

## Quad DC/DC Converter for Triple Output TFT Supply Plus LED Driver

#### **FEATURES**

- Triple Output TFT Supply with Built-In Power Sequencing
- Integrated White LED Backlight Driver
- Integrated Schottky Diodes
- Active Ballast Circuitry Ensures Precise Current Matching in White LEDs
- Low Noise 1MHz Fixed Frequency Operation
- 2.6V to 16V Input Voltage Range
- Soft-Start Limits Inrush Current
- TFT Supply Output Voltages Independently Set (Not Charge Pump Derived)
- Power Good and Output Disconnect for TFT Supply
- Built-In LED Dimming Capability
- Open LED Protection for LED Driver
- 24-Lead QFN Package (4mm × 4mm)

#### **APPLICATIONS**

- Poly-Silicon TFT Displays
- Amorphous Silicon TFT Displays

#### DESCRIPTION

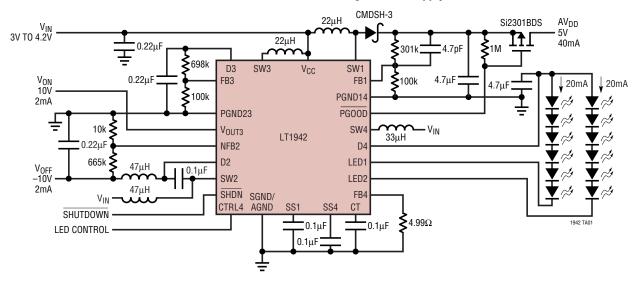
The LT®1942 is a highly integrated quad switching regulator designed to provide all necessary power supply functions for TFT displays, including the white LED backlight driver. The TFT supply incorporates two boost and one inverting DC/DC converters. The TFT supply output voltages are independently set, unlike charge pump solutions which have many limitations. Power sequencing for the TFT supply is built into the part and is user programmable.

The LED driver is a boost DC/DC converter that can be independently controlled. The LED driver has built-in dimming control for precise control of LED current. An 8:1 dimming range is achieved by adjusting the CTRL4 voltage. The user can elect to drive a single string or two strings of LEDs. An LED ballast circuit is included to precisely match the LED currents if two strings of LEDs are used. Soft-start is built into the LED driver as well as the primary TFT supply.

The LT1942 is available in a low profile (0.75mm) 24-lead QFN (4mm  $\times$  4mm) package.

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#### TFT Bias and White LED Backlight Power Supply

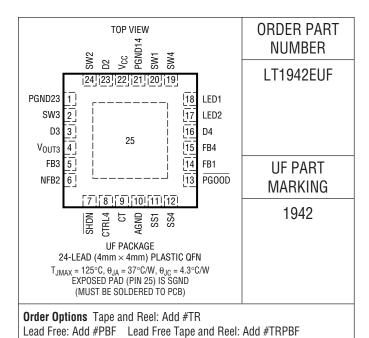




#### **ABSOLUTE MAXIMUM RATINGS**

(Note 1)
V <sub>CC</sub> Voltage16V
SW1, SW2, SW3 Voltage0.4V to 36V
SW40.4V to 45V
FB1, FB3 Voltage 2.5V
FB4 Voltage
NFB2 Voltage ————————————————————————————————————
SHDN Voltage
CTRL4 Voltage
SS1, SS4 Voltage 1.5V
Current Into D2
D2 Voltage –36V
Current Out of D3 1A
D3 Voltage 36V
Current Out of D4 4A
D4 Voltage 45V
LED1, LED2 Voltage
Current Into LED1, LED2
PGOOD Voltage
V <sub>OUT3</sub> Voltage
CT Current ±1mA
Maximum Junction Temperature
Operating Temperature Range (Note 2)40°C to 85°C
Storage Temperature Range65°C to 125°C

### PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

Lead Free Part Marking: http://www.linear.com/leadfree/

## **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25 \,^{\circ}\text{C}$ . $V_{IN} = 3.3V$ , $V_{\overline{SHDN}} = V_{IN}$ , unless otherwise noted. (Note 2)

PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
Quiescent Current	V <sub>SHDN</sub> = 3.3V, V <sub>CTRL4</sub> = 0V, Not Switching V <sub>SHDN</sub> = V <sub>CTRL4</sub> = 0V, In Shutdown	•		7 0	10 1	mA μA
Input Voltage Range	V <sub>CC</sub> Pin	•	2.6		16	V
AV <sub>DD</sub> Feedback Voltage	FB1 Pin	•	1.225	1.25	1.275	V
AV <sub>DD</sub> Feedback Pin Bias Current (FB1)	FB1 Pin (Note 3)			15	60	nA
AV <sub>DD</sub> Feedback Voltage for Power Good Signal	FB1 Pin Voltage as a Percent of Nominal Voltage	•	95	97	100	%
V <sub>OFF</sub> Feedback Voltage	NFB2 Pin		-160	-150	-140	mV
V <sub>OFF</sub> Feedback Pin Bias Current (NFB2)	NFB2 Pin (Note 3)			25	60	nA
V <sub>ON</sub> Feedback Voltage	FB3 Pin		1.225	1.25	1.275	V
V <sub>ON</sub> Feedback Pin Bias Current (FB3)	FB3 Pin (Note 3)			25	60	nA
LED Feedback Voltage	FB4 Pin, V <sub>CTRL4</sub> = 3.3V	•	190	200	210	mV
LED Feedback Pin Bias Current (FB4)	FB4 Pin (Note 3)			20	60	μΑ
LED Feedback Voltage with Dimming	FB4 Pin, V <sub>CTRL4</sub> = 1V		85	95	105	mV

