## Ayama ${ }^{\text {TM }} 10000$ Network Search Engine

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### 1.0 Features

- Up to 512K 36-bit entries in a single device for CYNSE10512
-256K entries in 72-bit configuration
-128K entries in 144-bit configuration
-64K entries in 288-bit configuration
- 32K entries in 576-bit configuration
-Up to 256K 36-bit entries in a single device for CYNSE10256
-128K entries in 72-bit configuration
-64K entries in 144-bit configuration
-32K entries in 288-bit configuration
-16K entries in 576-bit configuration
-Up to 128K 36-bit entries in a single device for CYNSE10128
-64K entries in 72-bit configuration
-32K entries in 144-bit configuration
-16K entries in 288-bit configuration
-8K entries in 576-bit configuration
- Multiple width tables in a single device
- Single-cycle Search operation on 72-/144-bit tables
- Mini-Key ${ }^{\top \mathrm{M}}$-programmable search key for fine grain table selection and power conservation
- Prioritized blocks with no overhead in table management using programmable Soft Priority ${ }^{\mathrm{TM}}$
- Parity support for reliable operation
- Non-Enhanced Mode and Enhanced Mode operation
—Up to 133 million searches per second in 72-/144-bit configuration
-Up to 66.5 million searches per second in 36-/288-bit configuration
—Up to 33.25 million searches per second in 576-bit configuration (Enhanced Mode Only)
- Enhanced Mode with MultiSearch ${ }^{T M}$ operation
—Up to $\mathbf{2 6 6}$ million searches per second in 72-/144-bit configuration
-Up to 133 million searches per second in 36-/288-bit configuration
-Up to 66.5 million searches per second in 576-bit configuration
- Cascadable for depth expansion
- Glueless interface to industry-standard SRAMs and SSRAMs
- Simple hardware instruction interface
- IEEE 1149.1 test access port
- 1.2 V core voltage supply
- Supports 1.5V HSTL and 1.8V/2.5V LVCMOS I/O Standards
- 388-pin BGA package


### 2.0 Overview

Cypress Semiconductor Corporation's (Cypress's) Ayama ${ }^{\text {TM }} 10000$ Network Search Engine (NSE) is designed to be a highperformance, pipelined, synchronous, $512 \mathrm{~K} / 256 \mathrm{~K} / 128 \mathrm{~K} 36$-bit entries NSE. This high-speed, high-capacity Ayama 10000 NSE can be deployed in a variety of networking and communications applications. It can be used to accelerate network protocols such as Longest-Prefix Match (CIDR), ARP, MPLS, and other layer 2, 3, and 4 protocols. The performance and features of the Ayama 10000 make it attractive in applications such as Enterprise LAN switches and routers, and broadband switching and/or routing equipment that supports multiple data rates at OC-48 and beyond. Ayama 10000 can operate at a maximum performance of 266 million searches per second (MSPS).
The Ayama 10000 is designed to be scalable in order to support network database sizes of up to 15872 K 36 -bit entries specifically for environments that require large network policy databases. It includes features that ease table management, reduce power consumption and improve data integrity. The device can have its features individually enabled or disabled for flexibility based on the needs of the applications. The Ayama 10000's Data and Mask arrays that make up the Core are organized into blocks that can be individually configured to optimize the device performance and provide even more flexibility.
Figure 2-1 below shows the block diagram of the Ayama 10000.


Figure 2-1. Ayama ${ }^{\text {TM }} 10000$ Block Diagram

Figure 2-2 shows how an NSE subsystem can be formed using a host ASIC, a bank of Ayama 10000 devices and a bank of SRAM devices. It presents an example of how the NSE subsystem is integrated in a switch or router. The example also shows two possible ways of connecting the devices in the NSE subsystem. In the Associative set-up, the host ASIC sends instructions to the NSE. Where applicable, the NSE drives the SRAM inputs and the SRAM then returns the requested data to the host ASIC. In the Index set-up, the NSE's SRAM address information is routed back to the host ASIC. The host ASIC then interacts with the SRAM bank after it receives the result from the NSE.


Figure 2-2. Example of Switch/Router Implementation Using Ayama 10000

### 3.0 Device Architecture Overview

### 3.1 Data Array, Mask Array and Table Widths

The Ayama 10000 device consists of $M \times 72$-bit ( $M=256 \mathrm{~K}$ for CYNSE10512, 128K for CYNSE10256, 64K for CYNSE10128) storage cells referred to as data bits. There is also a mask cell corresponding to each data cell. A database entry includes both the data and mask cells. Figure 3-1 shows the four possible table width sizes of the data and mask cells and the maximum possible table depth for each width.


M $=256 \mathrm{~K}$ for CYNSE 10512
128 K for CYNSE1 0256
64K for CYNSE10128


Figure 3-1. Ayama 10000 Database Table Widths
The Ayama 10000 can be configured to contain tables of different widths in one device up to a maximum equal to $512 \mathrm{~K} / 256 \mathrm{~K} / 128 \mathrm{~K}$ 72-bit entries. For example, a single Ayama 10000 device can have both a 5 -Tuple Flow table and an IPv6 forwarding table. Figure 3-2 shows a sample configuration of multiple table widths in a CYNSE10512 device.


Figure 3-2. Multi-Width Database Configuration Example
Note:

1. 576-bit table configuration is only supported in the Enhanced mode.

### 3.2 Data and Mask Addressing

Each 72-bit entry in the device can be accessed directly through its address index. The data and mask arrays addresses are as shown in Figure 3-3.


$$
\begin{gathered}
N=262144 \text { for CYNSE10512 } \\
131072 \text { for CYNSE10256 } \\
65536 \text { for CYNSE10128 }
\end{gathered}
$$

Figure 3-3. Addressing the Ayama 10000 Data and Mask Arrays

### 3.3 Successful Search and Multiple Match Arbitration

During a Search operation, the search data bit is masked with the corresponding global mask bit from the selected Global Mask Register and the mask array bit before being compared to the data array entry bit to check for a match at that bit position (see Table 3-1). The entry with a match on every bit position results in a successful Search. For example, in order for a successful Search within a device to make the device the local winner, all 72-bit positions must generate a match for a 72-bit entry in 72-bitconfigured quadrants. The same applies to 144 -bit, 288 -bit, and 576 -bit searches.
The on-chip priority encoder selects the first matching entry in the database that is nearest to memory address 0 . An arbitration mechanism using a cascade bus determines the global winning device among the local winning devices in a Search cycle. The global winning device then drives the output signals. When there is no successful Search, the device designated as the last device (Refer to Section 5.3 for more information on last device designation) will drive the output signals.
Table 3-1. Bit Position Match

| Global Mask Bit | Mask Array Bit | Data Array Bit | Search Key Bit | Match Result |
| :---: | :---: | :---: | :---: | :---: |
| 0 | $X$ | $X$ | $X$ | 1 |
| 1 | 0 | $X$ | 0 | 1 |
| 1 | 1 | 0 | 0 | 1 |
| 1 | 1 | 1 | 1 | 0 |
| 1 | 1 | 1 | 1 | 0 |
| 1 |  |  | 1 |  |

### 4.0 Signals Description

Table 4-1 lists and describes all Ayama 10000 signals.
Table 4-1. Ayama ${ }^{\text {TM }} 10000$ Signal Description

| Parameter | Type ${ }^{[2]}$ | Description |
| :---: | :---: | :---: |
| Clocks and Reset |  |  |
| CLK_MODE | I | Clock Mode. Selects the clock source for the device. When set to Low, the device uses both CLK2X and PHS_L for its clock sources. When pulled High (VDDASIC), the device uses CLK1X for its clock source (PHS_L must be externally grounded). |
| CLK2X/CLK1X | 1 | Master Clock. CLK_MODE selects either the CLK2X or CLK1X as the clock input signal. CLK1X <br> Input signals are sampled on both rising and falling edges. <br> Output signals can be driven on both falling and rising depending on the operation and the device configuration. <br> CLK2X <br> Input signals are sampled on the rising edge. <br> Output signals are driven on the rising edge. |
| PHS_L | I | Phase. An input signal that must switch at half the frequency of CLK2X. This signal should be pulled LOW when the device is in CLK1X mode. See Section 5.6, "Clocks," on page 42. |
| RST_L | I | Reset. Driving RST_L LOW initializes the device to the default state. The device becomes active stable 4 CLK1X (8 CLK2X) cycles after RST_L is driven High ( $90 \%$ threshold). |
| Configuration |  |  |
| CFG_L | 1 | Configuration. When CFG_L is set to Low, the device will tristate DQ[71:68]. |
| ID[4:0] | I | Device Identification. The binary-encoded device identification for a depth-cascaded system starts at "00000" and goes up to " 11110 ". "11111" is reserved as the broadcast address which selects all NSEs in the cascade. <br> On a broadcast Read, only the device with the LDEV bit set to '1' will respond. <br> Any ID bit that is to be set High must be connected to $V_{\text {DDQ_AsIc }}$. |
| ASICSEL | I | ASIC IO Select. When this signal is pulled High (1.8V or 2.5V LVCMOS), the Command, Data and Cascade buses will operate in LVCMOS mode. When tied to Low, the buses will operate in HSTL mode. <br> Signals affected by ASICSEL selection: <br> Clocks: CLK2X/CLK1X, PHS_L, RST_L <br> Command and Data: CMD[10:0], CM $\overline{\mathrm{D} V, ~ D Q[71: 0], ~ P A R[1: 0], ~ A C K, ~ E O T, ~ S S F, ~ S S V, ~}$ MULTI_HIT <br> Cascā̄e Interface: LHI[6:0], LHO[1:0], BHI[2:0], BHO[2:0], FULI[6:0], FULO[1:0], FULL |
| SRAMSEL | I | SRAM IO Select. When this signal is pulled High (1.8V or 2.5V LVCMOS), the SRAM Interface will operate in LVCMOS mode. When tied to Low, the interface will operate in HSTL mode. <br> Signals affected by SRAMSEL selection: <br> SADR[25:0], CE_L, WE_L, OE_L, ALE_L |
| HSVREF0 | I | HSTL Reference Voltage. When ASICSEL is set to GND, this signal must be connected to the HSTL reference voltage (VDDQ_ASIC/2). Otherwise, they should be left floating. |
| HSVREF1 | I | HSTL Reference Voltage. Refer to HSVREF0 description. |
| PARERR_L ${ }^{[3]}$ | 0 | Parity Error. This signal is updated when there is a Core parity error or DQ Bus parity error. It is an Active-Low Open-Drain signal that requires an external pull-up resistor to VDDQ_ASIC. <br> This signal is valid only after the device is fully initialized. |

ASIC Interface / Command and Data Buses (LVCMOS or HSTL I/II)

| CMD[10:0] | I | Command Bus. Bit[10:2] contains the command parameters and Bit[1:0] specifies the <br> command. |
| :--- | :---: | :--- |

Table 4-1. Ayama ${ }^{\text {TM }} 10000$ Signal Description (continued)

| Parameter | Type ${ }^{[2]}$ | Description |
| :---: | :---: | :---: |
| CMDV | I | Command Valid. This signal indicates valid command in the CMD bus when set to High. |
| DQ[71:0] | I/O | Address/Data Bus. This signal carries the following information: <br> Search operation: Compare Data (Search Key) <br> SRAM PIO operations: SRAM Address <br> Other operations to Register, Data, and Mask Array regions: Address and Data |
| PAR[1:0] | I/O | Parity Bus. These signals contain the even parity values for the DQ bus. On the Read return data, the NSE generates the parity bits. On all other operations these bits are externally driven. Bit [0] is the parity for all even DQ signals. Bit[1] is the parity for all odd DQ signals. |
| $\mathrm{ACK}^{[4]}$ | T | Read Acknowledge. This signal indicates that valid data is available on the DQ bus during register, data, and mask array Read operations, or that the data is available on the SRAM data bus during SRAM Read operations. |
| EOT ${ }^{[4]}$ | T | End of Transfer. This signal indicates the end of burst transfer to the data or mask array during Read or Write burst operations. |
| SSF ${ }^{[5]}$ | T | Search Successful Flag. When asserted, this signal indicates that the device is the global winner in a Search operation. |
| $\mathrm{SSV}^{[5]}$ | T | Search Successful Flag Valid. When asserted, it indicates valid SSF value. In Enhanced mode, this signal also indicates valid FULL and MULTI HIT values. |
| MULTI_HIT ${ }^{[5]}$ | 0 | Multiple Hit Flag. In a Search operation, this signal indicates that there are multiple entries in the array or in the selected blocks that match the Search key when it is set to 1. In a Learn operation, it indicates that there are multiple free entries. <br> In Non-Enhanced mode, it becomes valid 4 CLK1X cycles after the command is issued. In Enhanced mode, it becomes valid when SSV is 1. |
| FULL | T | Full Flag. When High, it indicates that the table in the array or in the selected blocks (Enhanced mode) is full. <br> In the Non-Enhanced mode, it becomes valid 4 CLK1X cycles after the command is issued. <br> In the Enhanced mode, it becomes valid when SSV is 1. |
| HIGH_SPEED1 | I | High Speed 1. This signal must be pulled High (VDD_ASIC) when the device operates at CLK2X frequency above 166 MHz . |
| HIGH_SPEED2 | I | High Speed 2. This signal must be pulled High (VDD_ASIC) when the device operates at CLK2X frequency above 200 MHz . |
| SRAM Interface (LVCMOS or HSTL I/II) |  |  |
| SADR[M:0] ${ }^{[5]}$ | T | SRAM Address. This bus contains address lines to access off-chip SRAMs that contain associative data. In a cascaded system of multiple Ayama 10000 NSEs, each corresponding SADR bit from all cascaded devices must be tied together. <br> M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128. |
| CE_L ${ }^{[5]}$ | T | SRAM Chip Enable. This is the chip enable (CE) control for external SRAMs. In a cascaded system of multiple Ayama 10000 NSEs, CE_L of all cascaded devices must be tied together. This signal is then driven by only one of the devices. |
| WE_L ${ }^{[5]}$ | T | SRAM Write Enable. This is the Write enable control for external SRAMs. In a cascaded system of multiple Ayama 10000 NSEs, WE_L of all cascaded devices must be tied together. This signal is then driven by only one of the devices. |
| OE_L ${ }^{[5]}$ | T | SRAM Output Enable. This is the output enable (OE) control for external SRAMs. Only the last device drives this signal (the device that has the LRAM bit set). |
| ALE_ ${ }^{[5]}$ | T | Address Latch Enable. When this signal is Low, the addresses are valid on the SRAM address bus. In a cascaded system of multiple Ayama 10000s, the ALE_L of all cascaded devices must be tied together. This signal is then driven by only one of the devices. |

Table 4-1. Ayama ${ }^{\text {TM }} 10000$ Signal Description (continued)

| Parameter | Type ${ }^{[2]}$ | Description |
| :---: | :---: | :---: |
| Cascade Interface (LVCMOS and HSTL) |  |  |
| $\begin{aligned} & \text { LHI[6:0] } \\ & \text { LHI_O[6:0] (MSE=1) } \end{aligned}$ | 1 | Local Hit In/Local Hit In Array 0. These signals are inputs from upstream devices in a cascade that indicate whether there is a hit in the upstream/previous device(s). <br> When MultiSearch is performed, LHI[6:0] becomes LHI_0[6:0] (Local Hit input signals for Array 0). |
| $\begin{aligned} & \mathrm{LHO}[1: 0] \\ & \mathrm{LHO} \mathrm{O}[1: 0](\mathrm{MSE}=1) \end{aligned}$ | 0 | Local Hit Out/ Local Hit Out Array 0. LHO[1] and LHO[0] are logically the same signal. One of these signal is connected to one input on the LHI bus of the downstream devices in a cascade. <br> When MultiSearch is performed, LHO[1:0] becomes LHO_0[1:0] (Local Hit output signals for Array 0). |
| BHI[2:0] | I | Block Hit In. These signals are inputs from the last device in the upstream blocks in a cascade that indicate whether there is a hit in the upstream/previous block(s). |
| BHO[2:0] | 0 | Block Hit Out. These signals are logically the same signal. One of these signals is connected to one input on the BHI bus of the downstream devices in the downstream blocks. |
| $\begin{aligned} & \text { FULI[6:0] } \\ & \text { LHI_1_L[6:0] (MSE=1) } \end{aligned}$ | 1 | Full In/Local Hit In Array 1. Each signal is driven by an upstream device's FULO output in a block to generate the FULL signal for that block. During a Search operation, these signals indicate whether an upstream device had a free entry for a future Learn. <br> When MultiSearch is performed, FULI[6:0] becomes active Low LHI_1_L[6:0] (Local Hit input signals for Array 1). |
| $\begin{aligned} & \text { FULO[1:0] } \\ & \text { LHO_1_L[1:0] (MSE=1) } \end{aligned}$ | 0 | Full Out/Local Hit Out Array 1. FULO[0] and FULO[1] are logically the same signal. One of these signal is connected to one input on the FULO bus of the downstream devices in a cascade. <br> When MultiSearch is performed, FULO[1:0] becomes active Low LHI_0_L[1:0] (Local Hit output signals for Array 1). |
| Supplies |  |  |
| $V_{\text {DD }}$ |  | Core Supply: 1.2V. |
| $\mathrm{V}_{\text {DD_PLL }}$ |  | PLL Block Supply: 1.2V. |
| V ${ }_{\text {DDQ_ASIC }}$ |  | ASIC and Cascade Interface I/O Supply: 1.5 V (HSTL) or 1.8V/2.5V (LVCMOS). |
| V ${ }_{\text {DDQ_SRAM }}$ |  | SRAM Interfaced I/O Supply: 1.5 V (HSTL) or 1.8V/2.5V (LVCMOS). |
| $V_{\text {DDQ_JTAG }}$ |  | JTAG Test Access Port I/O Supply: 2.5V (LVCMOS). |
| Test Access Port |  |  |
| TDI | I | Test access port test data in. |
| TCK | I | Test access port test clock. |
| TDO | T | Test access port test data out. |
| TMS | I | Test access port test mode select. |
| TRST_L | 1 | Test access port reset. |

## Notes:

2. I = Input only, $\mathrm{I} / \mathrm{O}=$ input or output, $\mathrm{O}=$ output only, $\mathrm{T}=$ three-state output.
3. The rise time of PARERR_L will depend on the value of the pull-up resistance. Sufficient delay should be allotted for in the error routine after clearing the parity error in the parity control register and before this pin is sampled as part of the next command. Recommended external pull-up resistance range: $4.7 \mathrm{~K} \Omega$ to $47 \mathrm{~K} \Omega$.
4. Require an external pull-down resistor such as $47 \mathrm{~K} \Omega$ or $100 \mathrm{~K} \Omega$.
5. These signals will output at the rising edge of CLK2X (both rising and falling edges of CLK1X) in a MultiSearch operation.

### 5.0 Functional Description

### 5.1 Modes of Operation

Ayama 10000 can operate in two different modes of operation: Non-Enhanced and Enhanced. The Non-Enhanced mode of operation is provided for backward compatibility with the CYNSE70000 device family. The Enhanced mode allows the Ayama 10000 to utilize the features that can be used to lower power consumption, ease table management, increase data integrity and increase Search throughput. These features are Mini-Key ${ }^{\top M}$, Soft Priority ${ }^{\top M}$, Parity, and MultiSearch ${ }^{\top \mathrm{M}}$. The following subsections provide more information on each of the modes and features.
The device powers-up in Non-Enhanced mode. A switch to Enhanced mode and activation of the features require the user to configure internal registers with appropriate values. Refer to Section 5.4 for detailed information on the internal registers.
Table 5-1 lists the features and functions that are different between the two modes of operation.
Table 5-1. Summary of Non-Enhanced and Enhanced Mode Features and Functions Differences

| Features/Functions | Non-Enhanced | Enhanced |
| :--- | :--- | :--- |
| Maximum Search Throughput | 133 MSPS | 266 MSPS |
| MultiSearch ${ }^{\text {TM }}$ | No | Yes |
| Soft Priority ${ }^{\text {TM }}$ | No | Yes |
| Mini-Key ${ }^{\text {TM }}$ | No | Yes |
| Parity | No | Yes |
| Learn Operation | Data from CMPR Register; <br> Target Data Array; <br> Supports x72 and x144 table widths | Data from CMPR Register or DQ Bus; <br> Target Data or Mask Array; <br> Supports all table widths |
| Where to Configure the Table Width | CONFIG Register | BMR Register |
| Table Widths Supported | x72, x144, x288 | x72, x144, x288, x576 |
| Data and Mask Array Organization | $32 / 16 / 8 ~ 8 K x 72-b i t ~ P a r t i t i o n s ~ f o r ~$ <br> CYNSE10512/256/128 respectively | $128 / 64 / 32$ 2Kx72-bit Blocks for <br> CYNSE10512/256/128 respectively |

### 5.1.1 Non-Enhanced Mode

In the Non-Enhanced mode of operation, the Ayama 10000 device is organized into $32 / 16 / 8$ partitions (corresponds to CYNSE10512/256/128, respectively) that each can be configured to be $8 \mathrm{~K} \times 72,4 \mathrm{~K} \times 144$, or $2 \mathrm{~K} \times 288$. The 576 -bit table width configuration is not supported in this operation mode. The LSB of each 72-bit is designated to indicate whether that entry is used or not. When the entry is empty, that bit must be set to 0 . When the entry is used, that bit must be set to 1 . For example, in a 288 -bit table a used entry will have bit[0], bit[72], bit[144], and bit[216] set to 1. When all bit[0] are set to 1, the Ayama 10000 will assert FULO[1:0] to "11."
References are present throughout this document to indicates features that are applicable when device is in this mode.
Internal registers for configuration: CONFIG and CMD.

### 5.1.2 Enhanced Mode

In Enhanced mode, Ayama 10000 is organized into 128/64/32 blocks (corresponds to CYNSE10512/256/128 respectively) of $2 \mathrm{~K} \times 72$ which can also be configured into $1 \mathrm{~K} \times 144,512 \times 288$, or $256 \times 576$. The Mini-Key, Soft Priority, Parity, and MultiSearch features can also be activated. Each block has internal block registers associated to it that needs to be initialized before the device goes into normal operation. Figure 5-1 shows the general overview of the block registers association.
References are present throughout this document to indicates features that are applicable when the device is in this mode.


Figure 5-1. Blocks and Block Registers Association

### 5.1.2.1 Mini-Key

The Mini-Key feature allows the device to power down blocks within the device that are not being selected to participate in the search operation. This results in lower power consumption. When a device has multiple tables, the block architecture of the device combined with Mini-Key can be used to ease table expansion or reorganization. There are four Mini-Keys that can be associated with each block (Figure 5-2) which supports each block to be a member of up to four logical tables. The block register that holds the Mini-Key values also includes the field that configures the block to be of a certain table width.


Figure 5-2. Mini-Key Register Contents
During a Search operation, the Search key width as well as the Search Mini-Key are used to selectively activate certain blocks. A block will participate in the Search operation only when the Search width matches the block's table width and the Search MiniKey matches one of the four Mini-Keys of the block.
Internal registers for configuration: CMD, CPR and BMR.

### 5.1.2.2 Soft Priority

Table management can become a time consuming process and slow down the performance of the system. In an edge router with multiple table of same widths in one or more Ayama 10000 devices, that constantly update the entries, one table may become full very quickly. The time it takes to process table expansion and data reorganization can be critical in a system that requires high performance and quality of service. Soft Priority feature in the Ayama 10000 can help avoid that problem. For Soft Priority purposes, each 2 Kx 72 block of Data/Mask array is arranged into four $512 \times 72$ sub-blocks. Each sub-block has a user-program-
mable Soft Priority value that becomes part of the search key when Soft Priority feature is enabled. This feature eases the management of the tables, especially for table expansion. Each sub-block also has a Priority Valid bit that can be used to set the Soft Priority value of the sub-block to invalid state which will also prevent the sub-block from participating in a search operation.
Figure 5-3 shows the associations of the sub-blocks and Soft Priority.

$\mathrm{N}=127$ for CYNSE10512
63 for CYNSE10256
31 for CYNSE10128
Figure 5-3. Sub-Blocks and Soft Priority Associations
Internal registers for configuration: CMD, CPR and BPR.

### 5.1.2.3 Parity

Ayama 10000 introduces parity to provide additional protection for data integrity. Parity checking can be performed both on the data transmission that passes through DQ bus and the data stored in the Core (data and mask arrays). The Parity feature can be enabled through the PARITY register. DQ Bus and Core parity checking can be independently enabled. When parity checking is enabled, a Write operation ignores any masking and all bits are written as presented in the DQ bus.
Even Parity is used in the parity checking. For example, if there is an odd number of logic-1 bits in a word, the corresponding parity bit will be set High in order for the combination (word and parity bit) to have even parity.
When an error is detected, the device will update the PARITY register and set the parity error flag (PARERR_L) to report the error. Parity status is not cascaded. However, PARERR_L is an open-drain signal to allow signals from cascaded Ayama 10000 devices to be connected together and provide cascaded parity error detection. Therefore, the AC timing parameters associated with the signal (rise time/fall time) will be dependent on the loading conditions. Note that all parity status fields in PARITY and BPAR registers needs to be cleared by the ASIC after fixing the errors.
Internal registers for configuration: PARITY and BPAR.

## DQ Bus Parity

The DQ bus is divided into even-bits and odd-bits groups for parity checking. Parity bits of both even- and odd-bits groups are provided in the bidirectional PAR[1:0]. When the ASIC is driving the DQ bus, the ASIC must generate the parity bits. When the NSE is driving the DQ bus, the NSE will generate the parity bits.
When the ASIC is driving the DQ bus, the NSE will calculate the data stream parity and compare it to PAR[1:0]. When there is an error, the NSE will update the PARITY register and set PARERR_L to O. PARERR_L is valid on the $(3+T)^{\text {th }}$ cycle of latency for a Read operation and $(4+\mathrm{T})^{\text {th }}$ cycle of latency for the other operations. $T$ is the cycle where the bus parity error is detected. When a DQ bus parity error is detected, the NSE must be reset and reinitialized.
Figure 5.4 shows the timing diagram of a DQ Bus Parity error during a 288-bit Search instruction. In cycle 1B the parity of the odd DQ bits is shown to be ' 1 ' while the corresponding parity bit (PAR[1]) is ' 0 ' (should be High for parity check to result in a ' 0 '). The PARERR_L signal goes Low 4 cycles after the error is detected.


## Core Parity

The Core includes a one-bit parity for each 72-bit entry in the data and mask arrays. When writing into the data or mask array, the NSE will calculate and generate the one-bit parity for each 72-bit data. Each block also has a block-associated internal register to enable the parity checking for the block (BPAR). When disabled, the block will ignore the Read Parity command.
To issue the Read Parity command, the ASIC issues a Read command and sets the Parity field in the parameters sent through the DQ bus as described in Table 5-25. Core parity checking is performed in parallel on four adjacent 72-bit entries per pair of blocks. At the beginning of each parity operation, an internal address counter is incremented. The new incremented address is then used for the parity check operation.
It will cycle through the data and mask arrays as well as odd and even blocks for both arrays for each Read Parity issued. If one or more parity errors are detected, the error is reported in the block's BPAR register. Then all errors are prioritized through an arbiter to select the highest priority parity error, which is then reported in the PARITY register. PARERR_L will also be set to 0 when there is a parity error. PARERR_L is valid on the ( $5+$ TLSZ) ${ }^{\text {th }}$ cycle of latency. For example, with TLSZ set to " 00 " and the command is issued at Cycle1, PARERR_L will be valid on Cycle6. Read Parity also responds to broadcast CHIPID selection.
Figure 5-5 shows the timing diagram of a Core Parity error during a Read Parity instruction. The PARERR_L signal goes Low 5 cycles after the error is detected.
There are two basic flows for parity error recovery. The first flow is by reading the highest priority parity error address stored in the PARITY register, fix the error, decrement the internal address counter and reissue Read Parity. The second flow is by reading the PARITY register to obtain the location, reading the BPAR registers to locate blocks that has the error and then fixing those locations.


Figure 5-5. Timing Diagram of a Core Parity Error (TLSZ=00)

### 5.1.2.4 MultiSearch

When MultiSearch is activated, the Core is divided into two separate arrays. Each array is organized into $64 / 32 / 16$ blocks (corresponds to CYNSE10512/CYNSE10256/CYNSE10128, respectively) of 2K 72-bit entries. Each block can be configured to be of width x 72 , $\mathrm{x} 144, \mathrm{x} 288$, or x 576 . This separation allows a Search operation to simultaneously perform the search across both arrays. The output signals will run at double data rate to effectively increase the throughput to a maximum of 266 million searches per second. Each array can have multiple tables with different widths. Single-Search operation outputs are driven at the rising edge of CLK1X. When the device has the MultiSearch feature enabled and MultiSearch operation is issued (SingleSearch can still be issued even when MultiSearch is enabled), the output is driven at both rising and falling edges of CLK1X (rising edge of CLK2X). Output from Array 0 is driven at the rising edge while output from Array 1 is driven at the falling edge of CLK1X. Figure $5-6$ shows an illustration of the MultiSearch operation.


Figure 5-6. MultiSearch Operation Overview
Both arrays will use the same Search key except for search operation on 72-bit wide tables. So does the selection of the Global Mask Register (GMR) and Comparand Register (CMPR) as listed in Table 5-2.

Table 5-2. Selection of Search Key, GMR, and CMPR in MultiSearch Operation

| Search Width | Array 0 |  |  | Array 1 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | GMR | Key | SRR | GMR | Key | SRR |
| x72/x72 | 1: GMR | 1: A | 1: CMPR | 1: GMR+1 | 1: B | 1: CMPR+1 |
| x144/x144 | 1: GMR | 1: A, B | 1: CMPR | 1: GMR+1 | 1: A, B | 1: CMPR+1 |
| x288/x288 | 1: GMR | 1: A, B | 1: CMPR | 1: GMR+1 | 1: A, B | 1: CMPR+1 |
|  | 2: GMR | 2: A, B |  | 2: GMR+1 | 2: A, B |  |
| x576/x576 | 1: GMR | 1: A, B | 1: CMPR | 1: GMR+1 | 1: A, B | 1: CMPR+1 |
|  | 2: GMR | 2: A, B |  | 2: GMR+1 | 2: A, B |  |
|  | 3: GMR | 3: A, B |  | 3: GMR+1 | 3: A, B |  |
|  | 4: GMR | 4: A, B |  | 4: GMR+1 | 4: A, B |  |

CLK2X runs at twice the frequency of CLK1X. "A" refers to the first CLK2X cycle in a CLK1X cycle. "B" refers to the second CLK2X cycle in a CLK1X cycle.
When MultiSearch is issued on $\times 72$ tables, Array 0 will use the GMR provided in cycle 1 and Array 1 will automatically selects to use the next-up GMR. For example, if GMR[0] is selected in cycle 1, GMR[1] will be selected automatically for Array 1. Array 0 uses the Search key provided in cycle 1-A while Array 1 uses the Search key provided in cycle 1-B.
When MultiSearch is issued on x288 tables, Array 0 and Array 1 will both use the same Search key selected in cycle 1 and cycle 2. The GMR selection for Array 0 is done per cycle and Array 1 automatically selects the next up GMR register. For example, if GMR[6] is selected for the first 144 bits of Array 0 in cycle 1, GMR[7] is automatically selected for the first 144 bits of Array 1. Then if GMR[15] is selected for the second 144 bits of Array 0 in cycle 2, GMR[0] is automatically selected for the second 144 bits of Array 1.
In all cases, the CMPR used in the Search operation for Array 0 is the one selected in cycle 1 and the next up CMPR is automatically selected for Array 1.
Internal register for configuration: CMD.

