# 40ns, Low-Power, 3V/5V, Rail-to-Rail Single-Supply Comparators 

## General Description

The MAX9140/MAX9141 are single and the MAX9142/ MAX9144 are dual/quad high-speed comparators optimized for systems powered from a 3 V or 5 V supply. The MAX9141 features latch enable and device shutdown. These devices combine high speed, low power, and Rail-to-Rail ${ }^{\circledR}$ inputs. Propagation delay is 40 ns , while supply current is only $150 \mu \mathrm{~A}$ per comparator.
The input common-mode range of the MAX9140/ MAX9141/MAX9142/MAX9144 extends beyond both power-supply rails. The outputs pull to within 0.3 V of either supply rail without external pullup circuitry, making these devices ideal for interface with both CMOS and TTL logic. All input and output pins can tolerate a continuous short-circuit fault condition to either rail. Internal hysteresis ensures clean output switching, even with slow-moving input signals.
The MAX9140/MAX9141/MAX9142/MAX9144 are high-er-speed, lower-power, and lower-cost upgrades to industry-standard comparators MAX941/MAX942/ MAX944.
The MAX9140 are offered in tiny 5-pin SC70 and SOT23 packages. The MAX9141 and MAX9142 are available in 8 -pin SOT23 and SO packages, while the MAX9144 is available in both 14-pin SO and TSSOP packages.

## Applications

Line Receivers
Battery-Powered Systems
Threshold Detectors/Discriminators
3V/5V Systems
Zero-Crossing Detectors
Sampling Circuits

Features

- Fast, 40ns Propagation Delay (10mV Overdrive)
- Low Power:
0.45mW Power Dissipation Per Comparator (3V) 150 HA Supply Current
- Optimized for 3V and 5V Applications (Operation Down to 2.7V)
- Rail-to-Rail Input Voltage Range
- Low, 500 $\mathbf{~ V}$ Offset Voltage
- Internal Hysteresis for Clean Switching
- Outputs Swing 300mV of Power Rails
- CMOS/TTL-Compatible Outputs
- Output Latch (MAX9141 only)
- Shutdown Function (MAX9141 only)
- Available in SC70 and SOT23 Packages

Ordering Information

| PART | TEMP. RANGE | PIN- <br> PACKAGE | TOP MARK |
| :---: | :---: | :---: | :---: |
| MAX9140EXK-T | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 5 SC70-5 | ACC |
| MAX9140EUK-T | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 5 SOT23-5 | ADQP |
| MAX9141EKA-T | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SOT23-8 | AAFD |
| MAX9141ESA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SO | - |
| MAX9142EKA-T | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SOT23-8 | AAFE |
| MAX9142ESA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SO | - |
| MAX9144EUD | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 14 TSSOP | - |
| MAX9144ESD | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 14 SO | - |

Pin Configurations


SC70/SOT23




Rail-to-Rail is a registered trademark of Nippon Motorola, Ltd.

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## ABSOLUTE MAXIMUM RATINGS

Power Supply Ranges
Supply Voltage (VCC to GND)

Differential Input Voltage .......................-0.3V to (VCC +0.3 V )
Common-Mode Input Voltage to GND ..- 0.3 V to (VCC +0.3 V )
LE Input Voltage (MAX9141 only) .........-0.3V to (VCC +0.3 V )
SHDN Input Voltage (MAX9141 only)....-0.3V to (VCC +0.3 V )
Input/Output Short-Circuit Duration to
VCC or GND $\qquad$ Continuous