LINEAR

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PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
LED1-LED2 Current Matching	Current Into Each Pin = 20mA, V <sub>LED1</sub> = V <sub>LED2</sub> Current Into Each Pin = 20mA,  V <sub>LED1</sub> -V <sub>LED2</sub>   = 1V			0 1	1.5 2.5	% %
LED1 Voltage	20mA Into Pin, V <sub>LED2</sub> = 2V, FB4 = 0			1	1.2	V
LED2 Voltage	20mA Into Pin, V <sub>LED1</sub> = 2V, FB4 = 0			1	1.2	V
LED1, LED2 Maximum Current	Into Either Pin, V <sub>LEDX</sub> < 1.5V, FB4 = 0		35			mA
LED1, LED2 Leakage Current	$V_{\overline{SHDN}} = 0V, V_{LEDX} = 45V$				1	μА
AV <sub>DD</sub> Maximum Duty Cycle	SW1 Pin		88	93		%
V <sub>OFF</sub> Maximum Duty Cycle	SW2 Pin		75	86		%
V <sub>ON</sub> Maximum Duty Cycle	SW3 Pin		75	86		%
LED Maximum Duty Cycle	SW4 Pin		88	93		%
AV <sub>DD</sub> Switch Current Limit at Minimum DC (SW1)	SW1 Pin (Note 4)		150	200	250	mA
AV <sub>DD</sub> Switch Current Limit at Maximum DC (SW1)	SW1 Pin (Note 4)		80	140	200	mA
V <sub>OFF</sub> Switch Current Limit at Minimum DC (SW2)	SW2 Pin (Note 4)		50	80	110	mA
V <sub>OFF</sub> Switch Current Limit at Maximum DC (SW2)	SW2 Pin (Note 4)		30	60	90	mA
V <sub>ON</sub> Switch Current Limit at Minimum DC (SW3)	SW3 Pin (Note 4)		50	80	110	mA
V <sub>ON</sub> Switch Current Limit at Maximum DC (SW3)	SW3 Pin (Note 4)		30	60	90	mA
LED Switch Current Limit at Minimum DC (SW4)	SW4 Pin (Note 4)		550	750	900	mA
LED Switch Current Limit at Maximum DC (SW4)	SW4 Pin (Note 4)		450	600	850	mA
V <sub>OFF</sub> Schottky Diode Forward Drop	D2 Pin, I = 60mA			0.65		V
V <sub>OFF</sub> Schottky Diode Leakage Current	D2 Pin, $V_{D2} = -36V$	•			1	μА
V <sub>ON</sub> Schottky Diode Forward Drop	D3 Pin, I = 60mA			0.65		V
V <sub>ON</sub> Schottky Diode Leakage Current	$V_{D3} = 36V$ , SW3 = 0V, Output Disconnect PNP Off	•			1	μА
LED Schottky Diode Forward Drop	I = 250mA			0.75		V
LED Schottky Diode Leakage Current	V <sub>D4</sub> = 36V, SW4 = 0V	•			1	μΑ
Switching Frequency			0.8	1	1.2	MHz
SHDN Pin Current	$V_{\overline{SHDN}} = 3V$ $V_{\overline{SHDN}} = 0V$	•		70 20	100 30	μA nA
CTRL4 Pin Current	V <sub>CTRL4</sub> = 3V V <sub>CTRL4</sub> = 0V	•		30 -60	60 -150	μA nA
SW1-SW4 leakage Current	V <sub>SWX</sub> = 36V	•		0.01	1	μА
SW1 V <sub>CESAT</sub>	I <sub>SW1</sub> = 100mA (Note 5)			200	300	mV
SW2 V <sub>CESAT</sub>	I <sub>SW2</sub> = 40mA (Note 5)			150	250	mV
SW3 V <sub>CESAT</sub>	I <sub>SW3</sub> = 40mA (Note 5)			150	250	mV
SW4 V <sub>CESAT</sub>	I <sub>SW4</sub> = 400mA (Note 5)			280	400	mV
SS1 Charging Current	V <sub>SS</sub> = 0.5V		1	3	5	μА
SS4 Charging Current	V <sub>SS</sub> = 0.5V		1	3	5	μА
SHDN Input Voltage High			1			V
SHDN Input Voltage Low					0.4	V
CTRL4 Input Voltage High			0.25			V
CTRL4 Input Voltage Low					0.1	V