### 5.1.2.5 Enhanced Learn Operation

Ayama 10000 extends the capability of the Learn function to allow users to select the data source for the operation. The data can be from the DQ bus or one of the Comparand (CMPR) registers. It also allows the data to be written to both the mask and data array while in Non-Enhanced mode it allows the data to be written only to the data array.
Internal register for configuration: CMD.

### 5.2 I/O Interfaces

Data flows in and out of the device through three separate I/O interfaces: ASIC, SRAM and Cascade Interface. Section 4.0, "Signals Description," on page 15 lists the signals that are part of each interface. Input signals are registered on the rising edge and falling edge of CLK1X or rising edge of CLK2X. Output signals are driven out on the rising edge of CLK1X or rising edge of CLK2X when PHS_L is Low. An exception is when MultiSearch operation is activated, the output will be driven out on both edges of CLK1X or rising edge of CLK2X. Refer to Section 5.6 for more information on clock signals. Figure 5-7 shows an example of ASIC, NSEs and SRAMs I/O interconnects. The SRAM Interface outputs may be connected to SRAM devices in associative applications mode or back to the ASIC in index applications mode.


Figure 5-7. Ayama 10000 I/O Interfaces
Internal register for configuration: HARDWARE.

### 5.2.1 ASIC Interface

The ASIC Interface includes all signals for data that comes in from and out to a system's processing unit, which could be an application specific (ASIC) or a more generic network processing unit (NPU and NCP). It supports LVCMOS and HSTL I/O standards. LVCMOS allows the I/O signals to run at a rate of up to 100 MHz (CLK1X; Double data rate in MultiSearch operation). With HSTL, the I/O signals can run at a rate of up to 133 MHz (CLK1X; Double data rate in MultiSearch operation).
The ASIC Interface includes the Command and DQ bus signal group. CMD[10:0] carries the command and its associated parameter. DQ[71:0] is used for data transfer to and from the database entries, which are comprised of data and mask fields that are organized as data and mask arrays. The DQ bus carries Search data (of the data and mask arrays and internal registers) during the Search command as well as the address and data during Read and/or Write operations. The DQ bus also carries address information for the direct accesses to the external SRAM.

### 5.2.2 SRAM Interface

The SRAM Interface includes output only signals that are used to interact with SRAM memory devices. As with the ASIC Interface, it supports LVCMOS and HSTL I/O standards. LVCMOS allows the I/O signals to run at a rate of up to $100 \mathrm{MHz}(200-\mathrm{MHz}$ double data rate with MultiSearch operation). With HSTL, the I/O signals can run at a rate of up to $133 \mathrm{MHz}(266-\mathrm{MHz}$ double date rate with MultiSearch operation).

### 5.2.3 Cascade Interface

The Cascade Interface is used for cascading multiple Ayama 10000 devices in a system. It supports LVCMOS and HSTL I/O standards that can run up to 133 MHz in all operation modes. The Cascade Interface power supply is the same power supply that the ASIC Interface uses. Thus the selection of the I/O standard used for the Cascade Interface depends on the I/O Standard selected for the ASIC Interface.

When multiple NSEs are cascaded to create large databases, the data being searched is presented to all NSEs in the cascaded system simultaneously. If multiple matches occur, arbitration logic on the NSEs will enable the winning device (the one with a matching entry closest to address 0 of the cascaded database) to drive the SRAM bus. User can set the default device to respond to an operation when a Search operation does not result in a Search Hit. Refer to Section 5.3 for more information.

### 5.3 Output Signals Default Driver/Last Device Designation (LRAM and LDEV)

When NSEs are cascaded using multiple Ayama 10000 devices, the SADR, CE_L, and WE_L (three-state signals) are all tied together. In order to eliminate external pull-up and pull-downs, one device in a bank is designated the default driver. For nonsearch or non-learn cycles (see Subsection 6.6, "Learn Command") or Search cycles with a global miss, the SADR, CE_L, and WE_L signals are driven by the device with the LRAM bit set. It is important that only one device in a bank of cascaded NSEs have this bit set. Failure to do so will cause contention on the SADR, CE_L, and WE_L, and can potentially cause damage to the device(s).
Similarly, when NSEs using multiple Ayama 10000 devices are cascaded, SSF and SSV (also three-state signals) are tied together. In order to eliminate external pull-up and pull-downs, one device in a bank is designated as the default driver. For nonsearch cycles or Search cycles with a global miss, the SSF and SSV signals are driven by the device with the LDEV bit set. It is important that only one device in a bank of cascaded NSEs have this bit set. Failure to do so will cause contention on the SSV and SSF, and can potentially cause damage to the device(s).

### 5.4 Registers

Table $5-3$ provides an overview of all the Ayama 10000 internal registers. Each register is 72 bits wide. The Ayama 10000 contains sixteen pairs of comparand storage registers, sixteen pairs of global mask registers, eight Search status index registers, sixteen Search control parameters registers, sixteen Search result registers and one each of command, information, burst Read, burst Write, next-free address register, partition configuration, hardware and parity control registers. Each of the blocks in the NSE device (128/64/32 2Kx72 blocks in CYNSE10512/256/128 respectively) also has one each of Block Mini-Key, Block Priority, Block Parity and Block Next-free Address registers. There are also four Block Priority Register Aliases registers for each Block Priority register that allows an alternative way to update the Block Priority registers. The registers are presented in ascending address order. Each register group is then described in the following subsections. Reserved fields in the registers are read as 0s. When writing to the registers, all Reserved fields must be written with 0s, unless specified otherwise in the field's description.

Table 5-3. List of Internal Registers

| Address (decimal) | Abbreviation | Type (Read/Write) | Description |
| :---: | :---: | :---: | :---: |
| 0-31 | CMPR0-15 | R | Comparand Register. Sixteen CMPR pairs (144 bits per pair) that store comparands from the DQ bus during a Search operation for later use with the Learn command. See Section 5.4.1. |
| $\begin{gathered} \hline 32-47 \\ 96-111 \end{gathered}$ | $\begin{gathered} \text { GMR0-7 } \\ \text { GMR8-15 } \end{gathered}$ | R/W | Global Mask Register. Sixteen GMR pairs (144 bits per pair) used for global mask bits on the DQ bus for all commands. See Section 5.4.2. |
| 48-55 | SSR0-7 | R | Search Successful Register. These registers store the result of Search operations. See Section 5.4.3. |
| 56 | COMMAND | R/W | Command Register. This register contains control fields that determine how the NSE operates. See Section 5.4.4. |
| 57 | INFO | R | Information Register. This Read-only register contains static information about the NSE device. See Section 5.4.5. |
| 58 | RBURREG | R/W | Burst-Read Register. This register contains the starting address and count for a Read Burst operation. See Section 5.4.6. |
| 59 | WBURREG | R/W | Burst-Write Register. This register contains the starting address and count for a Write Burst operation. See Section 5.4.7. |
| 60 | NFA | R | Next-free Address Register. This register contains the index of the next-free entry when the device is in the Non-Enhanced mode (Enhanced mode uses SRR registers to store the next-free entry information). See Section 5.4.8. |
| 61 | CONFIG | R/W | Partition Configuration Register. This register contains the partition type bits when the NSE device operates in the Non-Enhanced mode. It is not used in the Enhanced mode. See Section 5.4.9. |
| 62 | HARDWARE | R/W | Hardware Register. This register contains I/O drive strength settings. See Section 5.4.10. |
| 63 | PARITY | R/W | Parity Control Register. This register contains the control and address for parity checking of the Core and registers. See Section 5.4.11. |
| 64-79 | CPR0-15 | R/W | Control Register. These registers provide Mini-Key and Soft Priority for the associated operation. See Section 5.4.12. |
| 80-95 | SRR0-15 | R | Search Result Register. These registers provide information of the next-free entry when the device is in the Enhanced mode. (Non-Enhanced mode uses the NFA register to store the next-free entry information.) See Section 5.4.13. |

Table 5-3. List of Internal Registers (continued)

| Address <br> (decimal) | Abbreviation | Type <br> (Read/Write) | Description |
| :---: | :---: | :---: | :--- |
| $112-1023$ | - | - | Reserved. |
| 1024 | BMRx | R/W | Block Mini-Key Register. This register holds the four Mini-Keys associated with <br> a block. There is one BMR per block. See Section 5.4.14. |
| 1025 | BPRx | R/W | Block Priority Register. This register holds the four sub-block priorities. There is <br> one BPR per block. See Section 5.4.15. |
| 1026 | BPARx | R/W | Block Parity Register. This register contains the control and status bits for <br> controlling and detecting parity errors for a block. There is one BPAR per block. <br> See Section 5.4.16. |
| 1027 | BNFAx | R | Block Next-free Address Register. This register contains the next-free entry <br> information for the block that it is associated with. There is one BNFA per block. <br> See Section 5.4.17. |
| $1028-1031$ | BPRA0x- <br> BPRA3x | R/W | Block Priority Register Aliases. These locations are aliases for the corre- <br> sponding BPRx. See Section 5.4.18. |

### 5.4.1 Comparand Register (CMPR)

The device contains 16 pairs of comparand registers (one pair is 144 bits) dynamically selected in every Search operation to store the comparand presented on the DQ bus. The device may later use these registers when it executes a Learn operation. Search and Learn commands specify the comparand registers in pairs. The Ayama 10000 device stores the Search command's cycle A comparand in the even-numbered register and the cycle B comparand in the odd-numbered register, as shown in Figure 5-8. For wider width keys, pairs of comparand registers are concatenated together. The concatenation of the registers must be done by the user. On a 72-bit operation, both halves of the comparand register must be loaded with the same value. When performing MultiSearch operation, the NSE requires two comparand registers for an operation. The first comparand register is specified in the command and the NSE automatically selects the comparand register one index higher than the command specified register. When the device powers-up, the CMPR registers are initialized to 0 .


Figure 5-8. Comparand Register Selection During Search and Learn Instructions

### 5.4.2 Global Mask Register (GMR)

The device contains 16 pairs of GMRs (one pair is 144 bits) dynamically selected in every Search operation to select the Search subfield. The addressing of these registers is shown in Figure 5-9. The GMR index supplied on the command bus selects one of the sixteen pairs of global masks during Search and Write operations. In 72-bit Search and Write operations, the host ASIC must program both the even and odd mask registers with the same values. For a MultiSearch operation, two separate GMRs are used in the operation. The first one is specified in the command and the second one is one index higher.
Each mask bit in the GMRs is used during Search and Write operations. In a Search operation, setting the mask bit to 1 enables while setting the mask bit to 0 disables compares at the corresponding bit position (forced match). In Write operations to the data or mask array, setting the mask bit to 1 enables Write while setting the mask bit to 0 disables Write at the corresponding bit position. Write operation to internal registers does not use the GMR to mask the data and ignores the GMR selection when the command is issued.
When the device powers-up, the GMR registers are initialized to 0 . Figure 5-9 below shows each portion (Even, Odd) of each GMR, and what address (in binary) is required to access that register.

|  |  |  |
| :---: | :---: | :---: |
|  |  |  |
| 0 | 32 | 33 |
| 1 | 34 | 35 |
| 2 | 36 | 37 |
| 3 | 38 | 39 |
| 4 | 40 | 41 |
| 5 | 42 | 43 |
| 6 | 44 | 45 |
| 7 | 46 | 47 |
| 8 | 96 | 97 |
| 9 | 98 | 99 |
| 10 | 100 | 101 |
| 11 | 102 | 103 |
| 12 | 104 | 105 |
| 13 | 106 | 107 |
| 14 | 108 | 109 |
| 15 | 110 | 111 |

Figure 5-9. Addressing the Global Mask Register Array

### 5.4.3 Search Successful Register (SSR)

The device contains eight Search Successful Registers (SSR) to hold the index of the location at which a successful search occurred. The format of each SSR is described in Table 5-4. The Search command specifies which SSR stores the index of a specific Search command in cycle B of the Search instruction. Subsequently, the host ASIC can use this register to access that data array, mask array, or external SRAM using the index as part of the indirect access address. The selected register is updated when the device performs a Search operation regardless of the operation modes.


Figure 5-10. Search Successful Register

Table 5-4. Search Successful Register Description

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| INDEX | $[\mathrm{N}: 0]$ | 0 | Index. This is the address of the 72-bit entry where a successful search occurs. This index <br> is updated if the device is either a local or global winner in a Search operation. $\mathrm{N}=17$ for <br> CYNSE10512, 16 for CYNSE10256, 15 for CYNSE10128. <br> If a hit occurs in a 144-bit table, the least-significant bit (LSB) is cleared to 0. <br> If a hit occurs in a 288-bit table, the two LSBs are cleared to 0. <br> If a hit occurs in a 576-bit table, the three LSBs are cleared to 0. |
|  | $[29: \mathrm{N}+1]$ |  | Reserved. |
| GVAL | $[30]$ | 0 | Global Valid. Valid only in Enhanced mode. It is updated when the device performs a <br> Search operation. It is set to 1 when there is no hit anywhere in the cascade and this device <br> is the last one in the cascade (LDEV field in CMD register is set to 1). Otherwise it is cleared <br> to 0. When set to 1, the device is responsible for responding to broadcast PIO operation. |
| VAL | $[31]$ | 0 | Valid. This field is updated when the device performs a Search operation. It is set to 1 only <br> when the device is a global winner. Otherwise it will be cleared to 0. |
|  | $[71: 32]$ |  | Reserved. |

### 5.4.4 Command Register (COMMAND)

Table 5-5 describes the command register fields. This register is expected to be initialized by the user right after reset before performing any Read, Write, Learn, Search, or Parity operations and thereafter not changed during normal operation. The user must also wait for at least 32 CLK2X cycles after a write to the COMMAND register before issuing the next command.


Figure 5-11. Command Register
Table 5-5. Command Register Description
$\left.\begin{array}{|c|c|c|l|}\hline \text { Field } & \begin{array}{c}\text { Range } \\ \text { (decimal) }\end{array} & \begin{array}{c}\text { Initial Value } \\ \text { (binary) }\end{array} & \\ \hline \text { SRST } & {[0]} & 0 & \begin{array}{l}\text { Software Reset. If set to 1, this bit resets the device with the same effect as a hardware } \\ \text { reset. Internally, it generates a reset pulse lasting for eight CLK2X cycles. This bit automat- } \\ \text { ically resets to 0 after the reset pulse is deasserted. }\end{array} \\ \hline \text { DEVE } & {[1]} & 0 & \begin{array}{l}\text { Device Enable. If 0, it keeps the SRAM bus (SADR, WE_L, CE_L, OE_L and ALE_L), SSF, } \\ \text { and SSV signals in a three-state condition and forces the cascade interface output signals } \\ \text { LHO[1:0] and BHO[2:0] to 0. It also keeps the DQ bus in input mode. The purpose is to make } \\ \text { sure that there are no bus contentions when the device powers up. Set this bit to 1 when the } \\ \text { device is ready for operation. }\end{array} \\ \hline \text { TLSZ } & {[3: 2]} & 10 & \begin{array}{l}\text { Table Size. This field increases the pipeline latency of the Search and Learn operations as } \\ \text { well as the Read and Write accesses to the SRAM. Once programmed, it is expected to not } \\ \text { be changed. } \\ \text { Affected signals in both Enhanced and Non-Enhanced Modes: } \\ \text { SADR, CE_L, OE_L, WE_L, ALE_L, SSV, SSF, and ACK. }\end{array} \\ \text { Affected signals only in Enhanced Mode: } \\ \text { FULL and MULTI_HIT. } \\ \text { Latency in number of CLK cycles: }\end{array}\right\}$

Table 5-5. Command Register Description (continued)

| Field | Range (decimal) | Initial Value (binary) | Description |
| :---: | :---: | :---: | :---: |
| HLAT | [6:4] | 000 | Latency of Hit Signals. This field adds latency to the SSF, SSV, FULL and MULTI_HIT signals (in addition to the latency of TLSZ) during a Search operation and ACK signal during SRAM Read accesses as listed below: |
| LDEV | [7] | 0 | Last Device in the Cascade. When set, the device is the last device in a cascaded and is the default driver for the SSF and SSV signals. In the event of a Search failure, the device with this bit set drives the hit signals as follows: SSF = 0 (binary), SSV = 1 (binary). In an operation other than Search, the device with this bit set drives the hit signals as follows: SSF $=0$ (binary), SSV $=0$ (binary). When multiple devices are cascaded, one of the devices must have LDEV set to 1 . |
| LRAM | [8] | 0 | Last Device on the SRAM Bus. When set to 1, this is the last device on the SRAM bus in a cascade and is the default driver for the SADR, CE_L, WE_L, and ALE_L signals. <br> In cycles where none of the Ayama 10000 devices in a cascade drive these signals, this device drives the signals as follows: <br> For CYNSE10512: SADR $=0 \times 1$ FFFFFF <br> For CYNSE10256: SADR $=0 x F F F F F F$ <br> For CYNSE10128: SADR $=0 \times 7 F F F F F$ <br> For CYNSE10512/256/128: <br> CE_L = 1 <br> WE_L $=1$ <br> ALE_L $=1$ <br> The device with this field set to 1 always drives OE_L. When multiple devices are cascaded, one of the devices must have LRAM set to 1 . |
| CFGA | [24:9] | 0 | Database Configuration. The field is an alias for the first eight pairs of partition configuration bits of the configuration register. Reading and writing this field is reflected in the configuration register and vice versa. This field is only used when the device operates in the Non-Enhanced mode. |
|  | [55:25] |  | Reserved. |
| BEN | [56] | 0 | DQ Bus Parity Enable. When set to 1, it enables parity checking on the data transferred through DQ bus. |
| EN | [57] | 0 | Core Parity Enable. When set to 1, it enables Core parity checking. |
|  | [60:58] |  | Reserved. |
| LRN | [61] | 0 | Enhanced LEARN Enable. When set to 1, it allows the user to select the data source for the Learn operation from either the DQ bus or one of the CMPRs. It also allows the user to select whether the write is to the data or the mask array. This field is valid in the Enhanced mode. |
| MSE | [62] | 0 | MultiSearch Enable. When set to 1, it activates support for MultiSearch operation. The SRAM output operates at CLK2X rate instead of CLK1X. This field is valid only when the EMODE field of COMMAND register is set to 1. |
| EMODE | [63] | 0 | Enhanced Mode. When set to 1, the device operates in the Enhanced mode. When cleared to 0 , the device operates in the Non-Enhanced mode. |
|  | [71:64] |  | Reserved. |

### 5.4.5 Information Register (INFO)

Table 5-6 describes the information register fields.


Table 5-6. Information Register Description

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| REV | $[3: 0]$ | 0001 | Device Revision Number. |
| IMPL | $[6: 4]$ | 001 | Implementation Number. |
| Reserved | $[7]$ | 0 | Reserved. |
| DEVID | $[15: 8]$ | 00010100 | Device Identification Number for CYNSE10128. |
|  |  | 00010101 | Device Identification Number for CYNSE10256. |
|  |  | 00010110 | Device Identification Number for CYNSE10512. |
| MANID | $[33: 16]$ | 000000000011011100 | Manufacturer ID. |
| Reserved | $[71: 34]$ |  | Reserved. |

### 5.4.6 Read Burst Address Register (RBURREG)

Table 5-7 shows the Read Burst Address register fields. These must be programmed before issuing a Burst-Read operation.


Table 5-7. Read Burst Register Description

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| INDEX | $[\mathrm{N}: 0]$ | 0 | Index. This field is used to identify the starting address of the data or mask array in a Burst- <br> Read operation. The NSE will automatically increment the value by one after each <br> successive Read of the data or mask array. It must be reinitialized before the next Burst- <br> Read operation. $\mathrm{N}=17$ for CYNSE10512, 16 for CYNSE10256, 15 for CYNSE10128. |
|  | $[18: \mathrm{M}]$ |  | Reserved. M = 18 for CYNSE10512, 17 for CYNSE10256, 16 for CYNSE10128. |
| BLEN | $[27: 19]$ | 0 | Length of Burst Access. The device provides the capability to Read from 4 to 511 locations <br> in a single burst. The NSE automatically decrements the value by one after each successive <br> reading of the data or mask array. It must be reinitialized before the next Burst-Read <br> operation. |
|  | $[71: 28]$ |  | Reserved. |

### 5.4.7 Write Burst Address Register (WBURREG)

Table 5-8 describes the Write Burst Address register fields. These must be programmed before issuing a Burst-Write operation.


Figure 5-14. Write Burst Address Register

Table 5-8. Write Burst Register Description

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| INDEX | $[\mathrm{N}: 0]$ | 0 | Index. This field is used to identify the starting address of the data or mask <br> array in a Burst-Write operation. The NSE will automatically increment the <br> value by one after each successive Write of the data or mask array. It must <br> be reinitialized before the next Burst-Write operation. N = 17 for <br> CYNSE10512, 16 for CYNSE10256, 15 for CYNSE10128. |
|  | $[18: \mathrm{M}]$ |  | Reserved. M $=18$ for CYNSE10512, 17 for CYNSE10256, 16 for <br> CYNSE10128. |
| BLEN | $[27: 19]$ | 0 | Length of Burst Access. The device provides the capability to Write from <br> 4 to 511 locations in a single burst. The NSE automatically decrements the <br> value by one after each successive writing of the data or mask array. It must <br> be reinitialized before the next Burst-Write operation. |
|  | $[71: 28]$ |  | Reserved. |

### 5.4.8 Next-free Address Register (NFA)

The NFA register is used only when the device operates in the Non-Enhanced mode. The NFA register's Index field (Table 5-9) holds the address of the highest priority free entry in the table. When the table is full, the Index field will be set to all 1 s . When all entries in the device is full, the Ayama 10000 will assert FULO[1:0] to "11".


Figure 5-15. Next-free Address Register
Table 5-9. NFA Register Description

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| Index | $[\mathrm{N}: 0]$ | 0 | Index. The address index of the next-free entry location. N = 17 for <br> CYNSE10512, 16 for CYNSE10256, 15 for CYNSE10128. |
|  | $[71: \mathrm{M}]$ |  | Reserved. $\mathrm{M}=18$ for CYNSE10512, 17 for CYNSE10256, 16 for <br> CYNSE10128. |

### 5.4.9 Configuration Register (CONFIG)

The CONFIG register is valid only when the device operates in the Non-Enhanced mode. Table 5-10 describes the information register fields.


Figure 5-16. Configuration Register
Table 5-10. Configuration Register Description

| Field | Range (decimal) | Initial Value (binary) | Description |
| :---: | :---: | :---: | :---: |
| CFG | $[\mathrm{N}: \mathrm{N}-1]$ $[\mathrm{N}-2: \mathrm{N}-3]$ $\ldots$ $[1: 0]$ $\mathrm{N}=63$ for CYNSE10512, 31 for CYNSE10256, 15 for CYNSE10128. | 0 | Partition Configuration. In the Non-Enhanced mode, Ayama 10000 is internally divided into 32/16/8 partitions corresponding to CYNSE10512/256/128 respectively. Each two bits configures one partition as encoded below: <br> 00: $8 \mathrm{~K} \times 72$ <br> 01: $4 \mathrm{~K} \times 144$ <br> 10: $2 \mathrm{~K} \times 288$ <br> 11: Disabled (Does not reduce power consumption in a Search operation) Bit[1:0] configures the first partition, Bit[3:2] configures the second partition and so on. Bit [15:0] of this register is aliased in Bit[24:9] of the Command register. Modification to Bit[15:0] of this field will affect the CFGA field in the Command register and vice versa. |
| Reserved | [71:N + 1] |  | Reserved. |

## PRELIMINARY

### 5.4.10 Hardware Register (HARDWARE)

The Hardware register controls the drive strength of the groups of signals as listed in Section 6.0. Table 5-11 shows the fields that control each of the group and the output signals associated with it.


Figure 5-17. Hardware Register
Table 5-11. Hardware Register Description

| Field | Range (decimal) | Initial Value (binary) | Description |
| :---: | :---: | :---: | :---: |
|  | [1:0] |  | Reserved. |
| IOJTAG | [3:2] | 11 | JTAG I/Os. Sets the drive strength for the I/O. By default it is set to "11". The following output signal is part of this group: TDO. <br> The LVCMOS I/O drive strength for encoding is as listed below: $\begin{aligned} & \text { 00: } 2 \mathrm{~mA} \\ & \text { 01: } \mathrm{mA} \\ & \text { 10: } 16 \mathrm{~mA} \\ & \text { 11: } 24 \mathrm{~mA}\left(\mathrm{~V}_{\mathrm{DDQ}}=2.5 \mathrm{~V}\right) ; 20 \mathrm{~mA}\left(\mathrm{~V}_{\mathrm{DDQ}}=1.8 \mathrm{~V}\right) \end{aligned}$ <br> The HSTL I/O drive strength for encoding is as listed below: <br> 00: 8 mA (HSTL I) <br> 01: Reserved <br> 10: Reserved <br> 11: 17 mA (HSTL II) |
| IOCAS | [5:4] | 11 | Cascade I/Os. The following output signals are part of this group: LHO, BHO and FULO. <br> Refer to IOJTAG above for I/O drive strength encoding. |
| IOSRAM | [7:6] | 11 | SRAM I/Os. The following output signals are part of this group: SADR, CE_L, WE_L, OE_L and ALE_L. <br> Refer to IOJTAG above for I/O drive strength encoding. |
| IODQ | [9:8] | 11 | Command and DQ Bus I/Os. The following output signals are part of this group: DQ, ACK, EOT, SSF, SSV, PAR, PARERR_L, MULTI_HIT, and FULL. <br> Refer to IOJTAG above for I/O drive strength encoding. |
|  | [63:10] |  | Reserved. |
|  | [71:64] | 0 | Reserved. This field must be set to 0 . |

### 5.4.11 Parity Control Register (PARITY)

Table 5-12 describes the Parity Control Register fields. This register is only active when the device is in the Enhanced mode.


Figure 5-18. Parity Control Register
Table 5-12. Parity Control Register Description

| Field | Range <br> (decimal) | Initial Value (binary) | Description |
| :---: | :---: | :---: | :---: |
| INDEX | [18, N:0] | 0 | Index. This field contains the highest priority parity error index. When a parity error is detected, the global priority encoder selects the highest priority parity error out of the entire Core. <br> Note that if another Parity operation is performed, this field is updated based upon that operation. <br> N = 17 for CYNSE10512, 16 for CYNSE10256 (bit [17] is reserved), 15 for CYNSE10128 (bits [17:16] are reserved). Bit[18] is used to indicate whether a mask ( $=1$ ) or data (=0) entry contained the error. |
|  | [27:19] |  | Reserved. |
| BMULTI | [28] | 0 | Multi DQ Parity Error Status Bit. This field is set to 1 when multiple errors were detected during a bus transfer. It is also set to 1 when new parity error occurs and BERR is set. This bit can only be cleared by a user Write. |
| BERR | [29] | 0 | DQ Parity Error Status Bit. This bit is set when a parity error is detected during a data transfer across the DQ bus. This bit can only be cleared by a user Write. |
| MULTI | [30] | 0 | Multi-Parity Error Status Bit. This bit is set when more than one parity error in the Core is detected during the Parity operation. It also updates when a new parity error occurs and ERR is set. This bit can only be cleared by a user Write. |
| ERR | [31] | 0 | Parity Error Status Bit. This bit is set when any parity error in the Core is detected during the Parity operation. This bit can only be cleared by a user Write. |
| ADR | [50, M:32] | 0 | Current Address. After a parity check, the address in this field is incremented and is ready for the next address to check for parity. When the Parity operation finishes and an error is detected, assuming no intervening new Parity operations, this field will point to the next entry address to be checked. Bit[50] selects between mask (=1) or data (=0) array. As the address is incremented, this bit is treated as the LSB and toggles before Bit[34]. Bit[33:32] are always 0 because Read Parity operation checks 4 adjacent 72-bit entries. M = 49 for CYNSE10512, 48 for CYNSE10256 (bit [49] is reserved), 47 for CYNSE10128 (bits [49:48] are reserved). |
|  | [71:51] |  | Reserved. |

### 5.4.12 Control Register (CPR[0:15])

These registers are active only when the device is in the Enhanced Mode. During a Search operation, selecting a GMR will automatically select one of the CPRs to participate in the search as shown in Figure 5-19.

GMR Index CPR Addr CPR Index

| 0 | 64 | 0 |
| :---: | :---: | :---: |
| 1 | 65 | 1 |
| 2 | 66 | 2 |
| 3 | 67 | 3 |
| 4 | 68 | 4 |
| 5 | 69 | 5 |
| 6 | 70 | 6 |
| 7 | 71 | 7 |
| 8 | 72 | 8 |
| 9 | 73 | 9 |
| 10 | 74 | 10 |
| 11 | 75 | 11 |
| 12 | 76 | 12 |
| 13 | 77 | 13 |
| 14 | 78 | 14 |
| 15 | 79 | 15 |

Figure 5-19. Selection of the CPR through GMR Index
Table 5-13 shows the fields of the CPR.


Figure 5-20. Control Register

Table 5-13. Control Register

| Field | Index <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| PRIORITY | $[39: 0]$ |  | Reserved. |
| Mini-Key | $[47: 40]$ | 0 | Software Priority (Soft Priority). This field contains the software priority <br> for the associated command. Smaller numeric value is higher in priority. <br> 0x00 is the highest priority and 0xFF is the lowest priority. |
| FLG | $[48]$ | 0 | Mini-Key. This field contains the Mini-Key to be used for the associated <br> command. |
|  | $[71: 49]$ |  | Soft Priority Comparison Flag. When set to 1, Search comparison is with <br> entries that has Soft Priority value equal to or higher (lower priority) than <br> PRIORITY. When set to 0, comparison is only with equal value. |
|  |  | Reserved. |  |

### 5.4.13 Search Result Register (SRR[15:0])

The SRR register is only active when the device is in the Enhanced mode. It contains status information about where the nextfree entry is and what kind of entry it is. There are sixteen SRRs; one SRR associated with one CMPR. The SSR is updated on a Search operation regardless of hit or miss. Two SRRs are used in one Search operation when MSE is set. The second SRR is automatically selected to be the register one index higher. Table 5-14 below details the SRR fields.


