

## SGM2324

# 1MHz, Quad, General Purpose CMOS Operational Amplifier

### PRODUCT DESCRIPTION

The SGM2324 has quad rail-to-rail output voltage feedback amplifiers in one package. It takes the minimum operating supply voltage down to 3V and the maximum recommended supply voltage is 5.5V. SGM2324 is specified over the extended -40°C to +85°C temperature range.

The amplifier in SGM2324 provides 1MHz bandwidth; very low input bias currents of 10pA, these features enable SGM2324 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-rail output feature is useful for designers to buffer ASIC in single-supply systems.

Applications of SGM2324 include safety monitoring, portable equipment, battery and power supply control, signal conditioning and interfacing for transducers in low power systems.

The SGM2324 is offered in SO-16, TSSOP-16, TSSOP-14 and SO-14 packages.

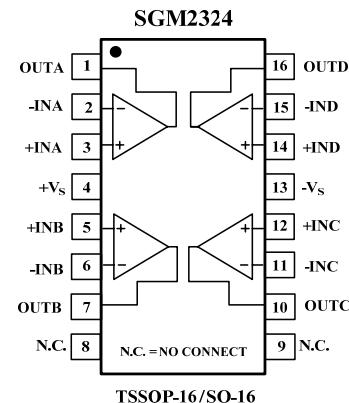
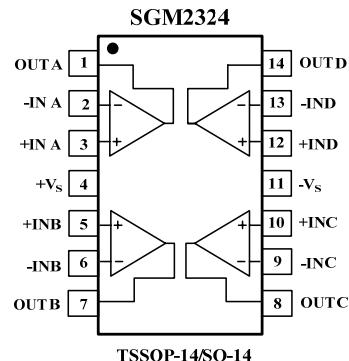
### APPLICATIONS

ASIC Input or Output Amplifier  
Sensor Interface  
Piezo Electric Transducer Amplifier  
Medical Instrumentation  
Mobile Communication  
Portable Systems  
Smoke Detectors  
Notebook PC  
PCMCIA Cards  
Battery-Powered Equipment  
DSP Interface

### FEATURES

- Low Cost
- Rail-to-Rail Output  
    1.7mV Typical V<sub>os</sub>
- Unity Gain Stable
- Gain Bandwidth Product: 1MHz
- Very Low Input Bias Currents: 10pA
- Input Common-Mode Voltage Range Includes Ground
- Operates from 3V to 5.5V
- Lead (Pb) Free Packages:  
    SO-16, TSSOP-16, SO-14 and TSSOP-14

### PIN CONFIGURATIONS (Top View)



SG Micro Ltd.  
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REV. B

# ELECTRICAL CHARACTERISTICS: $V_s = +5V$

(At  $R_L = 100\text{K}\Omega$  connected to  $V_s/2$ , and  $V_{OUT} = V_s/2$ , unless otherwise noted)