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PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
PGOOD Leakage Current	$\overline{PGOOD}$ NPN Off, $V_{\overline{PGOOD}} = 5V$		0	0.1	μА
PGOOD Sink Current	PGOOD NPN On, V <sub>PGOOD</sub> = 1V	0.5	0.7		mA
CT Charging Current			12		μА
CT Reference Voltage			0.8		V
V <sub>OUT3</sub> Leakage Current	Output Disconnect PNP Off, D3 = 36V, V <sub>OUT</sub> = 0V		0	1	μА
V <sub>OUT3</sub> Source Current	Output Disconnect PNP On, D3 = 3V, V <sub>OUT</sub> = 2.5V	10	15		mA
Minimum D3/V <sub>OUT3</sub> Operating Voltage	Minimum D3 Voltage for Proper Output Disconnect PNP Operation		3		V
D4 LED Open-Circuit Voltage		38	42	44	V

**Note 1:** Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

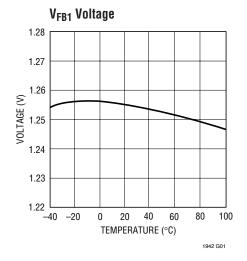
**Note 2:** The LT1942 is guaranteed to meet performance specifications from  $0^{\circ}$ C to  $70^{\circ}$ C. Specifications over the  $-40^{\circ}$ C to  $85^{\circ}$ C operating temperature range are assured by design, characterization and correlation with statistical controls.

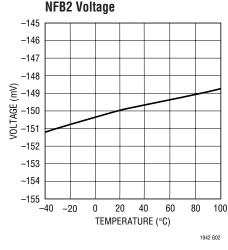
Note 3: Current flows out of the pin.

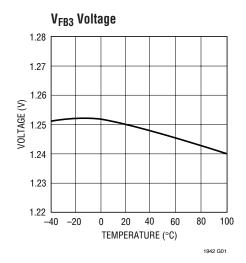
**Note 4:** Current limit guaranteed by design and/or correlation to static test.

Note 5: V<sub>CESAT</sub> 100% tested at wafer level.

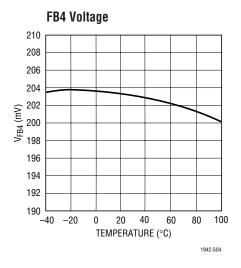
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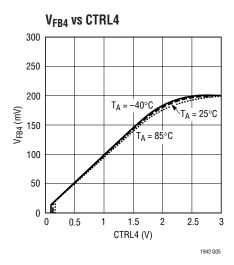


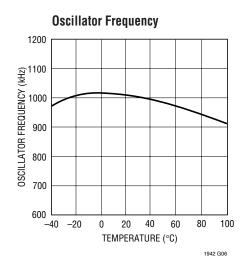


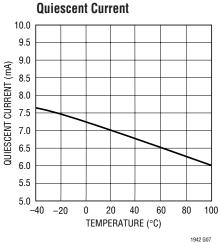


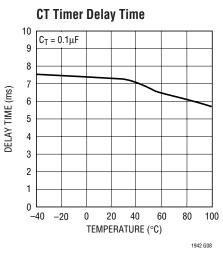
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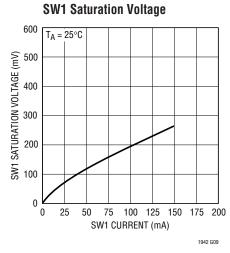


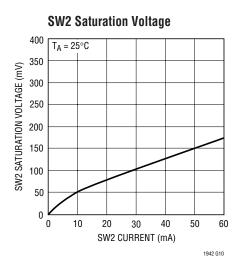


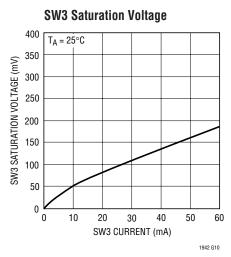


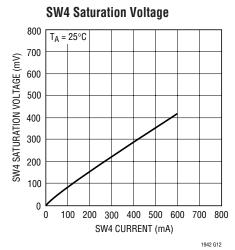




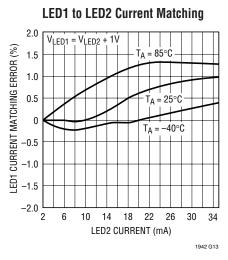


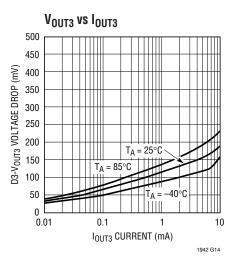


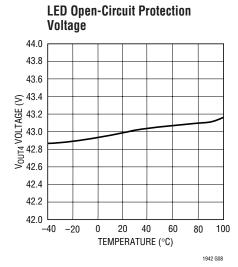


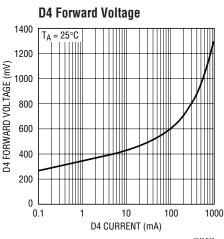


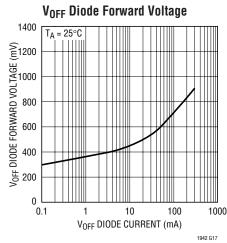
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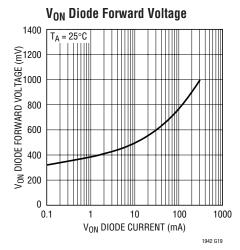


