = 30 \text{ mA}$)		—	—	1.4	
Base-Emitter Saturation Voltage (1) ($I_C = 150 \text{ mA}$, $I_B = 15 \text{ mA}$)	$V_{BE(sat)}$	—	—	1.3	Vdc
($I_C = 300 \text{ mA}$, $I_B = 30 \text{ mA}$)		—	—	2.0	
DYNAMIC CHARACTERISTICS					
Current-Gain-Bandwidth Product (1) ($I_C = 50 \text{ mA}$, $V_{CE} = 20 \text{ Vdc}$, $f = 100 \text{ MHz}$)	f_T	200	350	—	MHz
Output Capacitance ($V_{CB} = 10 \text{ Vdc}$, $I_E = 0$, $f = 100 \text{ kHz}$)	C_{ob}	—	6.0 4.5	8.0 8.0	pF
Input Capacitance ($V_{EB} = 2.0 \text{ Vdc}$, $I_C = 0$, $f = 100 \text{ kHz}$)	C_{ib}	—	20 17	30 30	pF
SWITCHING CHARACTERISTICS					
Turn-On Time ($V_{CC} = 30 \text{ Vdc}$, $V_{BE(on)} = 0.5 \text{ Vdc}$, $I_C = 150 \text{ mA}$, $I_B = 15 \text{ mA}$, Figure 1)	t_{on}	—	30	—	ns
Turn-Off Time ($V_{CC} = 30 \text{ Vdc}$, $I_C = 150 \text{ mA}$, $I_B = I_{B2} = 15 \text{ mA}$, Figure 2)	t_{off}	—	225	—	ns

(1) Pulse Test: Pulse Width $\leq 300 \mu\text{s}$, Duty Cycle = 2%.

NPN SATURATED SWITCHING TIME TEST CIRCUITS

For NPN Switching Tests, reverse the diodes, voltage polarities, and input pulses.

FIGURE 1 - NPN TURN-ON TIME

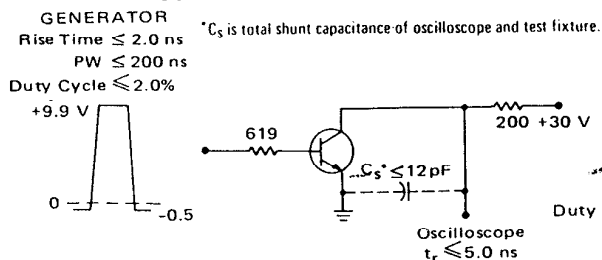
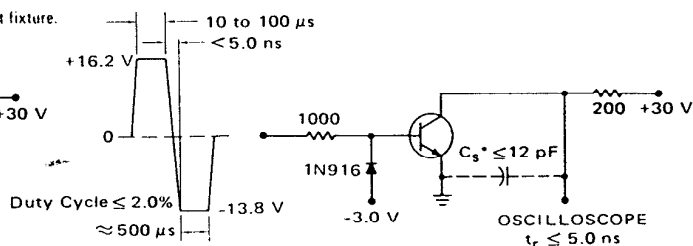