Table 5-14. Search Result Register

| Field | Index (decimal) | Initial Value (binary) | Description |
| :---: | :---: | :---: | :---: |
| INDEX | [ $\mathrm{N}: 0]$ | 0 | Index. This field contains the Hit or Miss index inside the Core. $\mathrm{N}=17$ for CYNSE10512, 16 for CYNSE10256, 15 for CYNSE10128. |
|  | [23:M] |  | Reserved. M = 18 for CYNSE10512, 17 for CYNSE10256, 16 for CYNSE10128. |
| Mini-Key | [31:24] | 0 | Mini-Key. This field contains a copy of the Mini-Key value selected for the Search operation. The value comes from the selected CPR. |
| PRIORITY | [39:32] | 0 | Soft Priority. This field holds the priority value of the sub-block where a successful search occurs. Otherwise it holds the priority of the next-free entry sub-block. If there are no free entries, this field is set to the selected CPR's Soft Priority value. This field is not valid when STATUS value is Taken. |
| STATUS | [43:40] | 0 | Next-free Entry Status. This field contains the status information for the next-free entry. <br> The STATUS value is encoded as described below: 0000: Single match. Search hit and there is a single match. 0010: Single free entry. Search miss and there is a single free entry. 0100: Single free sub-block. Search miss and there is a single free sub-block. 0110: Single free block. Search miss and there is a single free block. 0001: Multiple matches. Search hit and there are multiple matches. 0011: Multiple free entries. Search miss and there are multiple free entries 0101: Multiple free sub-blocks. Search miss and there are multiple free sub-blocks. 0111: Multiple free blocks. Search miss and there are multiple free blocks. 1000: Taken. Search miss and there are no free entries. |
|  | [62:44] |  | Reserved. |
| FLG | [63] | 0 | Flag. When set to 1, this flag indicates that a Search operation resulted in a miss, this device has a free entry and the upstream devices in a cascade have a Search miss with no free entries reported. Note that a device with no free entries can still have free blocks or sub-blocks. |
|  | [71:64] |  | Reserved. |

Table 5-15 below shows the different parts of the INDEX field of the SRR.
Table 5-15. SRR's INDEX Composition Based on STATUS

| STATUS | INDEX[17:11] | INDEX[10:9] | INDEX[8:0] |
| :--- | :---: | :---: | :--- |
| Hit | Block ID | Sub-block ID | Hit Entry Index |
| Free Entry | Block ID | Sub-block ID | Free Entry Index |
| Free Sub-block | Block ID | Sub-block ID | All zeros |
| Free Block | Block ID | All zeros | All zeros |
| Taken | Undefined | Undefined | Undefined |

### 5.4.14 Block Mini-Key Register (BMR)

The BMR is only accessible when the device is in the Enhanced mode. There is one BMR for each block in the device. The following table (Table 5-16) shows the BMR fields.


Figure 5-22. Block Mini-Key Register
Table 5-16. Block Mini-Key Register Description

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| NES | $[1: 0]$ | 0 | NSE Entry Size. This field selects the entry width for the associated block. A <br> Search operation that is of different size than the NES will cause the block to not <br> participate in the search. <br> The NES encoding is as follows: <br> 00: 72-bit <br> 01: 144-bit <br> 10: 288-bit <br> 11: 576-bit <br> For proper free-entry address computation, this NES field must be set before <br> initializing the entries in a block. The entries of a block must then be initialized to <br> a known value before accessing the block. |
| Mini-Key3 | $[39: 32]$ | 0 | Reserved. |
| Mini-Key2 | $[47: 40]$ | 0 | Mini-Key \#3. There are four Mini-Key fields in each BMR. When an operation <br> occurs, all four fields are checked against the Mini-Key in the selected CPR by the <br> command. If there is a match, the associated block is enabled to participate in the <br> operation. |
| Mini-Key1 | $[55: 48]$ | 0 | Mini-Key \#2. See Mini-Key \#3 description. |
| Mini-Key0 | $[63: 56]$ | 0 | Mini-Key \#1. See Mini-Key \#3 description. |
|  | $[71: 64]$ |  | Reserved. |

### 5.4.15 Block Priority Register (BPR)

The BPR is only accessible when the device is in the Enhanced mode. There is one BPR for each block in the device. For each Priority in the BPR, there is an alias address (Block Priority Register Address) which allows individual priorities to be updated. Table $5-17$ shows the BPR fields.


Figure 5-23. Block Priority Register
Table 5-17. Block Priority Register Description

| Field | Range (decimal) | Initial Value (binary) | Description |
| :---: | :---: | :---: | :---: |
| PRIORITY3 | [7:0] | 0 | Soft Priority \#3. There are four Priority fields in each BPR. Each Priority represents the priority of a sub-block located within the block associated with this register. Priority value of 00 (hex) is highest and FF (hex) is lowest in priority. <br> The addresses in a block associated with each Soft Priority is as follows: <br> Entry Index 0 to 511: Priority0 <br> Entry Index 512-1023: Priority1 <br> Entry Index 1024-1535: Priority2 <br> Entry Index 1536-2047: Priority3 |
| PRIORITY2 | [15:8] | 0 | Soft Priority \#2. See Soft Priority \#3 description. |
| PRIORITY1 | [23:16] | 0 | Soft Priority \#1. See Soft Priority \#3 description. |
| PRIORITY0 | [31:24] | 0 | Soft Priority \#0. See Soft Priority \#3 description. |
|  | [59:32] |  | Reserved. |
| V3 | [60] | 0 | V \#3. There are four V fields in each BPR. Each field represent the valid bit for a sub-block within the block associated with this register. If this bit is set to 1 and the Soft Priority in the CPR selected by the operation matches, the associated sub-block will participate in the operation. If this bit is set to 0 , the associated sub-block will not participate in a Search operation. |
| V2 | [61] | 0 | V \#2. See V \#3 description. |
| V1 | [62] | 0 | $V$ \#1. See V \#3 description. |
| V0 | [63] | 0 | V \#0. See V \#3 description. |
|  | [71:64] |  | Reserved. |

### 5.4.16 Block Parity Register (BPAR)

The BPAR is only accessible when the device is in the Enhanced mode. There is one BPAR for each block in the device. Table 518 shows the BPR fields.


Figure 5-24. Block Parity Register
Table 5-18. Block Parity Register Description

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| PERR | $[3: 0]$ | 0000 | Parity Error. This field contains the status of a Parity operation. It is set to 1 when any parity <br> error is detected during the Parity operation on the associated block. Each bit corresponds <br> to one of the four x72 entries checked during the Parity operation. Bit[0] corresponds to the <br> lowest address. This field is sticky, i.e., it can only be cleared by a user write to the register. <br> This field contains only this block's status. To clear this bit, the user must write a "1" to this <br> bit location. Writing a "0" will preserve the old value. |
|  | $[30: 4]$ |  | Reserved. |
| EN | $[31]$ | 0 | Enable Parity Checking. This field enables parity checking for the associated block. When <br> set to 1, the associated block will participate in Parity operation. |
|  | $[71: 32]$ |  | Reserved. |

### 5.4.17 Block NFA Register (BNFA)

The BNFA is only accessible when the device is in the Enhanced mode. There is one BNFA for each block in the device. Table 519 shows the BNFA fields.


Figure 5-25. Block NFA Register

Table 5-19. Block NFA Register Description

| Field | Range (decimal) | Initial Value (binary) | Description |
| :---: | :---: | :---: | :---: |
| NFA3 | [8:0] | 0 | Next-free Address for Sub-block \#3. This field contains the address/index of the nextfree entry within the sub-block of the block associated with this register. If the entry size is larger than $x 72$, the least significant bits will be set to 0 as follows: <br> x144: NFAx[0] = '0', <br> x288: NFAx[1:0] = "00", <br> x576: NFAx[2:0] = "000". |
|  | [13:9] |  | Reserved. |
| MULTI3 | [14] | 1 | Multiple Free Entries in Sub-block \#3. This field contains the multiple free entry status. If there are multiple free entries in the sub-block, this bit is set to 1 . |
| F3 | [15] | 0 | Free Entry in Sub-block \#3. This field indicates the sub-block full status. If this field is set to 1 , the sub-block is full and there are no free entries. If the field is set to 0 , the sub-block is not full and there is a free entry. |
| NFA2 | [24:16] | 0 | Next-free Address for Sub-block \#2. See NFA3 description. |
|  | [29:25] |  | Reserved. |
| MULTI2 | [30] | 1 | Multi Free Entry in Sub-block \#2. See MULTI3 description. |
| F2 | [31] | 0 | Free Entry in Sub-block \#2. See F3 description. |

Table 5-19. Block NFA Register Description (continued)

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |
| :---: | :---: | :---: | :--- |
| NFA1 | $[40: 32]$ | 0 | Next-free Address for Sub-block \#1. See NFA3 description. |
|  | $[45: 41]$ |  | Reserved. |
| MULTI1 | $[46]$ | 1 | Multi Free Entry in Sub-block \#1. See MULTI3 description. |
| F1 | $[47]$ | 0 | Free Entry in Sub-block \#1. See F3 description. |
| NFA0 | $[56: 48]$ | 0 | Next-free Address for Sub-block \#0. See NFA3 description. |
|  | $[61: 57]$ |  | Reserved. |
| MULTIO | $[62]$ | 1 | Multi Free Entry in Sub-block \#0. See MULTI3 description. |
| F0 | $[63]$ | 0 | Free Entry in Sub-block \#0. See F3 description. |
|  | $[71: 64]$ |  | Reserved. |

### 5.4.18 Block Priority Register Aliases (BPRA)

The BPRA is only accessible when the device is in the Enhanced mode. There are four BPRAs for each block. These pseudo registers provide an alternate means to update the associated block's BPR. The fields of these BPRAs exactly match the BPR fields. Please see the corresponding BPR fields for the descriptions of the BPRAs fields.
Table 5-20 shows the BPRA fields for BPR's Priority0.


Figure 5-26. Block Priority Register Aliases

Table 5-20. Block Priority Register Alias for Priority \#0 Fields

| Field | Range <br> (decimal) | Initial Value <br> (binary) |  | Description |
| :--- | :---: | :---: | :--- | :--- |
|  | $[23: 0]$ |  | Reserved. |  |
| PRIORITY0 | $[31: 24]$ | 0 | Priority \#0. |  |
|  | $[62: 32]$ |  | Reserved. |  |
| V0 | $[63]$ | 0 | V \#0. |  |
|  | $[71: 64]$ |  | Reserved. |  |

Table 5-21 shows the BPRA fields for BPR's Priority1.
Table 5-21. Block Priority Register Alias for Priority \#1 Fields

| Field | Range <br> (decimal) | Initial Value <br> (binary) |  | Description |
| :--- | :---: | :---: | :--- | :--- |
|  | $[15: 0]$ |  | Reserved. |  |
| PRIORITY1 | $[23: 16]$ | 0 | Priority \#1. |  |
|  | $[61: 24]$ |  | Reserved. |  |
| V1 | $[62]$ | 0 | V\#1. |  |
|  | $[71: 63]$ |  | Reserved. |  |

Table 5-22 shows the BPRA fields for BPR's Priority2.
Table 5-22. Block Priority Register Alias for Priority \#2 Fields

| Field | Range <br> (decimal) | Initial Value <br> (binary) |  | Description |
| :---: | :---: | :---: | :--- | :--- |
|  | $[7: 0]$ |  | Reserved. |  |
| PRIORITY2 | $[15: 8]$ | 0 | Priority \#2. |  |
|  | $[60: 16]$ |  | Reserved. |  |
| V2 | $[61]$ | 0 | V \#2. |  |
|  | $[71: 62]$ |  | Reserved. |  |

Table 5-23 shows the BPRA fields for BPR's Priority3.
Table 5-23. Block Priority Register Alias for Priority \#3 Fields

| Field | Range <br> (decimal) | Initial Value <br> (binary) | Description |  |
| :---: | :---: | :---: | :--- | :--- |
| PRIORITY3 | $[7: 0]$ | 0 | Priority \#3. |  |
|  | $[59: 8]$ |  | Reserved. |  |
| V3 | $[60]$ | 0 | V \#3. |  |
|  | $[71: 61]$ |  | Reserved. |  |

### 5.5 Multi-Hit Description

For a Search operation, Multi-Hit is set when there are multiple matching entries in the array (Non-Enhanced) or in the selected blocks (Enhanced). For a Learn operation, Multi-Hit is set when there are multiple free entries in the array (Non-Enhanced) or in the selected blocks (Enhanced). Multi-Hit maintains its value until another operation changes it.
In the Non-Enhanced mode, the Multi-Hit signal is valid four cycles after the command is issued, regardless of the setting of TLSZ and HLAT. In the Enhanced mode, the Multi-Hit signal is valid at the same time as SSV.

### 5.6 Clocks

If the CLK_MODE pin is LOW, Ayama 10000 receives the CLK2X and PHS_L signals. It uses the PHS_L signal to divide CLK2X and generate an internal clock (CLK ${ }^{[6]}$ ), as shown in Figure 5-27. If the CLK_MODE pin is HIGH, Ayama 10000 receives CLK1X only. Ayama 10000 uses an internal phase-locked loop (PLL) to lock the frequency of CLK1X and generates the internal clock CLK, as shown in Figure 5-28. Also noted on these figures are cycles A and B. In CLK2X mode, cycle A begins on the rising edge of CLK2X, when PHS_L is Low, and ends on the next rising edge. Cycle B begins on the rising edge of CLK2X when PHS_L is High, and ends on the subsequent CLK2X rising edge. For CLK1X mode, the falling edge of CLK1X is considered the end of cycle A, while the rising edge after that is considered the end of cycle B. Valid data must be available for the NSE at the END of any cycle. Note. For the purpose of showing timing diagrams, all such diagrams in this document will be shown in CLK2X mode. For a timing diagram in CLK1X mode, the following substitution can be made (see Figure 5-29).


Figure 5-27. Ayama 10000 Clocks (CLK2X and PHS_L)


Figure 5-28. Ayama 10000 Clocks (CLK1X)


Figure 5-29. Ayama 10000 Clocks for All Timing Diagrams

## Notes:

6. "CLK" is an internal clock signal.
7. Any reference to "CLK" cycles means one cycle of CLK.
8. Only supported in Non-Enhanced mode.

### 5.7 Phase-Locked Loop

When the device first powers up, it takes 0.5 milliseconds (ms) after the power supplies are stable to lock the internal PLL. During this time period, the RST_L must be held LOW for proper power-up. All signals to the device in CLK1X mode are sampled by a clock that is generated by multiplying CLK1X by two. Since the PLL has a locking range, the device will only work between the range of frequencies specified in the timing specification wave form section of this data sheet (see Section 10.0, "AC Timing Parameters, Waveforms and Test Conditions," on page 138).

### 5.8 Pipeline Latency

Pipeline latency is used to give enough time for a cascaded system's arbitration logic to determine the device that will drive the output of an operation on the SRAM bus. The Ayama 10000 has a default of 4 CLK1X pipeline latencies but more latency can be added as necessary. The number of additional pipeline stages is set in the TLSZ and HLAT fields of the COMMAND Register. The number of pipeline stages also controls the maximum operating speed for a single Ayama 10000 NSE. Table 5- 24 lists the additional pipeline stages and the maximum operating speed.
Table 5-24. Pipeline Stages and Maximum Operating Speed.

| TLSZ | Additional CLK1X Cycle Latency | Total Search CLK1X Cycle Latency | Maximum Operating Speed <br> (CLK1X/CLK2X) |
| :---: | :---: | :---: | :---: |
| 00 | 0 | 4 | $83 / 166 \mathrm{MHz}$ |
| 01 | 1 | 5 | $100 / 200 \mathrm{MHz}$ |
| 10 | 2 | 6 | $133 / 266 \mathrm{MHz}$ |
| 11 | Invalid | Invalid | Invalid |

Internal register for configuration: CMD

### 5.9 DQ Bus Encoding of Ayama 10000 Address Space

A set of parameters for an operation must be provided in the DQ bus to the NSE along with the command sent in the CMD bus. This section covers the encoding of the parameters expected in the DQ bus. There are two ways of addressing an entry location or an internal register within the device: Direct and Indirect. The internal registers can only use Direct addressing while Data array, Mask array and SRAM access operations can use either Direct or Indirect. Indirect addressing allows the use of the SSR register INDEX field as the address for a Read, Write and Learn operations. Indirect Read operation on the internal registers will return undefined values.

### 5.9.1 Addressing the Data Array, Mask Array and External SRAM

The following table (Table 5-25) lists the parameters for addressing the Data array, Mask array and external SRAM.


Figure 5-30. Data Array, Mask Array and External SRAM Address Space Encoding
Table 5-25. Data Array, Mask Array and External SRAM Address Space Encoding

| Field | Range (decimal) | Description |
| :---: | :---: | :---: |
| ADDRESS | [ $\mathrm{N}: 0]$ | Address. This field contains the location of the entry to be accessed on Direct Addressing operations. $\mathrm{N}=17$ for CYNSE10512, 16 for CYNSE10256, 15 for CYNSE10128. <br> Note that on a burst Read or Write operation, the appropriate burst register (WBURADR or RBURADR) INDEX field is used as the address. |
|  | [M] | Reserved. M = 18 for CYNSE10512, [18:17] for CYNSE10256, [18:16] for CYNSE10128. |
| TARGET | [20:19] | Target Area Select. This field indicates in what context the access takes place. It is encoded as follows: <br> 00: Access the Data array <br> 01: Access the Mask array <br> 10: Access the external SRAM <br> 11: Access the Internal Registers (Refer toSection 5.9.2) |
| CHIPID | [25:21] | Device ID. This field indicates which NSE device should respond to the READ or WRITE operation. CHIPID value "11111" indicates a broadcast operation. |
| SSR | [28:26] | SSR Index. This field selects the SSR for Indirect accesses. |
| INDIRECT | [29] | Indirect Addressing Enable. <br> 1: Indirect. <br> When DQ[30] is 0 , the selected SSR register INDEX field is used to generate the address as follow: \{SSR[17:3], SSR[2] \| DQ[2], SSR[1] | DQ[1], SSR[0] | DQ[0]\}. <br> To issue a Read Parity command, this bit must be set to 1 . The address of the entry location to be checked is taken from the PARITY control register's ADR field. Note that Read Parity command can be issued only on a Read command. Issuing an Indirect Write with Bit[30] set to 1 will result in NoOperation. <br> 0: Direct. DQ[17:0] contains the address for the operation. |
| PARITY | [30] | Read Parity. This bit must be set to 1 to issue a Read Parity command. It is valid only when DQ[29] is also set to 1 . |
|  | [71:31] | Reserved. |

The address generation of the SADR bits varies depending on the operation being performed. The following table (Table 5-26) shows the SRAM address generation for the various operations.
Table 5-26. SRAM Address Generation

| Command | SRAM Operation | SADR[M+8:M+6] $^{[9]}$ | SADR[M+5:M+1] | SADR[M:0] ${ }^{[9,10]}$ |
| :---: | :---: | :---: | :---: | :---: |
| Search | Read | CMD[8:6] | ID[4:0] | Index[M:0] |
| Learn | Write | CMD[8:6] | ID[4:0] | NFA/SRR[M:0] $]^{[11]}$ |
| SRAM PIO Read | Read | CMD[8:6] | ID[4:0] | DQ[M:0] |
| SRAM PIO Write | Write | CMD[8:6] | ID[4:0] | DQ[M:0] |
| Indirect Read | Read | CMD[8:6] | ID[4:0] | SSR[M:0] \| DQ[2:0] ${ }^{[12]}$ |
| Indirect Write | Write | CMD[8:6] | ID[4:0] | SSR[M:0] \| DQ[2:0] ${ }^{[12]}$ |

## Notes:

9. When MultiSearch feature is enabled, SADR $[M+8]$ is not used and SADR $[M]$ will be 0 to indicate Array 0 output or 1 to indicate Array 1 output.
10. $M=17$ for CYNSE10512; $M=16$ for CYNSE10256; $M=15$ for CYNSE10128.
11. Non-Enhanced mode uses NFA register. Enhanced mode uses SRR register.
12. $\operatorname{SSR}[2: 0]$ is OR-ed with $\operatorname{DQ}[2: 0]$ to generate the SADR[2:0] values.

### 5.9.2 Addressing the Internal Registers

The following table (Table 5-27) details the parameters expected (in DQ bus) to access the internal registers of the NSE.


Figure 5-31. Internal Register Address Space Encoding
Table 5-27. Internal Register Address Space Encoding

| Field | Range <br> (decimal) | Description |
| :---: | :---: | :--- |
| REGSEL | $[10: 0]$ | Register Address Selected. This field selects which internal register to address. Table 5- <br> 3 lists the registers that are available. |
| BLKNUM | $[17: 11]$ | Block Number. This field selects the block within the device that will participate in the <br> operation. It is only used when accessing block specific internal registers (BMR, BPR, <br> BPAR, BNFA and BPRA0-3). For other internal register accesses, this field must be set to 0. |
| RSEL | $[20: 19]$ | Reserved. <br> Register Area Select. This field indicates in what context the access takes place. It must <br> be set to "11". |
| CHIPID | $[25: 21]$ | Device ID. This field indicates which NSE device should respond to the READ or WRITE <br> operation. CHIPID value "11111" indicates a broadcast operation. |
|  | $[28: 26]$ | Reserved. |
| INDIRECT | $[29]$ | Indirect Addressing Enable. This bit must be cleared to 0. |
|  | $[71: 30]$ | Reserved. |

### 5.10 Depth Cascading

The NSE application can depth-cascade the devices to various table sizes of different widths (72-bit, 144-bit, 288-bit or 576-bit). The devices perform all the necessary arbitration to decide which device will drive the SRAM bus. Some operations and features are not cascadable, which means that the operation or feature is on a device-by-device basis and the results are not propagated to the next device. Table 5-28 lists those operations and features. The following subsections covers the interconnects when the devices in a cascade operates in the Non-Enhanced mode or Enhanced Mode with MSE set to 0 (MultiSearch disabled). For device interconnects when operating in Enhanced mode with MultiSearch enabled, please refer to Figure 6-14.

Table 5-28. Cascadability of Operations and Features

| Operations / \# of Devices | Non-Enhanced |  |  | Enhanced Mode with MSE = 0 |  |  | Enhanced Mode with MSE = 1 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 1 | 2-8 | 9-31 | 1 | 2-8 | 9-31 | 1 | 2-8 | 9-31 |
| MultiSearch Command | No | No | No | No | No | No | Yes | Yes | No |
| Learn Command | Yes | Yes | No | Yes | Yes | $\mathrm{No}{ }^{[13]}$ | Yes | $\mathrm{No}{ }^{[3]}$ | $\mathrm{No}^{[13]}$ |
| Soft Priority | No | No | No | Yes | $\mathrm{No}{ }^{[13]}$ | $\mathrm{No}{ }^{[13]}$ | Yes | $\mathrm{No}{ }^{[13]}$ | $\mathrm{No}{ }^{[13]}$ |
| FULL | Yes | Yes | No | Yes | Yes | No | Yes | No | No |
| MULTI_HIT | Yes | No | No | Yes | No | No | Yes | No | No |

### 5.10.1 Depth Cascading up to Eight Devices in One Block

Figure 5-32 shows the interconnection of up to eight devices in a cascade to form $2 \mathrm{M} \times 72,1 \mathrm{M} \times 144,512 \mathrm{~K} \times 288$, or $256 \mathrm{~K} \times$ 576 tables. Each NSE asserts the LHO[1] and LHO[0] signals to inform downstream devices of its result. LHI[6:0] signals for a device are connected to LHO signals of the upstream devices. The host ASIC must program the TLSZ to 01 (binary) for each of up to eight devices in a block. Only a single device drives the SRAM bus in any single cycle.
Note:
13. Software solutions are possible for these cases. Please refer to specific application notes.


Figure 5-32. Depth Cascading in a Single Block

### 5.10.2 Depth Cascading up to 31 Devices in 4 Blocks

Figure $5-33$ shows the cascading of up to four blocks. Each block except the last contains up to eight Ayama 10000 devices, and the interconnection within each with the cascading of up to eight devices in a block was shown in the previous subsection. Note. The interconnection between blocks for depth cascading is important. For each Search, a block asserts BHO[2], BHO[1], and $\mathrm{BHO}[0]$. The $\mathrm{BHO}[2: 0]$ signals for a block are taken only from the last device in that block. For all other devices within that block, these signals stay open. The host ASIC must program TLSZ to 10 (binary) in each of the devices for cascading up to 31 devices (in up to four blocks).


Figure 5-33. Depth Cascading 4 Blocks

### 5.10.3 Depth Cascading for a FULL Signal

Bit[0] of each of the 72-bit entries is designated as a special bit ( $1=$ occupied, $0=$ empty). For each Learn or PIO Write to the data array, each device asserts FULO[1] or FULO[0] depending on whether or not it has any empty locations within it (see Figure 5-34). Each device combines the FULO signals from the devices above it with its own full status to generate a FULL signal that gives the full status of the table up to the device asserting the FULL signal. Figure 5-34 shows the hardware connection diagram for generating the FULL signal that goes back to the ASIC. In a depth-cascaded block of up to eight devices, the FULL signal from the last device should be fed back to the ASIC controller to indicate the fullness of the table. The FULL signal of the other devices should be left open. Note. The Learn instruction is supported for only up to eight devices, whereas FULL cascading is allowed only for one block in tables containing more than eight devices. In tables for which a Learn instruction is not going to be used, the bit[0] of each 72-bit entry should always be set to 1 .


Figure 5-34. FULL Signal Generation in a Cascaded Table

### 5.11 Device Selection in a Cascaded System

On a Direct Read operation, if the CHIPID field matches the current device's ID[4:0], this device will respond to the read request. If the CHIPID field does not match, the device will not respond. Note that if the CHIPID does not match any device in the cascade, no read acknowledge will be generated. If the CHIPID is set to broadcast ("11111", binary), the device with the LDEV bit set to 1 will respond to the read request.
On an Indirect Read operation, if the CHIPID field matches the current device's ID[4:0], this device will respond to the read request. If the CHIPID field does not match, the device will not respond. Note that if the CHIPID does not match any device in the cascade, no read acknowledge will be generated. If the CHIPID is set to broadcast ("11111"), each device examines its SSR register's VAL and GVAL bits. The device with one of these bits set responds to the read request. If none of these bits are set (this occurs when a Search has not been done after a reset), no read acknowledge will be returned.
On a Direct Write operation, if the CHIPID field matches the current device's ID[4:0], this device will perform the write request. If the CHIPID field does not match, the device will not respond. If the CHIPID is set to broadcast, all devices write to the desired location.

On an Indirect Write operation, if the CHIPID field matches the current device, this device will perform the write request. If the CHIPID field does not match, the device will not respond. If the CHIPID is set to broadcast, different actions occur based upon the target. If the target is an internal register, the write request is ignored. If the target is a Data or Mask array, the device with the VAL field of the SSR register set performs the write request. If the target is external SRAM, the device with the LRAM field set will drive the SRAM signals.

### 5.12 Power-up Sequence

Ayama 10000 requires that the power supplies follow a known sequence to ensure successful device power-up to set the device to its initial state. RST_L should be held Low before the power supplies ramp-up and must be held Low for a duration of time afterward. Clock signals (CLK1X/CLK2X and PHS_L) should start running after the power supplies become stable. All IO voltages ( $V_{D D Q}$, which includes $V_{D D Q \_A S I C}$ and $V_{D D Q}$ SRAM $)$ should only ramp up only after the core voltage ( $V_{D D}$ ) level reaches $90 \%$ point. The following describes the proper power-up sequence required to correctly initialize the Cypress Network Search Engines before functional access to the device can begin. The following steps are presented in order of priority.

1. Hold RST_L and TRST_L signals low and power up $V_{D D}$. Then power up $V_{D D Q}$ when $V_{D D}$ is stable. TRST_L can be tied to RST_L, tied low permanently, or driven asynchronously (more information on resetting JTAG in the JTAG section of the datasheet).
2. Start running CLK2X/CLK1X and PHS_L (if applicable) after $\mathrm{V}_{\text {DDQ }}$ powers up.
3. Hold RST_L low for at least $0.5 \mathrm{~ms}+\mathrm{t}_{\mathrm{RSTL}}$ after the clock signal is stable, then drive high.

RST_L should be set High with sufficient hold time with respect to CLK2X. Following steps 1 through 3 will power up the device gracefully and ensure proper operation of the device. Figure 5-35 illustrates the proper sequences of the power-up operation.


Figure 5-35. Proper Power-up Sequence
Note: The PLL will lose lock if the CLK2X/CLK1X or PHS_L (if applicable) stop transitioning.

### 6.0 Operations and Timing Diagrams

A master device, such as an ASIC controller, issues commands to the Ayama 10000 device using the CMD bus and CMDV signals. The following subsections describe the operation of these commands.

### 6.1 Command Encoding

The Ayama 10000 device implements four basic commands, as shown in Table 6-1. The Search command is a non-blocking operation which allows another operation to be issued immediately on the following cycle. Read, Write and Learn are blocking operations. There are also other derivative commands that the device supports. The operation of basic commands as well as the derivative commands are explained in more detail in the following sections.
The command code must be presented to CMD[1:0] while keeping the CMDV signal HIGH for two CLK2X cycles (cycles A and B) when the CLK_MODE pin is LOW. In CLK2X mode, the controller ASIC must align the instructions using the PHS_L signal. The command code must be presented to CMD[1:0] while keeping the CMDV signal HIGH for one CLK1X cycle when the CLK_MODE pin is HIGH. In CLK1X mode, cycle A ends on the falling edge of CLK1X and cycle B ends on the rising edge of CLK제X. Valid data must be present at the edge ending any given cycle for valid inputs. The CMD[10:2] field passes command parameters in cycles A and B. All commands must begin with cycle A operations.

Table 6-1. Command Codes

| Command Code <br> (binary) | Command |  |
| :---: | :---: | :--- |
| 00 | Read | Reads from one of the following: data array, mask array, device registers, or external SRAM. <br> Read command is also used to issue Read Parity command. |
| 01 | Write | Writes to one of the following: data array, mask array, device registers, or external SRAM. |
| 10 | Search | Searches the data array for a desired pattern using the specified register from the GMR <br> array and local mask associated with each data cell. |
| 11 | Learn | The device has internal storage for up to sixteen comparands that it can learn. The device <br> controller can insert these entries at the next-free address (as specified by the NFA register) <br> using the Learn instruction. |

### 6.2 Command Bus Parameters

Table 6.2.1, Table 6.2.2 and Table 6.2.3 list the command bus fields that contain the Ayama 10000 command parameters and their respective cycles.

