

SEMICONDUCTOR

Direct Conversion FSK Data Receiver

Preliminary Information

Supersedes July 1996 version, DS3853 - 3.5

The SL6619 is an advanced Direct Conversion FSK Data Receiver for operation up to 450 MHz. The device integrates all functions to convert a binary FSK modulated RF signal into a demodulated data stream.

Adjacent channel rejection is provided using tuneable gyrator filters. RF and audio AGC functions assist operation when large interfering signals are present and an automatic frequency control (AFC) function is provided to extend centre frequency acceptance.

FEATURES

- Very Low Power Operation from Single Cell
- Superior Sensitivity
- Operation at 512, 1200 and 2400 Baud
- On Chip 1 Volt Regulator
- 1mm Height Miniature Package
- Automatic Frequency Control Function
- Programmable Post Detection Filter
- AGC Detection Circuitry
- Power Down Function
- Battery Strength Indicator

APPLICATIONS

- Pagers, including Credit Card, PCMCIA and Watch Pagers
- Low Data Rate Receivers, e.g. Security Systems

ORDERING INFORMATION

- SL6619/KG/TP1N
 - 1mm TQFP device, baked and dry packed, supplied in trays 1mm TQFP device, baked and dry
- SL6619/KG/TP1Q packed, supplied in tape and reel



Fig. 1 Pin identification diagram (top view). See Table 1 for pin descriptions

ABSOLUTE MAXIMUM RATINGS

Storage temperature -	-55°C to+150°C
Operating temperature	$-10^{\circ}C$ to+55°C
Maximum voltage on any pin w.r.t. any	+4V
other pin, subject to the following conditions:	
Current, pin 3 (MIXIP), pin 5 (MIXPB),	<5ma
pin 12 (LOIPI) and pin 14 (LOIPB)	
Most negative voltage on any pin	-0.5V w.r.t. gnd



Fig. 2 Block diagram of SL6619

Pin number	Pin name	Pin description						
1	IRF	LNA current source						
2	GND	Ground						
3	MIXIP A	Mixer input A						
4	MIX DEC	Mixer biasing decouple						
5	MIXIP B	Mixer input B						
6	REG CNT	1V regulator control external PNP drive						
7	VREG	1V regulator output voltage						
8	TPI	I channel pre-gyrator filter test point.						
9	11	Mixer output, I channel						
10	12	Mixer output, I channel						
11	VCC1	Positive supply 1						
12	LOIP I	LO input channel I						
13	GYRI	Gyrator current adjust pin						
14	LOIP Q	LO input channel Q						
15	Q1	Mixer output, Q channel						
16	Q2	Mixer output, Q channel						
17	TPQ	Q channel pre-gyrator filter test point						
18	VREF	Reference voltage						
19	AFC OP	AFC output						
20	BEC	Battery economy control						
21	DATA OP	Data output pin						
22	VCC2	Positive supply 2						
23	BATT FLAG	Battery flag output						
24	AFC1	AFC characteristic defining pin						
25	AFC2	AFC characteristic defining pin						
26	BRF CNT	Bit rate filter control						
27	BRF1	Bit rate filter 1, output from detector						
28	VBATT	Battery flag input voltage						
29	TP LIM I	I channel limiter (post gyrator filter) test point, output only						
30	IAGC OP	Audio AGC output current						
31	TC ADJ	Audio AGC time constant adjust						
32	GTH ADJ	Audio AGC gain and threshold adjust. RSSI signal indicator						

Table 1 SL6619 pin descriptions

ELECTRICAL CHARACTERISTICS (1)

Electrical Characteristics (1) are guaranteed over the following range of operating conditions unless otherwise stated $T_{AMB} = +25^{\circ}C$, $V_{CC}1 = 1.3V$, $V_{CC}2 = 2.7V$