PARAMETER	CONDITIONS	SGM2324					
		TYP	MIN/MAX OVER TEMPERATURE				UNITS
		+25	+25	-40 to +85			
<b>INPUT CHARACTERISTICS</b>							
Input Offset Voltage ( $V_{OS}$ )		1.7	10	12	mV	MAX	
Input Bias Current ( $I_B$ )		10			pA	TYP	
Input Offset Current ( $I_{OS}$ )		10			pA	TYP	
Common-Mode Rejection Ratio(CMRR)	$V_s = 5V, V_{CM} = -0.1V$ to $3.3V$	88	65	50	dB	MIN	
Open-Loop Voltage Gain ( $A_{OL}$ )	$R_L = 2\text{K}\Omega, V_o = 0.1V$ to $4.9V$	100	85	80	dB	MIN	
	$R_L = 10\text{K}\Omega, V_o = 0.035V$ to $4.965V$	110	90	85	dB	MIN	
Input Offset Voltage Drift ( $\Delta V_{OS}/\Delta T$ )		3.5			$\mu\text{V}/^\circ\text{C}$	TYP	
<b>OUTPUT CHARACTERISTICS</b>							
Output Voltage Swing from Rail	$R_L = 2\text{K}\Omega$	0.8			V	TYP	
	$R_L = 10\text{K}\Omega$	0.008			V	TYP	
Output Current ( $I_{OUT}$ )		43	28	24	mA	MIN	
<b>POWER SUPPLY</b>							
Operating Voltage Range			3.0	3.0	V	MIN	
			5.5	5.5	V	MAX	
Power Supply Rejection Ratio (PSRR)	$V_s = +3V$ to $+5.5V$						
	$V_{CM} = (-V_s) + 0.5V$						
Quiescent Current / Amplifier ( $I_Q$ )	$I_{OUT} = 0$	80	75	70	dB	MIN	
		0.65	1.2	1.3	mA	MAX	
<b>DYNAMIC PERFORMANCE</b>							
Gain-Bandwidth Product (GBP)		1			MHz	TYP	
Slew Rate (SR)	$G = +1, 2V$ Output Step	0.65			V/ $\mu\text{s}$	TYP	
Settling Time to 0.1% ( $t_S$ )	$G = +1, 2V$ Output Step	9.0			$\mu\text{s}$	TYP	
Overload Recovery Time	$V_{IN} \cdot \text{Gain} = V_s$	4.0			$\mu\text{s}$	TYP	
Crosstalk	1kHz	-80			dB	TYP	
	1MHz	-65			dB	TYP	
<b>NOISE PERFORMANCE</b>							
Voltage Noise Density ( $e_n$ )	$f = 1\text{kHz}$	42.0			$\text{nV}/\sqrt{\text{Hz}}$	TYP	
	$f = 10\text{kHz}$	38.0			$\text{nV}/\sqrt{\text{Hz}}$	TYP	

Specifications subject to changes without notice.

## PACKAGE/ORDERING INFORMATION

MODEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
SGM2324	SGM2324YS/TR	SO-16	Tape and Reel, 2500	SGM2324YS
	SGM2324YTS/TR	TSSOP-16	Tape and Reel, 3000	SGM2324YTS
	SGM2324YS14/TR	SO-14	Tape and Reel, 2500	SGM2324YS14
	SGM2324YTS14/TR	TSSOP-14	Tape and Reel, 3000	SGM2324YTS14

### ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V+ to V- .....	6V
Storage Temperature Range .....	-65 to +150
Junction Temperature .....	160
Operating Temperature Range .....	-40 to +85
Package Thermal Resistance @ $T_A = 25$	
SO-16, $\theta_{JA}$ .....	82 /W
TSSOP-16, $\theta_{JA}$ .....	105 /W
Lead Temperature Range (Soldering 10 sec) .....	260
ESD Susceptibility	
HBM .....	4000V
MM .....	400V

### NOTES

1. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### CAUTION

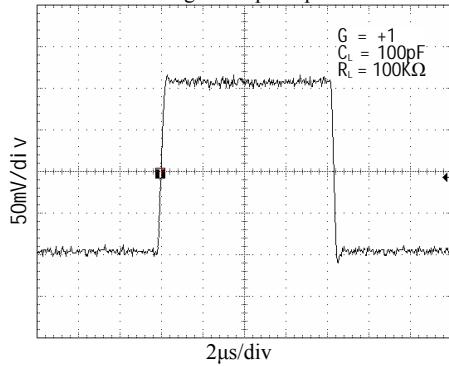
This integrated circuit can be damaged by ESD. SG Micro-electronics recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

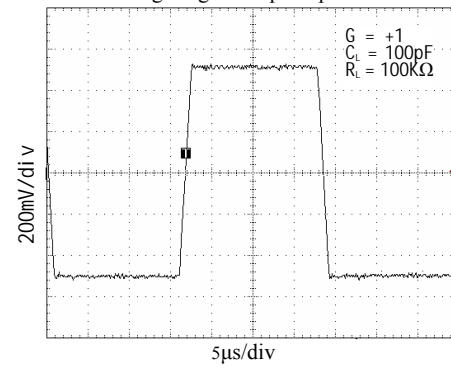
# TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ C$ ,  $V_S = +5V$ , and  $R_L = 100\text{K}\Omega$  connected to  $V_S/2$ , unless otherwise noted.