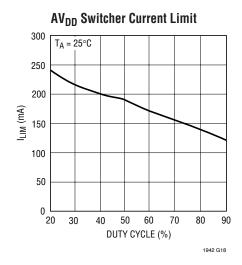


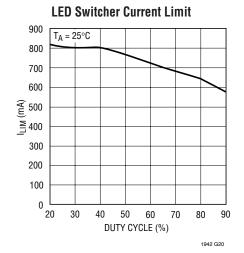
















#### PIN FUNCTIONS

**PGND23 (Pin 1):** Power Ground for V<sub>OFF</sub> and V<sub>ON</sub> Switcher. Tie directly to local ground plane.

**SW3 (Pin 2):** Switch Pin for V<sub>ON</sub> Switcher. This is the collector of an internal NPN power switch. Minimize the metal trace area connected to this pin to minimize EMI.

**D3 (Pin 3):** Output of V<sub>ON</sub> Switcher. Place output decoupling capacitor from this pin to local ground plane.

 $V_{OUT3}$  (Pin 4): Collector of Internal Output Disconnect PNP Transistor for the  $V_{ON}$  Switcher. Place a capacitor from this pin to local ground plane of at least  $0.1\mu F$ .

**FB3 (Pin 5):** Feedback Pin for  $V_{ON}$  Switcher. Reference voltage is 1.25V. Connect resistive divider tap here. Minimize trace area at FB3. Set  $V_{OUT3}$  according to  $V_{OUT3}$  = 1.25(1 + R5/R6).

**NFB2** (Pin 6): Feedback Pin for  $V_{OFF}$  Switcher. Reference voltage is -150mV. Connect resistive divider tap here. Minimize trace area at NFB2. Set  $V_{OUT2}$  according to  $V_{OUT2} = (-0.15)(1 + R3/R4)$ .

**SHDN** (Pin 7): Shutdown. Tie to 1V or greater to enable the part. Tie to 0.4V or lower to disable. This pin is the global shutdown pin for all four switching regulators.

**CTRL4 (Pin 8):** Shutdown for the LED Switcher. Tie to 0.25V or greater to enable the LED switcher. Tie to 0.1V or lower to disable. CTRL4 can also override the default 200mV reference for the LED switcher. If CTRL4 is tied to 2.5V or greater, the reference voltage is 200mV. If CTRL4 is less than 2.5V, the LED reference voltage is approximately  $V_{CTRL4}/10$ . This allows an 8:1 dimming range for the LEDs. Refer to the Applications Section for more information.

**CT (Pin 9):** Programmable Delay for the TFT Supplies. Place a capacitor from this pin to local ground plane. After the delay, the output disconnect PNP for the  $V_{ON}$  supply is turned on. The delay time is initiated when  $AV_{DD}$  reaches 97% of its final value.

**AGND (Pin 10):** Analog Ground. Tie directly to local ground plane.

**SS1** (Pin 11): Soft-Start Pin for  $AV_{DD}$  Switcher. Place a soft-start capacitor from this pin to ground. Upon start-up,  $3\mu A$  of current charges the capacitor to 1.25V. Use larger capacitor for slower start-up. Leave floating if not in use.

**SS4** (Pin 12): Soft-Start for LED Switcher. Place a soft-start capacitor from this pin to ground. Upon start-up,  $3\mu$ A of current charges the capacitor to 1.25V. Use larger capacitor for slower start-up. Leave floating if not in use.

**PG00D** (Pin 13): Power Good Indication for  $AV_{DD}$  Switcher. Open-collector NPN. Stays open until the  $V_{FB1}$  voltage reaches 97% of target value at which point the pin will pull down.

**FB1** (Pin 14): Feedback Pin for  $AV_{DD}$  Switcher. Connect resistive divider tap here. Minimize trace area at FB1. Set  $V_{OUT1}$  according to  $V_{OUT1} = 1.25(1 + R1/R2)$ .

**FB4 (Pin 15):** Feedback Pin for LED Switcher. Place a resistor from this node to ground. Choose R7 to set LED current ( $I_{LED}$ ) according to the following formula:

if 
$$V_{CTRL4} \ge 2.5V$$
, R7 =  $0.2/I_{LED}$ 

In this case,  $I_{LED}$  is the total LED current. If two strings of LEDs are used,  $I_{LED}$  is the sum of the current in the two strings. If only a single string of LEDs is used,  $I_{LED}$  is simply the single string current. For dimming LED current, drive  $V_{CTRL4}$  to a lower voltage. See the Applications Section for more information.