### 6.2.1 Non-Enhanced Mode (EMODE = 0)



## Note:

14. The NSE density determines to which SADR field EADR[2:0] is mapped. In Ayama10128, SADR[23:21] gets EADR[2:0]; In Ayama10256, SADR[24:22] gets EADR[2:0]; In Ayama10512, SADR[25:23] gets EADR[2:0].

### 6.2.2 Enhanced Mode (EMODE =1) with MultiSearch Disabled $(M S E=0)$



### 6.2.3 Enhanced Mode (EMODE =1) with MultiSearch Enabled $(M S E=1)$



### 6.3 Read Command

In both the Non-Enhanced and Enhanced mode, the Read command can be issued to read data from the data array, mask array, NSE-associated SRAMs or internal registers. The Read can be a single or burst Read (Table 6-2). Burst Read can only be issued for accesses to the data or mask array locations. SRAM Read operation is covered in Section 6.7.1 to Section 6.7.3. In the Enhanced mode, the Read command is also used to issue the Read Parity command, which is issued to perform parity check on the data and mask array entries.
Read is a blocking operation and must be completed before the next operation can be issued.
Table 6-2. Single/Burst Read Command Parameters

| CMD Parameter <br> CMD[2] | Read Command | Description |
| :---: | :---: | :--- |
| 0 | Single Read | Reads a single location of the data array, mask array, NSE-associated SRAM or internal <br> registers. All access information is applied on the DQ bus. |
| 1 | Burst Read | Reads a block of locations from the data or mask array as a burst. RBURREG specifies <br> the starting address and the length of the data transfer from the data or mask array; it <br> also auto-increments the address for each access. All other access information is <br> applied on the DQ bus. |

### 6.3.1 Single Read

A single Read operation lasts six cycles (CLK1X) with the data driven out by the NSE on cycle 5 as illustrated in Figure 6-1.


Figure 6-1. Single-Location Read Cycle Timing
Read operation sequence:

- Cycle 1: The host ASIC applies the Read instruction on CMD[1:0] (CMD[2] $=0$ ) using CMDV $=1$, and the DQ bus supplies the address. The host ASIC selects the Ayama 10000 device for which ID[4:0] matches the DQ[25:21] lines. If DQ[25:21] = 11111, the host ASIC selects the Ayama 10000 with the LDEV bit set. The host ASIC also supplies SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6] in cycle A of the Read instruction if the Read is directed to the external SRAM.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive the DQ[71:0] bus and drives the ACK signal from $Z$ to LOW.
- Cycle 5: The selected device drives the Read data from the addressed location on the DQ[71:0] bus, and drives the ACK signal HIGH.
- Cycle 6: The selected device floats the DQ[71:0] to a three-state condition and drives the ACK signal LOW.

At the termination of cycle 6, the selected device releases the ACK line to a three-state condition. The Read instruction is complete and the next operation can begin.

### 6.3.2 Burst Read

The burst Read operation lasts $4+2 n$ CLK1X cycles, where n is the number of the burst length as specified by the BLEN field of the RBURREG. The BLEN field is automatically decremented after each Read of the burst, so the register must be reinitialized before another burst Read is issued. Instead of the address provided by the user, the address in the INDEX field of the RBURREG is used and incremented each cycle.
Figure 6-2 illustrates the timing diagram for the burst Read of the data or mask array.


Figure 6-2. Burst Read of the Data and Mask Arrays (BLEN = 4)
Burst Read operation sequence:

- Cycle 1: The host ASIC applies the Read instruction on CMD[1:0] (CMD[2] =1) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC selects the Ayama 10000 device where ID[4:0] matches the DQ[25:21] lines. If $D Q[25: 21]=11111$, the host ASIC selects the Ayama 10000 device with the LDEV bit set.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive the DQ[71:0] bus and drives ACK and EOT from Z to LOW.
- Cycle 5: The selected device drives the Read data from the address location on the $D Q[71: 0]$ bus and drives the ACK signal HIGH.
Cycles 4 and 5 repeat for each additional access until all the accesses specified in the BLEN field of RBURREG are complete. On the last data transfer, the Ayama 10000 drives the EOT signal HIGH.
Cycle (4+2n): The selected device drives the DQ[71:0] to a three-state condition, and drives ACK and EOT signals LOW.
At the termination of cycle $(4+2 n)$, the selected device floats ACK and EOT to a three-state condition. The burst Read operation is complete and the next operation can begin.


### 6.3.3 Read Parity

Data output of the Read Parity command should be ignored. Read Parity is a blocking operation only on the cycles as the normal Read operation even though the parity status signal (PARERR) is valid TLSZ cycles later. Figure shows an example of the PARERR update timing diagram with TLSZ set to "10" (two additional cycles of latency) to a total of eight cycles (six Read cycles plus two TLSZ cycles).

### 6.4 Write Command

The Write command can be issued to write to the data array, mask array, NSE-associated SRAMs or internal registers. The Write can be a single or burst Write (Table 6-3). Burst Write can only be issued for accesses to the data or mask array locations. SRAM Write operation is covered in Section 6.7.4 to Section 6.7.6. The Write command is also used to issue the Parallel Write command. Note that when Parity feature is enabled masks will be ignored and all bits will be written as presented in the DQ bus.

Write is a blocking operation and must be completed before the next operation can be issued.
Table 6-3. Single/Burst Write Command Parameters

| CMD Parameter <br> CMD[2] | Write Command | Description |
| :---: | :---: | :--- |
| 0 | Single Write | Writes a single location of the data array, mask array, NSE-associated SRAM or internal <br> registers. All access information is applied on the DQ bus. |
| 1 | Burst Write | Writes a block of locations to the data or mask array as a burst. WBURREG specifies the <br> starting address and the length of the data transfer from the data or mask array; it also <br> auto-increments the address for each access. All other access information is applied on <br> the DQ bus. |

### 6.4.1 Single Write

A single Write operation lasts 3 cycles (CLK1X) as illustrated in Figure 6-3.


Figure 6-3. Single Write Cycle Timing
Write operation sequence:

- Cycle 1A: The host ASIC applies the Write instruction to CMD[1:0] (CMD[2] = 0) using CMDV = 1, and the target address supplied on the DQ bus. The host ASIC also supplies the GMR index to mask the Write to the data or mask array location on \{CMD[10], CMD[5:3]\}. For SRAM WRITEs, the host ASIC must supply the SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6]. The host ASIC sets CMD[9] to 0 for a normal Write.
- Cycle 1B:The host ASIC continues to apply the Write instruction to CMD[1:0] (CMD[2] = 0) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC continues to supply the GMR index to mask the Write to the data or mask array locations in \{CMD[10], CMD[5:3]\}. The host ASIC selects the device where ID[4:0] matches the DQ[25:21] lines, or it selects all the devices when DQ[25:21] = 11111.
- Cycle 2: The host ASIC drives DQ[71:0] with the data to be written to the data array, mask array, or register location of the selected device.
- Cycle 3: Idle cycle. DQ bus should be driven to 0 .

At the termination of cycle 3, another operation can begin.

### 6.4.2 Burst Write

The burst Write operation lasts $2+n$ CLK1X cycles, where $n$ is the number of the burst length as specified by the BLEN field of the WBURREG. The BLEN field is automatically decremented after each Write of the burst, so the register must be re-initialized before another burst Write is issued. Instead of the address provided by the user, the address in the INDEX field of the WBURREG is used and incremented each cycle.

Figure 6-4 illustrates the timing diagram for the burst Write to the data or mask array.


Figure 6-4. Burst Write of the Data and Mask Arrays (BLEN = 4)
Burst Write operation sequence:

- Cycle 1A: The host ASIC applies the Write instruction to CMD[1:0] (CMD[2] = 1) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC also supplies the GMR index to mask the Write to the data or mask array locations in \{CMD[10], CMD[5:3]\}. The host ASIC sets CMD[9] to 0 for the normal Write.
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] (CMD[2] = 1) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC continues to supply the GMR index to mask the Write to the data or mask array locations in \{CMD[10], CMD[5:3]\}. The host ASIC selects the device for which ID[4:0] matches the DQ[25:21] lines. It selects all devices when DQ[25:21] = 11111.
- Cycle 2: The host ASIC drives the DQ[71:0] with the data to be written to the data or mask array location of the selected device. The Ayama 10000 device writes the data from the DQ[71:0] bus only to the subfield with the corresponding mask bit set to 1 in the GMR that is specified by the index \{CMD[10],CMD[5:3]\} supplied in cycle 1.
- Cycles 3 to $\mathbf{n + 1}$ : The host ASIC drives the DQ[71:0] with the data to be written to the next data or mask array location of the selected device (addressed by the auto-increment ADR field of the WBURREG register).

The Ayama 10000 device writes the data on the DQ[71:0] bus only to the subfield that has the corresponding mask bit set to 1 in the GMR specified by the index supplied in cycle 1 \{CMD[10],CMD[5:3]\}. The Ayama 10000 device drives the EOT signal LOW from cycle 3 to cycle $n$; the Ayama 10000 device drives the EOT signal HIGH in cycle $n+1$ ( $n$ is specified in the BLEN field of the WBURREG).

- Cycle n + 2: The Ayama 10000 device drives the EOT signal LOW.

At the termination of cycle $n+2$, the Ayama 10000 device floats the EOT signal to a three-state operation and the next instruction can be issued.

### 6.4.3 Parallel Write

In order to write the Data or Mask array faster for initialization, testing, or diagnostics, the user can issue a Parallel Write command. Parallel Write allows the user to specify one address and write multiple locations in the Core with the same data. Parallel Write only works with Direct addressing. If Indirect addressing is used, the operation will result in No-Operation. Parallel Write can also be done in burst operation.
In Non-Enhanced Mode, address bits $D Q[10: 1]$ specify which location to perform parallel write. $D Q[17: 11]$ defines a set of partitions, all of which write two $x 72$ entries (DQ[0] is ignored). For Ayama 10512, this corresponds to 64 parallel locations (32 8 Kx 72 partitions, 2 locations per partition).
In Enhanced Mode, address bits DQ[10:0] specify the location within a block. Parallel write only occurs on those blocks which match the Mini-Key(s) selected by the GMR field. For Ayama 10512, this corresponds to 128 parallel locations (128 blocks, 1 location per block).

### 6.5 Search Command

One of the key parameters that controls SEARCH operation is TLSZ (Statically programmed in Command Register). TLSZ controls the maximum number of devices that can be cascaded and the latency of SEARCH instruction, as shown in Table 6-4.
Table 6-4. TLSZ[1:0] Description

| TLSZ[1:0] | Max. Number of Devices <br> Allowed in a Cascaded <br> NSEs | SEARCH Latencies (CLK1X Cycles) | Max. Table Size (With Max. Number of |
| :---: | :---: | :---: | :---: |
| Devices |  |  |  |$|$

The following is a list of SEARCH operations described in the data sheet:

- Mixed-size Single Search for a single device on tables configured with different widths (TLSZ[1:0] = 01)
- Mixed-size MultiSearch for a single device on tables configured with different widths (TLSZ[1:0] = 01)
-72-bit Single Search for single device or cascade up to eight devices (TLSZ[1:0] = 01)
-72-bit MultiSearch for single device or cascade up to eight devices (TLSZ[1:0] = 01)
- 144-bit Single Search for cascade up to 31 devices (TLSZ[1:0] = 10)
- 576-bit Single Search for single device or cascade up to eight devices (TLSZ[1:0] = 01)
- 576-bit MultiSearch for single device or cascade up to eight devices (TLSZ[1:0] = 01)
- Mixed-size Single Search for cascade up to 31 devices on tables configured with different widths (TLSZ[1:0] = 10)
- Mixed-size MultiSearch for cascade up to eight devices on tables configured with different widths (TLSZ[1:0] = 01).


### 6.5.1 Mixed-size Single Searches with One Device on Tables Configured with Different Widths

This subsection covers single-searches with a single device configured with tables of different widths ( $\times 72, \times 144, \times 288$ ). Figure 6-5 shows three sequential searches: first, a 72 -bit Search on a $\times 72$-configured table; a 144 -bit Search on a $\times 144$-configured table; and a 288-bit Search on a $\times 288$-configured table that each results in a hit. Figure 6-6 shows the sample table.
Note: If the mixed-size tables include a 576-bit table, then the device can only operate in the Enhanced Mode, as the maximum table width allowed in the Non-Enhanced Mode is 288 bits.
One way to create multiple tables of different widths in an NSE is by having table designation bits. It is assumed that bits [71:70] for each entry will be assigned such table designation bits. DQ[71:70] will be 00 in each of the two $A$ and $B$ cycles of the $\times 72$-bit Search (Search1). DQ[71:70] is 01 in each of the $A$ and $B$ cycles of the $\times 144$-bit Search (Search2). DQ[71:70] is 10 in each of the A, B, C, and D cycles of the $\times 288$-bit Search (Search3).


HLAT = 001 (binary), TLSZ = 00 (binary), LRAM $=1$ (binary), LDEV $=1$ (binary).
M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-5. Timing Diagram for Mixed Single Search (One Device)
TLSZ = 01 (binary), LRAM = 1 (binary), LDEV = 1 (binary) for this particular example.
The following is the sequence of operation for a single Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

## - Cycle A:

- Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10". The CMD[2] and CMD[9] signals must be driven to logic 0 for the 72-bit search, but for 144-bit search, CMD[9] = 1 and CMD [2] = 0 . For 288 -bit search, CMD[9] is don't care, whereas CMD[2] $=1$ for the first " $A$ " cycle and 0 for the last " $A$ " cycle. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 if it has a hit.
-DQ Bus: At the same time in cycle A, DQ[71:0] must be driven with the 72-bit data to be compared.


## - Cycle B:

—Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10". CMD[5:2] must now be driven by the index of the comparand register pair for storing the search key presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 27 for a description of SSR[0:7]). CMD[10:9] are don't cares for this cycle.
-DQ Bus: The DQ[71:0] continues to carry the search key to be compared.
Note. For 72-bit searches, the host ASIC must supply the same 72-bit data on DQ[71:0] during both cycles A and B. Also, the even and odd pairs of GMRs selected for the comparison must be programmed with the same value. For 144-bit, 288-bit or 576bit searches, each 72-bit presented on each cycle A and B will together form the 144 -bit or 288 -bit or 576 -bit search key respectively.
When an N -bit search key, K , is presented on the DQ bus, the entire table of N -bit entries is compared to the search key using the GMR and local mask bits. The GMR is selected by the GMR Index in the command's cycle A. K is also stored in both even and odd comparand register pairs (selected by the comparand register index in command cycle B). K is compared with each entry in the table, starting at location 0. A matching entry that satisfies the Soft Priority and Mini-Key scheme (for Enhanced Mode) will be the winning entry, and its location address $L$ will be driven as part of the SRAM address on the SADR[N:0] lines (see Section 6.7, "SRAM PIO Access," on page 121), N = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128.
The latency of the Search from command to SRAM access cycle is 5 for up to eight devices in the table (TLSZ[1:0] = 01). SSV and SSF also shift further to the right for different values of HLAT, as specified in Table 6-5.
Figure 6-6 shows an example of multiple table configuration with a CYNSE10512 device.
Table 6-5. Shift of SSF and SSV from SADR

| HLAT (binary) | Number of CLK Cycles | HLAT | Number of CLK Cycles |
| :---: | :---: | :---: | :---: |
| 000 | 0 | 100 | 4 |
| 001 | 1 | 101 | 5 |
| 010 | 2 | 110 | 6 |
| 011 | 3 | 111 | 7 |



Figure 6-6. Multiwidth Configurations Using CYNSE10512 as an Example
Referring to Figure 6-6, if the CYNSE10512 device is used in the Non-Enhanced Mode, the CFG field in the Configuration Register should be configured to "1010101010101010010101010101010100000000000000000000000000000000" in order to have three individual tables within a device. If the device is used in the Enhanced Mode, the NES field in the Block Mini-Key Register (BMR) should be configured as follows:

- For the first 64 blocks in the data array, NES $=00$ for 72 -bit table width.
- For the next 32 blocks, NES $=01$ for 144-bit table width.
- For the final 32 blocks, NES $=10$ for 288 -bit table width.


### 6.5.2 Mixed-size Multi Searches with One Device on Tables Configured with Different Widths

The multiple search operates the search commands in parallel on the upper half (array 0) and lower half (array 1) of the data array in the device. The results from the two parallel searches are then driven on the SRAM bus at twice that rate relative to single-search. This subsection covers multi searches with a single device configured with tables of different widths $(\times 72, \times 144$, $\times 288$ ) in each of the two arrays. Figure $6-7$ shows three sequential searches: first, a 72 -bit Search on a $\times 72$-configured table; a 144-bit Search on a $\times 144$-configured table; and a 288 -bit Search on a $\times 288$-configured table.
Note: MultiSearch is only available in the Enhanced Mode, not in the Non-Enhanced Mode. Figure 6-8 shows the sample table.
One way to create multiple tables of different widths in an NSE is by having table designation bits. It is assumed that bits [71:70] for each entry will be assigned such table designation bits. $\mathrm{DQ}[71: 70$ ] will be 00 in each of the two A and B cycles of the $\times 72$-bit MultiSearch (M-Search1). DQ[71:70] is 01 in each of the $A$ and $B$ cycles of the $\times 144$-bit MultiSearch ( $M$-Search2). DQ[71:70] is 10 in each of the A, B, C, and D cycles of the $\times 288$-bit MultiSearch (M-Search3).


HLAT $=000$ (binary), TLSZ = 00 (binary), LRAM = 1 (binary), LDEV = 1 (binary).
M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-7. Timing Diagram for Mixed MultiSearch (One Device)
The MSE bit in the Command Register must be set high to enable the MultiSearch feature. The same with the Enhanced Mode (EMODE) bit. TLSZ = 01 (binary), LRAM = 1 (binary), LDEV = 1 (binary) for a single-device configuration. HLAT = 000 (binary) for this example. The following is the sequence of operation for a single Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

## - Cycle A:

- Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10". The CMD[2] and CMD[9] signals must be driven to logic 0 for the 72-bit search, but for 144-bit search, CMD[9] = 1 and CMD [2] = 0 . For 288 -bit search, CMD[9] is don't care, whereas CMD[2] $=1$ for the first " $A$ " cycle and 0 for the last " $A$ " cycle.
\{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[7:6] signals must be driven with the same bits that will be driven on SADR[24:23] for CYNSE10512, SADR[23:22] for CYNSE10256, SADR[22:21] for CYNSE10128 by this device if it has a hit. CMD[8] must be driven high for MultiSearch operation.
—DQ Bus: At the same time in cycle A, DQ[71:0] must be driven with the 72-bit data to be compared.


## - Cycle B:

—Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10". CMD[5:2] must now be driven by the index of the comparand register pair for storing the search key presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 27 for a description of SSR[0:7]). CMD[10:9] are don't cares for this cycle.
-DQ Bus: The DQ[71:0] continues to carry the search key to be compared.
Note. For 72-bit searches, the host ASIC can supply different 72-bit data on DQ[71:0] during both cycles A and B to be compared with the tables in array 0 and 1 of the data array. The even and odd pairs of GMRs selected for the comparison need not be programmed with the same value. For 144 -bit, 288 -bit or 576 -bit searches, each 72 -bit presented on each cycle A and B will together form the 144 -bit or 288 -bit or 576 -bit search key respectively. These search keys are compared to both array 0 and 1 during cycles A and B .
When an N-bit search key, K, is presented on the DQ bus, both arrays of N-bit entries are compared to the search key using the GMR and local mask bits. The GMR is selected by the GMR Index in the command's cycle A. K is also stored in both even and odd comparand register pairs (selected by the comparand register index in command cycle B ). K is compared with each entry in the table, starting at location 0. A matching entry from each array that satisfies the Soft Priority and Mini-Key scheme will be the winning entries, and their location addresses La and Lb will be driven as part of the SRAM address on the SADR[N:0] lines (see Section 6.7, "SRAM PIO Access," on page 121), N = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128.
The latency of the Search from command to SRAM access cycle is 5 for a single device (or up to eight devices) configuration in the table (TLSZ[1:0] = 01). SSV and SSF also shift further to the right for different values of HLAT, as specified in Table 6-5.
Figure 6-8 shows a multiwidth configuration when multisearch is enabled using CYNSE10512 as an example.


Figure 6-8. Multiwidth Configurations Using CYNSE10512 as an Example
The NES field in the Block Mini-Key Register (BMR) should be configured as follows:

- For the first 32 blocks in the data array, NES = 00 (binary) for 72-bit table width. For the next 16 blocks, NES = 01 (binary) for 144 -bit table width. For the following 16 blocks, NES $=10$ (binary) for 288 -bit table width. These will configure the tables in array 0.
- Setting NES $=00$ (binary) for the next 32 blocks will configure those blocks to be 72-bit table in array 1 . Setting NES = 01 (binary) for the next 16 blocks will configure those blocks to be 144-bit table. Setting the final 16 blocks' NES field will configure those blocks to be 288-bit table.


### 6.5.3 72-bit Single Search for 1 device or cascade up to eight devices

The hardware diagram of the Search subsystem of up to eight devices is shown in Figure 6-9. The MultiSearch Mode (MSE) bit in the Command Register must be set LOW to perform single-search. The following are the rest of the parameters programmed into the eight devices.

- In Non-Enhanced Mode, first seven devices (devices 0-6) must reset all bits of the CFG field in Configuration register to zeroes. In Enhanced Mode, these devices should have the NES field of each block within a device configured to 00 for 72 -bit table width. TLSZ = 01 (binary), HLAT = 010 (binary), LRAM = 0 (binary), and LDEV = 0 (binary) for both modes.
- In Non-Enhanced Mode, the eighth device (device 7) should still reset all bits of the CFG field in Configuration register to zeroes. In Enhanced Mode, NES should still be 00 (binary). But TLSZ = 01 (binary), HLAT = 010 (binary), LRAM = 1 (binary), and LDEV = 1 (binary) for the last device
Note: The device receiving all the LHO signals from the other devices is the last device.
For a single-device configuration, the parameters are the same as device 7 . $\mathrm{BHI}[2: 0]$ and $\mathrm{LHI}[6: 0]$ should be tied to ground.

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Figure 6-9. Hardware Diagram for a Table with Eight Devices
The following three figures show the response of three of the eight devices having a hit at different time according to a Hit/Miss assumption shown below in Table 6-6. For these timing diagrams, four 72-bit searches are performed sequentially. Figure 6-10 shows the timing diagram for a Search command in the 72 -bit-configured table of eight devices for device number 0. Figure 611 and Figure $6-12$ shows the same for device number 1 and number 7 (the last device in this specific table) respectively.
Note: All the shared signals showing tri-stated condition ("z") indicate that, that particular device is not driving the shared signals. The shared signals are not three-stated in a real life because other devices will be driving them.

## Table 6-6. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ |
| :---: | :---: | :---: | :---: | :---: |
| Device 0 | Hit | Miss | Hit | Miss |
| Device 1 | Miss | Hit | Hit | Miss |
| Devices 2-6 | Miss | Miss | Miss | Miss |
| Device 7 | Miss | Miss | Hit | Hit |



Figure 6-10. Timing Diagram for 72-bit Search Device Number 0
M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128 CFG[N:0] are all zeroes for Non-Enhanced Mode, $\mathrm{N}=63$ for CYNSE10512, 31 for CYNSE10256, 15 for CYNSE10128.
NES $=00$ (binary) in each block for Enhanced Mode.
HLAT = 010 (binary), TLSZ = 01 (binary), LRAM = 0 (binary), LDEV = 0 (binary).
Note: |(LHI[6:0]) stands for the boolean 'OR' of the entire bus LHI[6:0].
Note: Each bit in LHO[1:0] is the same logical signal.
Search1
(Miss
Search3
(Local
on this
device)
winner
but not
global
Search2
winner)

Figure 6-11. Timing Diagram for 72-bit Search Device Number 1


Figure 6-12. Timing Diagram for 72-bit Search Device Number 7 (Last Device)
The following is the sequence of operation for a single 72-bit Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

## - Cycle A:

-Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10". The CMD[2] and CMD[9] signals must be driven to logic 0 for this 72 -bit search. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 by this device if it has a hit.
-DQ Bus: At the same time in cycle A, DQ[71:0] must be driven with the 72-bit data to be compared.

## - Cycle B:

—Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10". CMD[5:2] must now be driven by the index of the comparand register pair for storing the two 72-bit word presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 27 for a description of SSR[0:7]). CMD[10:9] are don't cares for this cycle.
-DQ Bus: The DQ[71:0] continues to carry the 72-bit data to be compared.
Note. For 72-bit searches, the host ASIC must supply the same 72-bit data on DQ[71:0] during both cycles A and B. Also, the even and odd pairs of GMRs selected for the comparison must be programmed with the same value.

The logical 72-bit Search operation is shown in Figure 6-13. The entire table of 72-bit entries (eight devices) is compared to a 72-bit word K (presented on the DQ bus in both cycles A and B of the command) using the GMR and local mask bits. The effective GMR is the 72-bit word specified by the identical value in both even and odd GMR pairs, in each of the eight devices, and selected by the GMR Index in the command's cycle A. The 72-bit word K (presented on the DQ bus in both cycles A and B of the command) is also stored in both even and odd comparand register pairs (selected by the comparand register index in command cycle $B$ ) in each of the eight devices. In the $\times 72$ configuration, only the even comparand register can subsequently be used by the Learn command in one of the devices (the first non-full device only). The word K (presented on the DQ bus in both cycles A and B of the command) is compared with each entry in the table, starting at location 0. A matching entry that satisfies the Soft Priority and Mini-Key scheme (for Enhanced Mode) will be the winning entry, and its location address $L$ will be driven as part of the SRAM address on the SADR[N:0] lines (see Section 6.7, "SRAM PIO Access," on page 121), N = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128. The global winning device will drive the bus in a specific cycle. On a global miss cycle, the device with LRAM = 1 (default driving device for the SRAM bus) and LDEV = 1 (default driving device for SSF and SSV signals) will be the default driver for such missed cycles.
The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 72-bit searches in $\times 72$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 72 -bit Search command cycle (two CLK2X cycles) is shown in Table 6-4.
The latency of the Search from command to SRAM access cycle is 5 for up to eight devices in the table (TLSZ $=01$ ). SSV and SSF also shift further to the right for different values of HLAT, as specified in Table 6-5.


### 6.5.4 72-bit MultiSearch for One Device or Cascade Up to Eight Devices

The multiple search operates the search commands in parallel on the upper half (array 0 ) and lower half (array 1 ) of the data array in the device. The results from the two parallel searches are then driven on the SRAM bus at twice that rate relative to single-search. The hardware diagram of the Search subsystem of up to eight devices is shown in Figure 6-14 below.

## Note:

- MultiSearch feature is only available in the Enhanced Mode, not in the Non-Enhanced Mode.
- Comparing the hardware diagrams shown in Figure 6-9 and Figure 6-14, enabling MultiSearch does not mean that a board layout change is required. The LHO_1_L and LHI_1_L share the same pin with the Full In and Full Out signals, which are not shown in Figure 6-9. Cascading multiple devices together still allow the user to configure the devices through software to perform single-search or MultiSearch operations without any board change.
- The device receiving all the LHO signals from the other devices is the last device.
- All the shared signals showing three-stated condition ("z") indicate that, that particular device is not driving the shared signals. The shared signals are not three-stated in a real life because other devices will be driving them.


Figure 6-14. Hardware Diagram for a Table with Eight Devices for MultiSearch

## Notes:

- In MultiSearchMode, there is a separate set of the LHO and LHI signals corresponding to memory array 0 and array 1. LHO_0[1:0] and LHI_0[6:0] corresponds to array 0 whereas LHO_1_L[1:0] and LHI_1_L[6:0] corresponds to array 1. The latter share the same pins as FULO[1:0] and FULI[6:0] respectively.
- Both $\mathrm{LHO}_{-0}[1]$ and $\mathrm{LHO}_{2} 0[0]$ are exact same signals so that the loads can be shared by two outputs. The same is true for LHO_1_L[1] and LHO_1_L[0].
- Unused LHI_0 signals should be tied to ground whereas unused LHI_1_L signals should be tied to $V_{\text {DDQ_ASIC }}$, which is either 1.8 V or 2.5 V only.
- LHI_1_L signals are active LOW while LHI_0 are active HIGH.

The MultiSearch Enable (MSE) bit in the Command Register must be set HIGH when the Command Register is programmed. The same with the Enhanced Mode (EMODE) bit. The following are the rest of the parameters programmed into the eight devices.

- First seven devices (devices 0-6): NES = 00 (binary) for each block in each device, TLSZ = 01 (binary), HLAT = 001 (binary), LRAM $=0$ (binary), and LDEV $=0$ (binary).
- Eighth device (device 7): NES = 00 (binary) for each block in each device, $\mathrm{TLSZ}=01$ (binary), HLAT = 001 (binary), LRAM = 1 (binary), and LDEV = 1 (binary).
For a single-device configuration, the parameters are the same as device $7 . \mathrm{BHI}[2: 0]$ and $\mathrm{LHI}[6: 0]$ should be tied to ground.
The following three figures show the response of 3 of the 8 devices having a hit at different time according to a Hit/Miss assumption shown below in Table 6-7. For these timing diagrams, five 72-bit searches are performed sequentially. Figure 6-15 shows the timing diagram for a Search command in the 72-bit-configured table of eight devices for device number 0. Figure 6-16 and Figure 6-17 shows the same for device number 1 and number 7 (the last device in this specific table) respectively.

Table 6-7. Hit/Miss Assumption for MultiSearch Mode

| Search \# | $\mathbf{1}$ |  | $\mathbf{2}$ |  | $\mathbf{3}$ |  | $\mathbf{4}$ |  | $\mathbf{5}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Device 0 | Hit | Miss | Hit | Miss | Hit | Miss | Miss | Hit | Hit | Hit |
| Device 1 | Miss | Hit | Hit | Miss | Hit | Hit | Miss | Miss | Miss | Miss |
| Devices <br> $2-6$ | Miss | Miss | Miss | Miss | Miss | Miss | Miss | Miss | Miss | Miss |
| Device 7 | Miss | Miss | Hit | Hit | Miss | Hit | Hit | Miss | Miss | Miss |



HLAT = 001 (binary), TLSZ = 01 (binary), LRAM = 0 (binary), LDEV = 0 (binary).
Note: |(LHI_0[6:0]) and \&(LHI_1_L[6:0]) stands for the boolean 'OR' and 'AND' respectively of the entire LHI bus.
Note: Each bit in LHO_0[1:0] and LHO_1_L[1:0] is the same logical signal.
This timing diagram is for device \#0 only, High-Z means this device is not driving, but other device in the cascade may be driving the bus.