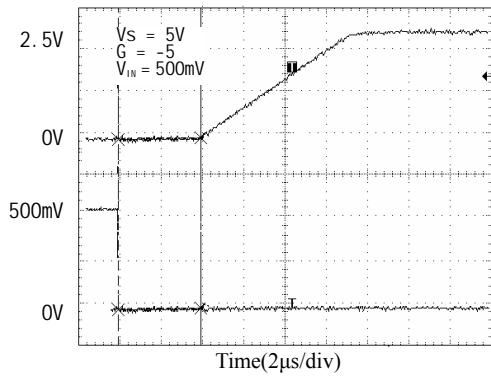
Small-Signal Step Response



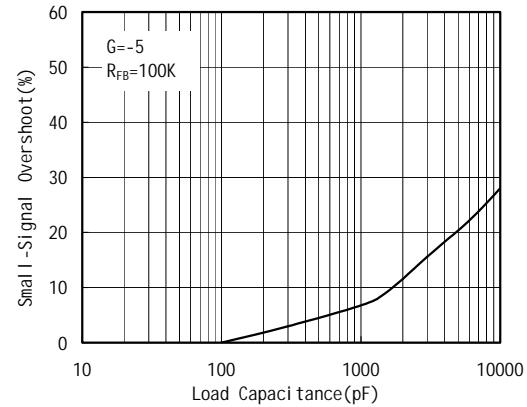
Large-Signal Step Response



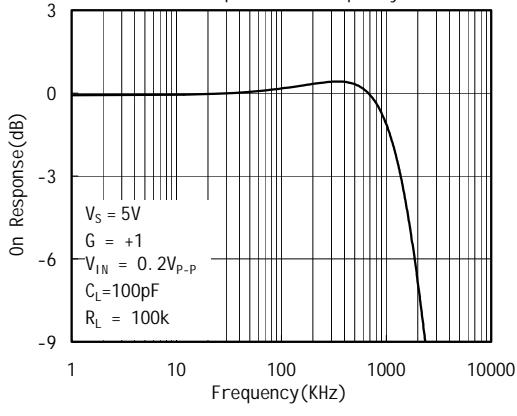
Overload Recovery Time



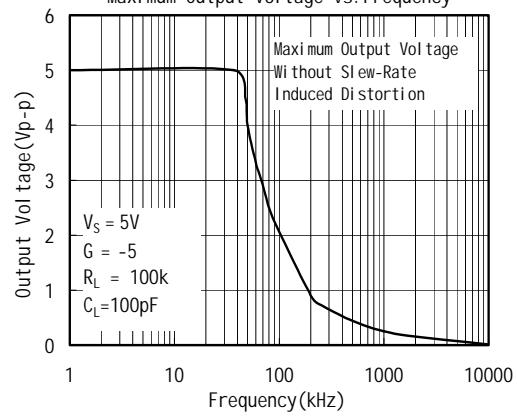
Small Signal Overshoot vs. Load Capacitance