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(V_{C C}=5 \mathrm{~V}, \mathrm{~V}_{C M}=0, \overline{\mathrm{SHDN}}=\overline{\mathrm{LE}}=\mathrm{V}_{C C}\right.$ (MAX9141 only), $C_{L}=15 \mathrm{pF}, \mathrm{T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{MIN}}$ to $T_{M A X}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Operating Supply Voltage | $\mathrm{V}_{\text {CC }}$ | (Note 2) |  | 2.7 |  | 5.5 | V |
| Input Voltage Range | VCMR | (Note 3) |  | -0.2 |  | $\begin{gathered} V_{C C}+ \\ 0.2 \end{gathered}$ | V |
| Input Offset Voltage | Vos | (Note 4) | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | 0.5 | 2 | mV |
|  |  |  | $\mathrm{T}_{\text {A }}=\mathrm{T}_{\text {MIN }}$ to $\mathrm{T}_{\text {MAX }}$ |  |  | 4.5 |  |
| Input Hysteresis | VHYST | (Note 5) |  |  | 1.5 |  | mV |
| Input Bias Current | IB | (Note 6) |  |  | 90 | 320 | nA |
| Input Offset Current | los |  |  |  | 8 | 120 | nA |
| Common-Mode Rejection Ratio | CMRR | $\mathrm{V}_{\mathrm{CC}}=5.5 \mathrm{~V}$ (Note 7) |  |  | 80 | 800 | $\mu \mathrm{V} / \mathrm{V}$ |
| Power-Supply Rejection Ratio | PSRR | $2.7 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 5.5 \mathrm{~V}$ |  |  | 80 | 750 | $\mu \mathrm{V} / \mathrm{V}$ |
| Output High Voltage | VOH | ISOURCE $=4 \mathrm{~mA}$ |  | $\begin{aligned} & V_{\text {CC }}- \\ & 0.425 \end{aligned}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}- \\ 0.3 \end{gathered}$ |  | V |
| Output Low Voltage | VOL | ISINK $=4 \mathrm{~mA}$ |  |  | 0.3 | 0.425 | V |
| Output Leakage Current | ILEAK | $\overline{\text { SHDN }}=$ GND, MAX9141 only (Note 8) |  |  | 0.04 | 1 | $\mu \mathrm{A}$ |
| Supply Current (Per Comparator) | ICC | $V_{C M}=V_{C C}=3 \mathrm{~V}$ | MAX9141 |  | 165 | 275 | $\mu \mathrm{A}$ |
|  |  |  | MAX9140/MAX9142/ MAX9144 |  | 150 | 250 |  |
|  |  | $V_{C M}=V_{C C}=5 \mathrm{~V}$ | MAX9141 |  | 200 | 320 |  |
|  |  |  | MAX9140/MAX9142/ MAX9144 |  | 165 | 300 |  |
|  |  | $\begin{aligned} & \text { MAX9141 only, } \overline{\text { SHDN }}=\text { GND; } \\ & \text { VCC }=\mathrm{V}_{\mathrm{CM}}=3 \mathrm{~V} \\ & \hline \end{aligned}$ |  |  | 12 | 30 |  |
| Propagation Delay | $\begin{gathered} \text { tPD }+, \\ \text { tPD- } \\ \hline \end{gathered}$ | $V_{C C}=3 V, V_{O D}=10 \mathrm{mV}$ |  |  | 40 |  | ns |
| Differential Propagation Delay | dtpD | $\mathrm{V}_{\text {OD }}=10 \mathrm{mV}$ (Note 9) |  |  | 2 |  | ns |
| Propagation Delay Skew |  | $V_{O D}=10 \mathrm{mV}$ ( Note 10) |  |  | 2 |  | ns |
| Logic Input Voltage High | $\mathrm{V}_{\text {IH }}$ | (Note 11) |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC} / 2}+ \\ 0.4 \end{gathered}$ | $\mathrm{VCc} / 2$ |  | V |
| Logic Input Voltage Low | VIL | (Note 11) |  |  | $\mathrm{V}_{\mathrm{cc} / 2}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC} / 2}- \\ 0.4 \end{gathered}$ | V |
| Logic Input Current | IIL, IIH | V LOGIC $=0$ to VCC ( Note 11) |  |  | 2 | 10 | $\mu \mathrm{A}$ |