Figure 6-15. Timing Diagram for 72-bit MultiSearch Device Number 0
Notes:

- Each "cycle" consists of 2 CLK2x cycles, which is effectively one CLK1x cycle.
- The latency of SSV and SSF specified by HLAT refers to CLK1x cycles.
- LHO_0[1:0] will be valid 4 (CLK1x) cycles after the search parameter is sampled, regardless of the number of searches entered; e.g., the search parameter A entered on DQ bus is sampled at cycle 1, LHO_0[1:0] will be available at cycle 5 .
- For TLSZ[1:0] $=01$, all signals on SRAM interface will be driven 5 (CLK1x) cycles after the search parameter is sampled, regardless of the number of searches; e.g., the search parameter A sampled on cycle 1 is a hit, thus the address value sent to SADR bus and the rest of the SRAM control signals will be driven at cycle 6.


Figure 6-16. Timing Diagram for 72-bit MultiSearch Device Number 1


Figure 6-17. Timing Diagram for 72-bit MultiSearch Device Number 7 (Last Device)

- When more than one device is cascaded, the last device is always the default driver of the SRAM interface signals, i.e., when none of the devices is driving, the last device will set the SRAM interface signals to a known default state. When other devices are driving, the last device will set its I/Os on the SRAM interface to High-Z.
- Referring to Figure 6-17, the last device drives the SRAM interface signals until the end of cycle \#5. From cycle \#6 onwards, its I/Os are three-stated to allow other devices to drive the SRAM interface signals, except when it's its turn to drive. This goes on until the end of cycle \#10, and at the beginning of cycle \#11, it drives the SRAM interface to a known default state again when no other devices are driving.

The following is the sequence of operation for a single 72-bit MultiSearch command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

- Cycle A:
—Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10". The CMD[2] signal must be driven to logic 0 . \{CMD[10], CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[7:6] signals must be driven with the same bits that will be driven on SADR[24:23] for CYNSE10512, SADR[23:22] for CYNSE10256, SADR[22:21] for CYNSE10128 by this device if it has a hit. CMD[8] must be set to logic 1, and CMD[9] must be set to logic 0 .
—DQ Bus: DQ[71:0] must be driven with the 72-bit data to be compared against the upper half (array 0 ) of the device entries.
- Cycle B:
—Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10". CMD[5:2] must be driven with the index of the Comparand Register. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 27 for information on SSR[0:7]). CMD[10:9] are don't cares during this cycle.
-DQ Bus: The DQ[71:0] is driven with the data that needs to compared with lower half (array 1) of the device entries.
The logical 72-bit Search operation is shown in Figure 6-18. The upper half of the device consisting of 72-bit entries is compared to a 72-bit word that is presented on the DQ bus in cycles A using the GMR and local mask bits. The GMR used is the 72-bit word specified in the even GMR selected by the GMR Index in the command's cycle A. The lower half of the device consisting of 72bit entries is compared to a 72-bit word that is presented on the DQ bus in cycles $B$ using the GMR and local mask bits. The GMR used is the 72-bit word specified in the odd GMR selected by the GMR Index in the command's cycle A. The result of the two searches from the two halves are driven as two SRAM cycles as shown in the timing diagram. A matching entry from each array that satisfies the Soft Priority and Mini-Key scheme will be the winning entries, and their location addresses La and Lb will be driven as part of the SRAM address on the SADR[N:0] lines (see Section 6.7, "SRAM PIO Access," on page 121), $\mathrm{N}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128.
The Search command is a pipelined operation and executes a Search at the frequency of CLK2X for 72 -bit searches in $\times 72$ configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 72-bit Search command cycle (two CLK2X cycles) is shown in Table 6-4. Search latency from command to SRAM access cycle is 5 from a single device upto eight devices in the table with $T L S Z=01$. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 6-5.

(72-bit configuration)

Figure 6-18. $\times 72$ Table with in MultiSearchMode

### 6.5.5 144-bit Single Search for Cascade Up to 31 Devices

The hardware diagram of the Search subsystem of 31 devices is shown in Figure 6-19. Each of the four blocks in the diagram represents eight Ayama 10000 devices (except the last, which has seven devices). The diagram for a block of eight devices is very similar to the hardware diagram in Figure 6-9, except that the BHI[2:0] signals are connected to BHO of the previous block (rather than being grounded) as shown in Figure 6-9. The following are the parameters programmed into the 31 devices:

- First thirty devices (devices $0-29$ ): TLSZ = 10 (binary), HLAT = 001 (binary), LRAM = 0 (binary), and LDEV = 0 (binary).
- Thirty-first device (device 30): TLSZ = 10 (binary), HLAT = 001 (binary), LRAM = 1 (binary), and LDEV = 1 (binary).
- For Non-Enhanced Mode, CFG[63:0] = 5555555555555555 (hex) for all devices for CYNSE10512. CFG[31:0] $=55555555$ (hex) for all devices for CYNSE10256, and CFG[15:0] = 5555 (hex) for all devices for CYNSE10128. For Enhanced Mode, NES in each block for all devices should be set to "01" to create 144-bit table.
- The device receiving all the LHO signals from the other devices is considered the last device.
- All the shared signals showing tri-stated condition ("z") indicate that, that particular device is not driving the shared signals. The shared signals are not three-stated in a real life because other devices will be driving them.
- Comparing the hardware diagrams shown in Figure 6-9 and Figure 6-14, enabling MultiSearch does not mean that a board layout change is required. The LHO_1_L and LHI_1_L share the same pin with the Full In and Full Out signals, which are not shown in Figure 6-9. Cascading multiple devices together still allow the user to configure the devices through software to perform single-search or MultiSearch operations without any board change.
The timing diagrams referred to in this paragraph reference the Hit/Miss assumptions defined in Table 6-8. For the purpose of illustrating the timings, it is further assumed that there is only one device with a matching entry in each of the blocks. Figure 620 shows the timing diagram for a Search command in the 144-bit-configured table of 31 devices for each of the eight devices in block number 0 . Figure $6-21$ shows the same for the all the devices in block number 1 (above the winning device in that block). Figure 6-22 shows the timing diagram for the globally winning device (defined as the final winner within its own and all blocks) in block number 1. Figure 6-23 shows the timing diagram for all the devices below the globally winning device in block number 1. Figure 6-24, Figure 6-25, and Figure 6-26 show the timing diagrams of the devices above the globally winning device, the globally winning device, and the devices below the globally winning device, respectively, for block number 2. Figure 6-27, Figure 6-28, Figure 6-29, and Figure 6-30 show the timing diagrams of the devices above globally winning device, the globally winning device, and the devices below the globally winning device except the last device (device 30 ), respectively, for block number 3.
The 144-bit Search operation is pipelined and executes as follows:
- Four cycles from the Search command, each of the devices knows the outcome internal to it for that operation.
- On the fifth cycle, the devices arbitrate for a winner within a block (a "block" is defined as less than or equal to eight devices resolving the winner within them using the LHI[6:0] and LHO[1:0] signalling mechanism).
- On the sixth cycle after the Search command, the blocks (of devices) resolve the winning block through the BHI[2:0] and BHO [2:0] signalling mechanism. The winning device within the winning block is the global winning device for a Search operation.

Table 6-8. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ |
| :---: | :---: | :---: | :---: | :---: |
| Block 0 | Miss | Miss | Miss | Miss |
| Block 1 | Miss | Miss | Hit | Miss |
| Block 2 | Miss | Hit | Hit | Miss |
| Block 3 | Hit | Hit | Miss | Miss |



Figure 6-19. Hardware Diagram for a Table with 31 Devices


Note: Each bit in $\mathrm{BHO}[2: 0]$ is the same logical signal.
Note: Each bit in LHO[1:0] is the same logical signal.

Figure 6-20. 144-bit Search for Devices in Block \#0 and Above Block \#1 Winning Device


Figure 6-21. 144-bit Search Timing Diagram for Block \#1 Global Winning Device


Figure 6-22. 144-bit Search Timing Diagram for Devices Below Block \#1 Winning Device


Figure 6-23. 144-bit Search Timing Diagram for Devices Above Block \#2 Winning Device


Figure 6-24. 144-bit Search Timing Diagram for Block \#2 Global Winning Device


Figure 6-25. 144-bit Search Timing Diagram for Devices Below Block \#2 Winning Device


Figure 6-26. 144-bit Search Timing Diagram for Devices Above Block \#3 Winning Device


Figure 6-27. 144-bit Search Timing Diagram for Block \#3 Global Winning Device
 HLAT = 001 (binary), TLSZ = 10 (binary), LRAM = 0 (binary), LDEV = 0 (binary).

Figure 6-28. 144-bit Search Diagram Below Block \#3 Winning Device Except the Last Device


Figure 6-29. 144-bit Search Timing Diagram for Device Number 6 in Block \#3
The following is the sequence of operation for a single 144-bit Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

- Cycle A:
—Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10" (binary). CMD[2] must be driven to logic low. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 by this device if it has a hit. CMD[9] must be driven to logic high to indicate a 144-bit search.
—DQ Bus: DQ[71:0] must be driven with the 72-bit data to be compared.


## - Cycle B:

—Command Bus: The host ASIC continues to drive CMDV HIGH and applies Search command CMD[1:0] = "10" (binary). CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus
during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 27 for the description of SSR[0:7]). CMD[10:9] are don't cares in this cycle.
—DQ Bus: The DQ[71:0] continues to carry the 72-bit data to be compared.
The logical 144-bit Search operation is shown in Figure 6-30. The entire table of 31 devices (consisting of 72-bit entries) is compared to a 144-bit word K presented on the DQ bus in both cycles A and B of the command using the GMR and local mask bits. The GMR is the 144 -bit word specified by the even and odd GMR pairs selected by the GMR index in the command's cycle A. The 144-bit word K (presented on the DQ bus in both cycles A and B of the command) is also stored in both even and odd comparand register pairs selected by the comparand register index in command cycle B. In the $\times 144$ configuration, the even and odd comparand register can be subsequently used by the Learn command only in the first non-full device.
Note. The Learn command is supported for only one of the blocks consisting of up to eight devices in a depth-cascaded table of more than one block.
The word K (presented on the DQ bus in both cycles A and B of the command) is compared with each entry in the table, starting at location 0 (decimal). A matching entry that satisfies the Soft Priority and Mini-Key scheme (for Enhanced Mode) will be the winning entry, and its location address L will be driven as part of the SRAM address on the SADR[N:0] lines (see "SRAM PIO Access" on page 121), $\mathrm{N}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128. The global winning device will drive the bus in a specific cycle. On global miss cycles, the device with LRAM = 1 (binary) and LDEV = 1 (binary) will be the default driver for such missed cycles.
Note. During 144-bit searches of 144-bit-configured tables, the Search hit will always be at an even address.
The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 144-bit searches in $\times 144$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 144-bit Search command cycle (two CLK2X cycles) is shown in Table 6-4.

For up to 31 devices in the table (TLSZ = 10 (binary)), Search latency is 6 from command to SRAM access cycle. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 6-5.


Figure 6-30. $\times 144$ Table with 31 Devices

### 6.5.6 576-bit Single Search for One Device or Cascade up to Eight Devices

The hardware diagram of the Search subsystem of up to eight devices is shown in Figure 6-9. The MultiSearch Enable (MSE) bit in the Command Register must be set LOW to perform single-search. The following are the rest of the parameters programmed into the eight devices.

- First seven devices (devices $0-6$ ): TLSZ = 01 (binary), HLAT = 000 (binary), LRAM = 0 (binary), and LDEV = 0 (binary).
- Eighth device (device 7): TLSZ = 01 (binary), HLAT = 000 (binary), LRAM = 1 (binary), and LDEV = 1 (binary).
- 576 -bit search is only available in the Enhanced Mode, not in the Non-Enhanced Mode. NES should be set to "11" (binary) in all blocks of all devices to create a 576-bit table.
For a single-device configuration, all parameters will be the same as device 7 . $\mathrm{BHI}[2: 0]$ and $\mathrm{LHI}[6: 0]$ should be tied to ground.


## Notes:

- All eight devices must be programmed with the same values for TLSZ and HLAT. Only the last device in the table (device number 7 in this case) must be programmed with LRAM $=1$ (binary) and LDEV = 1 (binary). All other upstream devices (devices 0 through 6 in this case) must be programmed with LRAM $=0$ (binary) and LDEV $=0$ (binary).
- The device receiving all the LHO signals from the other devices is considered the last device.
- All the shared signals in the following timing diagrams showing tri-stated condition ("z") indicate that, that particular device is not driving the shared signals. The shared signals are not three-stated in a real life because other devices will be driving them.
- Comparing the hardware diagrams shown in Figure 6-9 and Figure 6-14, enabling MultiSearch does not mean that a board layout change is required. The LHO_1_L and LHI_1_L share the same pin with the Full In and Full Out signals, which are not shown in Figure 6-9. Cascading multiple devices together still allow the user to configure the devices through software to perform single-search or MultiSearch operations without any board change.
The following three figures show the response of three of the eight devices having a hit at different time according to a Hit/Miss assumption shown below in Table 6-9. For these timing diagrams, three 576-bit searches are performed sequentially. Figure 631 shows the timing diagram for a Search command in the 576 -bit-configured table of eight devices for device number 0 . Figure 632 and Figure 6-33 shows the same for device number 1 and number 7 (the last device in this specific table) respectively.


## Table 6-9. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ |
| :---: | :---: | :---: | :---: |
| Device 0 | Hit | Miss | Miss |
| Device 1 | Miss | Hit | Miss |
| Devices 2-6 | Miss | Miss | Miss |
| Device 7 | Miss | Miss | Miss |





M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
HLAT $=000$ (binary), TLSZ $=01$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).
Note: |(LHI[6:0]) stands for the boolean 'OR' of the entire bus LHI[6:0].
Note: Each bit in LHO[1:0] is the same logical signal.
Figure 6-31. Timing Diagram for 576-bit Single Search Device Number 0


M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
HLAT $=000$ (binary), $\mathrm{TLSZ}=01$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).
Note: |(LHI[6:0]) stands for the boolean 'OR' of the entire bus LHI[6:0].
Note: Each bit in LHO[1:0] is the same logical signal.

Figure 6-32. Timing Diagram for 576-bit Single Search Device Number 1


M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
HLAT = 000 (binary), TLSZ = 01 (binary), LRAM = 1 (binary), LDEV = 1 (binary).
Note: |(LHI[6:0]) stands for the boolean 'OR' of the entire bus LHI[6:0].
Note: Each bit in $\mathrm{LHO}[1: 0]$ is the same logical signal.
Figure 6-33. Timing Diagram for 576-bit Single Search Device Number 7 (Last Device)
The following is the sequence of operation for a single 576-bit Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

## - Cycle A:

—Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10" (binary). CMD[2] must be driven to logic 1 for the first three A-cycles and then driven to logic 0 for the final A-cycle for 576-bit search.
\{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. Each of the four A-cycles provide a GMR index to mask 144 bits of the data to be compared (each A-cycle provide a pair of GMR, which is 144 bits, for A -cycles will result in a total of 576 bits of GMR). CMD[8:6] signals must be driven with the same bits that will be driven on SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 by this device if it has a hit. CMD[9] is don't care for this cycle.
—DQ Bus: At the same time in cycle A, DQ[71:0] must be driven with 72-bit data (which is part of the 576-bit data) to be compared.

## - Cycle B:

—Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10" (binary). CMD[5:2] must now be driven by the index of the comparand register pair for storing the two 72-bit word presented on the DQ bus during cycles $A$ and $B$. Each of the four $B$-cycles provide an index for a pair of comparand register. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 27 for a description of $\operatorname{SSR}[0: 7]$ ). CMD[10:9] are don't cares for this cycle.
-DQ Bus: The DQ[71:0] continues to carry the 72-bit data (which is part of the 576-bit data) to be compared.
Note. For 576-bit searches, the host ASIC must supply individual 72-bit data on DQ[71:0] during cycles A and B. Also, four individual pairs of GMR and CMPR registers may be involved in the comparison.
The logical 576-bit Search operation is shown in Figure 6-34. The entire table of 576-bit entries (eight devices) is compared to a 576 -bit word K that is presented on the DQ bus in eight cycles using the GMR and local mask bits. The GMR is the 576-bit word specified by four pairs of GMRs selected by GMR indices in each of the eight devices. The 576-bit word K (presented on the DQ bus in all eight cycles of the command) is also stored in both even and odd comparand register pairs (selected by the comparand register index in command cycle $B$ ) in each of the eight devices. The word $K$ is compared with each entry in the table, starting at location 0 (decimal). A matching entry that satisfies the Soft Priority and Mini-Key scheme will be the winning entry, and its location address $L$ will be driven as part of the SRAM address on the SADR[N:0] lines (see Section 6.7, "SRAM PIO Access," on page 121), $\mathrm{N}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128. The global winning device will drive the bus in a specific cycle. On a global miss cycle, the device with LRAM $=1$ (binary) (default driving device for the SRAM bus) and LDEV $=1$ (binary) (default driving device for SSF and SSV signals) will be the default driver for such missed cycles.
The Search command is a pipelined operation and executes a Search at one-eighth the rate of the frequency of CLK2X for 576-bit searches in $\times 576$-configured tables. The latency of the Search from command to SRAM access cycle is 5 for up to eight devices in the table (TLSZ = "01" (binary)). SSV and SSF also shift further to the right for different values of HLAT, as specified in Table 6-5.


### 6.5.7 576-bit MultiSearch for One Device or Cascade up to Eight Devices

The MultiSearch operates the search commands in parallel on the upper and lower half (array 0 and 1) of the device. The results from the two parallel searches are then driven on the SRAM bus at twice that rate relative to a single-search.

## Notes:

- For $x 72$ multi searches, two individual 72 -bit search keys can be searched in array 0 and array 1 simultaneously. For $\times 144$, x288 and x576 multi searches, both arrays will be searched with the same 144-bit, 288-bit or 576-bit search keys respectively.
- In MultiSearch Mode, there is a separate set of the LHO and LHI signals corresponding to memory array 0 and array 1 . LHO_0[1:0] and LHI_0[6:0] corresponds to array 0 whereas LHO_1_L[1:0] and LHI_1_L[6:0] corresponds to array 1. The latter share the same pins as FULO[1:0] and FULI[6:0] respectively.
- Both LHO_0[1] and LHO_0[0] are exact same signals so that the loads can be shared by two outputs. The same is true for LHO_1_L[1] and LHO_1_L[0].
- Unused LHI_0 signals should be tied to ground whereas unused LHI_1_L signals should be tied to $V_{\text {DDQ_ASIC }}$, which is either 1.8 V or 2.5 V only.
- LHI_1_L signals are active LOW while LHI_0 are active HIGH.

The hardware diagram of the MultiSearch subsystem of up to eight devices is shown in Figure 6-14. The MultiSearch Enable (MSE) bit in the Command Register must be set HIGH to perform multi search. The same with Enhanced Mode (EMODE) bit. The following are the rest of the parameters programmed into the eight devices.

- First seven devices (devices $0-6$ ): $\mathrm{TLSZ}=01$ (binary), HLAT $=000$ (binary), LRAM $=0$ (binary), and LDEV $=0$ (binary).
- Eighth device (device 7): TLSZ = 01 (binary), HLAT = 000 (binary), LRAM = 1 (binary), and LDEV = 1 (binary).
- NES (in the Block Mini-Key Register) field in each block of all devices must be set to "11" (binary) to make a 576-bit table.

For a single-device configuration, all parameters will be the same as device 7 . $\mathrm{BHI}[2: 0]$ and all LHI should be tied to ground.

## Notes:

- The device receiving all the LHO signals from the other devices is considered the last device.
- All the shared signals in the following timing diagrams showing tri-stated condition (" $z$ ") indicate that, that particular device is not driving the shared signals. The shared signals are not three-stated in a real life because other devices will be driving them.
- Comparing the hardware diagrams shown in Figure 6-9 and Figure 6-14, enabling MultiSearch does not mean that a board layout change is required. The LHO_1_L and LHI_1_L share the same pin with the Full In and Full Out signals, which are not shown in Figure 6-9. Cascading multiple devices together still allows the user to configure the devices through software to perform single-search or MultiSearch operations without any board change.

The following three figures show the response of three of the eight devices having a hit at different time according to a Hit/Miss assumption shown below in Table 6-10. For these timing diagrams, three 576-bit searches are performed sequentially. Figure 635 shows the timing diagram for a MultiSearch command in the 576 -bit-configured table of eight devices for device number 0. Figure 6-36 and Figure 6-37 shows the same for device number 1 and number 7 (the last device in this specific table) respectively.

Table 6-10. Hit/Miss Assumptions for 576-bit Multi Search

| Search Number | $\mathbf{1}$ |  | $\mathbf{2}$ |  | 3 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Device 0 | Hit | Hit | Miss | Miss | Miss | Miss |
| Device 1 | Miss | Miss | Hit | Miss | Miss | Miss |
| Devices 2-6 | Miss | Miss | Miss | Miss | Miss | Miss |
| Device 7 | Miss | Miss | Miss | Hit | Miss | Miss |



M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
HLAT $=000$ (binary), $\mathrm{TLSZ}=01$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).
Note: |(LHI_0[6:0]) and \&(LHI_1_L[6:0]) stands for the boolean 'OR' and 'AND' of the entire LHI bus.
Note: Each bit in LHO_0[1:0] and LHO_1_L[1:0] is the same logical signal.

Figure 6-35. Timing Diagram for 576-bit MultiSearch Device Number 0


M $=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
HLAT $=000$ (binary), $\mathrm{TLSZ}=01$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).
Note: |(LHI_0[6:0]) and \&(LHI_1_L[6:0]) stands for the boolean 'OR' and 'AND' of the entire LHI bus.
Note: Each bit in LHO_0[1:0] and LHO_1_L[1:0] is the same logical signal.
Figure 6-36. Timing Diagram for 576-bit MultiSearch Device Number 1


M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
HLAT $=000$ (binary), $\mathrm{TLSZ}=01$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).
Note: |(LHI_0[6:0]) and \&(LHI_1_L[6:0]) stands for the boolean 'OR' and 'AND' of the entire LHI bus.
Note: Each bit in LHO_0[1:0] and LHO_1_L[1:0] is the same logical signal.
Figure 6-37. Timing Diagram for 576-bit MultiSearch Device Number 7 (Last Device)

The following is the sequence of operation for a single 576-bit Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

- Cycle A:
—Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10" (binary). CMD[2] must be driven to logic 1 for the first three A-cycles and then driven to logic 0 for the final A-cycle for 576-bit search. $\{C M D[10], C M D[5: 3]\}$ signals must be driven with the index to the GMR pair for use in this Search operation. Each of the four A-cycles provide a GMR index to mask 144 bits of the data to be compared. CMD[7:6] signals must be driven with the same bits that will be driven on SADR[24:23] for CYNSE10512, SADR[23:22] for CYNSE10256, SADR[22:21] for CYNSE10128 by this device if it has a hit. CMD[8] must be driven HIGH for every A-cycle. CMD[9] is don't care for this cycle.
-DQ Bus: At the same time in cycle A, DQ[71:0] must be driven with 72-bit data (which is part of the 576-bit data) to be compared.
- Cycle B:
- Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10" (binary). CMD[5:2] must now be driven by the index of the comparand register pair for storing the two 72-bit word presented on the DQ bus during cycles $A$ and $B$. Each of the four $B$-cycles provide an index for a pair of comparand register. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 27 for a description of SSR[0:7]). CMD[10:9] are don't cares for this cycle.
-DQ Bus: The DQ[71:0] continues to carry the 72-bit data (which is part of the 576-bit data) to be compared.
Note. For 576-bit searches, the host ASIC must supply individual 72-bit data on DQ[71:0] during cycles A and B. Also, four individual pairs of GMR and CMPR registers may be involved in the comparison.
The logical 576-bit Search operation is shown in Figure 6-38. The upper half of the device consisting of 576-bit entries is compared to a 576 -bit search key, $K$ that is presented on the DQ bus in eight CLK2x cycles using the GMR and local mask bits. The same also happens in the lower half of the device. The GMR is the 576 -bit word specified by four pairs of GMRs selected by GMR indices in each of the eight devices. The 576 -bit word K (presented on the DQ bus in all eight cycles of the command) is also stored in both even and odd comparand register pairs (selected by the comparand register index in command cycle B) in each of the eight devices. The word K is compared with each entry in the table in both arrays. The winning addresses from both arrays will be determined based on the Soft Priority and Mini-Key scheme, and the result of the two searches from the two halves are driven as part of the SRAM address on the SADR[N:0] lines ( $N=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128) with two SRAM cycles as shown in the timing diagram (see Section 6.7, "SRAM PIO Access," on page 121). On a global miss cycle, the device with LRAM = 1 (binary) (default driving device for the SRAM bus) and LDEV = 1 (binary) (default driving device for SSF and SSV signals) will be the default driver for such missed cycles.
The Search command is a pipelined operation and executes a Search at one-eighth the rate of the frequency of CLK2X for 576-bit searches in $\times 576$-configured tables. The latency of the Search from command to SRAM access cycle is 5 for up to eight devices in the table (TLSZ = 01 (binary)). SSV and SSF also shift further to the right for different values of HLAT, as specified in Table 6-5.


Figure 6-38. $\times 576$ Table with Eight Devices

### 6.5.8 Mixed-size Single Searches with 31 Devices on Tables Configured with Different Widths

This subsection will cover mixed searches ( $\times 72, \times 144$, and $\times 288$ ) with tables of different widths $(\times 72, \times 144, \times 288)$. Note: NonEnhanced Mode does not support 576-bit tables. The sample operation shown is for 31 devices, with devices 0 to 7 containing x72 tables (CFG field in Command register all zeroes for Non-Enhanced Mode, NES = "00" (binary) for all blocks for Enhanced Mode), devices 8 to 15 containing x144 tables (CFG[63:0] field in Command register for CYNSE10512 = 5555555555555555 (hex), CFG[31:0] = 55555555 (hex) for CYNSE10256, CFG[15:0] $=5555$ (hex) for CYNSE10128 for Non-Enhanced Mode, NES $=$ "01" (binary) in all blocks for Enhanced Mode), and the rest of the devices containing x288 tables (CFG[63:0] = AAAAAAAAAAAAAAAA (hex) for CYNSE10512, CFG[31:0] = AAAAAAAA (hex) for CYNSE10256, CFG[15:0] = AAAA (hex) for CYNSE10128 for Non-Enhanced Mode, NES = "10" (binary) in all blocks for Enhanced Mode). The following figures show three sequential searches: first, a 72 -bit Search on a $\times 72$-configured table; a 144-bit Search on a $\times 144$-configured table; and a 288 -bit Search on a $\times 288$-configured table that each results in a hit.
The 31 cascaded devices can be viewed as three blocks of 8 devices and a fourth block of 7 devices, as shown in Figure 6-19. Each individual block of 8 or 7 devices is connected very similarly to the connection shown in Figure 6-9, except that the BHI[2:0] signals are connected to BHO of the previous block rather than being grounded. Figure 6-39 shows a graphical example of the tables using CYNSE10512s.


Figure 6-39. Multiwidth Configurations Example with CYNSE10512s

## Notes:

- The "Block" in the figure above refers to a block of 8 devices, not a block within a single device.
- All 31 devices must be programmed with the same values for TLZ ("10" (binary)) and HLAT ("000" (binary) in this example). Only the last device in the table must be programmed with LRAM $=1$ (binary) and LDEV $=1$ (binary) (device 30 in this case). All other upstream devices must be programmed with LRAM $=0$ (binary) and LDEV $=0$ (binary) (devices 0 through 29 in this case).
- The device receiving all the LHO signals from the other devices is considered the last device.
- All the shared signals in the following timing diagrams showing tri-stated condition ("z") indicate that, that particular device is not driving the shared signals. The shared signals are not three-stated in a real life because other devices will be driving them.
- One way to create many tables of different widths in a bank of NSEs is by having table designation bits. It is assumed that bits [71:70] for each entry will be assigned such table designation bits. DQ[71:70] will be 00 in each of the two $A$ and $B$ cycles of the $\times 72$-bit Search (Search1). DQ[71:70] is 01 in each of the $A$ and $B$ cycles of the $\times 144$-bit Search (Search2). DQ[71:70] is 10 in each of the A, B, C, and D cycles of the $\times 288$-bit Search (Search3).
The timing diagrams below corresponds to the Hit/Miss assumptions defined in Table 6-11. For the purpose of illustrating the timings, it is further assumed that there is only one device with a matching entry in each of the blocks.

Table 6-11. Hit/Miss Assumptions

| Search Number | \#1 (x72) | \#2 (x144) | \#3 (x288) |
| :---: | :---: | :---: | :---: |
| Block 0 | Hit | Miss | Miss |
| Block 1 | Miss | Hit | Miss |
| Block 2 | Miss | Miss | Hit |
| Block 3 | Miss | Miss | Miss |



Figure 6-40. Timing Diagram for Mixed Search for Devices Above Block 0 Winning Device


Figure 6-41. Timing Diagram for Mixed Search for Block 0 Winning Device


Figure 6-42. Timing Diagram for Mixed Search for Devices Below Block 0 Winning Device


M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
For Non-Enhanced Mode:
CYNSE10512: CFG[63:0] = 5555555555555555 h ;
CYNSE10256: CFG[31:0] $=55555555 \mathrm{~h}$;
CYNSE10128: CFG[15:0] $=5555 \mathrm{~h}$.
NES $=01$ (binary) in all blocks for Enhanced Mode, x144 search
HLAT $=000$ (binary), $\mathrm{TLSZ}=10$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).
Figure 6-43. Timing Diagram for Mixed Search Above Block 1 Winning Device


Figure 6-44. Timing Diagram for Mixed Search for Block 1 Winning Device


Figure 6-45. Timing Diagram for Mixed Search Below Block 1 Winning Device


Figure 6-46. Timing Diagram for Mixed Search Above Block 2 Winning Device


Figure 6-47. Timing Diagram for Mixed Search for Block 2 Winning Device


Figure 6-48. Timing Diagram for Mixed Search Below Block 2 Winning Device


For Non-Enhanced Mode:
CYNSE10512: CFG[63:0] = AAAAAAAAAAAAAAAAh;
CYNSE10256: CFG[31:0] = AAAAAAAAh;
CYNSE10128, CFG[15:0] = AAAAh.
NES $=10$ (binary) in all blocks for Enhanced Mode, x288 search.
HLAT $=000$ (binary), $\mathrm{TLSZ}=10$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).