On Response vs. Frequency

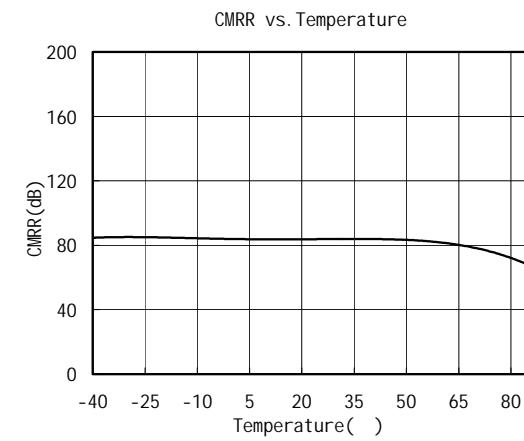
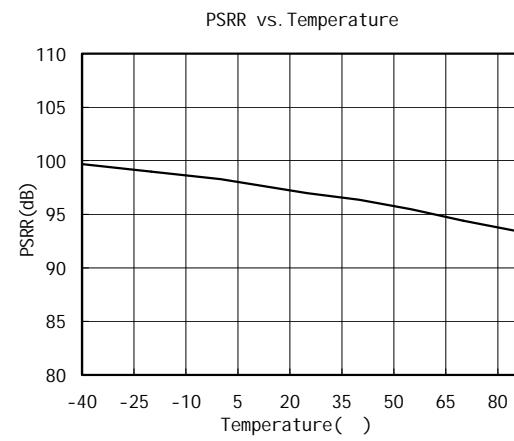
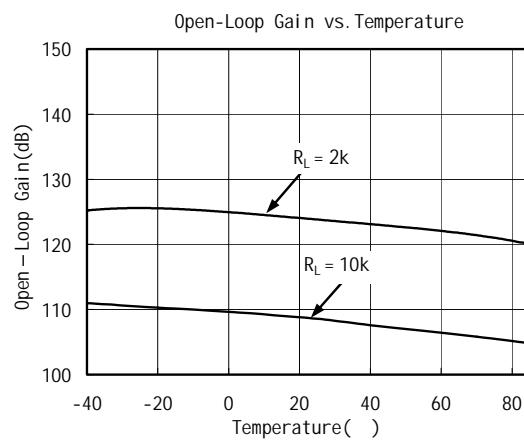
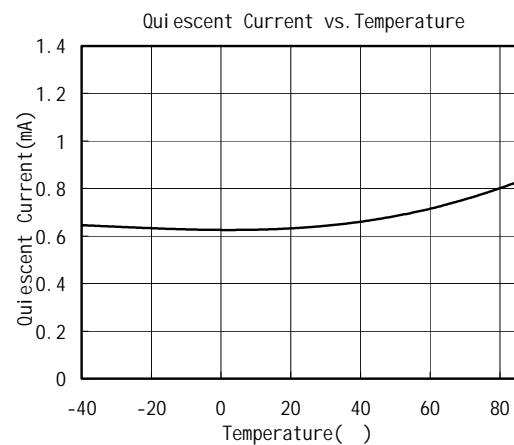
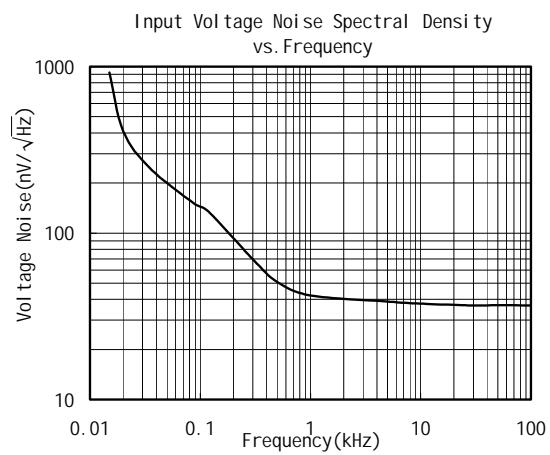
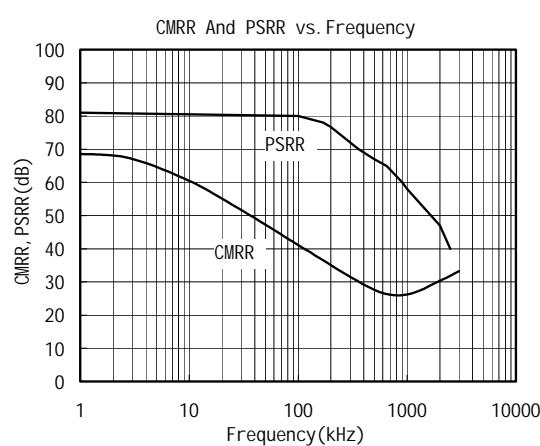


Maximum Output Voltage vs. Frequency



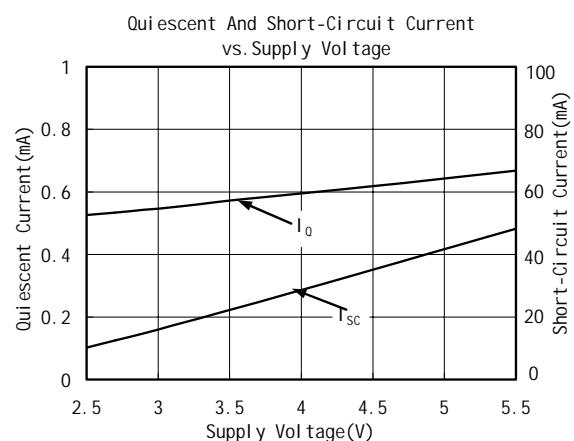
## TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 100\text{K}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



## TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 100\text{K}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



# APPLICATION NOTES

## Driving Capacitive Loads

The SGM2324 can directly drive 250pF in unity-gain without oscillation. The unity-gain follower (buffer) is the most sensitive configuration to capacitive loading. Direct capacitive loading reduces the phase margin of amplifiers and this results in ringing or even oscillation. Applications that require greater capacitive drive capability should use an isolation resistor between the output and the capacitive load like the circuit in Figure 1. The isolation resistor  $R_{ISO}$  and the load capacitor  $C_L$  form a zero to increase stability. The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. Note that this method results in a loss of gain accuracy because  $R_{ISO}$  forms a voltage divider with the  $R_{LOAD}$ .

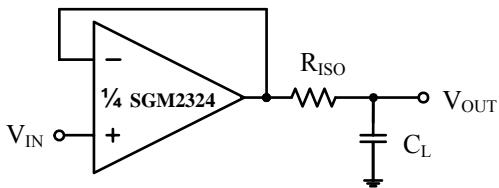


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability.  $R_F$  provides the DC accuracy by connecting the inverting signal with the output.  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

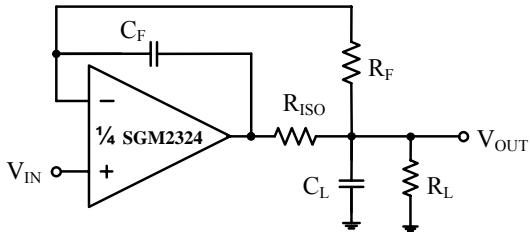


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For no-buffer configuration, there are two others ways to increase the phase margin: (a) by increasing the amplifier's gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node.

## Power-Supply Bypassing and Layout

The SGM2324 operates from a single +3V to +5.5V supply or dual ±1.5V to ±2.75V supplies. For single-supply operation, bypass the power supply  $V_{DD}$  with a 0.1µF ceramic capacitor which should be placed close to the  $V_{DD}$  pin. For dual-supply operation, both the  $V_{DD}$  and the  $V_{SS}$  supplies should be bypassed to ground with separate 0.1µF ceramic capacitors. 2.2µF tantalum capacitor can be added for better performance.

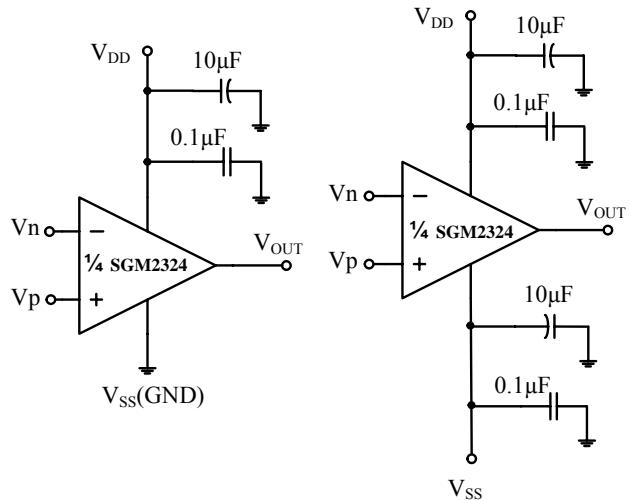


Figure 3. Amplifier with Bypass Capacitors

# Typical Application Circuits

## Differential Amplifier

The circuit shown in Figure 4 performs the difference function. If the resistors ratios are equal ( $R_4 / R_3 = R_2 / R_1$ ), then  $V_{OUT} = (V_p - V_n) \times R_2 / R_1 + V_{ref}$ .