FIGURE 2 - NPN TURN-OFF TIME



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NPN DATA

FIGURE 3 – NORMALIZED DC CURRENT GAIN

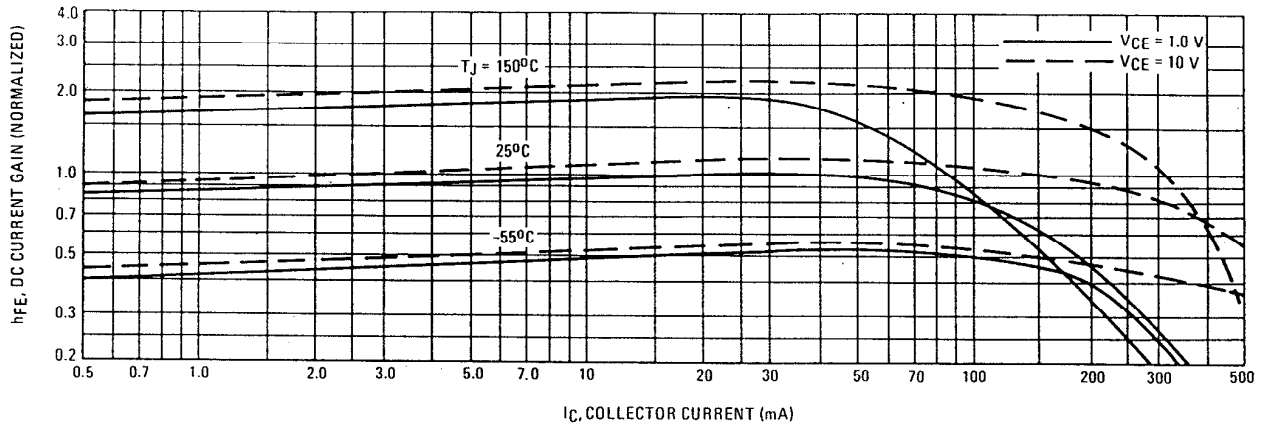


FIGURE 4 – "ON" VOLTAGES

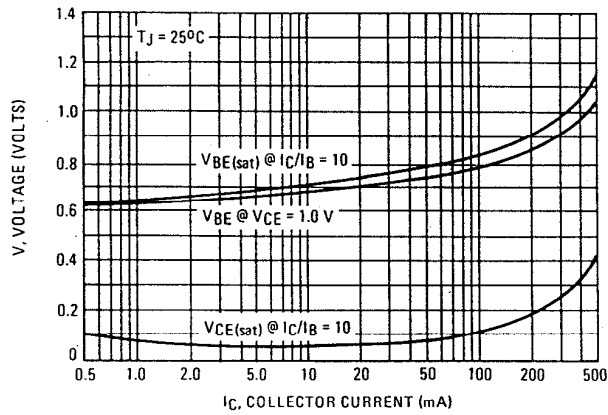
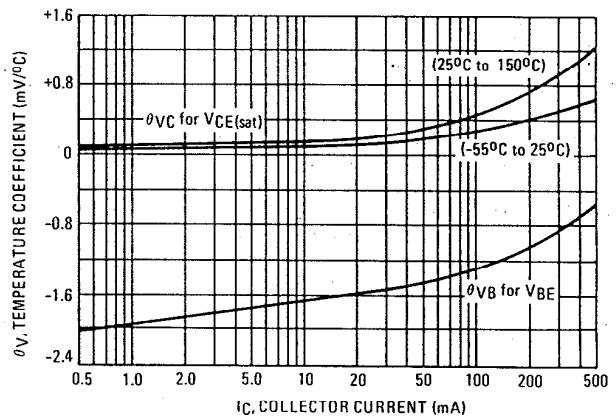


FIGURE 5 – TEMPERATURE COEFFICIENTS



NOISE FIGURE
($V_{CE} = 10\text{ Vdc}$, $T_A = 25^\circ\text{C}$)