Figure 6-49. Timing Diagram for Mixed Search for All Except the Last Device in Block 3


M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
For Non-Enhanced Mode:
CYNSE10512: CFG[63:0] = AAAAAAAAAAAAAAAAh;
CYNSE10256: CFG[31:0] = AAAAAAAAh;
CYNSE10128, CFG[15:0] = AAAAh.
NES = 10 (binary) in all blocks for Enhanced Mode, x288 search
HLAT = 000 (binary), TLSZ = 10 (binary), LRAM = 1 (binary), LDEV = 1 (binary).
Figure 6-50. Timing Diagram for Mixed Search for the Last Device in Block 3

The following is the sequence of operation for a single mixed-width Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

- Cycle A:
—Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10" (binary). The CMD[2] and CMD[9] signals must be driven to logic 0 for the 72 -bit search, but for 144 -bit search, $\mathrm{CMD}[9]=1$ and CMD [2] = 0 . For 288-bit search, CMD[9] is don't care, whereas CMD[2] = 1 for the first " A " cycle and 0 for the last " A " cycle. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 by this device if it has a hit.
-DQ Bus: At the same time in cycle A, DQ[71:0] must be driven with the 72-bit data to be compared.


## - Cycle B:

—Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10" (binary). CMD[5:2] must now be driven by the index of the comparand register pair for storing the search key presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 27 for a description of SSR[0:7]). CMD[10:9] are don't cares for this cycle.
-DQ Bus: The DQ[71:0] continues to carry the search key to be compared.
Note. For 72-bit searches, the host ASIC must supply the same 72-bit data on DQ[71:0] during both cycles A and B. Also, the even and odd pairs of GMRs selected for the comparison must be programmed with the same value. For 144-bit, 288-bit or 576bit searches, each 72 -bit presented on each cycle A and B will together form the 144 -bit or 288 -bit or 576 -bit search key respectively.
When an N-bit search key, K, is presented on the DQ bus, the entire table of N-bit entries is compared to the search key using the GMR and local mask bits. The GMR is selected by the GMR Index in the command's cycle A. K is also stored in both even and odd comparand register pairs (selected by the comparand register index in command cycle B). K is compared with each entry in the table, starting at location 0. A matching entry that satisfies the Soft Priority and Mini-Key scheme (for Enhanced Mode) will be the winning entry, and its location address $L$ will be driven as part of the SRAM address on the SADR[N:0] lines (see Section 6.7, "SRAM PIO Access," on page 121), $\mathrm{N}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128. Note. The Learn command is supported for only one of the blocks consisting of up to eight devices in a depth-cascaded table of more than one block.
For up to 31 devices in the table (TLSZ = 10 (binary)), Search latency is 6 from command to SRAM access cycle. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 6-5.

### 6.5.9 Mixed-size Multi Searches with 8 Devices on Tables Configured with Different Widths

This subsection will cover mixed searches ( $\times 72, \times 144$, and $\times 288$ ) with tables of different widths $(\times 72, \times 144, \times 288)$ when MultiSearch is enabled. The sample operation shown is for 8 -device-cascade, with devices 0 and 1 containing $\times 72$ tables (NES $=00$ (binary) in all blocks), devices 2 and 3 containing $x 144$ tables (NES $=01$ (binary) in all blocks), and devices 4 to 7 containing $\times 288$ tables (NES $=10$ (binary) in all blocks). The following figures show three sequential searches: first, a 72 -bit Search on a $\times 72$ configured table; a 144-bit Search on a $\times 144$-configured table; and a 288 -bit Search on a $\times 288$-configured table that each results in a hit.
The hardware connection of the 8 cascaded devices is shown in Figure 6-19. A graphical representation of the tables is shown in Figure 6-51 using CYNSE10512s as an example.


Figure 6-51. Multiwidth Configurations Example for MultiSearch with CYNSE10512s

## Note:

- When MultiSearch is enabled, the maximum number of devices that can be cascaded is 8 if CLK2x is less than or equal to 200 MHz . The number of devices will be 4 if CLK2x operates above 200 MHz but up to 266 MHz .
- All eight devices must be programmed with the same values for TLZ ("01" (binary)) and HLAT ("000" (binary) in this example). Only the last device in the table must be programmed with LRAM = 1 (binary) and LDEV = 1 (binary) (device 7 in this case). All other upstream devices must be programmed with LRAM $=0$ (binary) and LDEV $=0$ (binary) (devices 0 through 6 in this case).
- The device receiving all the LHO signals from the other devices is considered the last device.
- All the shared signals in the following timing diagrams showing tri-stated condition (" $z$ ") indicate that, that particular device is not driving the shared signals. The shared signals are not three-stated in a real life because other devices will be driving them.
- Comparing the hardware diagrams shown in Figure 6-9 and Figure 6-14, enabling MultiSearch does not mean that a board layout change is required. The LHO_1_L and LHI_1_L share the same pin with the Full In and Full Out signals, which are not shown in Figure 6-9. Cascading multiple devices together still allow the user to configure the devices through software to perform single-search or MultiSearch operations without any board change.
- One way to create many tables of different widths in a bank of NSEs is by having table designation bits. It is assumed that bits [71:70] for each entry will be the table designation bits. The DQ[71:70] will be 00 in each of the two $A$ and $B$ cycles of the $\times 72-$ bit MultiSearch (M-Search1). DQ[71:70] is 01 in each of the $A$ and $B$ cycles of the $\times 144$-bit MultiSearch (M-Search2). DQ[71:70] is 10 in each of the $A, B, C$, and D cycles of the $\times 288$-bit MultiSearch (M-Search3).
The timing diagrams below corresponds to the Hit/Miss assumptions defined in Table 6-12.
Table 6-12. Hit/Miss Assumptions in MultiSearchMode

| Search Number | \#1 (x72) |  | \#2 (x144) |  | \#3 (x288) |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Device 0 | Miss | Hit | Miss | Miss | Miss | Miss |
| Device 1 | Miss | Hit | Miss | Miss | Miss | Miss |
| Device 2 | Miss | Miss | Miss | Hit | Miss | Miss |
| Device 3 to 6 | Miss | Miss | Miss | Miss | Miss | Miss |
| Device 7 | Miss | Miss | Miss | Miss | Hit | Miss |



NES $=00$ (binary) in all blocks for Enhanced Mode, x72 search.
HLAT $=000$ (binary), TLSZ $=01$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).

Figure 6-52. Timing Diagram for Mixed MultiSearch (Eight Devices) for Device 0


Figure 6-53. Timing Diagram for Mixed MultiSearch (Eight Devices) for Device 1


NES = 01 (binary) in all blocks for Enhanced Mode, x144 search.
HLAT = 000 (binary), TLSZ = 01 (binary), LRAM $=0$ (binary), LDEV $=0$ (binary).
Figure 6-54. Timing Diagram for Mixed MultiSearch (Eight Devices) for Device 2


Figure 6-55. Timing Diagram for Mixed MultiSearch (Eight Devices) for Device 7

The MSE bit in the Command Register must be set high to enable the MultiSearch feature. The same with the Enhanced Mode (EMODE) bit. The following is the sequence of operation for a single mixed-width Search command (also refer to Subsection 6.2, "Command Bus Parameters," on page 50).

## - Cycle A:

—Command Bus: The host ASIC drives CMDV HIGH and applies Search command CMD[1:0] = "10" (binary). The CMD[2] and CMD[9] signals must be driven to logic 0 for the 72-bit search, but for 144-bit search, CMD[9] = 1 and CMD [2] = 0 . For 288-bit search, CMD[9] is don't care, whereas CMD[2] = 1 for the first " A " cycle and 0 for the last " A " cycle. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[7:6] signals must be driven with the same bits that will be driven on SADR[24:23] for CYNSE10512, SADR[23:22] for CYNSE10256, SADR[22:21] for CYNSE10128 by this device if it has a hit. CMD[8] must be set high for MultiSearch operation.
—DQ Bus: At the same time in cycle A, DQ[71:0] must be driven with the 72-bit data to be compared.

- Cycle B:
- Command Bus: The host ASIC continues to drive CMDV HIGH and to apply Search command CMD[1:0] = "10" (binary). CMD[5:2] must now be driven by the index of the comparand register pair for storing the search key presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 27 for a description of SSR[0:7]). CMD[10:9] are don't cares for this cycle.
-DQ Bus: The DQ[71:0] continues to carry the search key to be compared.
Note. For 72-bit multi-searches, the host ASIC can provide different 72-bit data on DQ[71:0] on each of the A and B cycles. The even and odd pairs of GMRs selected for the comparison need not be programmed with the same value. For 144-bit, 288-bit or 576 -bit searches, each 72 -bit presented on each cycle A and B will together form the 144-bit or 288-bit or 576-bit search key respectively. Each search key will be compared to both arrays 0 and 1 during cycles A and B when MultiSearch is enabled.
When an N -bit search key, K , is presented on the DQ bus, both arrays of N -bit entries are compared to the search key using the GMR and local mask bits. The GMR is selected by the GMR Index in the command's cycle A. K is also stored in both even and odd comparand register pairs (selected by the comparand register index in command cycle $B$ ). K is compared with each entry in the table, starting at location 0. A matching entry from each array that satisfies the Soft Priority and Mini-Key scheme will be the winning entries, and their location addresses La and Lb will be driven as part of the SRAM address on the SADR[N:0] lines (see Section 6.7, "SRAM PIO Access," on page 121), N = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128. Note. The Learn command is supported for only one of the blocks consisting of up to eight devices in a depth-cascaded table of more than one block.
The latency of the MultiSearch from command to SRAM access cycle is 5 for a configuration of up to eight devices (TLSZ = 01 (binary)). SSV and SSF also shift further to the right for different values of HLAT, as specified in Table 6-5.


### 6.6 Learn Command

The device contains sixteen pairs of Comparand (CMPR) registers that store the search key as the device executes searches. On a Search miss, signalled to the ASIC through the SSV and SSF signals (SSV = 1 (binary), SSF = 0 (binary)), the host ASIC can apply the Learn command to learn the entry from a CMPR register to the next-free location. However, it is recommended that the host ASIC first check the FULL signal, to determine if the device is full. If the device is not full, and the Search was a miss, a Learn can be applied. If the device is already full, and the Learn is issued, the operation will be suppressed.
The Learn command is a pipelined operation and lasts for two CLK cycles. Figure 6-60, Figure 6-61 and Figure 6-62 show the timing diagram of Learn operations with the address taken from the NFA or SRR register. Learn operations with the address taken from the DQ bus follow the same diagrams except that the DQ bus contains the address instead of Don't Cares. Figure 6-61 and Figure 6-62 assume that the device performing the Learn operation is not the last device in the table and will therefore have its LRAM bit set to 0 . The OE_L for the device with the LRAM bit set goes HIGH for two cycles for each Learn (one during the SRAM Write cycle and one during the cycle before). The SRAM Write cycle latency from the second cycle of the instruction is shown in Table 6-13. The Learn command also generates a Write cycle to the external SRAM (see Section 6.7, "SRAM PIO Access," on page 121).
Note that mismatched entry-width Learn operation is not supported. For example, the result of a 72-bit Search miss stored in one of the SRR registers cannot be used for a 144-bit Learn operation.

### 6.6.1 Non-Enhanced Mode

The Learn command in the Non-Enhanced mode supports x72 and x144 table widths. The operation uses the data stored in the user selected CMPR register for writing to an entry in the Data array. Non-Enhanced mode Learn operation ignores the DQ bus and cannot perform a write to the Mask array. The address for the target data entry is the INDEX field of the Next-free Address (NFA) register.
Once the operation is completed the NFA register's INDEX field is updated with next highest priority free entry in the Data array. The LSB of each $x 72$ entry is treated as a valid bit and used to indicate whether that entry is free ( $=0$ (binary)) or not (=1 (binary)). For a 144-bit entry, bit [72] and bit[0] must be set to the same value.

Note that Learn command for x 144 entry width in Non-Enhanced can only be issued when all the tables in the device is of $\times 144$ table width.

### 6.6.2 Enhanced Mode

The Learn command in the Enhanced mode supports all table widths ( $x 72, x 144, x 288$ and $x 576$ ). The user can select whether the data stored in the user selected CMPR register or the data presented in the DQ bus be used for the learn operation. The user can also select to write to an entry in either the Data or Mask array. The address for the target entry is the INDEX field of the userselected Search Result Register (SRR). Each SRR is one-to-one associated to a Comparand (CMPR) register. So the selection of the SRR is accomplished by selecting the corresponding (CMPR) register.
The SRR register is updated after a Search operation. Only the LSB of each entry is used, regardless of width, to indicate whether that entry is free (=0 (binary)) or not (=1 (binary)).


TLSZ = 00 (binary), LRAM = 1 (binary), LDEV = 1 (binary). $\mathrm{M}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128

Figure 6-56. Timing Diagram of 72-bit Learn from DQ Bus and CMPR Registers (One Device)


TLSZ $=10$ (binary), LRAM = 1 (binary), LDEV = 1 (binary). $\mathrm{M}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-57. Timing Diagram of 288-bit Learn from DQ Bus and CMPR Registers (One Device)


TLSZ $=10$ (binary), LRAM = 1 (binary), LDEV = 1 (binary). $\mathrm{M}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-58. Timing Diagram of 576-bit Learn from DQ Bus (One Device)


TLSZ $=10$ (binary), LRAM = 1 (binary), LDEV = 1 (binary). $\mathrm{M}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-59. Timing Diagram of 576-bit Learn from CMPR Register (One Device)

### 6.6.3 Learn Operation on Depth-Cascaded Table

When all entries in a device are occupied, the device asserts FULO to inform the downstream devices that it is full. The result of this communication between depth-cascaded devices determines the global FULL signal for the entire table. The FULL signal in the last device determines the fullness of the depth-cascaded table.
In a depth-cascaded table, only a single device will Learn the entry through the application of a Learn instruction. The determination as to which device will Learn is based on the FULI and FULO signals between the devices. The first non-full device learns the entry by storing the content of the selected CMPR register to the location pointed to by the NFA or SRR register.
The global FULL signal indicates to the table controller (the host ASIC) that all entries within a block are occupied and that no more entries can be learned. The Ayama 10000 device updates the signal after each Write or Learn command to a data array.


TLSZ $=00$ (binary), LRAM = 1 (binary), LDEV = 1 (binary). $\mathrm{M}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-60. Timing Diagram of Learn (TLSZ = 00 (binary), LDEV = 1 (binary))


TLSZ $=01$ (binary), LRAM = 0 (binary), LDEV = 0 (binary). $\mathrm{M}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-61. Timing Diagram of Learn (Except on the Last Device [TLSZ = 01 (binary)])


TLSZ $=01$ (binary), LRAM = 1 (binary), LDEV = 1 (binary). $\quad \mathrm{M}=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128
Figure 6-62. Timing Diagram of Learn on Device Number 7 (TLSZ = 01 (binary))
Table 6-13. SRAM Write Cycle Latency from Second Cycle of Learn Instruction

| Number of Devices | Latency in CLK Cycles |
| :---: | :---: |
| $1($ TLSZ $=00$ (binary)) | 4 |
| $1-8($ TLSZ $=01$ (binary)) | 5 |
| $1-31($ TLSZ $=10$ (binary)) | 6 |

The Learn operation lasts two CLK cycles. The sequence of operation is as follows.

- Cycle 1A: The host ASIC applies the Learn instruction on CMD[1:0] using CMDV = 1 (binary). The CMD[5:2] field specifies the index of the comparand register pair that will be written in the data array in the 144-bit-configured table. For a Learn in a 72 -bit-configured table, the even-numbered comparand specified by this index will be written. CMD[8:6] carries the bits that will be driven on SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 in the SRAM Write cycle.
- Cycle 1B: The host ASIC continues to drive CMDV to 1 (binary), CMD[1:0] to 11 (binary), and CMD[5:2] with the comparand pair index. CMD[6] must be set to 0 if the Learn is being performed on a 72 -bit-configured table, and to 1 if the Learn is being performed on a 144-bit-configured table.
- Cycle 2: The host ASIC drives CMDV to 0 .

At the end of cycle 2, a new instruction can begin. SRAM Write latency is the same as the Search to the SRAM Read cycle. It is measured from the second cycle of the Learn instruction.

### 6.7 SRAM PIO Access

SRAM Read enables read access to the off-chip SRAM containing associative data. The latency from the issuance of the Read instruction to the appearance of the address on the SRAM bus is the same as the Search instruction latency, and will depend on the value programmed for the TLSZ parameter in the device configuration register. The latency of the ACK from the Read instruction is the same as that from the Search instruction to the SRAM address latency, plus the HLAT programmed in the configuration register. Note. SRAM Read is a blocking operation-no new instruction can begin until the ACK is returned by the selected device performing the access.
SRAM Write enables write access to the off-chip SRAM containing associative data. The latency from the second cycle of the Write instruction to the appearance of the address on the SRAM bus is the same as the Search instruction latency, and will depend on the TLSZ value parameter programmed in the device configuration register. Note: SRAM Write is a pipelined operation-new instruction can begin right after the previous command has ended.

### 6.7.1 SRAM Read with a Table of One Device

SRAM Read enables read access to the off-chip SRAM containing associative data. The latency from the issuance of the Read instruction to the appearance of the address on the SRAM bus is the same as Search instruction latency, and will depend on the TLSZ value parameter programmed into the device configuration register. ACK latency from the Read instruction is the same as that from the Search instruction to the SRAM address, plus the HLAT programmed in the configuration register. The following explains the SRAM Read operation in a table with only one device that has the following parameters: TLSZ = 00 (binary), HLAT = 000 (binary), LRAM $=1$ (binary), and LDEV $=1$ (binary). Figure 6-63 shows the associated timing diagram. For the following description, the selected device refers to the only device in the table because it is the only device to be accessed.

- Cycle 1A: The host ASIC applies the Read instruction on CMD[1:0] using CMDV = 1 (binary). The DQ bus supplies the address, with DQ[20:19] set to 10 (binary), to select the SRAM address. The host ASIC selects the device for which ID[4:0] matches the DQ[25:21] lines. During this cycle, the host ASIC also supplies SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6].
- Cycle 1B: The host ASIC continues to apply the Read instruction on CMD[1:0] using CMDV = 1 (binary). The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive DQ[71:0] and drives ACK from High-Z to LOW.
- Cycle 5: The selected device drives the Read address on SADR[N:0] lines ( $N=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128) and drives ACK HIGH, CE_L LOW, and ALE_L LOW.
- Cycle 6: The selected device drives CE_L HIGH, ALE_L HIGH, the SADR bus, the DQ bus in a three-state condition, and ACK LOW.
At the end of cycle 6, the selected device floats ACK in a three-state condition, and a new command can begin.


TLSZ $=00$ (binary), HLAT = 000 (binary), LRAM = 1 (binary), LDEV = 1 (binary)
Figure 6-63. SRAM Read Access (TLSZ = 00 (binary), HLAT = 000 (binary), LRAM =1 (binary), LDEV =1 (binary))

### 6.7.2 SRAM Read with a Table of up to Eight Devices

The following explains the SRAM Read operation completed through a table of up to eight devices using the following parameter: TLSZ $=01$ (binary). Figure $6-64$ diagrams a block of eight devices. The following assumes that SRAM access is successfully achieved through Ayama 10000 device number 0 . Figure $6-65$ and Figure 6-66 show timing diagrams for device number 0 and device number 7 , respectively.

- Cycle 1A: The host ASIC applies the Read instruction on CMD[1:0] using CMDV $=1$. The DQ bus supplies the address, with DQ[20:19] set to 10 , to select the SRAM address. The host ASIC selects the device for which ID[4:0] matches the $D Q[25: 21]$ lines. During this cycle the host ASIC also supplies SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6].
- Cycle 1B: The host ASIC continues to apply the Read instruction on CMD[1:0], using CMDV $=1$. The DQ bus supplies the address, with DQ[20:19] set to 10 , to select the SRAM address.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive DQ[71:0].
- Cycle 5: The selected device continues to drive DQ[71:0] and drives ACK from High-Z to LOW.
- Cycle 6: The selected device drives the Read address on SADR[N:0] lines ( $N=25$ for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128) and drives ACK HIGH, CE_L LOW, WE_L HIGH, and ALE_L LOW.
- Cycle 7: The selected device drives CE_L, ALE_L, WE_L, and the DQ bus in a three-state condition. It continues to drive ACK LOW.
At the end of cycle 7, the selected device floats ACK in a three-state condition. A new command can begin.


Figure 6-64. Hardware Diagram of a Block of Eight Devices


TLSZ $=01$ (binary), HLAT $=000$ (binary), LRAM $=0$ (binary), LDEV $=0$ (binary)
Figure 6-65. SRAM Read of Device \#0 in a Block of Eight Devices


TLSZ = 01 (binary), HLAT = 000 (binary), LRAM = 1 (binary), LDEV = 1 (binary)
Figure 6-66. SRAM Read Timing of Device \#7 in a Block of Eight Devices

### 6.7.3 SRAM Read with a Table of up to 31 Devices

The following explains the SRAM Read operation accomplished through a table of up to 31 devices, using the following parameter: TLSZ = 10 (binary). The hardware diagram is shown in Figure 6-67. The following assumes that SRAM access is being accomplished through Ayama 10000 device number 0, and that device number 0 is the selected device. Figure 6-68 and Figure 6-69 show the timing diagrams for device number 0 and device number 30, respectively.

- Cycle 1A: The host ASIC applies the Read instruction to CMD[1:0] using CMDV $=1$. The DQ bus supplies the address, with DQ[20:19] set to 10 , to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. During this cycle, the host ASIC also supplies SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6].
- Cycle 1B: The host ASIC continues to apply the Read instruction to CMD[1:0], using CMDV = 1 . The DQ bus supplies the address, with DQ[20:19] set to 10, to select the SRAM address.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive DQ[71:0].
- Cycles 5 to 6: The selected device continues to drive DQ[71:0].
- Cycle 7: The selected device continues to drive DQ[71:0], and drives an SRAM Read cycle.
- Cycle 8: The selected device drives ACK from Z to LOW.
- Cycle 9: The selected device drives ACK to HIGH.
- Cycle 10: The selected device drives ACK from HIGH to LOW.

At the end of cycle 10, the selected device floats ACK to High-Z and a new command can begin.


Figure 6-67. Hardware Diagram of 31 Devices Using Four Blocks


Figure 6-68. SRAM Read of Device \#0 in a Bank of 31 Devices


TLSZ $=10$ (binary), HLAT $=010$ (binary), LRAM = 1 (binary), LDEV = 1 (binary) M = 25 for CYNSE10512, 24 for CYNSE10256, 23 for CYNSE10128

Figure 6-69. SRAM Read of Device \#0 in a Bank of 31 Devices

### 6.7.4 SRAM Write with a Table of One Device

SRAM Write enables Write access to the off-chip SRAM containing associative data. The latency from the second cycle of the Write instruction to the appearance of the address on the SRAM bus is the same as Search instruction latency, and will depend on the TLSZ value parameter programmed in the device configuration register. The following explains the SRAM Write operation accomplished through a table of only one device with the following parameters: TLSZ = 00 (binary), HLAT = 000 (binary), LRAM $=1$ (binary), and LDEV = 1 (binary). Figure 6-70 shows the timing diagram. For the following description, the selected device refers to the only device in the table because it is the only device that will be accessed.

- Cycle 1A: The host ASIC applies the Write instruction on CMD[1:0] using CMDV $=1$. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. The host ASIC also supplies SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6] in this cycle. Note. CMD[2] must be set to 0 for SRAM Write because burst WRITEs into the SRAM are not supported.
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] using CMDV $=1$. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. Note. CMD[2] must be set to 0 for SRAM Write because burst WRITEs into the SRAM are not supported.
- Cycle 2 and cycle 3: wait states. Data not used by NSE.

At the end of cycle 3, a new command can begin. The Write is a pipelined operation; the Write cycle appears at the SRAM bus, however, with the same latency as the Search instruction, as measured from the second cycle of the Write command.


Figure 6-70. SRAM Write Access (TLSZ = 00 (binary), HLAT = 000 (binary), LRAM = 1 (binary), LDEV = 1 (binary))

### 6.7.5 SRAM Write with a Table of up to Eight Devices

The following explains the SRAM Write operation accomplished through a table(s) of up to eight devices with the following parameters (TLSZ = 01 (binary)). The hardware diagram for this table is shown in Figure 6-71. The following assumes that SRAM access is achieved through Ayama 10000 device number 0. Figure 6-72 and Figure 6-73 show the timing diagram for device number 0 and device number 7 , respectively.


Figure 6-71. Hardware Diagram of a Block of Eight Devices

- Cycle 1A: The host ASIC applies the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. The host ASIC also supplies SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6] in this cycle. Note. CMD[2] must be set to 0 for SRAM Write because burst WRITEs into the SRAM are not supported.
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] using CMDV $=1$. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. Note. CMD[2] must be set to 0 for SRAM Write because burst WRITEs into the SRAM are not supported.
- Cycle 2: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the Ayama 10000 device.
- Cycle 3: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the Ayama 10000 device.

At the end of cycle 3, a new command can begin. Write is a pipelined operation, but the Write cycle appears at the SRAM bus with the same latency as that of a Search instruction, as measured from the second cycle of the Write command.


Figure 6-72. SRAM Write of Device \#0 in a Block of Eight Devices


Figure 6-73. SRAM Write Timing of Device \#7 in Block of Eight Devices

### 6.7.6 SRAM Write with Table(s) Consisting of up to 31 Devices

The following explains the SRAM Write operation accomplished through a table of up to 31 devices with the following parameter: TLSZ = 10 (binary). The hardware diagram is shown in Figure 6-74. The following assumes that SRAM access is accomplished through Ayama 10000 device number 0-the selected device. Figure 6-75 and Figure 6-76 show timing diagrams for device number 0 and device number 30, respectively.

- Cycle 1A: The host ASIC applies the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. The host ASIC also supplies SADR[25:23] for CYNSE10512, SADR[24:22] for CYNSE10256, SADR[23:21] for CYNSE10128 on CMD[8:6] in this cycle. Note. CMD[2] must be set to 0 for SRAM Write because burst WRITEs into the SRAM are not supported.
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. Note. CMD[2] must be set to 0 for SRAM Write because burst WRITEs into the SRAM are not supported.
- Cycle 2: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the Ayama 10000 device.
- Cycle 3: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the Ayama 10000 device.

At the end of cycle 3, a new command can begin. The Write is a pipelined operation, but the Write cycle appears at the SRAM bus with the same latency as that of a Search instruction, as measured from the second cycle of the Write command.


Figure 6-74. Table of 31 Devices (Four Blocks)


Figure 6-75. SRAM Write of Device \#0 in Bank of 31 Devices


### 6.8 Timing Sequences for Back-to-Back Operations

Table 6-14 shows the idle cycle requirements between operations. The operations in the second column represent operations already performed, and the operations in the first row are those we would like to perform next.

## Example calculations:

1. Read after Write: The Write takes two 2 cycles, and one 1 idle cycle is required. Thus if the Write is issued in cycle 1 , the Read cannot be issued until cycle 4. Note, all cycles after an SRAM Read or an NSE Read (blocking) operation are considered blocked until the ACK signal is returned.
2. Learn from SRR after Search $\times 288$, with $T L S Z=10$ (binary): The Search takes 2 cycles, and ( $2+T L S Z$ ) idle cycles are required. Thus if the Search is issued in cycle 1, the Learn cannot be issued until cycle 7.

## Table 6-14. Required Idle Cycles Between Commands

| \# of Cycles | OPERATIONS | SEARCH | READ | WRITE | LEARN | SRAM |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \hline x 72 / \times 144=1 \text { Cycle } \\ \times 288=2 \text { Cycles } \\ \times 576=4 \text { Cycles } \end{gathered}$ | SEARCH | No Wait | $\begin{aligned} & \text { No Wait / } \\ & 2+\text { TLSZ }^{15} \end{aligned}$ | No Wait | $\begin{aligned} & \text { No Wait / } \\ & \text { 2+TLSZ }{ }^{16,19} \end{aligned}$ | $\begin{gathered} \text { TLSZ/ } \\ 2+\text { TLSZ }^{17} \end{gathered}$ |
| 1 Cycle | READ | 5 | 5 | 5 | 5 | 5 |
| 2 Cycles | WRITE | 1 | 1 | 1 | $1 / 1+$ TLSZ $^{\text {18 }}$ | 1 |
| $\begin{gathered} \text { x72/x144 = } 1 \text { Cycle } \\ \text { x288 }=2 \text { Cycles } \\ \text { x576 }=4 \text { Cycles } \end{gathered}$ | LEARN | 1 | 1 | 1 | 1 | 1 |
| 1 Cycle | SRAM READ | 5+TLSZ+HLAT | 5+TLSZ+HLAT | 5+TLSZ+HLAT | 5+TLSZ+HLAT | 5+TLSZ+HLAT |
| 2 Cycles | SRAM WRITE | 1 | 1 | 1 | 1 | 1 |

## Notes:

15. When the register being read is SSR/SRR and it matches the target location of the previous search, a READ operation cannot be issued for $2+T L S Z$ idle cycles to avoid reading the old value. Otherwise there is no idle cycle requirement.
16. In Non-Enhanced Mode there is no idle cycle requirement. In Enhanced Mode, an SRR is updated on a SEARCH miss and is used as the address for the LEARN. Must wait for $2+$ TLSZ cycles after the last Search, before issuing a subsequent Learn that uses the same SRR as the last Search.
17. The SRAM operation needs to insert idle cycles to avoid SADR bus contention with previous SEARCH.
18. In non-Enhanced Mode, a WRITE operation updates the NFA register used for LEARN operation. Must wait for $1+$ TLSZ cycles before issuing LEARN to avoid learning with the old NFA value.
19. If the Learn is issued $2+$ TLSZ after the corresponding Search that updated the SRR, the Learn will be issued before the Search result or the updated FULL signal is returned. If the Search resulted in a hit, the Learn will be suppressed. If there was a miss, but the device is already full, the Learn will also be suppressed.

### 6.9 Full Signal Timing Diagram

FULL indicates when the array (Non-Enhanced Mode) or selected blocks (Enhanced Mode) is full.
In Non-Enhanced Mode, FULL is valid four CLK1X cycles after the command is issued, regardless of TLSZ and HLAT. At all other times, FULL maintains its value until another operation changes it.
In Enhanced Mode, FULL is valid when SSV is high. Figure 6-77 is a timing diagram of the FULL signal in Enhanced (top) and non-Enhanced (bottom) Modes.


Figure 6-77. Timing Diagram for Full Signal (TLSZ = 10)

### 7.0 JTAG (IEEE 1149.1)

The Ayama 10000 device supports the Test Access Port and Boundary Scan Architecture as specified in IEEE JTAG Standard Number 1149.1. The pin interface to the chip consists of five signals with the standard definitions: TCK, TMS, TDI, TDO, and TRST_L. Table 7-1 describes the operations that the test access port controller supports, and Table 7-2 describes the TAP Device ID Register. JTAG can also be reset by driving TMS HIGH, and then holding it for three (3) TCK rising edges.