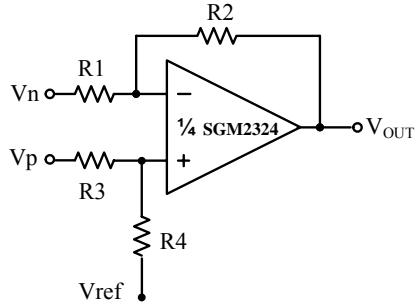


Figure 4. Differential Amplifier

## Low Pass Active Filter

The low pass filter shown in Figure 6 has a DC gain of  $(-R_2 / R_1)$  and the  $-3\text{dB}$  corner frequency is  $1/2\pi R_2 C$ . Make sure the filter is within the bandwidth of the amplifier. The Large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

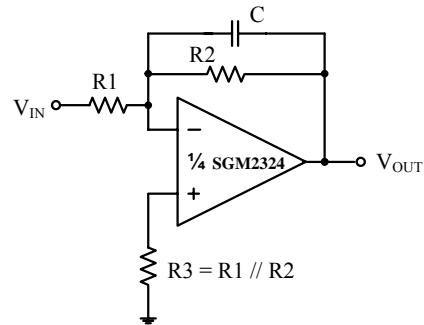


Figure 6. Low Pass Active Filter

## Instrumentation Amplifier

The circuit in Figure 5 performs the same function as that in Figure 4 but with the high input impedance.

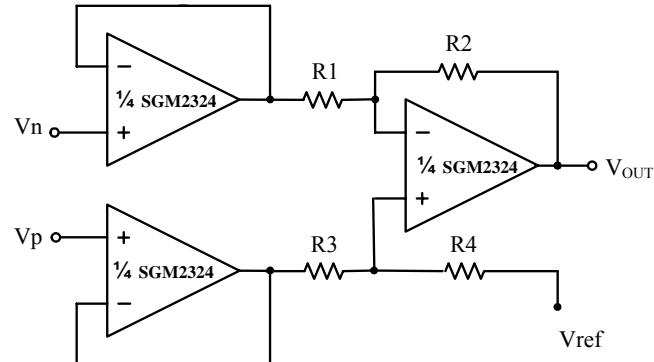
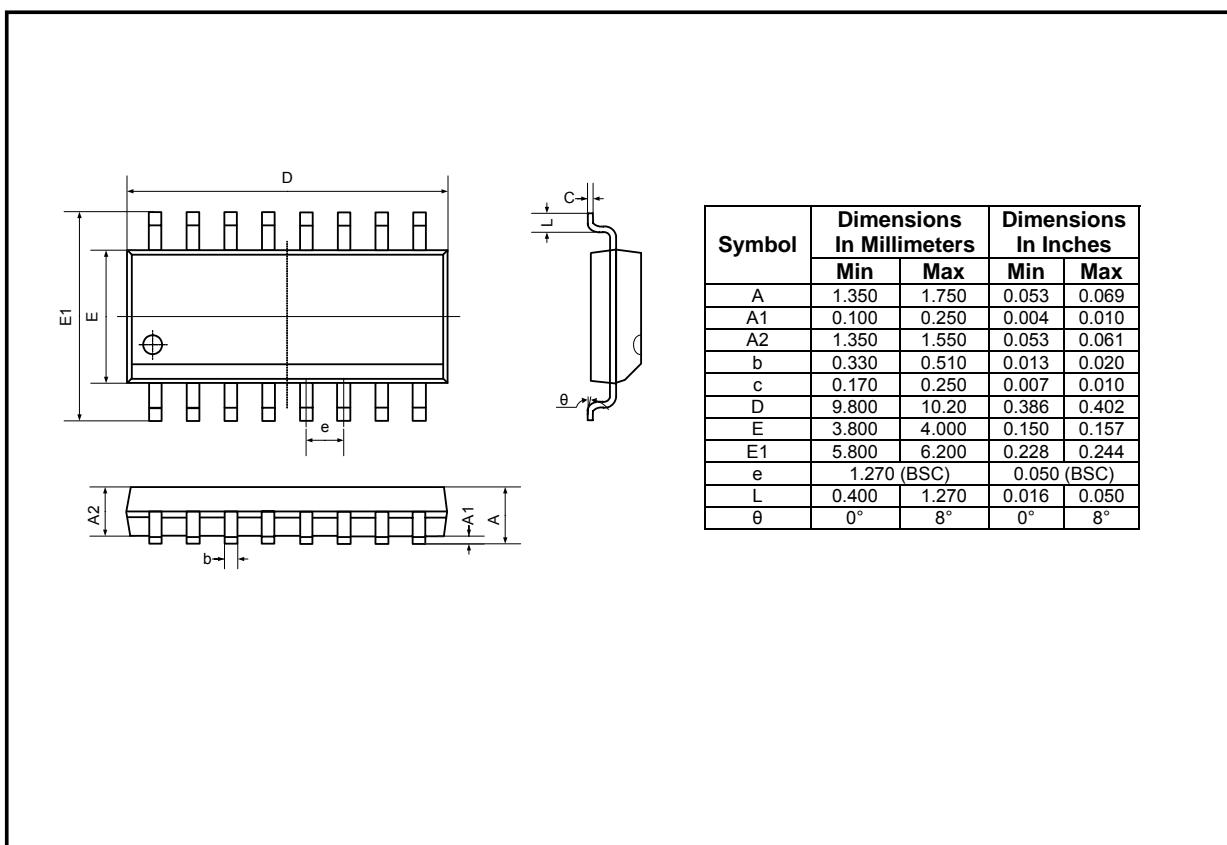