FIGURE 6 – FREQUENCY EFFECTS

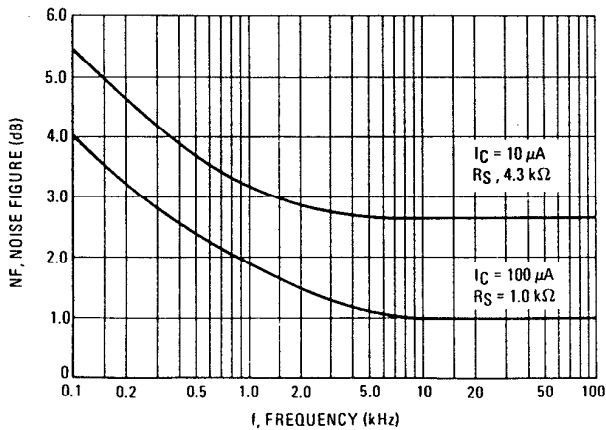
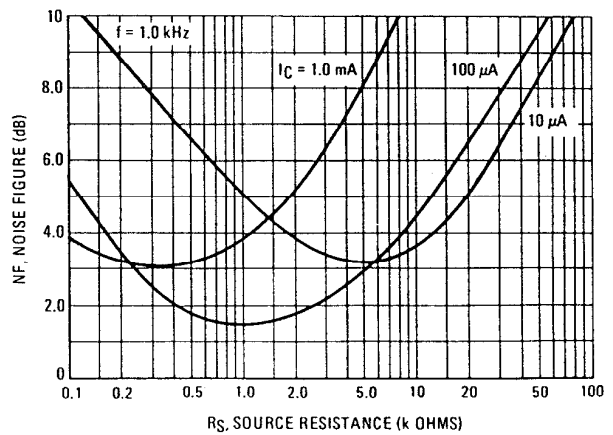
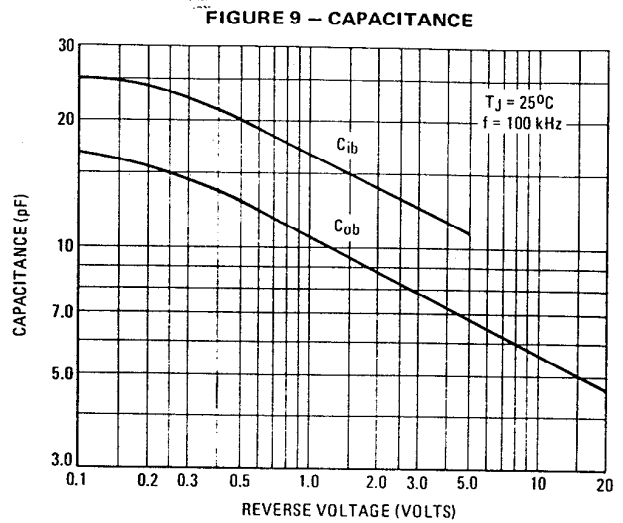
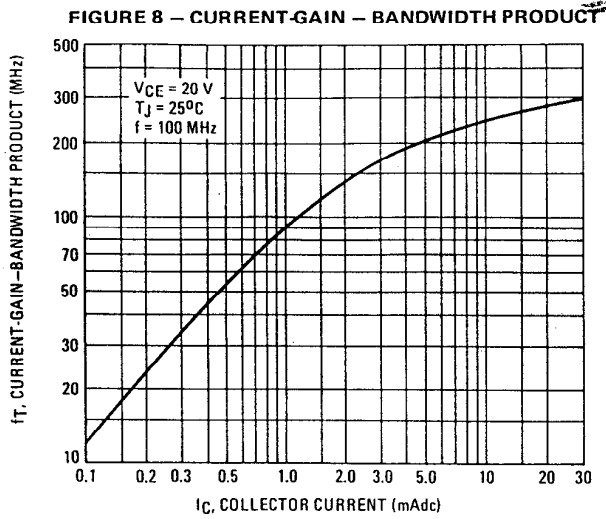