Note. To disable JTAG functionality, connect the TCK, TMS, and TDI pins to $V_{\text {DDQ }}$ through pull-resistors and hold TRST_L Low.
Table 7-1. Supported Operations

| Instruction | Type | Description |
| :---: | :--- | :--- |
| SAMPLE/PRELOAD | Mandatory | Sample/Preload. This operation loads the values of signals going to and from I/O pins <br> into the boundary scan shift register to provide a snapshot of the normal functional <br> operation. |
| EXTEST | Mandatory | External Test. This operation uses boundary scan values shifted in from the TAP to test <br> connectivity external to the device. |
| BYPASS | Mandatory | Bypass. This operation bypasses the device in a JTAG chain by loading a single bit shift <br> register between TDI and TDO and provides a minimum-length serial path when no test <br> operation is required. |
| IDCODE | Optional | Device JTAG ID Code. This operation selects the JTAG Identification register and output <br> the IDCODE field serially through TDO. |
| CLAMP | Optional | Output Clamp. This operation drives preset values onto the outputs of the device. |
| HIGHZ | Optional | High-Z Output. This operation sets the device output signals in high impedance state. |

Table 7-2. TAP Device ID Register

| Field | Range | Initial Value | Description |
| :---: | :---: | :---: | :--- |
| Revision | $[31: 28]$ | 0001 | Revision Number. This is the current device revision number. Numbers <br> start from one and increment by one for each revision of the device. |
| Part Number | $[27: 12]$ | 0000000000010100 | Part Number. This is the part number for CYNSE10128. |
|  |  | 0000000000010101 | This is the part number for CYNSE10256. |
|  | 0000000000010110 | This is the part number for CYNSE10512. |  |
| MFID | $[11: 1]$ | $000 \_1101 \_1100$ | Manufacturer ID. This field is the same as the manufacturer ID used in the <br> TAP controller. |
| LSB | $[0]$ | 1 | Least Significant Bit. |

### 8.0 Power Consumption

Figure 8-1 depicts the power consumption of Ayama 10000 devices based on $80 \%$ Searches, $50 \%$ I/O switching, 10-pF output load, 1.5 V HSTLII VDDQ_ASIC/VDDQ_SRAM, and 1.2 V VDD. The power data is with all the blocks in the device active. A device that operates in Enhanced mode and utilizes Mini-Key may have lower power consumption depending on the configuration.


Figure 8-1. Typical Power Consumption of Ayama 10000
Note: These values were determined through our power estimation model. Please contact Cypress to get an application specific power estimation.

### 9.0 Electrical Specifications

## Maximum Ratings

(Above which the useful life may be impaired. For user guidelines, not tested.)
Storage Temperature: $-65^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$
Ambient Temperature with Power Applied: $-55^{\circ} \mathrm{C}$ to $+112^{\circ} \mathrm{C}\left(\mathrm{Q}_{\mathrm{JA}}=1.2^{\circ} \mathrm{C} / \mathrm{W}\right)$
Maximum Junction Temperature: $125^{\circ} \mathrm{C}$
Static Discharge Voltage (per JEDEC EIA./JESD22-A114A): >2000V
Latch-up Current: > 500 mA
Table 9-1. DC Electrical Characteristics for Ayama 10000

| Parameter | Description | Conditions | $\mathrm{V}_{\text {DDQ }}=1.5 \mathrm{~V}$ |  | $\mathrm{V}_{\text {DDQ }}=1.8 \mathrm{~V}$ |  | $\mathrm{V}_{\text {DDQ }}=2.5 \mathrm{~V}$ |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Max. | Min. | Max. | Min. | Max. |  |
| $\mathrm{I}_{\text {LI }}$ | Input leakage current | $\mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\mathrm{DDQ}}$ Max., $\mathrm{V}_{\text {IN }}=0$ to $\mathrm{V}_{\mathrm{DDQ}}$ Max. | -10 | 10 | -10 | 10 | -10 | 10 | $\mu \mathrm{A}$ |
| lo | Output leakage current ${ }^{[25]}$ | $V_{D D Q}=V_{D D Q}$ Max., $\mathrm{V}_{\mathrm{IN}}=0$ to $\mathrm{V}_{\mathrm{DDQ}}$ Max. | -10 | 10 | -10 | 10 | -10 | 10 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\mathrm{IL}}$ | Input LOW voltage ${ }^{[21]}$ |  | -0.3 | $\begin{gathered} \mathrm{V}_{\text {REF }}- \\ 0.1 \end{gathered}$ | -0.3 | $\begin{gathered} 0.35 \\ \mathrm{~V}_{\mathrm{DDQ}} \end{gathered}$ | -0.3 | 0.7 | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input HIGH voltage ${ }^{[20]}$ |  | $\underset{\mathrm{V}_{\mathrm{REF}}}{ }{ }^{+1}$ | $\begin{gathered} \mathrm{V}_{\mathrm{DDQ}}+ \\ 0.3 \end{gathered}$ | $\begin{gathered} 0.65 \\ \mathrm{~V}_{\mathrm{DDQ}} \end{gathered}$ | $\underset{0.3}{\mathrm{~V}_{\mathrm{DDQ}}+}$ | 1.7 | $\begin{gathered} \mathrm{V}_{\mathrm{DDQ}} \\ 0.3 \end{gathered}$ | V |
| $\mathrm{V}_{\mathrm{OL}}$ | Output LOW voltage | $\begin{aligned} & \mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\mathrm{DDQ}} \text { Min., } \\ & \mathrm{I}_{\mathrm{OL}}=2 \mathrm{~mA} \end{aligned}$ |  | 0.4 |  | 0.45 |  | 0.7 | V |
| $\mathrm{V}_{\mathrm{OH}}$ | Output HIGH voltage | $\begin{aligned} & \mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\mathrm{DDQ}} \mathrm{Min} ., \\ & \mathrm{I}_{\mathrm{OH}}=2 \mathrm{~mA} \end{aligned}$ | $\begin{gathered} \mathrm{V}_{\mathrm{DDQ}}- \\ 0.4 \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{DDQ}}- \\ 0.45 \end{gathered}$ |  | 1.7 |  | V |
| $\mathrm{I}_{\mathrm{CC}}$ | CYNSE10000 Operating Current | The operating current for NSE devices is highly application dependent, and can vary widely due to a number of system configurations. Please contact Cypress and provide system characteristics to receive application specific values. |  |  |  |  |  |  |  |


| Parameter | Description | Max. | Unit |
| :--- | :--- | :---: | :---: |
| $\mathrm{C}_{\text {IN }}{ }^{[22]}$ | Input capacitance | $6 / 12^{[24]}$ | pF |
| $\mathrm{C}_{\text {OUT }}{ }^{[23]}$ | Output capacitance | 6 | pF |

Table 9-2. Operating Conditions for Ayama 10000

| Parameter | Description | Min. | Typ. | Max. | Unit |
| :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{DDQ}}=2.5 \mathrm{~V}$ | Operating voltage for I/O (2.5V LVCMOS) | 2.3 | 2.5 | 2.7 | V |
| $\mathrm{~V}_{\mathrm{DDQ}}=1.8 \mathrm{~V}$ | Operating voltage for I/O (1.8V LVCMOS) | 1.65 | 1.8 | 1.95 | V |
| $\mathrm{~V}_{\mathrm{DDQ}}=1.5 \mathrm{~V}$ | Operating voltage for I/O (HSTLI/II) | 1.4 | 1.5 | 1.6 | V |
| $\mathrm{~V}_{\mathrm{REF}}$ | Reference voltage for I/O (HSTLI/II) | 0.68 | 0.75 | 0.9 | V |
| $\mathrm{~V}_{\mathrm{DD}}$ | Operating supply voltage | 1.14 | 1.2 | 1.26 | V |
| $\mathrm{~T}_{\mathrm{A}}$ | Ambient operating temperature (C) | 0 |  | 70 | ${ }^{\circ} \mathrm{C}$ |
|  | Ambient operating temperature (I) | -40 |  | 85 | ${ }^{\circ} \mathrm{C}$ |

## Notes:

20. Maximum allowable applies to overshoot only.
21. Minimum allowable applies to undershoot only.
22. $\mathrm{f}=1 \mathrm{MHz}, \mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$.
23. $f=1 \mathrm{MHz}, \mathrm{V}_{\mathrm{OUT}}=0 \mathrm{~V}$.
24. CMD bus signals has an input capacitance of $12 p F, V_{R E F} 30 p F$, and all others $6 p F$
25. Output leakage current does not cover cascade ( $\mathrm{LHO}, \mathrm{BHO}, \mathrm{FULO}$ ) signals because these are always driven and are not measurable

### 10.0 AC Timing Parameters, Waveforms and Test Conditions

### 10.1 AC Timing Parameters and Waveforms with CLK2X

Table 10-1. AC Timing Parameters with CLK2X

| Parameter | Description | $\begin{gathered} \text { Ayama 10000- } \\ 083 \end{gathered}$ |  | $\begin{gathered} \text { Ayama 10000- } \\ 100 \end{gathered}$ |  | $\begin{gathered} \text { Ayama } 10000- \\ 133 \end{gathered}$ |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | Max. | Min. | Max. | Min. | Max. |  |
| fCLOCK | CLK2X frequency | 100 | 166 | 100 | 200 | 100 | 266 | MHz |
| t LOCK | PLL lock time |  | 0.5 |  | 0.5 |  | 0.5 | ms |
| $\mathrm{t}_{\text {CKHI }}$ | CLK2X HIGH pulse ${ }^{[23]}$ | 2.4 |  | 2.0 |  | 1.5 |  | ns |
| $\mathrm{t}_{\text {CKLO }}$ | CLK2X LOW pulse ${ }^{[23]}$ | 2.4 |  | 2.0 |  | 1.5 |  | ns |
| $\mathrm{t}_{\text {ISCH }}$ | Input set-up time to CLK2X rising edge ${ }^{[23]}$ | 1.8 |  | 1.5 |  | 1.0 |  | ns |
| $\mathrm{t}_{\mathrm{IHCH}}$ | Input hold time to CLK2X rising edge ${ }^{[23]}$ | 0.6 |  | 0.5 |  | 0.3 |  | ns |
| $\mathrm{t}_{\mathrm{ICSCH}}$ | Cascaded input set-up time to CLK2X rising edge ${ }^{\text {[24, 27] }}$ |  |  |  |  |  |  |  |
| ticsch_HIT | LHI, BHI signals | 4.5 |  | 4 |  | 3.5 |  | ns |
| ticsch_FUL | FULI signals | 1.8 |  | 1.5 |  | 1 |  | ns |
| $\mathrm{t}_{\mathrm{ICHCH}}$ | Cascaded input hold time to CLK2X rising edge ${ }^{[24,27]}$ |  |  |  |  |  |  |  |
| $\mathrm{t}_{\text {ICHCH_HIT }}$ | LHI, BHI signals | 0 |  | 0 |  | 0 |  | ns |
| $\mathrm{t}_{\text {ICHCH_FUL }}$ | FULI signals | 0.8 |  | 0.7 |  | 0.5 |  | ns |
| $\mathrm{t}_{\text {CKHOV }}$ | Rising edge of CLK2X to cascade output valid ${ }^{[27}$, 28] |  |  |  |  |  |  |  |
| teKHOV_HIT | LHO, BHO signals |  | 3.9 |  | 3.4 |  | 2.5 | ns |
| $\mathrm{t}_{\text {CKHOV_FUL }}$ | FULO signals |  | 7 |  | 6.5 |  | 6 | ns |
| $\mathrm{t}_{\mathrm{CKCOH}}$ | Rising edge of CLK2X to cascade output invalid (output hold) ${ }^{[30]}$ |  |  |  |  |  |  |  |
| $\mathrm{t}_{\text {CKCOH_HIT }}$ | LHO, BHO signals | 0.75 |  | 0.75 |  | 0.75 |  | ns |
| $\mathrm{t}_{\text {CKCOH_FUL }}$ | FULO signals | 1.2 |  | 1.2 |  | 1 |  | ns |
| $\mathrm{t}_{\text {CKHOVFE }}$ | Rising edge of CLK2X to FULL signal valid (Enhanced mode) |  | 3.2 |  | 3 |  | 2.5 | ns |
| $\mathrm{t}_{\text {CKHOV_FNE }}$ | Rising edge of CLK2X to FULL signal valid (NonEnhanced mode) |  | 7 |  | 6.5 |  | 6 | ns |
| $\mathrm{t}_{\text {CKCOHFE }}$ | Rising edge of CLK2X to FULL invalid (Enhanced mode) ${ }^{[30]}$ | 0.75 |  | 0.75 |  | 0.75 |  | ns |
| tekCOH_FNE | Rising edge of CLK2X to FULL invalid (NonEnhanced mode) ${ }^{[30]}$ | 1.2 |  | 1.2 |  | 1 |  | ns |
| $\mathrm{t}_{\text {CKHDV }}$ | Rising edge of CLK2X to DQ valid ${ }^{[26]}$ |  | 3.5 |  | 3.0 |  | 2.5 | ns |
| $\mathrm{t}_{\text {CKHDZ }}$ | Rising edge of CLK2X to DQ High-Z ${ }^{[26,30]}$ | 0.5 | 1.8 | 0.5 | 1.8 | 0.5 | 1.8 | ns |
| $\mathrm{t}_{\text {CKHSV }}$ | Rising edge of CLK2X to SRAM bus valid ${ }^{[26]}$ |  | 3.5 |  | 3.0 |  | 2.5 | ns |
| $\mathrm{t}_{\text {CKHSHZ }}$ | Rising edge of CLK2X to SRAM bus High-Z ${ }^{[26,30]}$ | 0.5 | 1.8 | 0.5 | 1.8 | 0.5 | 1.8 | ns |
| $\mathrm{t}_{\text {CKHSLZ }}$ | Rising edge of CLK2X to SRAM bus Low-Z ${ }^{[26,30]}$ | 2.2 |  | 1.9 |  | 1.9 |  | ns |
| ${ }^{\text {toH }}$ | Rising edge of CLK2X to DQ or SRAM bus invalid (output hold) | 0.5 |  | 0.5 |  | 0.75 |  | ns |
| $\mathrm{t}_{\text {RSTL }}$ | Minimum LOW pulse width for RST_L | 100 |  | 100 |  | 100 |  | us |

## Notes:

26. Values are based on $50 \%$ signal levels.
27. Values are based on $50 \%$ signal levels and a $50 \% / 50 \%$ duty cycle of CLK1X/CLK2X
28. Based on an AC load of 6 pF
29. Cascade signals only transition on CLK2X Cycle A rising edge
30. Based on an AC load of 30 pF . This parameter is guaranteed by design and is not production tested.


Signal Group 1A: DQ, CMD
Signal Group 1B: CMDV
Signal Group 2: LHI, BHI, FULI
Signal Group 3: LHO, BHO, FULO, FULL (Non-Enhanced mode)
Signal Group 4: SADR, CE_L, OE_L, WE_L, ALE_L, SSF, SSV
Signal Group 5: DQ, ACK, EOT, PAR, MULTI_HIT, FULL (Enhanced mode)

Figure 10-1. AC Timing Wave Forms with CLK2X

### 10.2 AC Timing Parameters and Waveforms with CLK1X

Table 10-2. AC Timing Parameters with CLK1X

| Parameter | Description | $\begin{gathered} \text { Ayama 10000- } \\ 083 \end{gathered}$ |  | $\begin{gathered} \text { Ayama 10000- } \\ 100 \end{gathered}$ |  | $\begin{gathered} \text { Ayama } 10000- \\ 133 \end{gathered}$ |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | Max. | Min. | Max. | Min. | Max. |  |
| $\mathrm{f}_{\text {CLOCK }}$ | CLK1X frequency | 50 | 83 | 50 | 100 | 50 | 133 | MHz |
| t LOCK | PLL lock time |  | 0.5 |  | 0.5 |  | 0.5 | ms |
| $\mathrm{t}_{\text {CKHI }}$ | CLK1X HIGH pulse; worst-case duty cycle ${ }^{[23]}$ | 5.4 |  | 4.5 |  | 3.4 |  | ns |
| $\mathrm{t}_{\text {CKLO }}$ | CLK1X LOW pulse; worst-case duty cycle ${ }^{[23]}$ | 5.4 |  | 4.5 |  | 3.4 |  | ns |
| $\mathrm{t}_{\text {ISCH }}$ | Input set-up time to CLK1X edge ${ }^{[23]}$ | 1.8 |  | 1.5 |  | 1.0 |  | ns |
| $\mathrm{t}_{\mathrm{IHCH}}$ | Input hold time to CLK1X edge ${ }^{[23]}$ | 0.6 |  | 0.5 |  | 0.3 |  | ns |
| $\mathrm{t}_{\text {ICSCH }}$ | Cascaded input set-up time to CLK1X rising edge ${ }^{[24,27]}$ |  |  |  |  |  |  |  |
| tICSCH_HIT | LHI, BHI signals | 4.5 |  | 4 |  | 3.5 |  | ns |
| ticsCH_FUL | FULI signals | 1.8 |  | 1.5 |  | 1 |  | ns |
| $\mathrm{t}_{\mathrm{ICHCH}}$ | Cascaded input hold time to CLK1X rising edge ${ }^{[24,}$ 27] |  |  |  |  |  |  |  |
| tichCH_HIT | LHI, BHI signals | 0 |  | 0 |  | 0 |  | ns |
| $\mathrm{t}_{\text {ICHCH_FUL }}$ | FULI signals | 0.8 |  | 0.7 |  | 0.5 |  | ns |
| $\mathrm{t}_{\text {CKHOV }}$ | Edge of CLK1X to cascade output valid ${ }^{[27,28]}$ |  |  |  |  |  |  |  |
| $\mathrm{t}_{\text {CKHOV_HIT }}$ | LHO, BHO signals |  | 3.9 |  | 3.4 |  | 2.5 | ns |
| $\mathrm{t}_{\text {CKHOV_FUL }}$ | FULO signals |  | 7 |  | 6.5 |  | 6 | ns |
| $\mathrm{t}_{\mathrm{CKCOH}}$ | Edge of CLK1X to cascade output invalid (output hold) ${ }^{[30]}$ |  |  |  |  |  |  |  |
| ${ }^{\text {t }}$ CKCOH_HIT | LHO, BHO signals | 0.75 |  | 0.75 |  | 0.75 |  | ns |
| $\mathrm{t}_{\text {CKCOH_FUL }}$ | FULO signals | 1.2 |  | 1.2 |  | 1 |  | ns |
| $\mathrm{t}_{\text {CKHOVFE }}$ | Edge of CLK1X to FULL signal valid (Enhanced mode) |  | 3.2 |  | 3 |  | 2.5 | ns |
| tCKHOV_FNE | Rising edge of CLK1X to FULL signal valid (NonEnhanced mode) |  | 7 |  | 6.5 |  | 6 | ns |
| $\mathrm{t}_{\text {CKCOHFE }}$ | Edge of CLK1X to FULL invalid (Enhanced mode) ${ }^{[30]}$ | 0.75 |  | 0.75 |  | 0.75 |  | ns |
| t ${ }_{\text {CKCOH_FNE }}$ | Rising edge of CLK1X to FULL invalid (NonEnhanced mode) ${ }^{[30]}$ | 1.2 |  | 1.2 |  | 1 |  | ns |
| $\mathrm{t}_{\text {CKHDV }}$ | Rising edge of CLK1X to DQ valid ${ }^{[24]}$ |  | 3.5 |  | 3.0 |  | 2.5 | ns |
| $\mathrm{t}_{\text {CKHDZ }}$ | Rising edge of CLK1X to DQ High-Z ${ }^{[28]}$ | 0.5 | 1.8 | 0.5 | 1.8 | 0.5 | 1.8 | ns |
| $\mathrm{t}^{\text {CKHSV }}$ | Edge of CLK1X to SRAM bus valid. ${ }^{24]}$ |  | 3.5 |  | 3.0 |  | 2.5 | ns |
| $\mathrm{t}_{\text {CKHSHZ }}$ | Edge of CLK1X to SRAM bus high-Z. ${ }^{[28]}$ | 0.5 | 1.8 | 0.5 | 1.8 | 0.5 | 1.8 | ns |
| $\mathrm{t}_{\text {CKHSLZ }}$ | Edge of CLK1X to SRAM bus LOW-Z. ${ }^{[28]}$ | 1.9 |  | 1.9 |  | 1.9 |  | ns |
| $\mathrm{t}_{\mathrm{OH}}$ | Rising edge of CLK1X to DQ or SRAM bus invalid (output hold) | 0.5 |  | 0.5 |  | 0.75 |  | ns |
| $\mathrm{t}_{\text {RSTL }}$ | Minimum LOW pulse width for RST_L | 100 |  | 100 |  | 100 |  | us |

Table 10-3. JTAG Timing Parameters

| Parameter | Description | $\begin{gathered} \text { Ayama 10000- } \\ 083 \end{gathered}$ |  | $\begin{gathered} \text { Ayama 10000- } \\ 100 \end{gathered}$ |  | $\begin{gathered} \text { Ayama 10000- } \\ 133 \end{gathered}$ |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | Max. | Min. | Max. | Min. | Max. |  |
| $\mathrm{f}_{\text {JTAG }}$ | Maximum JTAG TAP Controller Frequency |  | 10 |  | 10 |  | 10 | MHz |
| ${ }^{\text {t }}$ TCYC | TCK Clock Cycle Time | 100 |  | 100 |  | 100 |  | ns |
| $\mathrm{t}_{\mathrm{IH}}$ | TCK Clock HIGH Time | 40 |  | 40 |  | 40 |  | ns |
| $\mathrm{t}_{\text {TL }}$ | TCK Clock LOW Time | 40 |  | 40 |  | 40 |  | ns |
| $\mathrm{t}_{\text {TMSS }}$ | TMS Set-up to TCK Clock Rise | 10 |  | 10 |  | 10 |  | ns |
| $\mathrm{t}_{\text {TMSH }}$ | TMS Hold After TCK Clock Rise | 10 |  | 10 |  | 10 |  | ns |
| $\mathrm{t}_{\text {TDIS }}$ | TDI Set-up to TCK Clock Rise | 10 |  | 10 |  | 10 |  | ns |
| $\mathrm{t}_{\text {TDIH }}$ | TDI Hold After TCK Clock Rise | 10 |  | 10 |  | 10 |  | ns |
| ${ }_{\text {tidov }}$ | TCK Clock LOW to TDO Valid |  | 10 |  | 10 |  | 10 | ns |
| $\mathrm{t}_{\text {TDOX }}$ | TCK Clock LOW to TDO Invalid |  | 10 |  | 10 |  | 10 | ns |



Signal Group 1A: DQ, CMD
Signal Group 1B: CMDV
Signal Group 2: LHI, BHI, FULI
Signal Group 3: LHO, BHO, FULO, FULL (Non-Enhanced)
Signal Group 4: SADR, CE_L, OE_L, WE_L, ALE_L, SSF, SSV
Signal Group 5: DQ, ACK, EOT, MULTI_HIT, PAR, FULL (Enhanced)

Figure 10-2. AC Timing Wave Forms with CLK1X

### 10.3 AC Test Conditions and Output Loads

The following test conditions are the equivalent of the actual tester measurement condition. The effect of transmission line is removed from the results.
10.3.1 LVCMOS 2.5V/1.8V

Table 10-4. 2.5V / 1.8V AC Table for LVCMOS Test Condition of Ayama 10000

| Conditions | Results |
| :--- | :--- |
| Input pulse levels | GND to $2.5 \mathrm{~V} / 1.8 \mathrm{~V}$ |
| Input rise and fall times measured for 2.5V LVCMOS at 0.25 V and 2.25 V | $\leq 1.2 \mathrm{~ns}$ (see Figure 10-3) |
| Input rise and fall times measured for 1.8 V LVCMOS at 0.18 V and 1.98 V | $\leq 1.2 \mathrm{~ns}$ (see Figure 10-3) |
| Input timing reference levels $(2.5 \mathrm{~V} / 1.8 \mathrm{~V})$ | $1.25 \mathrm{~V} / 0.9 \mathrm{~V}$ |
| Output reference levels $(2.5 \mathrm{~V} / 1.8 \mathrm{~V})$ | $1.25 \mathrm{~V} / 0.9 \mathrm{~V}$ |



Figure 10-3. LVCMOS I/O Input Waveform


Figure 10-4. Test Condition of 2.5V LVCMOS I/O Output Load Equivalent


For high-Z

Figure 10-5. Test Condition of 2.5V High-Z LVCMOS I/O Output Load Equivalent


Figure 10-6. Test Condition of 1.8 V High-Z LVCMOS I/O Output Load Equivalent

### 10.3.2 HSTL I/II

Table 10-5. 1.5V AC Table for HSTL Test Condition of Ayama 10000

| Conditions | Results |
| :--- | :--- |
| Input pulse levels | 0.25 to 1.25 V |
| Input rise and fall times measured at 20\% and 80\% of input pulse | Faster than 1 V/ns (see Figure 10-7) |
| Input timing reference levels | 0.75 V |
| Output reference levels | 0.75 V |



Figure 10-7. HSTL I/II I/O Input Waveform


Figure 10-8. Test Condition of HSTL I I/O Output Load Equivalent


Figure 10-9. Test Condition of HSTL II I/O Output Load Equivalent


Figure 10-10. Test Condition of HSTLI/II I/O High-Z Output Load Equivalent

### 11.0 Pin Assignment and Pinout Diagram

Figure 11-1 shows the pinout diagram and Table 11-1 lists the pins assignment for Ayama 10000.

|  | AF | AE | AD | A C | А ${ }^{\text {b }}$ | A A | Y | w | v | $\cup$ | T | R | P | N | m | $\llcorner$ | к | J | н | G | F | E | D | c | в | A |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | NC | vss | RST_L | vss | FuLL | FuLor 1 ] | FUL[6] | vDDQ_A | FUL[2] | Fuli 0$]$ | вНо[2] | vDDQ_A | вно[0] | BHI[1] | vdod_A | 냉0] | 내[6] | LH[2] | 내[0] | ${ }_{\text {ID }}(3)$ | [D[1] | ${ }^{1} \mathrm{D}[0]$ | TRST_L | тCK | TDI | NC | 1 |
| 2 | nc | vss | VDDQ_A | еот | АСк | Asicsel | fulol0] | ful[5] | ful[3] | vDdo_A | vss | вно[1] | MULTI_H17 | 8H1[2] | вні[0] | LHO[ 1 ] | LHH 44$]$ | LHI[3] | LHH[1] | 10[4] | ${ }_{1012}$ | vDDQ_ ${ }^{\text {d }}$ | tdo | тms | vss | Dal7 1] | 2 |
| 3 | Da[68] | DQI70] | vod | vod | vod | vDD | vod | nc | ful[4] | fuli [1] | vod | vDD | vDD | vDo | vDo | vDo | LHH15] | vDdo_A | Parerr_L | vod | vDo | vdo | vod | vod | Da[69] | voda_A | 3 |
| 4 | DO[66] | vdda_A | vod | vss | vss | vss | vss | vss | vss | vss | vod | vod | vDD | vod | vDD | vod | vss | vss | vss | vss | vss | vss | vss | vDD | Do[65] | Dal67] | 4 |
| 5 | DO[62] | DO[64] | vod | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | vod | DQ[61] | Da[63] | 5 |
| 6 | vodoa | DQ[60] | vod | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | vod | DQ[59] | vodo-A | 6 |
| 7 | DO[56] | DQ[58] | vod | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | vod | DQ[55] | DQ[57] | 7 |
| 8 | DQ[52] | DQ[54] | hsvrefo | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | hsVref 1 | vddo.a | Da[53] | 8 |
| 9 | Da[48] | DQ[50] | vodo_a | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | DO[49] | Do[47] | Da[51] | 9 |
| 10 | vodea | DQ[44] | DQ[46] | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | voda_a | DQ[45] | Da[43] | 10 |
| 11 | Do[40] | DQ[42] | vod | vod |  |  |  |  |  |  | vss | vss | vss | vss | vss | vss |  |  |  |  |  |  | vdo | vod | Do[39] | Do[41] | 11 |
| 12 | Da[ 36$]$ | DQ[38] | vod | vod |  |  |  |  |  |  | vss | vss | vss | vss | vss | vss |  |  |  |  |  |  | vdo | vod | vddo_A | Dal37] | 12 |
| 13 | DQ[34] | vodo_A | vod | vod |  |  |  |  |  |  | vss | vss | vss | vss | vss | vss |  |  |  |  |  |  | vod | vod | Do[33] | Da[35] | 13 |
| 14 | DQ 30$]$ | DQ[32] | vod | vod |  |  |  |  |  |  | vss | vss | vss | vss | vss | vss |  |  |  |  |  |  | vod | vod | Da[29] | do[31] | 14 |
| 15 | vodea | DQ[28] | vod | vod |  |  |  |  |  |  | vss | vss | vss | vss | vss | vss |  |  |  |  |  |  | vod | vod | DQ[27] | vodoa | 15 |
| 16 | DQ[24] | DQ[26] | vod | vod |  |  |  |  |  |  | vss | vss | vss | vss | vss | vss |  |  |  |  |  |  | vod | vod | DQ[23] | Da[25] | 16 |
| 17 | Da[2] | vdod_A | Da[20] | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | da[19] | vddo.a | Da[21] | 17 |
| 18 | Da[14] | Da[18] | Da[ 16] | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | da[13] | do[15] | Da[ 17$]$ | 18 |
| 19 | vodoa | DQ[12] | Par[0] | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | PAR[1] | Da[ 11$]$ | voda $A$ | 19 |
| 20 | DO[8] | DQ[ 10$]$ | vod | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | vod | Do[7] | Dol9] | 20 |
| 21 | DO[4] | DO[6] | vod | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | GH_SPEE | vod | vdod_A | DO[5] | 21 |
| 22 | Da[2] | vodoa | vod | vss |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | vss | vod | do[1] | Do[3] | 22 |
| 23 | ssv | Do\0] | vod | vss | vss | vss | vss | vss | vss | vss_pul | vod | vod | vod | vod | vod | vod | vss | vss | vss | vss | vss | vss | vss | vod | vss | vss | 23 |
| 24 | ssF | vdod_A | vod | vod | vod | vod | vod | SADR[24] | CE_L | oe_L | vDD_PL | vod | vod | vDD | vDD | vDD | SADR[ 13$]$ | SADR[11] | SADR[25] | vDD | vod | vDD | vod | vod | CFG_L | SRAmSEL | 24 |
| 25 | CMD[10] | vss | cmD[8] | cmD[6] | CMD[5] | сmD[3] | CMD[ ${ }^{1]}$ | cmov | voda_s | PHS_L | CடK_MOO | SADR[22] | SADR[21] | SADR[19] | voda_s | SADR [ 15 ] | vdol_s | SADR[ 12] | vDDQ_s | SADR[8] | SADR[6] | SADR[5] | SADR[3] | SADR[1] | vss | GH_SPEE | 25 |
| 26 | cmD[9] | nc | CMD[7] | VDD_A | CMD[4] | CMD[2] | CmD[0] | ALE_L | wE_L | K1x/ CLK2 | SADR[23] | vdoo_s | SADR[20] | SADR[ 18$]$ | SADR[ 17$]$ | SADR[16] | SADR[14] | SADR 10$]$ | SADR[9] | SAdR[7] | vodo_s | SADR[4] | SADR[2] | vdoa_s | SADR[0] | nc | 26 |
|  | AF | AE | AD | A | AB | A A | Y | w | $\checkmark$ | $\cup$ | T | R | p | N | M | L | к | J | H | ${ }^{6}$ | F | E | D | c | в | A |  |