		\ \	/alue				
Characteristic	Pin	Min.	Тур.	Max.	Units	Conditions	
Supply voltage, V _{CC} 1	11	0.95	1.3	2.7	V	V _{CC} 1≤V _{CC} 2−0⋅8V	
Supply voltage, V _{CC} 2	22	1.9	2.7	3.5	V		
Supply current, I _{CC} 1	11	1.20	1.60	2.2	mA	Including IRF	
Supply current, I _{CC} 2	22	260	350	460	μA		
1 volt regulator, V _{REG}	7	0.95	1.0	1.05	V	$I_{LOAD} = 3mA$, external PNP($\beta \ge 100$, $V_{CE} = 0.1V$)	
1 volt regulator load current	7	0.25		3	mA	External PNP (h _{FE} ≥100, V _{CE} = 0·1V)	
LNA current source, IRF	1	375	500	700	μA	PTAT, voltage on pin $1 = 0.3V$ and $1.3V$	
Reference voltage, V _{REF}	18	1.15	1.25	1.31	V	Typical temperature coefficient = $+0.1$ mV/°C	
V _{REF} source current	18			20	μA		
V _{REF} sink current	18			1.0	μA		
Data Amplifier DATA OP sink current DATA OP leakage current Output mark:space ratio	21 21 21	25 7:9		1.0 9:7	μΑ μΑ	Output logic low, pin 21 voltage = $0.3V$ Output logic high, pin 21 voltage = $V_{CC}2$ Preamble at 1200 baud, $\Delta f = 4kHz$, pin 26 = 0V, BRF capacitor = 560pF, DATA OP pullup resistor = 200k Ω	
Battery Economy							
Power down I _{CC} 1	11		0.5	10	μA	Pin 20 = logic low	
Power down I _{CC} 2	22		2.0	10	μA	Pin 20 = logic low	
BEC input logic high	20	V _{cc} 2-0.3V		V _{CC} 2	V	Powered up	
BEC input logic low	20	0		0.3	V	Powered down	
BEC input current	20	-1.0		1.0	μA	Powered up	
BEC input current	20	-1.0		1.0	μA	Powered down	
Battery Flag							
V _{BATT} trigger point	28	1.04	1.08	1.12	V	Current sunk by pin 23 = $1\mu A$	
BATT FLAG sink current	23			1.0	μA	Pin 28 voltage = 1.04V	
BATT FLAG sink current	23	1.0			μA	Pin 28 voltage = 1.12V	
BATT FLAG sink current	23	25			μA	Pin 28 voltage = 1.14V	
V _{BATT} input voltage	28			2.0	V		
V _{BATT} input current	28	-1.0		1.0	μA	V _{BATT} = 1.14V	
V _{BATT} input current	28	-1.0		1.0	μA	$V_{BATT} = 1.04V$	

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ELECTRICAL CHARACTERISTICS (1) (Cont.)

Electrical Characteristics (1) are guaranteed over the following range of operating conditions unless otherwise stated $T_{AMB} = +25^{\circ}C$, $V_{CC}1 = 1.3V$, $V_{CC}2 = 2.7V$

		Value				
Characteristic	Pin	Min.	Тур.	Max.	Units	Conditions
Mixers LO DC bias voltage Gain to TPI	12,14 3,5,8,12	38	V _{cc} 1 42	46	V dB	LO inputs (12, 14) driven in quadrature: 45mVrms at 450MHz, CW. Mixer inputs (3, 5) driven differentially:
Gain to TPQ	3,5,14, 17	38	42	46	dB	As gain to TPI
Match of gain to TPI and TPQ	3,5,8, 12,14,17	-1	0	+1	dB	As gain toTPI
Audio AGC IAGC OP max. sink current IAGC OP leakage current	30 30		40	1	μΑ μΑ	TPI, TPQ signals limiting No signal applied
AFC AFC DC current, I _{AFC4k5} AFC DC current AFC DC current	19 19 19	I _{AFC4k5} +0·2	0.0 I _{AFC4k5} +0.7 I _{AFC4k5} -0.9	I _{AFC4k5} 0•2	μΑ μΑ μΑ	$f_{C} = f_{LO} + 4.5 \text{kHz}, CW$ $f_{C} = f_{LO} + 2.5 \text{kHz}, CW$ $f_{C} = f_{LO} + 6.5 \text{kHz}, CW$
Bit Rate Filter Control BRF CNT input logic high BRF CNT input logic low Tristate I/P current window BRF 1 output current BRF 1 output current BRF 1 output current BRF CNT input high current BRF CNT input low current	26 26 27 27 27 27 26 26	V_{CC}^2 -0.3 0 -0.4	3.5 1.7 0.74	V _{CC} 2 0·1 +0·4 +7·5 +7.5	V μΑ μΑ μΑ μΑ	2400 baud 1200 baud 512 baud Pin 26 logic high Pin 26 logic low Pin 26 logic tristate (open circuit)