**D4 (Pin 16):** Output of LED Switcher. Place output decoupling capacitor from this pin to local ground plane.

**LED2 (Pin 17):** Second Input of Current Mirror for an LED String. Connect the negative end of second string of LEDs to this pin. If not in use, leave floating.



## PIN FUNCTIONS

**LED1 (Pin 18):** First Input of Current Mirror for an LED String. Connect the negative end of first string of LEDs to this pin. If not in use, leave floating.

**SW4 (Pin 19):** Switch Pin for LED Switcher. This is the collector of an internal NPN power switch. Minimize the metal trace area connected to this pin to minimize EMI.

**SW1 (Pin 20):** Switch Pin for AV<sub>DD</sub> Switcher. This is the collector of an internal NPN Power switch. Minimize the metal trace area connected to this pin to minimize EMI.

**PGND14 (Pin 21):** Power Ground for LED Switcher and AV<sub>DD</sub> Switcher. Tie directly to local ground plane.

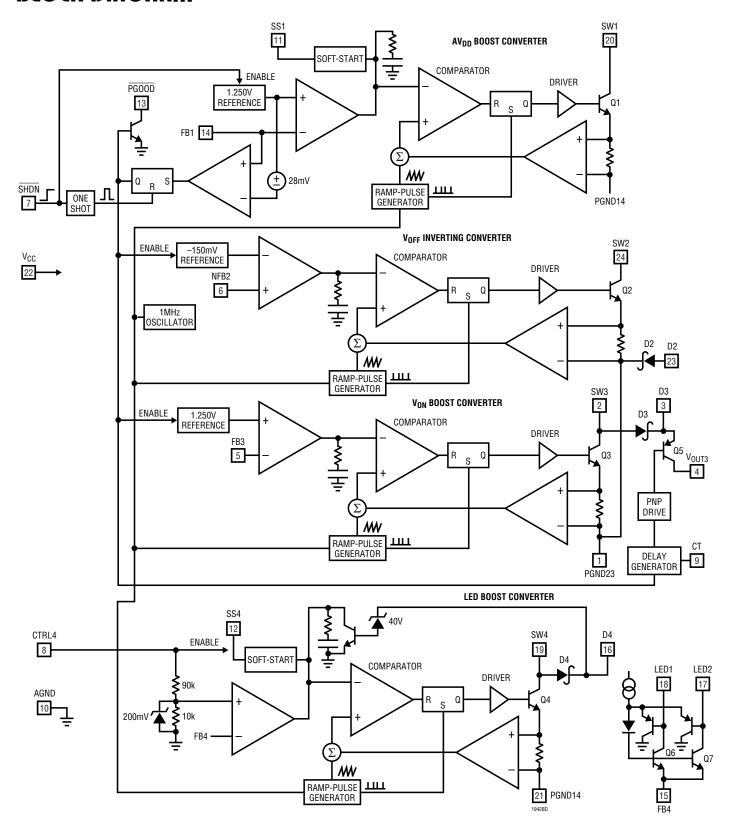
**V<sub>CC</sub> (Pin 22):** Input Supply. Must be locally bypassed.

**D2 (Pin 23):** Anode Connection of Internal Schottky Diode for  $V_{\text{OFF}}$  Switcher.

**SW2 (Pin 24):** Switch Pin for  $V_{OFF}$  Switcher. This is the collector of an internal NPN Power switch. Minimize the metal trace area connected to this pin to minimize EMI.

**SGND (Pin 25):** Signal Ground. This is the Exposed Pad, which must be soldered directly to a local ground plane to achieve optimum thermal performance.

## **BLOCK DIAGRAM**



#### **OPERATION**

Refer to the figure on the first page of this data sheet and the Block Diagram when following the description of the operation of the LT1942. The LT1942 contains four switching regulators. Three switchers are designed to be used as the power supply for driving small to medium sized TFT displays. The fourth switcher is designed to power an LED backlight. All four switching regulators use a constant frequency, current mode control scheme to provide excellent line and load regulation.

#### **TFT Supply**

The three switching regulators for the TFT supply are named AV<sub>DD</sub>,  $V_{OFF}$  and  $V_{ON}$ . Both the AV<sub>DD</sub> and  $V_{ON}$ switchers are boost type, while the  $V_{OFF}$  is an inverting type. Power sequencing circuitry is integrated into the LT1942 to properly power up the TFT power supply. When the SHDN pin is driven above 1V initially, only the  $AV_{DD}$ switcher is enabled. After the output of the AV<sub>DD</sub> switcher reaches 97% of its final value, the PGOOD pin is driven low and both the  $V_{OFF}$  and  $V_{ON}$  switchers are enabled. This event is latched rejecting any later drops on AV<sub>DD</sub> below 97%. The output-disconnect circuitry (Q5 in the Block Diagram) for the V<sub>ON</sub> switcher is not activated until the programmable delay set by the CT pin has elapsed. Once the output-disconnect circuitry is enabled, the  $V_{OUT3}$  pin is driven to nearly the same voltage as D3. The  $V_{OUT3}$  pin can only source current and is limited to 15mA nominally. All three switching regulators are internally compensated and operated at a fixed frequency of 1MHz. The AV<sub>DD</sub> switching regulator has integrated soft-start. Simply place a capacitor from the SS1 pin to GND to implement soft-start.