# 40ns, Low-Power, 3V/5V, Rail-to-Rail Single-Supply Comparators 

## ELECTRICAL CHARACTERISTICS (continued)

$\left(V_{C C}=5 \mathrm{~V}, ~ V C M=0, \overline{S H D N}=\overline{L E}=V_{C C}\right.$ (MAX9141 only), $C_{L}=15 \mathrm{pF}, T_{A}=T_{M I N}$ to $T_{M A X}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP |
| :--- | :---: | :--- | :---: | :---: |
| Data-to-Latch Setup Time | ts | (Note 12) | 16 | UNITS |
| Latch-to-Data Hold Time | th | (Note 12) | 16 | ns |
| Latch Pulse Width | tLPW | (Note 12) | 45 | ns |
| Latch Propagation Delay | tLPD | (Note 12) | 60 | ns |
| Shutdown Enable Time |  | (Note 13) | 1 | ns |
| Shutdown Disable Time |  | (Note 13) | 5 | $\mu \mathrm{~s}$ |

Note 1: All devices are $100 \%$ production tested at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$. Specifications over temperature are guaranteed by design.
Note 2: Inferred from PSRR test.
Note 3: Inferred from CMRR test. Note also that either or both inputs can be driven to the absolute maximum limit ( 0.3 V beyond either supply rail) without damage or false output inversion.
Note 4: Vos is defined as the center of the input-referred hysteresis zone. See Figure 1.
Note 5: The input-referred trip points are the extremities of the differential input voltage required to make the comparator output change state. The difference between the upper and lower trip points is equal to the width of the input-referred hysteresis zone. See Figure 1.
Note 6: The polarity of $\mathrm{I}_{\mathrm{B}}$ reverses direction as $\mathrm{V}_{\mathrm{CM}}$ approaches either supply rail.
Note 7: Specified over the full common-mode voltage range (VCMR).
Note 8: Specification is for current flowing into or out of the output pin for VOUT driven to any voltage from VCc to GND while the part is in shutdown.
Note 9: Specified between any two channels in the MAX9142/MAX9144.
Note 10: Specified as the difference between tPD+ and tpD- for any one comparator.
Note 11: Applies to the MAX9141 only for both $\overline{\mathrm{SHDN}}$ and $\overline{\mathrm{LE}}$.
Note 12: Applies to the MAX9141 only. Comparator is active with $\overline{L E}$ driven high and is latched with $\overline{\mathrm{LE}}$ driven low (VOD $=10 \mathrm{mV})$. See Figure 2.
Note 13: Applicable to the MAX9141 only. Comparator is active with the $\overline{\mathrm{SHDN}}$ driven high and is shutdown with $\overline{\mathrm{SHDN}}$ driven low. Shutdown enable time is the delay when the $\overline{\mathrm{SHDN}}$ is driven high to the time the output is valid. Shutdown disable time is the delay when the $\overline{\mathrm{SHDN}}$ is driven low to the time the comparator shuts down.

## 40ns, Low-Power, 3V/5V, Rail-to-Rail Single-Supply Comparators

$\left(\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=0, \mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{V}_{\mathrm{OD}}=10 \mathrm{mV}, \mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}\right.$, unless otherwise noted.)


OUTPUT HIGH VOLTAGE vs. TEMPERATURE


OUTPUT SHORT-CIRCUIT (SOURCE) CURRENT vs. TEMPERATURE


OUTPUT LOW VOLTAGE vs. TEMPERATURE


OUTPUT SHORT-CIRCUIT (SINK) CURRENT vs. TEMPERATURE


INPUT BIAS CURRENT vs. TEMPERATURE


# 40ns, Low-Power, 3V/5V, Rail-to-Rail Single-Supply Comparators 

Typical Operating Characteristics (continued)
$\left(\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=0, \mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{V}_{\mathrm{OD}}=10 \mathrm{mV}, \mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}\right.$, unless otherwise noted.)