ELECTRICAL CHARACTERISTICS (2)

Electrical Characteristics (2) are guaranteed over the following range of operating conditions unless otherwise stated. Characteristics are tested at room temperature only and are guaranteed by characterisation test or design. $T_{AMB} = -10^{\circ}C$ to $+55^{\circ}C$, $V_{CC}1 = 1.4V$ to 2.0V, $V_{CC}2 = 2.3V$ to 3.2V. $V_{CC}1 < V_{CC}2 - 0.8V$

	Valu		/alue	alue			
Characteristic	Pin	Min.	Тур.	Max.	Units	Conditions	
Supply voltage, V _{CC} 1 Supply voltage, V _{CC} 2 Supply current, I _{CC} 1 Supply current, I _{CC} 2	11 22 11 22	0.95 1.9	1·3 2·7 1·60 350	2·7 3·5 2·4 510	V V mA μA	$V_{CC}1$ ≤ $V_{CC}2$ -0.8V at ≥25°C only Including IRF	
1 volt regulator, V_{REG} 1 volt regulator load current LNA current source, IRF Reference voltage, V_{REF} V_{REF} source current V_{REF} sink current	7 7 1 18 18 18	0.93 0.25 375 1.13	1.0 500 1.25	1.05 3 800 1.33 18 0.8	V mA μA V μA μA	$\begin{split} I_{LOAD} &= 3mA, \text{ external } PNP(\beta \geqslant 100, V_{CE} = 0.1V) \\ \text{External } PNP(h_{FE} \geqslant 100, V_{CE} = 0.1V) \\ \text{PTAT, voltage on pin } 1 = 0.3V \text{ and } 1.3V \\ \text{Typical temperature coefficient} = +0.1mV/^{\circ}C \end{split}$	
Turn-on time Turn-off time			5 1		ms ms	Stable data O/P when 3dB above sensitivity. $C_{VREF} = 2.2 \mu F$ Fall to 10% of steady state I _{CC} 1. $C_{VREF} = 2.2 \mu F$	
Data Amplifier DATA OP sink current DATA OP leakage current Output mark:space ratio	21 21 21	22 7:9		1.5 9:7	μΑ μΑ	Output logic low, pin 21 voltage = $0.3V$ Output logic high, pin 21 voltage = $V_{CC}2$ Preamble at 1200 baud, $\Delta f = 4kHz$, pin 26 = 0V, BRF capacitor = 560pF, DATA OP pullup resistor = 200k Ω	
Battery Economy Power down I _{CC} 1 Power down I _{CC} 2 BEC input logic high BEC input logic low BEC input current BEC input current	11 22 20 20 20 20 20	V _{CC} 2-0·3V 0 -1·0 -1·0	0.5 2.0	12 12 V _{CC} 2 0·3 1·5 1·5	μΑ μΑ V ν μΑ μΑ	Pin 20 = logic low Pin 20 = logic low Powered up Powered down Powered up Powered down	
Battery Flag V _{BATT} trigger point BATT FLAG sink current BATT FLAG sink current BATT FLAG sink current V _{BATT} input voltage V _{BATT} input current V _{BATT} input current	28 23 23 23 28 28 28 28 28	1.04 2 20 -1.5 -1.5	1-08	1.12 2 2.0 1.5 1.5	V μΑ μΑ V μΑ μΑ	Current sunk by pin 23 = 1 μ A Pin 28 voltage = 1.04V Pin 28 voltage = 1.12V Pin 28 voltage = 1.12V V _{BATT} = 1.14V V _{BATT} = 1.04V	

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ELECTRICAL CHARACTERISTICS (2) (Cont.)

Electrical Characteristics (2) are guaranteed over the following range of operating conditions unless otherwise stated. Characteristics are tested at room temperature only and are guaranteed by characterisation test or design. $T_{AMB} = -10^{\circ}C$ to $+55^{\circ}C$, $V_{CC}1 = 1.4V$ to 2.0V, $V_{CC}2 = 2.3V$ to 3.2V. $V_{CC}1 < V_{CC}2 = 0.8V$