#### **LED Supply**

The fourth switching regulator in the LT1942 is designed to drive up to 20 LEDs (2 strings of 10) to power a backlight. The LED switcher has an integrated current ballaster (Q6 and Q7 in the Block Diagram) which enables two strings of LEDs to have active current matching. The current into LED1 and LED2 will always be actively matched, regardless of which LED string has the highest voltage drop. The LED ballast circuitry can be left unused when driving a single string of LEDs. Simply leave the LED1 and LED2 inputs floating and return the LED string directly to the FB4 pin. The LED switcher has an independent control pin which serves two functions: shutdown and reference adjustment. The CTRL4 pin needs to be driven above 0.25V to enable the LED switcher. When CTRL4 is between 0.25V and 2.5V, the internal reference for the LED switcher is attenuated. If CTRL4 is above 2.5V, the internal reference is limited to 200mV no matter how high CTRL4 is. This feature allows an 8:1 dimming range for the LEDs. The LED switching regulator is internally compensated and operates at a fixed frequency of 1MHz. The LED switching regulator has LED open-circuit protection. If any LED fails open, D4 is not allowed to exceed 42V (typical). This prevents damage to the power switch. The LED switching regulator also has built in soft-start. By placing a capacitor from the SS4 pin to GND, the user can program the soft-start time.



#### APPLICATIONS INFORMATION

#### **Setting the Output Voltages**

Unlike charge-pump-based TFT supplies, the LT1942 TFT outputs can all be independently set. The AV<sub>DD</sub>, V<sub>OFF</sub> and V<sub>ON</sub> outputs are all inductively based DC/DC converters. As such, the output voltages are set by choosing the right feedback divider ratio. The AV<sub>DD</sub> feedback pin is FB1 and the reference voltage is 1.25V. The V<sub>OFF</sub> feedback pin is NFB2 and the reference voltage is  $-150 \, \text{mV}$ . The V<sub>ON</sub> feedback pin is FB3 and the reference voltage is 1.25V. The LED driver is also an inductively based DC/DC converter.

#### **Output Disconnect**

The  $V_{ON}$  switching regulator has built in output disconnect. When the  $V_{ON}$  supply is enabled, the boost converter will charge up the D3 node. With the output disconnect PNP (Q5) turned off, the voltage on  $V_{OUT3}$  will be zero. Once the power sequencing circuitry turns on the PNP,  $V_{OUT3}$  will go to nearly the same voltage as D3.

The  $V_{OFF}$  switching regulator naturally achieves output disconnect due to the switching topology (inverting configuration). When the LT1942 is in shutdown,  $V_{OFF}$  will go to zero volts.

The  $AV_{DD}$  switching regulator provides a  $\overline{PGOOD}$  pin to be used for an external output disconnect function. This pin can drive the gate of a PMOS device. When the LT1942 is first enabled, the  $AV_{DD}$  switching regulator begins to charge up the output capacitor. Since  $\overline{PGOOD}$  is still an open collector at this point, the PMOS device is turned off, leaving the  $AV_{DD}$  output at zero volts. Once the output capacitor charges to 97% of the final value,  $\overline{PGOOD}$ 

latches low turning on the PMOS device and bringing the  $AV_{DD}$  output quickly to the desired voltage.

#### Soft-Start

The higher current regulators,  $AV_{DD}$  and LED, have user programmable soft-start functions built into the part. Simply place a capacitor from the SS1 pin to GND to achieve soft-start for the  $AV_{DD}$  switcher. For the LED switcher, place a capacitor from the SS4 pin to GND. A good value for both soft-start capacitors is  $0.1\mu F$ . The soft-start time is dependent on the load characteristics which will vary depending on the type and size of the TFT display. It is a good practice to experiment to determine the best value for the soft-start capacitors.

#### **Power Sequencing**

The TFT supply has integrated supply sequencing built-in to ensure proper initiation of the TFT display. Figure 1 shows a state diagram of the power sequencing circuitry. The power sequencing circuitry insures that the  $AV_{DD}$  supply comes up first, then the negative  $V_{OFF}$  supply and lastly the  $V_{ON}$  positive supply. To pick the CT capacitor for a particular delay, use the following equation:

$$C_{T} = \frac{10\mu A \cdot t_{DELAY}}{0.7V}$$

Figure 2 shows the power sequencing during start-up of the TFT power supply for the circuit shown in the figure on the first page of this data sheet. All bias supplies start up in a well controlled and well timed manner. The LED backlight driver is independently controlled via the CTRL4 pin.