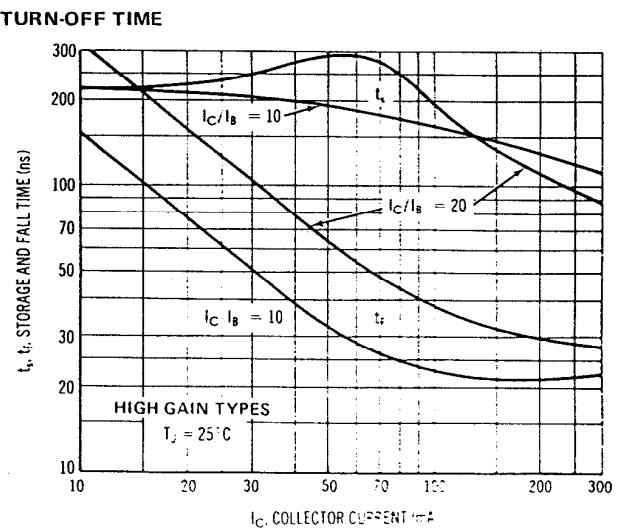
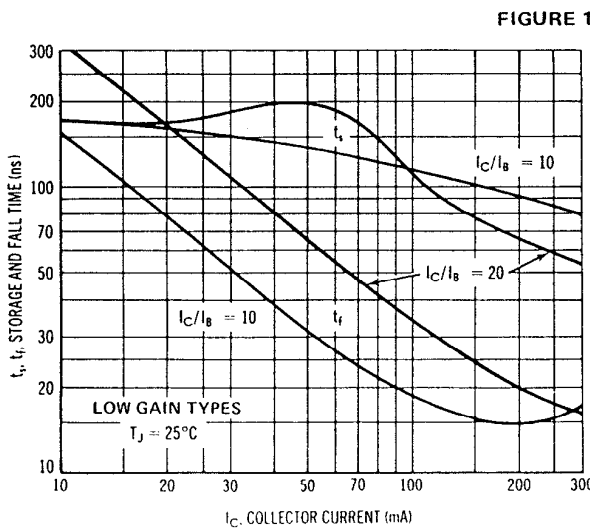
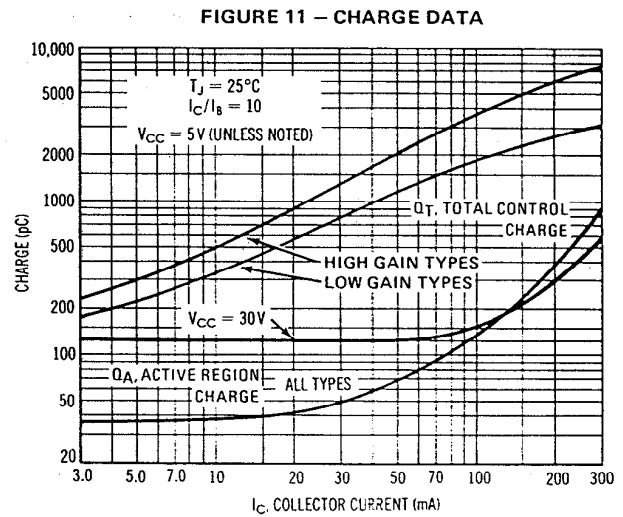
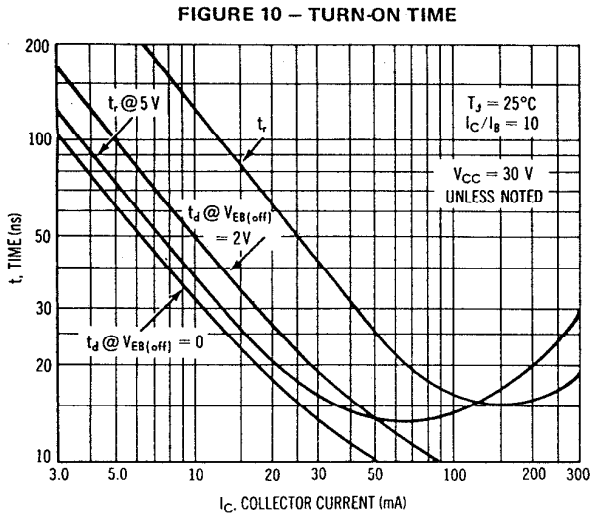
FIGURE 7 – SOURCE RESISTANCE EFFECTS