			Value			
Characteristic	Pin	Min.	Тур.	Max.	Units	Conditions
Mixers LO DC bias voltage Gain to TPI	12,14 3,5,8,12	35	V _{cc} 1 42	46	V dB	LO inputs (12, 14) driven in quadrature: 45mVrms at 450MHz, CW. Mixer inputs (3, 5) driven differentially: 0.45mVrms at 450.004MHz, CW
Gain to TPQ	3,5,14, 17	35	42	46	dB	As gain to TPI
Match of gain to TPI and TPQ	3,5,8, 12,14,17	-1.5	0	+1.5	dB	As gain toTPI
Audio AGC IAGC OP max. sink current IAGC OP leakage current	30 30	15	40	80 2	μΑ μΑ	TPI, TPQ signals limiting No signal applied
AFC AFC DC current, I _{AFC4k5} AFC DC current AFC DC current	19 19 19	I _{AFC4k5} +0∙1	0.0 _{AFC4k5} +0.7 _{AFC4k5} -0.9	I _{AFC4k5} −0·1	μΑ μΑ μΑ	$f_{C} = f_{LO} + 4.5 \text{kHz}, CW$ $f_{C} = f_{LO} + 2.5 \text{kHz}, CW$ $f_{C} = f_{LO} + 6.5 \text{kHz}, CW$
Bit Rate Filter Control BRF CNT input logic high BRF CNT input logic low Tristate I/P current window BRF 1 output current BRF 1 output current BRF 1 output current BRF CNT input high current BRF CNT input low current	26 26 27 27 27 26 26	V _{CC} 2 -0·3 0 -0·4	3.5 1.7 0.74	V _{cc} 2 0.1 +0.4 +10 +10	V μΑ μΑ μΑ μΑ	2400 baud 1200 baud 512 baud Pin 26 logic high Pin 26 logic low Pin 26 logic tristate (open circuit)

RECEIVER CHARACTERISTICS (450MHz)

Receiver Characteristics (450MHz) are guaranteed over the following range of operating conditions unless otherwise stated. Characteristics are not tested but are guaranteed by characterisation test or design. All measurements made using the characterisation circuit Fig. 5. See Application Note AN137 for details of test method. $T_{AMB} = -10^{\circ}C$ to $+55^{\circ}C$, $V_{CC}1 = 1.04V$ to 2.0V, $V_{CC}2 = 2.3V$ to 3.2V, $V_{CC}1 < V_{CC}2 - 0.8V$, carrier frequency = 450MHz, BER = 1 in 30, AFC open loop. LNA gain set such that an RF signal G-73dBm at the LNA input, offset from the LO

by 4kHz, gives a typical IF signal level of 300mV p-p at TPI and TPQ. LNA noise figure<2dB

	Value					
Characteristic	Min.	Тур.	Max.	Units	Conditions	
Sensitivity		-128 -126 -123	-122 -119	dBm dBm dBm	512bps, Δf = 4·5kHz 1200bps, Δf = 4·0kHz 2400bps, Δf = 4·5kHz. LO = -15dBm	
Intermodulation, IP3	50 48	57 55 53		dB dB dB	512bps, Δf = 4-5kHz 1200bps, Δf = 4-0kHz 2400bps, Δf = 4-5kHz. LO = -15dBm. Channel spacing 25kHz	
Adjacent Channel	62•5 60	74 72 69		dB dB dB	512bps, Δf = 4·5kHz 1200bps, Δf = 4·0kHz 2400bps, Δf = 4·5kHz. LO = -15dBm. Channel spacing 25kHz	
Deviation Acceptance						
Up		+1.9		kHz	512bps, Δf = 4⋅5kHz, no AFC	
Down		-2.5		kHz	512bps, $\Delta f = 4.5 \text{kHz}$, no AFC	
Up	+1.8	+3.0	+4.6	kHz	1200bps, $\Delta f = 4.0$ kHz, no AFC	
Down	-2.7	-2.3	-1.7	kHz	1200bps, $\Delta f = 4.0$ kHz, no AFC	
Up	+1.7	+2.5	+4.6	kHz	2400bps, $\Delta f = 4.5 \text{kHz}$, no AFC	
Down	-3	-2.3	-1.7	kHz	2400bps, $\Delta f = 4.5 \text{kHz}$, no AFC	
Centre Frequency Acceptance	±2•0 ±2•0	±2·8 ±2·5 ±2·5	±2·9 ±3·2	kHz kHz kHz	512bps, Δf = 4·5kHz, no AFC 1200bps, Δf = 4·0kHz, no AFC 2400bps, Δf = 4·5kHz, no AFC	
AFC Capture Range (AFC Closed Loop)		±4 ±3·5 ±4		kHz kHz kHz	512bps, $\Delta f = 4.5$ kHz. All at sensitivity +3dB or above 1200bps, $\Delta f = 4.0$ kHz. All at sensitivity +3dB or above 2400bps, $\Delta f = 4.5$ kHz. All at sensitivity +3dB or above	