Figure 5. Instrumentation Amplifier

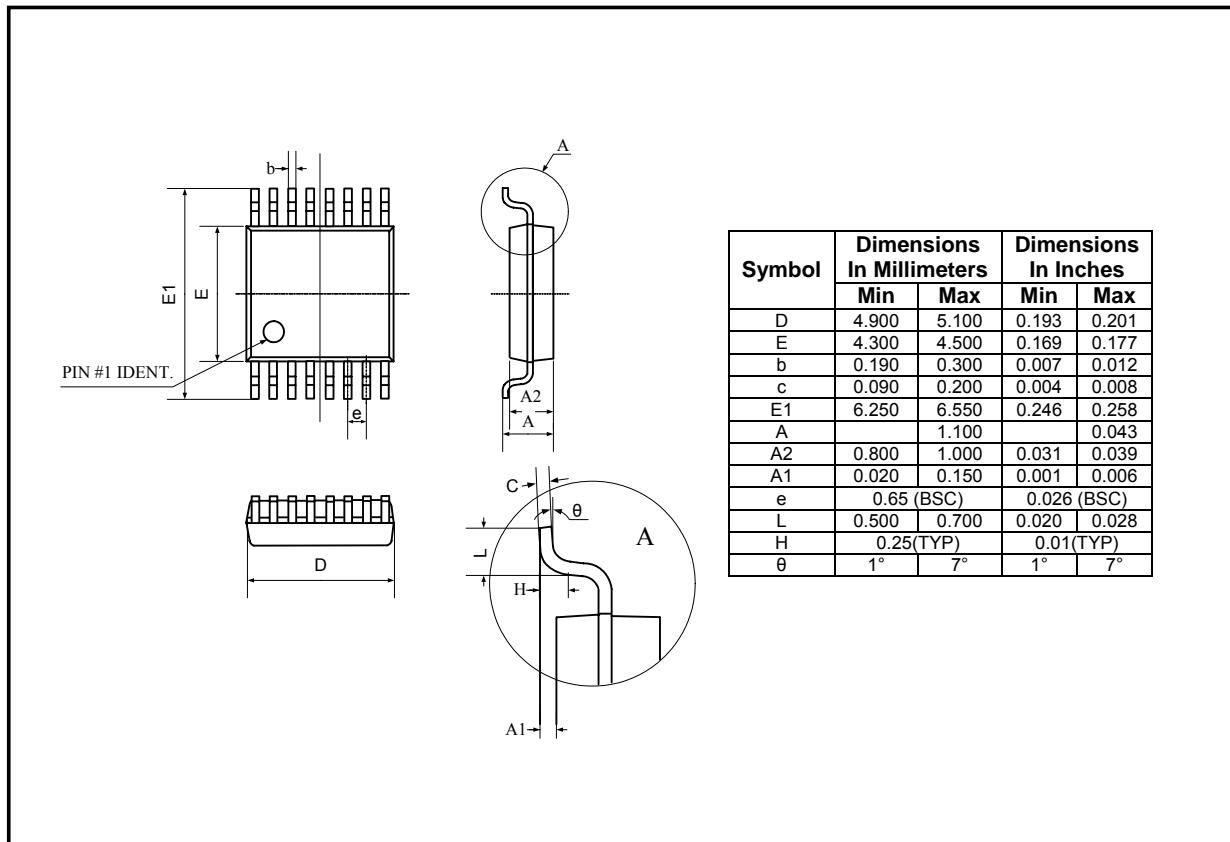