INPUT VOLTAGE RANGE
vs. TEMPERATURE


PROPAGATION DELAY
vs. TEMPERATURE


TRIP POINT vs. TEMPERATURE


PROPAGATION DELAY
vs. INPUT OVERDRIVE


PROPAGATION DELAY
vs. CAPACITIVE LOAD


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Typical Operating Characteristics (continued)
( $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=0, \mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{V}_{\mathrm{OD}}=10 \mathrm{mV}, \mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)




Pin Description

| PIN |  |  |  | NAME | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| MAX9140 | MAX9141 | MAX9142 | MAX9144 |  |  |
| - | - | 1 | 1 | OUTA | Comparator A Output |
| - | - | 2 | 2 | INA- | Comparator A Inverting Input |
| - | - | 3 | 3 | INA+ | Comparator A Noninverting Input |
| 5 | 1 | 8 | 4 | VCC | Positive Supply |
| - | - | 5 | 5 | INB+ | Comparator B Noninverting Input |
| - | - | 6 | 6 | INB- | Comparator B Inverting Input |
| - | - | 7 | 7 | OUTB | Comparator B Output |
| - | - | - | 8 | OUTC | Comparator C Output |
| - | - | - | 9 | INC- | Comparator C Inverting Input |
| - | - | - | 10 | INC+ | Comparator C Noninverting Input |
| 2 | 4 | 4 | 11 | GND | Ground |
| - | - | - | 12 | IND+ | Comparator D Noninverting Input |
| - | - | - | 13 | IND- | Comparator D Inverting Input |
| - | - | - | 14 | OUTD | Comparator D Output |
| 3 | 2 | - | - | $\mathrm{IN}+$ | Noninverting Input |
| 4 | 3 | - | - | IN- | Inverting Input |
| - | 6 | - | - | $\overline{\text { SHDN }}$ | Shutdown: MAX9141 is active when SHDN is driven high; MAX9141 is in shutdown when $\overline{\text { SHDN }}$ is driven low. |
| - | 5 | - | - | $\overline{\mathrm{LE}}$ | The output is latched when $\overline{\mathrm{LE}}$ is low. The latch is transparent when $\overline{\mathrm{LE}}$ is high. |
| 1 | 7 | - | - | OUT | Comparator Output |
| - | 8 | - | - | N.C. | No Connection. Not internally connected. |

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## Detailed Description

The MAX9140／MAX9141／MAX9142／MAX9144 single－ supply comparators feature internal hysteresis，high speed，and low power．Their outputs are pulled to with－ in 300 mV of either supply rail without external pullup or pulldown circuitry．Rail－to－rail input voltage range and low－voltage single－supply operation make these devices ideal for portable equipment．The MAX9140／MAX9141／MAX9142／MAX9144 interface directly to CMOS and TTL logic．
Most high－speed comparators oscillate in the linear region because of noise or undesired parasitic feed－ back．This tends to occur when the voltage on one input is at or equal to the voltage on the other input．To counter the parasitic effects and noise，the MAX9140／ MAX9141／MAX9142／MAX9144 have an internal hystere－ sis of 1.5 mV ．
The hysteresis in a comparator creates two trip points： one for the rising input voltage and one for the falling input voltage（Figure 1）．The difference between the trip points is the hysteresis．The average of the trip points is the offset voltage．When the comparator＇s input volt－ ages are equal，the hysteresis effectively causes one comparator input voltage to move quickly past the other，thus taking the input out of the region where oscillation occurs．Standard comparators require hys－ teresis to be added with external resistors．The MAX9140／MAX9141／MAX9142／MAX9144＇s fixed internal hysteresis eliminates these resistors．To increase hys－ teresis and noise margin even more，add positive feed－ back with two resistors as a voltage divider from the output to the noninverting input．
Figure 1 illustrates the case where IN －is fixed and $\mathrm{IN}+$ is varied．If the inputs were reversed，the figure would look the same，except the output would be inverted．
The MAX9141 includes an internal latch that allows storage of comparison results．The $\overline{\mathrm{LE}}$ pin has a high input impedance．If $\overline{\mathrm{LE}}$ is high，the latch is transparent （i．e．，the comparator operates as though the latch is not present）．The comparator＇s output state is latched when $\overline{\mathrm{LE}}$ is pulled low（Figure 2）．