RECEIVER CHARACTERISTICS (280MHz)

Receiver Characteristics (280MHz) are guaranteed over the following range of operating conditions unless otherwise stated. Characteristics are not tested but are guaranteed by characterisation test or design. All measurements made using the characterisation circuit Fig. 5. See Application Note AN137 for details of test method. $T_{AMB} = -10^{\circ}C$ to $+55^{\circ}C$, $V_{CC}1 = 1.04V$ to 2.0V, $V_{CC}2 = 2.3V$ to 3.2V, $V_{CC}1 < V_{CC}2 - 0.8V$, carrier frequency = 280MHz, BER = 1 in 30, AFC open loop. LNA gain set such that an RF signal of -73dBm at the LNA input, offset from the LO

by 4kHz, gives a typical IF signal level of 300mV p-p at TPI and TPQ. LNA noise figure<2dB

	Value					
Characteristic	Min.	Тур.	Max.	Units	Conditions	
Sensitivity	-128 -127	-129 -127 -124	-124 -121	dBm dBm dBm	512bps, Δf = 4·5kHz 1200bps, Δf = 4·0kHz 2400bps, Δf = 4·5kHz. LO = -15dBm	
Intermodulation, IP3	52 49	57 56 53·5	60 57	dB dB dB	512bps, Δf = 4-5kHz 1200bps, Δf = 4-0kHz 2400bps, Δf = 4-5kHz. LO = -15dBm. Channel spacing 25kHz	
Adjacent Channel	62-5 60	74 72 70	80 77	dB dB dB	512bps, $\Delta f = 4.5$ kHz 1200bps, $\Delta f = 4.0$ kHz 2400bps, $\Delta f = 4.5$ kHz. LO = -15dBm. Channel spacing 25kHz	
Deviation Acceptance Up Down Up Down Up Down	+1.8 -2.7 +1.7 -3.0	+1.9 -2.5 +3.0 -2.3 +2.5 -2.3	+4.6 -1.7 +4.6 -1.7	kHz kHz kHz kHz kHz kHz	512bps, Δf = 4·5kHz, no AFC 512bps, Δf = 4·5kHz, no AFC 1200bps, Δf = 4·0kHz, no AFC 1200bps, Δf = 4·0kHz, no AFC 2400bps, Δf = 4·5kHz, no AFC 2400bps, Δf = 4·5kHz, no AFC	
Centre Frequency Acceptance	±2•0 ±2•0	±2·8 ±2·5 ±2·5	±2·9 ±3·2	kHz kHz kHz	512bps, Δf = 4·5kHz, no AFC 1200bps, Δf = 4·0kHz, no AFC 2400bps, Δf = 4·5kHz, no AFC	
AFC Capture Range (AFC Closed Loop)		±4 ±3·5 ±4		kHz kHz kHz	512bps, $\Delta f = 4.5$ kHz. All at sensitivity +3dB or above 1200bps, $\Delta f = 4.0$ kHz. All at sensitivity +3dB or above 2400bps, $\Delta f = 4.5$ kHz. All at sensitivity +3dB or above	
1MHz Blocking	67 65	75 75 75	78 76	dB dB dB	512bps, Δf = 4·5kHz 1200bps, Δf = 4·0kHz 2400bps, Δf = 4·5kHz. LO = -15dBm	

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OPERATION OF SL6619

Low Noise Amplifier

To achieve optimum performance it is necessary to incorporate a Low Noise RF Amplifier at the front end of the receiver. This is easily biased using the on-chip voltages and current source provided. All voltages and current sources used for bias of the RF amplifier, receiver and mixers should be RF decoupled using 1nF capacitors. The receiver also requires a stable Local Oscillator at the required channel frequency.

Local Oscillator

The Local Oscillator signal is applied to the device in phase quadrature. This can be achieved with the use of two RC networks operating at their $-3dB/45^{\circ}$ transfer characteristic. The RC characteristics for I and Q channels are combined to give a full 90° phase differential between the LO ports of the device. Each LO port also requires an equal level of drive from the oscillator. This is achieved by forming the two RC networks into a power divider.