Figure 11-1. Pinout Diagram (Top View)

Table 11-1. Pin Assignment

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A1 | NC | NO CONNECT | AA26 | CMD[2] | INPUT |
| A10 | DQ[43] | I/O | AA3 | $V_{D D}$ | 1.2 V |
| A11 | DQ[41] | I/O | AA4 | $\mathrm{V}_{\text {SS }}$ | GND |
| A12 | DQ[37] | I/O | AB1 | FULL | OUTPUT-T |
| A13 | DQ[35] | I/O | AB2 | ACK | OUTPUT-T |
| A14 | DQ[31] | I/O | AB23 | $\mathrm{V}_{\text {SS }}$ | GND |
| A15 | V ${ }_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | AB24 | $V_{D D}$ | 1.2 V |
| A16 | DQ[25] | I/O | AB25 | CMD[5] | INPUT |
| A17 | DQ[21] | I/O | AB26 | CMD[4] | INPUT |
| A18 | DQ[17] | I/O | AB3 | $V_{\text {DD }}$ | 1.2 V |
| A19 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | AB4 | $\mathrm{V}_{\text {SS }}$ | GND |
| A2 | DQ[71] | I/O | AC1 | $\mathrm{V}_{S S}$ | GND |
| A20 | DQ[09] | I/O | AC10 | $V_{S S}$ | GND |
| A21 | DQ[05] | I/O | AC11 | $V_{\text {DD }}$ | 1.2 V |
| A22 | DQ[03] | I/O | AC12 | $V_{\text {DD }}$ | 1.2 V |
| A23 | $\mathrm{V}_{\text {SS }}$ | GND | AC13 | $V_{\text {DD }}$ | 1.2 V |
| A24 | SRAMSEL | $\mathrm{V}_{\text {DDQ_SRAM }} / \mathrm{V}_{\text {SS }}$ | AC14 | $V_{\text {DD }}$ | 1.2 V |
| A25 | HIGH_SPEED1 | INPUT | AC15 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| A26 | NC | NO CONNECT | AC16 | $V_{\text {DD }}$ | 1.2 V |
| A3 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | AC17 | $V_{S S}$ | GND |
| A4 | DQ[67] | I/O | AC18 | $\mathrm{V}_{\text {SS }}$ | GND |
| A5 | DQ[63] | I/O | AC19 | $\mathrm{V}_{\mathrm{SS}}$ | GND |
| A6 | V ${ }_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | AC2 | EOT | OUTPUT-T |
| A7 | DQ[57] | I/O | AC20 | $\mathrm{V}_{\text {SS }}$ | GND |
| A8 | DQ[53] | I/O | AC21 | $V_{S S}$ | GND |
| A9 | DQ[51] | I/O | AC22 | $V_{S S}$ | GND |
| AA1 | FULO[1] | OUTPUT-T | AC23 | $\mathrm{V}_{\text {SS }}$ | GND |
| AA2 | ASICSEL | $\mathrm{V}_{\text {DDQ_ASIC }} / \mathrm{V}_{\text {SS }}$ | AC24 | $V_{D D}$ | 1.2 V |
| AA23 | $\mathrm{V}_{\text {SS }}$ | GND | AC25 | CMD[6] | INPUT |
| AA24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 | AC26 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V |
| AA25 | CMD[3] | INPUT | AC3 | $V_{\text {DD }}$ | 1.2 V |
| AC4 | $\mathrm{V}_{\text {SS }}$ | GND | AE10 | DQ[44] | I/O |
| AC5 | $V_{S S}$ | GND | AE11 | DQ[42] | I/O |
| AC6 | $\mathrm{V}_{\text {SS }}$ | GND | AE12 | DQ[38] | I/O |
| AC7 | $V_{S S}$ | GND | AE13 | V ${ }_{\text {DDQ_ASIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ |
| AC8 | $V_{S S}$ | GND | AE14 | DQ[32] | I/O |
| AC9 | $\mathrm{V}_{S S}$ | GND | AE15 | DQ[28] | I/O |
| AD1 | RST_L | INPUT | AE16 | DQ[26] | 1/O |
| AD10 | DQ[46] | I/O | AE17 | V ${ }_{\text {DDQ_ASIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ |
| AD11 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | AE18 | DQ[18] | I/O |
| AD12 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | AE19 | DQ[12] | I/O |
| AD13 | $V_{\text {DD }}$ | 1.2 V | AE2 | $\mathrm{V}_{\mathrm{SS}}$ | GND |
| AD14 | $V_{D D}$ | 1.2 V | AE20 | DQ[10] | I/O |

Table 11-1. Pin Assignment (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| AD15 | $V_{\text {DD }}$ | 1.2V | AE21 | DQ[06] | I/O |
| AD16 | $V_{D D}$ | 1.2 V | AE22 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V |
| AD17 | DQ[20] | I/O | AE23 | DQ[00] | I/O |
| AD18 | DQ[16] | I/O | AE24 | V ${ }_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V |
| AD19 | PAR[0] | I/O | AE25 | $\mathrm{V}_{\text {SS }}$ | GND |
| AD2 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ | AE26 | NC | NO CONNECT |
| AD20 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | AE3 | DQ[70] | I/O |
| AD21 | $V_{\text {DD }}$ | 1.2 V | AE4 | V ${ }_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V |
| AD22 | $V_{\text {DD }}$ | 1.2 V | AE5 | DQ[64] | I/O |
| AD23 | $V_{\text {DD }}$ | 1.2 V | AE6 | DQ[60] | I/O |
| AD24 | $V_{\text {DD }}$ | 1.2 V | AE7 | DQ[58] | I/O |
| AD25 | CMD[8] | INPUT | AE8 | DQ[54] | I/O |
| AD26 | CMD[7] | INPUT | AE9 | DQ[50] | I/O |
| AD3 | $V_{D D}$ | 1.2 V | AF1 | NC | NO CONNECT |
| AD4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | AF10 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V2.5V |
| AD5 | $V_{D D}$ | 1.2 V | AF11 | DQ[40] | I/O |
| AD6 | $V_{\text {DD }}$ | 1.2 V | AF12 | DQ[36] | I/O |
| AD7 | $V_{\text {DD }}$ | 1.2 V | AF13 | DQ[34] | I/O |
| AD8 | HSVREF0 | INPUT | AF14 | DQ[30] | I/O |
| AD9 | V ${ }_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | AF15 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} 2.5 \mathrm{~V}$ |
| AE1 | VSS | GND | AF16 | DQ[24] | I/O |
| AF17 | DQ[22] | I/O | B23 | $\mathrm{V}_{\text {SS }}$ | GND |
| AF18 | DQ[14] | 1/O | B24 | CFG_L | INPUT |
| AF19 | V ${ }_{\text {DDQ_ASIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ | B25 | $V_{S S}$ | GND |
| AF2 | NC | NO CONNECT | B26 | SADR[0] | OUTPUT-T |
| AF20 | DQ[08] | I/O | B3 | DQ[69] | I/O |
| AF21 | DQ[04] | I/O | B4 | DQ[65] | I/O |
| AF22 | DQ[02] | 1/O | B5 | DQ[61] | I/O |
| AF23 | SSV | OUTPUT-T | B6 | DQ[59] | I/O |
| AF24 | SSF | OUTPUT-T | B7 | DQ[55] | I/O |
| AF25 | CMD[10] | INPUT | B8 | $V_{\text {DDQ_AsIC }}$ | 1.5V/1.8V2.5V |
| AF26 | CMD[9] | INPUT | B9 | DQ[47] | I/O |
| AF3 | DQ[68] | I/O | C1 | TCK | INPUT |
| AF4 | DQ[66] | I/O | C10 | V ${ }_{\text {DDQ_ASIC }}$ | 1.5V/1.8V2.5V |
| AF5 | DQ[62] | 1/O | C11 | $V_{\text {DD }}$ | 1.2 V |
| AF6 | $V_{\text {DDQ_AsIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ | C12 | $V_{D D}$ | 1.2 V |
| AF7 | DQ[56] | I/O | C13 | $V_{\text {DD }}$ | 1.2 V |
| AF8 | DQ[52] | 1/0 | C14 | $V_{D D}$ | 1.2 V |
| AF9 | DQ[48] | I/O | C15 | $V_{D D}$ | 1.2 V |
| B1 | TDI | INPUT | C16 | $V_{D D}$ | 1.2 V |
| B10 | DQ[45] | I/O | C17 | DQ[19] | I/O |
| B11 | DQ[39] | I/O | C18 | DQ[13] | I/O |
| B12 | V ${ }_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | C19 | PAR[1] | I/O |

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## PRELIMINARY

Table 11-1. Pin Assignment (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| B13 | DQ[33] | I/O | C2 | TMS | INPUT |
| B14 | DQ[29] | I/O | C20 | $V_{\text {DD }}$ | 1.2 V |
| B15 | DQ[27] | I/O | C21 | $V_{\text {DD }}$ | 1.2 V |
| B16 | DQ[23] | I/O | C22 | $V_{\text {DD }}$ | 1.2 V |
| B17 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ | C23 | $V_{\text {DD }}$ | 1.2 V |
| B18 | DQ[15] | I/O | C24 | $V_{\text {DD }}$ | 1.2 V |
| B19 | DQ[11] | I/O | C25 | SADR[1] | OUTPUT-T |
| B2 | $\mathrm{V}_{\mathrm{SS}}$ | GND | C26 | $\mathrm{V}_{\text {DDQ_SRAM }}$ | 1.5V/1.8V/2.5V |
| B20 | DQ[7] | I/O | C3 | $V_{\text {DD }}$ | 1.2 V |
| B21 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ | C4 | $V_{\text {DD }}$ | 1.2 V |
| B22 | DQ[1] | I/O | C5 | $V_{\text {DD }}$ | 1.2 V |
| C6 | $V_{\text {DD }}$ | 1.2 V | E24 | $V_{\text {DD }}$ | 1.2 V |
| C7 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | E25 | SADR[5] | OUTPUT-T |
| C8 | HSVREF1 | INPUT | E26 | SADR[4] | OUTPUT-T |
| C9 | DQ[49] | I/O | E3 | $V_{D D}$ | 1.2 V |
| D1 | TRST_L | INPUT | E4 | $\mathrm{V}_{\text {SS }}$ | GND |
| D10 | $\mathrm{V}_{\text {SS }}$ | GND | F1 | ID[1] | INPUT |
| D11 | $V_{\text {DD }}$ | 1.2 V | F2 | ID[2] | INPUT |
| D12 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | F23 | $\mathrm{V}_{\text {SS }}$ | GND |
| D13 | $V_{D D}$ | 1.2 V | F24 | $V_{D D}$ | 1.2 V |
| D14 | $V_{\text {DD }}$ | 1.2 V | F25 | SADR[6] | OUTPUT-T |
| D15 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | F26 | $\mathrm{V}_{\text {DDQ_SRAM }}$ | 1.5V/1.8V/2.5V |
| D16 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | F3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| D17 | $\mathrm{V}_{\mathrm{SS}}$ | GND | F4 | $\mathrm{V}_{S S}$ | GND |
| D18 | $V_{\text {SS }}$ | GND | G1 | ID[3] | INPUT |
| D19 | $\mathrm{V}_{\text {SS }}$ | GND | G2 | ID[4] | INPUT |
| D2 | TDO | OUTPUT-T | G23 | $\mathrm{V}_{S S}$ | GND |
| D20 | $\mathrm{V}_{\text {SS }}$ | GND | G24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| D21 | HIGH_SPEED2 | INPUT | G25 | SADR[8] | OUTPUT-T |
| D22 | $V_{S S}$ | GND | G26 | SADR[7] | OUTPUT-T |
| D23 | $\mathrm{V}_{S S}$ | GND | G3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| D24 | $V_{\text {DD }}$ | 1.2V | G4 | $V_{S S}$ | GND |
| D25 | SADR[3] | OUTPUT-T | H1 | LHI[0] | INPUT |
| D26 | SADR[2] | OUTPUT-T | H2 | LHI[1] | INPUT |
| D3 | $V_{\text {DD }}$ | 1.2V | H23 | $\mathrm{V}_{\text {SS }}$ | GND |
| D4 | $V_{S S}$ | GND | H24 | SADR[25] ${ }^{[31]}$ | OUTPUT-T |
| D5 | $\mathrm{V}_{\text {SS }}$ | GND | H25 | $V_{\text {DDQ_SRAM }}$ | 1.5V/1.8V/2.5V |
| D6 | $V_{S S}$ | GND | H26 | SADR[9] | OUTPUT-T |
| D7 | $\mathrm{V}_{\mathrm{SS}}$ | GND | H3 | PARERR_L | OUTPUT-Open Drain |
| D8 | $\mathrm{V}_{\text {SS }}$ | GND | H4 | $\mathrm{V}_{\text {SS }}$ | GND |
| D9 | $\mathrm{V}_{\mathrm{SS}}$ | GND | J1 | LHI[2] | INPUT |
| E1 | ID[0] | INPUT | J2 | LHI[3] | INPUT |
| E2 | $\mathrm{V}_{\text {DDQ_JTAG }}$ | 2.5 V | J23 | $\mathrm{V}_{\text {SS }}$ | GND |

Table 11-1. Pin Assignment (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| E23 | $\mathrm{V}_{\text {SS }}$ | GND | J24 | SADR[11] | OUTPUT-T |
| J25 | SADR[12] | OUTPUT-T | M2 | BHI[0] | INPUT |
| J26 | SADR[10] | OUTPUT-T | M23 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| J3 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | M24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| J4 | $\mathrm{V}_{\text {SS }}$ | GND | M25 | $V_{\text {DDQ_SRAM }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ |
| K1 | LHI[6] | INPUT | M26 | SADR[17] | OUTPUT-T |
| K2 | LHI[4] | INPUT | M3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| K23 | $\mathrm{V}_{\mathrm{SS}}$ | GND | M4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2V |
| K24 | SADR[13] | OUTPUT-T | N1 | BHI[1] | INPUT |
| K25 | $V_{\text {DDQ_SRAM }}$ | 1.5V/1.8V/2.5V | N11 | $\mathrm{V}_{\text {SS }}$ | GND |
| K26 | SADR[14] | OUTPUT-T | N12 | $\mathrm{V}_{\text {SS }}$ | GND |
| K3 | LHI[5] | INPUT | N13 | $\mathrm{V}_{\mathrm{SS}}$ | GND |
| K4 | $\mathrm{V}_{\mathrm{SS}}$ | GND | N14 | $\mathrm{V}_{S S}$ | GND |
| L1 | LHO[0] | OUTPUT-T | N15 | $V_{S S}$ | GND |
| L11 | $\mathrm{V}_{\mathrm{SS}}$ | GND | N16 | $V_{S S}$ | GND |
| L12 | $\mathrm{V}_{S S}$ | GND | N2 | BHI[2] | INPUT |
| L13 | $V_{S S}$ | GND | N23 | $V_{\text {DD }}$ | 1.2 V |
| L14 | $V_{S S}$ | GND | N24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| L15 | $V_{S S}$ | GND | N25 | SADR[19] | OUTPUT-T |
| L16 | $\mathrm{V}_{\mathrm{SS}}$ | GND | N26 | SADR[18] | OUTPUT-T |
| L2 | LHO[1] | OUTPUT-T | N3 | $V_{\text {DD }}$ | 1.2 V |
| L23 | $V_{\text {DD }}$ | 1.2 V | N4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| L24 | $V_{\text {DD }}$ | 1.2 V | P1 | BHO[0] | OUTPUT-T |
| L25 | SADR[15] | OUTPUT-T | P11 | $\mathrm{V}_{\mathrm{SS}}$ | GND |
| L26 | SADR[16] | OUTPUT-T | P12 | $\mathrm{V}_{S S}$ | GND |
| L3 | $V_{\text {DD }}$ | 1.2 V | P13 | $\mathrm{V}_{\text {SS }}$ | GND |
| L4 | $V_{D D}$ | 1.2 V | P14 | $V_{S S}$ | GND |
| M1 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | P15 | $\mathrm{V}_{S S}$ | GND |
| M11 | $\mathrm{V}_{\text {SS }}$ | GND | P16 | $\mathrm{V}_{\mathrm{SS}}$ | GND |
| M12 | $V_{S S}$ | GND | P2 | MULTI_HIT | OUTPUT-T |
| M13 | $V_{S S}$ | GND | P23 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| M14 | $V_{S S}$ | GND | P24 | $V_{\text {DD }}$ | 1.2 V |
| M15 | $V_{S S}$ | GND | P25 | SADR[21] | OUTPUT-T |
| M16 | $\mathrm{V}_{\text {SS }}$ | GND | P26 | SADR[20] | OUTPUT-T |
| P3 | $V_{\text {DD }}$ | 1.2 V | U24 | OE_L | OUTPUT-T |
| P4 | $V_{\text {DD }}$ | 1.2 V | U25 | PHS_L | INPUT |
| R1 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V | U26 | CLK1X/CLK2X | INPUT |
| R11 | $V_{\text {SS }}$ | GND | U3 | FULI[1] | INPUT |
| R12 | $\mathrm{V}_{S S}$ | GND | U4 | $\mathrm{V}_{\text {SS }}$ | GND |
| R13 | $V_{S S}$ | GND | V1 | FULI[2] | INPUT |
| R14 | $V_{S S}$ | GND | V2 | FULI[3] | INPUT |
| R15 | $V_{S S}$ | GND | V23 | $\mathrm{V}_{\text {SS }}$ | GND |
| R16 | $\mathrm{V}_{\mathrm{SS}}$ | GND | V24 | CE_L | OUTPUT-T |

Table 11-1. Pin Assignment (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R2 | BHO[1] | OUTPUT-T | V25 | $V_{\text {DDQ }}$ SRAM | 1.5V/1.8V/2.5V |
| R23 | $V_{\text {DD }}$ | 1.2 V | V26 | WE_L | OUTPUT-T |
| R24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | V3 | FULI[4] | INPUT |
| R25 | SADR[22] | OUTPUT-T | V4 | $\mathrm{V}_{\text {SS }}$ | GND |
| R26 | $V_{\text {DDQ_SRAM }}$ | $1.5 \mathrm{~V} / 1.8 \mathrm{~V} / 2.5 \mathrm{~V}$ | W1 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V |
| R3 | $V_{\text {DD }}$ | 1.2 V | W2 | FULI[5] | INPUT |
| R4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | W23 | $\mathrm{V}_{\text {SS }}$ | GND |
| T1 | BHO[2] | OUTPUT-T | W24 | SADR[24] ${ }^{[32]}$ | OUTPUT-T |
| T11 | $\mathrm{V}_{\mathrm{SS}}$ | GND | W25 | CMDV | INPUT |
| T12 | $V_{S S}$ | GND | W26 | ALE_L | OUTPUT-T |
| T13 | $\mathrm{V}_{\text {SS }}$ | GND | W3 | NC | NO CONNECT |
| T14 | $V_{S S}$ | GND | W4 | $\mathrm{V}_{\mathrm{SS}}$ | GND |
| T15 | $\mathrm{V}_{\mathrm{SS}}$ | GND | Y1 | FULI[6] | INPUT |
| T16 | $\mathrm{V}_{\text {SS }}$ | GND | Y2 | FULO[0] | OUTPUT-T |
| T2 | $V_{S S}$ | GND | Y23 | $\mathrm{V}_{\mathrm{SS}}$ | GND |
| T23 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V | Y24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |
| T24 | $\mathrm{V}_{\text {DD_PLL }}$ | 1.2 V | Y25 | CMD[1] | INPUT |
| T25 | CLK_MODE | INPUT | Y26 | CMD[0] | INPUT |
| T26 | SADR[23] | OUTPUT-T | Y3 | $V_{\text {DD }}$ | 1.2V |
| T3 | $V_{\text {DD }}$ | 1.2 V | Y4 | $\mathrm{V}_{\text {SS }}$ | GND |
| T4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.2 V |  |  |  |
| U1 | FULI[0] | INPUT |  |  |  |
| U2 | $\mathrm{V}_{\text {DDQ_ASIC }}$ | 1.5V/1.8V/2.5V |  |  |  |
| U23 | $\mathrm{V}_{\text {SS_PLL }}$ | GND |  |  |  |

## Notes:

31. No-Connect in CYNSE10256 and CYNSE10128.
32. No-Connect in CYNSE10256.

### 12.0 Package Diagrams

## 388-Ball HFC-BGA FG388A



REFERENCE JEDEC MS-034

### 13.0 Ordering Information

Table 13-1 provides ordering information.
Table 13-1. Ordering Information

| Part Number | Description | I/O Voltage | Max. Freq. | Temp. Range |
| :---: | :---: | :---: | :---: | :---: |
| CYNSE10512-83FGC/I | 83-MSPS 512K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 83 MHz | Comm/Ind |
| CYNSE10512-100FGC | 100-MSPS 512K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 100 MHz | Comm |
| CYNSE10512-133FGC | 133-MSPS 512K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 133 MHz | Comm |
| CYNSE10256-83FGC/I | 83-MSPS 256K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 83 MHz | Comm/Ind |
| CYNSE10256-100FGC | 100-MSPS 256K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 100 MHz | Comm |
| CYNSE10256-133FGC | 133-MSPS 256K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 133 MHz | Comm |
| CYNSE10128-083FGC/I | 83-MSPS 128K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 83 MHz | Comm/Ind |
| CYNSE10128-100FGC | 100-MSPS 128K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 100 MHz | Comm |
| CYNSE10128-133FGC | 133-MSPS 128K x 36-bit Entries NSE | 1.5V/1.8V/2.5V | 133 MHz | Comm |

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## Document History Page

## Document Title: CYNSE10512/CYNSE10256/CYNSE10128 Ayama 10000 Network Search Engine

Document Number: 38-02069

| REV. | $\begin{aligned} & \hline \text { ECN } \\ & \text { NO. } \end{aligned}$ | Issue Date | Orig. of Change | Description of Change |
| :---: | :---: | :---: | :---: | :---: |
| ** | 119954 | 01/16/03 | BGT | New Data Sheet |
| *A | 123910 | 02/13/03 | KHS for ITL | Added the following information: <br> 32 CLK2X cycles wait after a write to COMMAND register <br> Enhanced mode Block configuration (in BMR) prior to entries initialization <br> Blocks initialization prior to accessing the blocks <br> Clarification on static signals voltage level <br> External pull-up resistor requirement for PARERR_L signal |
| *B | 126318 | 05/09/03 | KHS | Section 5.1.2.3: PARERR changed to Active-Low PARERR_L <br> Updated the timing of PARERR_L signal validity <br> Section 5.3.14: Updated SRR's PRIORITY field definition for when Search operation resulted in a miss <br> Section 5.12: Modified the statement on how to control TRST_L <br> Section 6.2: Fixed Learn Command B-cycle Bit[7] on Enhanced Mode for x288/x576 from ' X ' to ' 0 ' <br> Figure 6.3: Updated PARERR_L signal valid per Section 5.1.2.3 update <br> Figure 6-6 and 6-8: SSV to reflect idle cycles on 288-bit searches <br> Section 6.6: Note on mismatched entry width Learn operation <br> Figure 6-57, 6-58 and 6-59: Added ALE_L signal <br> Section 11 and Section 12: Added package orientations <br> Figure 11-1 and Table 11-1: Pin AA24 changed from VDDQ_SRAM to VDD |
| *C | 129255 | 10/08/03 | KHS | Table 4-1: Added CFG_L signal and description (signal is bonded out to B24) <br> Table 4-1: Corrected the DQ direction to I/O <br> Section 4.0, Note 4: Updated PARERR_L pull-up resistance recommendation <br> Table 4-1: Corrected PAR[1:0] direction to I/O <br> Table 4-1: Updated HIGH_SPEED1 and HIGH_SPEED2 description <br> Table 4-1: Corrected polarity mode of FULI[6:0] and FULO[1:0] <br> Section 5.1.2.3: Updated description of Parity feature <br> Figure 5-4; Figure 5-5: Added timing diagrams of DQ and Core Parity error <br> Section 5.1.2.4: Corrected description of CLK2x <br> Section 5.4.1: Clarify CMPR registers width <br> Section 5.4.2: Clarify GMR registers width <br> Table 5-5: Renamed register name from Status to Successful and updated INDEX field <br> Figure 5-11, Table 5-6: Removed READ bit. It is now RESERVED. <br> Table 5-8 and 5-9: Replaced ADR with INDEX <br> Table 5-12: Updated Hardware register fields <br> Table 5-13: Update ADR description <br> Table 5-17: Corrected Minikey3 range <br> Table 5-20: Clarify F3 field description <br> Table 5-26: Corrected INDIRECT description <br> Section 5.10.1: Corrected typo on $64 \mathrm{~K} \times 288$ <br> Section 6.2.1-3: Replaced SADR with EADR, added note <br> Section 6.3.2: Added to description of Burst Read <br> Table 6-3: Corrected typos <br> Section 6.4.2: Added to description of Burst Write <br> Section 6.4.3: Added to description of Parallel Write <br> Figure 6-15: Rewording on one of the notes <br> Section 6-6: Added explanation of blind Learn and Learn suppression. <br> Figure 6-56/57: Added timing diagrams for x72 and x288 Learn <br> Section 6.8: Added section on Timing Sequences for Back-to-Back Operations <br> Section 6.9: Added section on FULL signal: description and timing diagram <br> Section 9.0: Added Maximum Ratings data and rearranged the parameters <br> Table 10-1: Updated the descriptions of the timing parameters and notes <br> Figure 10-1: Added PHS_L and clarify boundaries of parameters <br> Figure 10-2: Clarify boundaries of parameters <br> Table 11-1: Added CFG L to Pin List and removed DQ[72] |
| *D | 205841 |  | XBM | Minor Change: Upload MPN to external website |


| Document Title: CYNSE10512/CYNSE10256/CYNSE10128 Ayama 10000 Network Search Engine Document Number: 38-02069 |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| REV. | $\begin{aligned} & \hline \text { ECN } \\ & \text { NO. } \end{aligned}$ | $\begin{gathered} \hline \text { Issue } \\ \text { Date } \end{gathered}$ | Orig. of Change | Description of Change |
| *E | 212292 | $\begin{aligned} & \text { See } \\ & \text { ECN } \end{aligned}$ | KHS | p10: 576-bit configuration is supported in Enhanced Mode only <br> p15-p17, Table 4-1: General signal description clarification <br> p27, Figure 5-9: Added addresses value of the GMR Registers <br> p28, Table5-4: GVAL is valid only in Enhanced Mode <br> p29, Table 5-5: Clarified operations that are affected by HLAT <br> p33, Table 5-11: Corrected locations of the IO Interfaces control bits and drive strengths <br> p41, Section 5.6: Clarified phase cycles of CLK2X <br> p42, Figure 5-27, 5-28, 5-29: Added cycle A and cycle B references to diagrams <br> p43, Table 5-24: Clarified operating speed clock references <br> p48, Section 5.11: Clarified description of indirect read <br> p50, Section 6.1: Clarified cycle A, cycle B descriptions relative to command encoding <br> p53, Figure 6-2: Changed EOT low cycle time <br> p95, Section 6.5.8: Removed redundant note on Multisearch <br> p113-114, Section 6.6: Clarified Learn description <br> p115, Figure 6-57: Corrected timing diagram to add a cycle between commands <br> p116, Figure 6-58: Added 576-bit Learn DQ timing diagram <br> p117, Figure 6-59: Added 576-bit Learn CMPR timing diagram <br> p121-p133, Various: Corrected SRAM operation timing diagram <br> p133, Section 6-8: Added examples of back-to-back operations <br> p133, Note 16: Clarified note description <br> p133, Note 19: New note on Learn operation <br> p136, Section 9.0: Modified storage temperature range and Latch-Up Current rating <br> p136-p137, Table 9-1 part 1: Expanded Standby and Operating current parameters <br> p137, Table 9-1 part 2: Added Load Capacitance for CMD bus <br> p137, Table 9-2: Added Commercial and Industrial ambient temperature ranges <br> p138, Table 10-1: Updated the following parameters: fCLOCK, tlHCH, tICHCH, tCKHDV, <br> tCKHSV, tCKHDZ, tCKHSHZ, tISCH <br> p139, Figure 10-1: Added FULL, PAR, MULTI_HIT <br> p140, Table 10-2: Updated the following parameters: fCLOCK, tlHCH, tICHCH, tCKHDV, tCKHSV, tCKHDZ, tCKHSHZ, tISCH <br> p141, Figure 10-2: Added FULL, PAR, MULTI_HIT <br> p142, Figure 10-4, 10-5, 10-6: Updated diagrams to best represent actual test condition <br> p143, Figure 10-8, 10-9, 10-10: Updated diagrams to best represent actual test condition <br> p149, Table 11-1: Added notes to identify pins that are not valid for smaller densities <br> Other: <br> Removed all references to cascade performance. |
| *F | 239580 | $\begin{aligned} & \hline \text { See } \\ & \text { ECN } \end{aligned}$ | DCU | p134, Figure 6-77: Added diagram for Non-Enhanced mode FULL, corrected TLSZ error p140, Figure 10-1: Added signal group for MultiSearch operation timing, clarified cycle relationship <br> p138, Table 10-1: Expanded CLK2X AC timing parameters table for cascade signals, added minimum pulse width for RST_L <br> p139, Table 10-2: Expanded CLK1X $\bar{A}$ C timing parameters table for cascade signals, added minimum pulse width for RST_L <br> p142, Figure 10-2: Added signal group for MultiSearch operation timing, clarified cycle relationship, removed reference to internal clock <br> p49, Figure 5-35: Clarified power-up sequence figure and description <br> p15, Table 4-1: Clarified cycle timing for device activation after RST_L <br> p141, Table 10-3: Added JTAG AC Timing Table <br> p137, Table 9-1: Removed table cells and replaced with a note <br> p55, Section 6.4.3: Clarified Parallel write description |