## **APPLICATIONS INFORMATION**

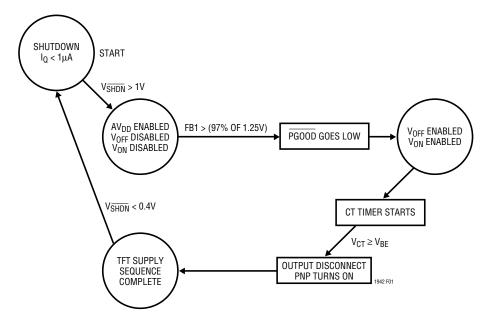


Figure 1. Power Sequencing State Diagram

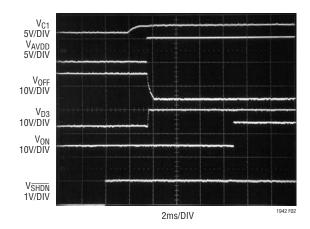


Figure 2. Power Sequencing During TFT Supply Start-Up (Refer to Figure 4 Node Names)

#### APPLICATIONS INFORMATION

#### **Layout Hints**

Proper layout techniques must be followed to achieve advertised performance of the part. Keep all SW and FB traces short and small. There are several power ground pins on the LT1942 and it is important to implement the grounding properly. PGND14 is the power ground pin for the  $AV_{DD}$  and LED switchers. The output capacitor for both the  $AV_{DD}$  and LED switcher must be returned back to this pin before being allowed to mix into the ground plane.

Similarly, PGND23 is the power ground pin for the  $V_{ON}$  and  $V_{OFF}$  switcher. Again, both output capacitors for the  $V_{ON}$  and  $V_{OFF}$  switchers need to return to the PGND23 pin before mixing with the ground plane. Place the input decoupling capacitor, which ties from the  $V_{CC}$  pin to GND, as close to the part as possible. Please refer to Figure 3 for a recommended layout of the power path components. The component names are those shown in the circuit of Figure 4.

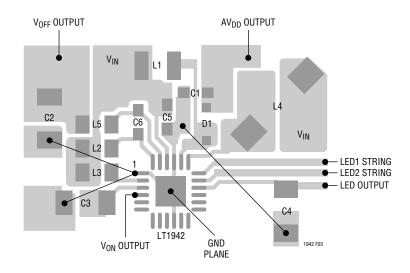


Figure 3: Power Path Component Recommended Layout (Refer to Figure 4 Component Names)

## TYPICAL APPLICATIONS

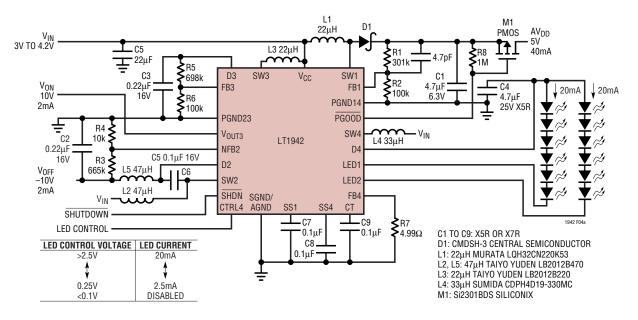
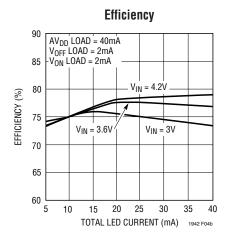
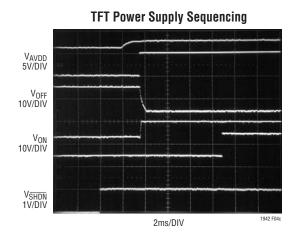
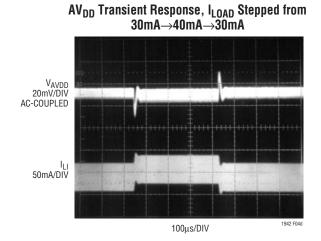


Figure 4. TFT Bias (5V, 10V, -10V) and White LED Backlight Supply from Single Li-lon Cell



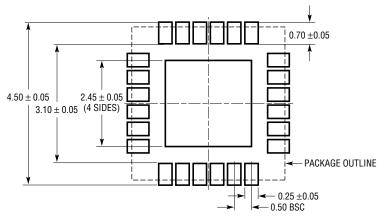




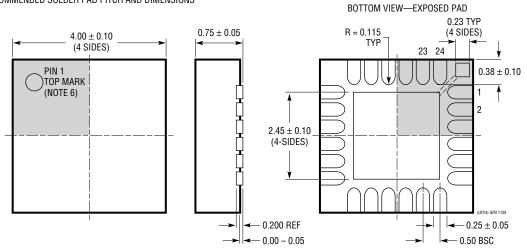
#### PACKAGE DESCRIPTION

#### **UF Package** 24-Lead Plastic QFN (4mm × 4mm)

(Reference LTC DWG # 05-08-1697)



RECOMMENDED SOLDER PAD PITCH AND DIMENSIONS



#### NOTE:

- 1. DRAWING PROPOSED TO BE MADE A JEDEC PACKAGE OUTLINE MO-220 VARIATION (WGGD-X)—TO BE APPROVED 2. DRAWING NOT TO SCALE