Gyrator Filters

The on-chip filters include an adjustable gyrator filter. This may be adjusted by changing the value of the resistor connected between pin 13 and GND. This allows adjustment of the filters' cutoff frequency and allows for compensation for possible process variations.

Audio AGC (Fig. 3)

The Audio AGC consists of a current sink which is controlled by the audio (baseband) signal. It has three parameters that may be controlled by the user. These are the attack (turn on) time, decay (duration) time and threshold level. The attack time is simply determined by the value of the external capacitor connected to TCADJ. The external capacitor is in series with an internal 100k Ω resistor and the time constant of this circuit dictates the attack time of the AGC.

i.e. $t_{ATTACK} = 100k\Omega \times C18$

The decay time is determined by the external resistor connected in parallel with the capacitor CTC. The decay time is simply

$t_{DECAY} = R17 \times C18$

When a large audio (baseband) signal is incident on the input to the AGC circuit, the variable current source is turned on. This causes a voltage drop across R13. The voltage potential between V_{REF} and the voltage on pin 31 causes a current to flow in pin 30. This charges up C18 through the 100k Ω internal resistor. As the voltage across the capacitor increases, a current source is turned on and this sinks current from pin 32. The current sink on pin 32 can be used to drive

the external AGC circuit by causing a PIN diode to conduct, reducing the signal to the RF amplifier.

RF AGC

The RF AGC is an automatic gain control loop that protects the mixer's RF inputs, Pins 3 and 5, from large out of band RF signals. The loop consists of an RF received signal strength indicator which detect the signal at the inputs of the mixers. This RSSI signal is then used to control the LNA current source (pin 1).

Regulator

The on-chip regulator should be used in conjunction with a suitable PNP transistor to achieve regulation. As the transistor forms part of the regulator feedback loop the transistor should exhibit the following characteristics:

 H_{FE} >100 for V_{CE} = 0.1V

If no external transistor is used, the maximum current sourcing capability of the regulator is limited to 30µA.

Automatic Frequency Control (Fig. 4)

The Automatic Frequency Control consists of a detection circuit which gives a current output at AFC OP whose magnitude and sign is a function of the difference between the local oscillator (f_{LO}) and carrier frequencies (f_C). This output current is then filtered by an off-chip integrating capacitor. The integrator's output voltage is used to control a voltage control crystal oscillator. This closes the AFC feedback loop giving the automatic frequency control function. For an FSK modulated incoming RF carrier, the AFC OP current's polarity is positive, i.e. current is sunk, for $f_{LO} > f_C > f_{LO} - 4kHz$. The magnitude of the AFC OP current is a function of frequency offset and the transmitted data's bit stream. If the carrier frequency, (f_C), equals the local oscillator frequency, (f_{LO}) then the magnitude of the current is zero.

BIT RATE FILTER CONTROL

The logic level on pin 26 controls the cutoff frequency of the 1st order bit rate for a given bit rate filter capacitor at pin 27. This allows the cutoff frequency to be changed between f_C , $2f_C$ and $0.43f_C$ through the logic level on pin 26. This function is achieved by changing the value of the current in the 4φ detector's output stage. A logic zero (0V to 0.1V) on pin 26 gives a cutoff frequency of f_C alogic one ($V_{CC}2-0.3V$ to $V_{CC}2$) gives a cutoff frequency of $2f_C$ and an open circuit at pin 26 gives a cutoff frequency of $0.43f_C$.



Fig.3 AGC schematic



Fig. 4 AFC schematic

Peak deviation	Baud rate	Component (Fig. 4)			
(kHz)	(bps)	C22	C21	R11	
3.5	512, 1200, 2400	750pF	2∙0nF	15kΩ	
4	512, 1200, 2400	560pF	1.5nF	$15k\Omega$	
4.5	512, 1200, 2400	510pF	1⋅3nF	$15k\Omega$	
5	512, 1200, 2400	470pF	1.2nF	$15k\Omega$	
5.5	512, 1200, 2400	430pF	1∙1nF	$15k\Omega$	

Table 2 AFC defining components



Fig. 5 SL6619 characterisation circuit (see Tables 3 and 4 for component values)