## Shutdown Mode（MAX9141 Only）

The MAX9141 shuts down when the SHDN pin is low． When shut down，the supply current drops to less than $12 \mu \mathrm{~A}$ ，and the three－state output becomes high imped－ ance．The $\overline{\text { SHDN }}$ pin has a high－input impedance． Connect $\overline{S H D N}$ to VCC for normal operation．Exit shut－ down with $\overline{\mathrm{LE}}$ high（transparent state）；otherwise，the output will be indeterminate．


Figure 1．Input and Output Waveform，Noninverting Input Varied

Input Stage Circuitry
The MAX9140／MAX9141／MAX9142／MAX9144 include internal protection circuitry that prevents damage to the precision input stage from large differential input volt－ ages．This protection circuitry consists of two back－to－ back diodes between IN＋and IN－as well as two series $4.1 \mathrm{k} \Omega$ resistors（Figure 3）．The diodes limit the differen－ tial voltage applied to the internal circuitry of the com－ parators to be no more than $2 \mathrm{~V}_{\mathrm{F}}$ ，where $\mathrm{V}_{\mathrm{F}}$ is the for－ ward voltage drop of the diode（about 0.7 V at $+25^{\circ} \mathrm{C}$ ）．
For a large differential input voltage（exceeding $2 \mathrm{~V}_{\mathrm{F}}$ ）， this protection circuitry increases the input bias current at $\mathrm{IN}+$（source）and IN －（sink）．

$$
\text { Input Current }=\frac{(\mathrm{IN}+-\mathrm{IN}-)-2 \mathrm{~V}_{\mathrm{F}}}{2 \times 4.1 \mathrm{k} \Omega}
$$

Input current with large differential input voltages should not be confused with input bias current（IB）．As long as the differential input voltage is less than $2 \mathrm{~V}_{\mathrm{F}}$ ， this input current is equal to $\mathrm{IB}_{\mathrm{B}}$ ．The output is in the cor－ rect logic state if one or both inputs are within the com－ mon－mode range．

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Figure 2. MAX9141 Timing Diagram with Latch Operator

Output Stage Circuitry
The MAX9140/MAX9141/MAX9142/MAX9144 contain a current-driven output stage as shown in Figure 4. During an output transition, ISOURCE or ISINK is pushed or pulled to the output pin. The output source or sink current is high during the transition, creating a rapid slew rate. Once the output voltage reaches VOH or VOL, the source or sink current decreases to a small value, capable of maintaining the VOH or $\mathrm{VOL}_{\mathrm{OL}}$ static condition. This significant decrease in current conserves power after an output transition has occurred.
One consequence of a current-driven output stage is a linear dependence between the slew rate and the load capacitance. A heavy capacitive load will slow down a voltage output transition. This can be useful in noisesensitive applications where fast edges may cause interference.

## Applications Information

## Circuit Layout and Bypassing

The high-gain bandwidth of the MAX9140/MAX9141/ MAX9142/MAX9144 requires design precautions to realize the full high-speed capabilities of these comparators. The recommended precautions are:

1) Use a printed circuit board with a good, unbroken, low-inductance ground plane.
2) Place a decoupling capacitor (a $0.1 \mu \mathrm{~F}$ ceramic capacitor is a good choice) as close to $\mathrm{V}_{\mathrm{Cc}}$ as possible.


Figure 3. Input Stage Circuitry
3) Pay close attention to the decoupling capacitor's bandwidth, keeping leads short.
4) On the inputs and outputs, keep lead lengths short to avoid unwanted parasitic feedback around the comparators.
5) Solder the device directly to the printed circuit board instead of using a socket.

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Figure 4．Output Stage Circuitry


Figure 6．Line Receiver Application


Figure 5．3．3V Digitally Controlled Threshold Detector
Chip Information
MAX9140 TRANSISTOR COUNT： 158
MAX9141 TRANSISTOR COUNT： 192
MAX9142 TRANSISTOR COUNT： 314
MAX9144 TRANSISTOR COUNT： 620
PROCESS：Bipolar

## 40ns, Low-Power, 3V/5V, Rail-to-Rail Single-Supply Comparators



# 40ns，Low－Power，3V／5V，Rail－to－Rail Single－Supply Comparators 

Package Information（continued）


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