NPN DATA



SWITCHING TIME CHARACTERISTICS



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PNP DATA

FIGURE 13 – DC CURRENT GAIN

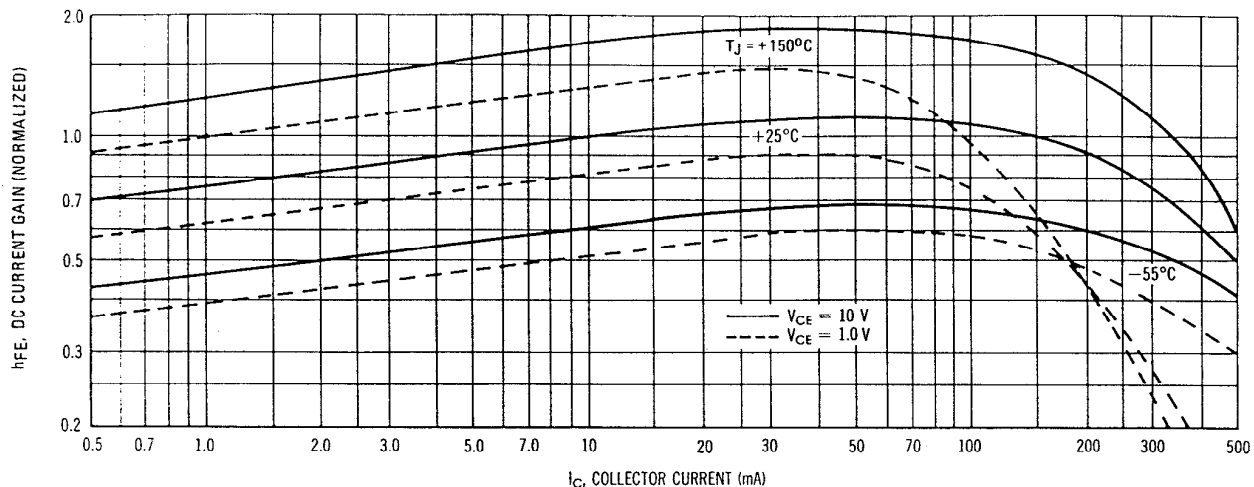


FIGURE 14 – "ON" VOLTAGES

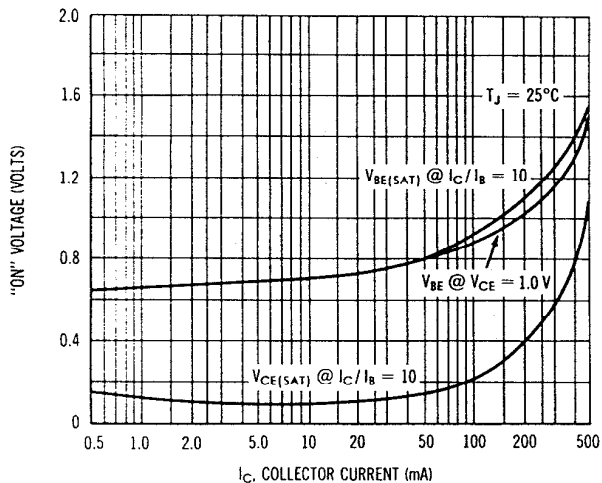
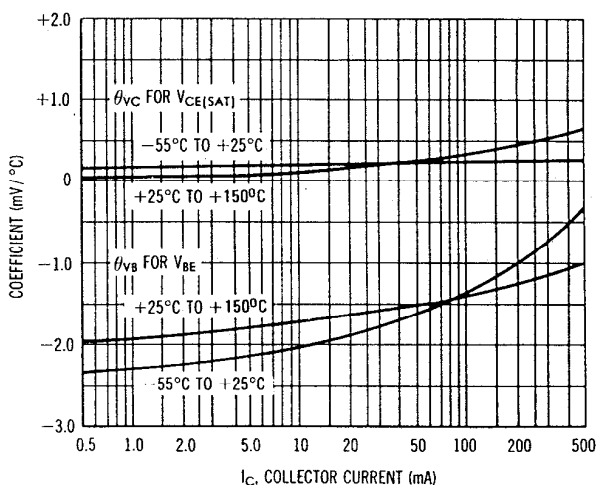


FIGURE 15 – TEMPERATURE COEFFICIENTS



NOISE FIGURE
($V_{CE} = 10\text{ V}$, $T_A = 25^\circ\text{C}$)