Resistors		Capacitors		Capacitor	Capacitors (cont.)		Inductors
R1	4·7kΩ	C1	12pF	C18	100nF	L1	56nH
R2	4·7kΩ	C2	O/C	C19	1nF	T1	30nH 1:1, Coilcraft M1686-A
R3	2kΩ	C3	220nF	C20	2•2µF		Transistors
R4	100Ω	C4	1nF	C21	1.5nF		114115151015
R5	100Ω	C5	1nF	C22	560pF	TR1	Toshiba 2SC5065
R6	100Ω	C6	1nF	C23	1nF	TR2	Toshiba 2SC5065
R7	100Ω	C7	1nF	C24	2·2µF	TR3	FMMT589 (Zetex ZTX550)
R8	$430 k\Omega$	C8	3-3pF	C25	100nF		
R9	$220k\Omega$	C9	4.7nF	C26	100nF		
R10	S/C	C10	4.7nF	C27	560pF		
R11	15kΩ	C11	4.7pF	C28	1nF		
R12	2kΩ	C12	5-6pF	C29	1nF		
R13	33kΩ	C13	1nF	C30	1nF		
R14	$180 k\Omega$	C14	1nF	C32	100nF		
R15	$430 \mathrm{k}\Omega$	C15	1nF	C33	100nF		
R16	$220k\Omega$	C16	1nF	C34	100nF		
R17	$220k\Omega$	C17	2•2µF	VC1	3-10pF		

Table 3 Component list for 280MHz characterisation board

Resis	Resistors		Capacitors		Capacitors (cont.)		Inductors
R1	4·7kΩ	C1	O/C	C18	100nF	L1	47nH
R2	4.7kΩ	C2	O/C	C19	1nF	T1	16nH 1:1, Coilcraft Q4123-A
R3	1.8kΩ	C3	1nF	C20	2•2µF		Transistors
R4	100Ω	C4	1nF	C21	1.5nF		110113131013
R5	100Ω	C5	1nF	C22	560pF	TR1	Philips BFT25A
R6	100Ω	C6	1nF	C23	1nF	TR2	Philips BFT25A
R7	100Ω	C7	1nF	C24	2•2µF	TR3	FMMT589 (Zetex ZTX550)
R8	$430 \text{k}\Omega$	C8	3.3pF	C25	100nF		
R9	$220 k\Omega$	C9	4∙7nF	C26	100nF		
R10	S/C	C10	4∙7nF	C27	560pF		
R11	15kΩ	C11	3∙9pF	C28	1nF		
R12	2kΩ	C12	3∙3pF	C29	1nF		
R13	33kΩ	C13	1nF	C30	1nF		
R14	$180 k\Omega$	C14	1nF	C32	100nF		
R15	$430 \text{k}\Omega$	C15	1nF	C33	100nF		
R16	$220k\Omega$	C16	1nF	C34	100nF		
R17	$220 k\Omega$	C17	2∙2µF	VC1	3-10pF		

Table 4 Component list for 450MHz characterisation board



Fig. 6 Typical I_{CC}1 and I_{CC}2 v. supply and temperature



Fig. 7 Typical V_{REF} and V_{REG} v. supply and temperature

SL6619



Fig. 8 Typical I_{RF} v. supply and temperature



Fig. 9 Typical battery flag trigger voltage ($V_{BATTFLAG} = V_{CC}/2$) v. supply and temperature

TYPICAL AC PARAMETERS (FIGS. 10 TO 13)



Fig. 10 Typical sensitivity v. supply and temperature

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Fig. 11 Typical IP3 and adjacent channel v. supply and temperature



Fig. 12 Typical deviation acceptance v. supply and temperature



Fig. 13 Typical centre frequency acceptance and 1MHz blocking v. supply and temperature



Notes:

- 1. Pin 1 indicator may be a corner chamfer, dot or both.
- 2. Controlling dimensions are in millimeters.

3. The top package body size may be smaller than the bottom package body size by a max. of 0.15 mm. 4. Dimension D1 and E1 do not include mould protusion.

- 5. Dimension b does not include dambar protusion.
- 6. Coplanarity, measured at seating plane G, to be 0.10 mm max.

This drawing supersedes 418/ED/51612/001 (Swindon)

© Mitel						ORIGINATING SITE: SWINDON
ISSUE	1	2				Title: Package Outline Drawing for
ACN	201348	207076			SEMICONDUCTOR	(7x7x1.0) mm, Body+2.0 mm
DATE	250CT96	6JUL99			SLIVICONDOCTOR	Drawing Number
APPROVED						GPD00233



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