- 3. ALL DIMENSIONS ARE IN MILLIMETERS
   4. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE, IF PRESENT
- 5. EXPOSED PAD SHALL BE SOLDER PLATED
- 6. SHADED AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION ON THE TOP AND BOTTOM OF PACKAGE



## **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LT1613	550mA (I <sub>SW</sub> ), 1.4MHz High Efficiency Step-Up DC/DC Converter	$V_{IN}$ : 0.9V to 10V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 3mA, $I_{SD}$ < 1 $\mu$ A, ThinSOT Package
LT1615/LT1615-1	300mA/80mA (I <sub>SW</sub> ), Constant Off-Time, High Efficiency Step-Up DC/DC Converter	$V_{IN}$ : 1.2V to 15V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 20 $\mu$ A, $I_{SD}$ < 1 $\mu$ A, ThinSOT Package
LT1930/LT1930A	1A (I <sub>SW</sub> ), 1.2MHz/2.2MHz High Efficiency Step-Up DC/DC Converter	$V_{IN}\!\!: 2.6V$ to 16V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 4.2mA/5.5mA, $I_{SD}$ < 1 $\mu$ A, ThinSOT Package
LT1932	Constant Current, 1.2MHz, High Efficiency White LED Boost Regulator	$V_{IN}\!\!: 1V$ to 10V, $V_{OUT(MAX)}=34V,\ I_Q=1.2mA,\ I_{SD}<1\mu A,$ ThinSOT Package
LT1943 (Quad)	Quad Output, 2.6A Buck, 2.6A Boost, 0.3A Boost, 0.4A Inverter, 1.2MHz TFT DC/DC Converter	$V_{\text{IN}}\!\!:4.5\text{V to 22V},V_{\text{OUT}(\text{MAX})}$ = 40V, $I_{\text{Q}}$ = 10mA, $I_{\text{SD}}$ $<35\mu\text{A},$ TSS0P28E Package
LT1944/LT1944-1 (Dual)	Dual Output 350mA/100mA (I <sub>SW</sub> ), Constant Off-Time, High Efficiency Step-Up DC/DC Converter	$V_{IN}$ : 1.2V to 15V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 20 $\mu$ A, $I_{SD}$ < 1 $\mu$ A, MS10 Package
LT1945	Dual Output, Pos/Neg, 350mA (I <sub>SW</sub> ), Constant Off-Time, High Efficiency Step-Up DC/DC Converter	$V_{IN}$ : 1.2V to 15V, $V_{OUT(MAX)} = \pm 34V$ , $I_Q = 20\mu A$ , $I_{SD} < 1\mu A$ , MS10 Package
LT1947	Adjustable Output TFT-LCD Triple Switching Regulator	$V_{IN}$ : 2.7V to 8V, $V_{OUT(MAX)}$ = 30V, $I_Q$ = 9.5mA, $I_{SD}$ < 1 $\mu$ A, MS10 Package
LTC3450	Triple Output Power Supply for Small TFT-LCDs	$V_{IN}$ : 1.5V to 4.6V, $V_{OUT(MAX)} = \pm 15$ V, $I_Q = 75\mu$ A, $I_{SD} < 1\mu$ A, DFN Package
LT3461/LT3461A	0.3A (I <sub>SW</sub> ), 1.3MHz/3MHz, High Efficiency Step-Up DC/DC Converter with Integrated Schottky	$V_{IN}\!\!: 2.5V$ to 16V, $V_{OUT(MAX)}=38V,\ I_Q=2.8mA,\ I_{SD}<1\mu A,$ SC70, ThinSOT Packages
LT3464	0.08A (I <sub>SW</sub> ), High Efficiency Step-Up DC/DC Converter with Integrated Schottky, Output Disconnect	$V_{IN}\!\!: 2.3V$ to 10V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 25 $\mu$ A, $I_{SD}$ < 1 $\mu$ A, ThinSOT Package
LT3465/LT3465A	Constant Current, 1.2MHz/2.7MHz High Efficiency White LED Boost Regulator with Integrated Schottky	$V_{IN}\!\!: 2.7V$ to 16V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 1.9mA, $I_{SD}$ < 1 $\mu$ A, ThinSOT Package
LT3466	Dual Constant Current, 2MHz, High Efficiency White LED Boost Regulator with Integrated Schottky	$V_{IN}$ : 2.7V to 24V, $V_{OUT(MAX)}$ = 40V, $I_Q$ = 5mA, $I_{SD}$ < 16 $\mu$ A, DFN Package
LT3467/LT3467A	1.1A (I <sub>SW</sub> ), 1.3MHz/2.1MHz, High Efficiency Step-Up DC/DC Converter with Integrated Soft-Start	$V_{IN}\!\!: 2.4V$ to 16V, $V_{OUT(MAX)}$ = 40V, $I_Q$ = 1.2mA, $I_{SD}$ < 1 $\mu\text{A}$ , ThinSOT Package