FIGURE 16 – FREQUENCY EFFECTS

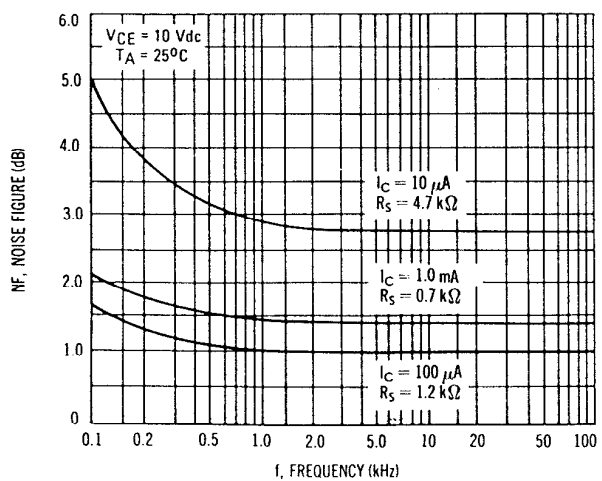
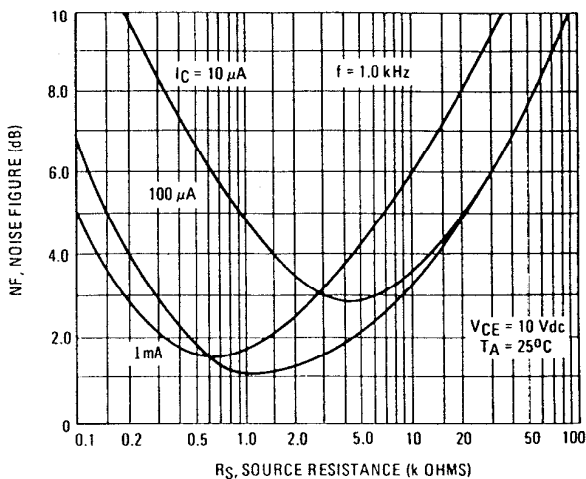


FIGURE 17 – SOURCE RESISTANCE EFFECTS



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PNP DATA

FIGURE 18 – CURRENT-GAIN BANDWIDTH PRODUCT

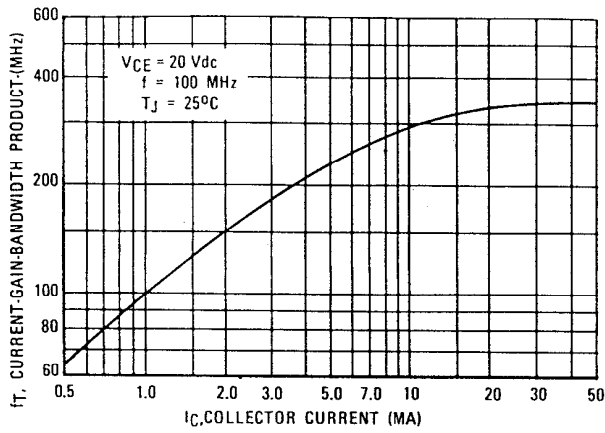


FIGURE 19 – CAPACITANCE

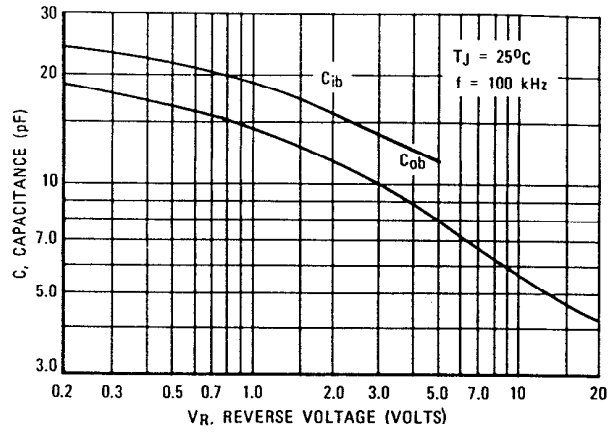


FIGURE 20 – TURN ON TIME

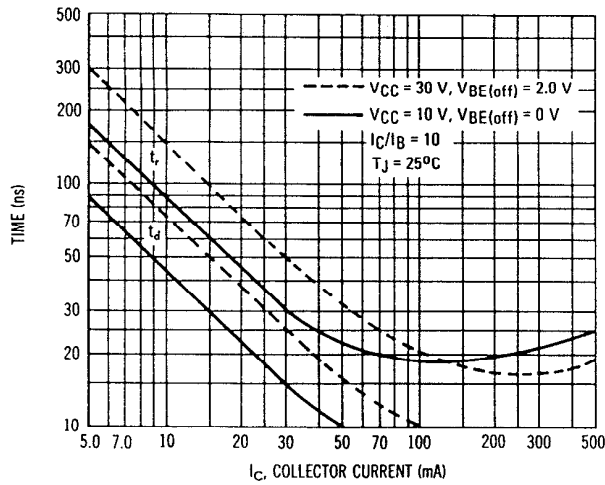


FIGURE 21 – CHARGE DATA

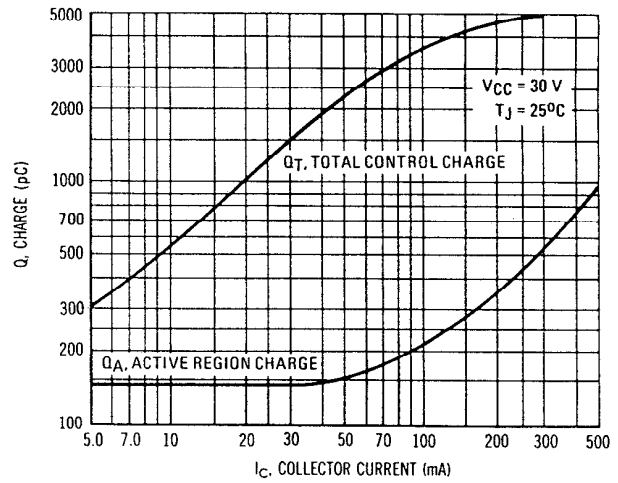


FIGURE 22 – STORAGE TIME

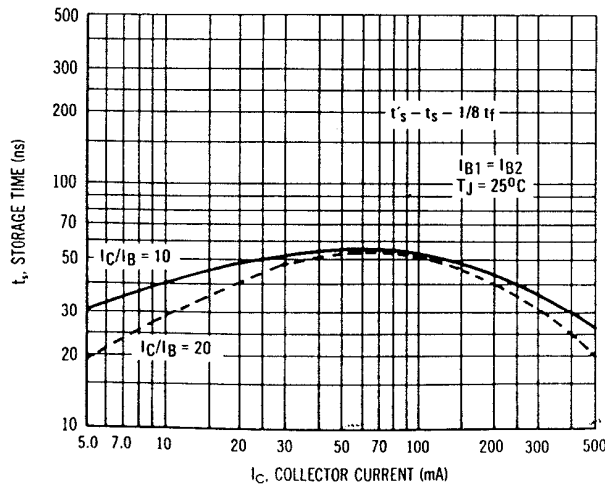
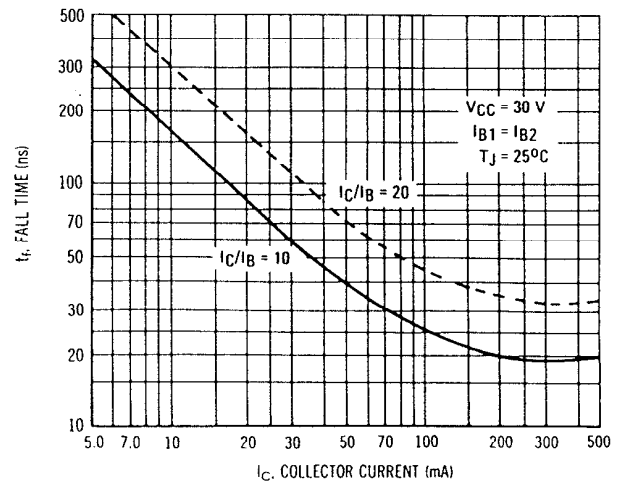


FIGURE 23 – FALL TIME



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