# PACKAGE OUTLINE DIMENSIONS

SO-16



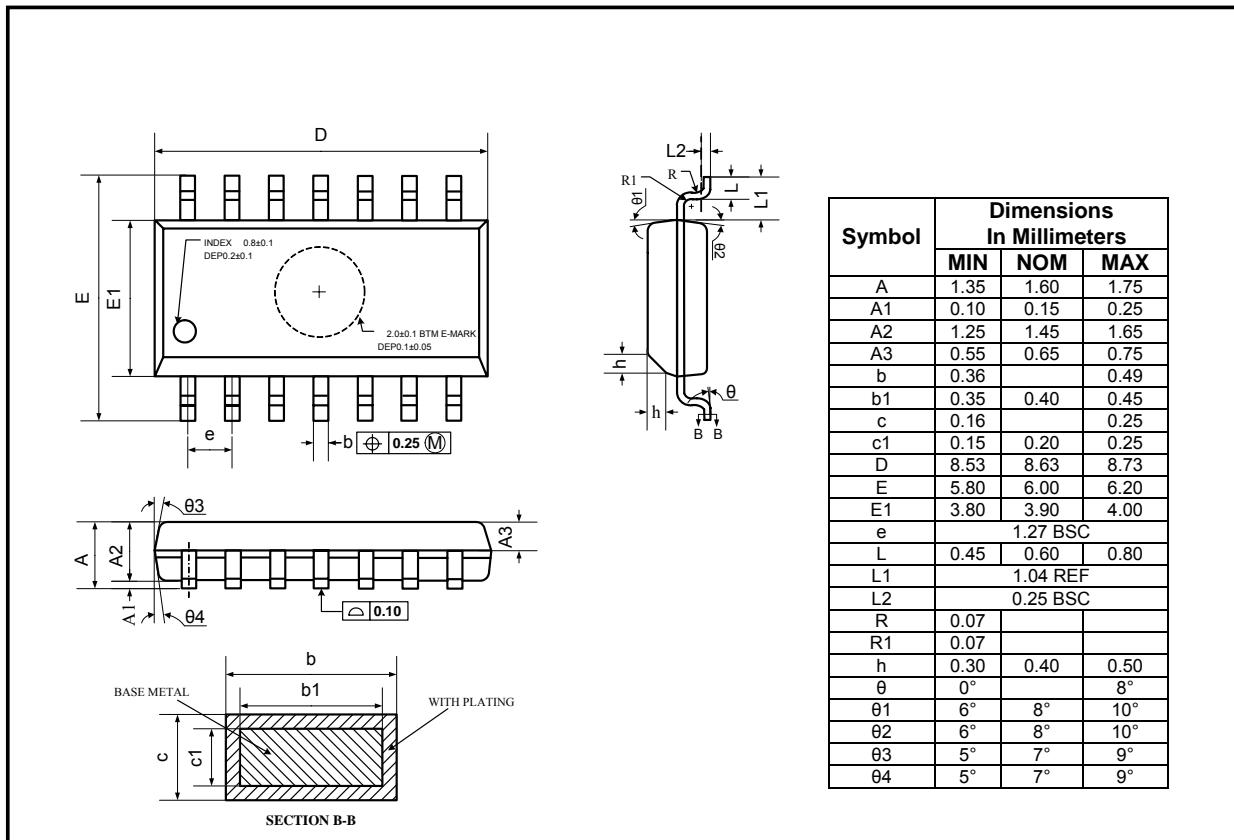
# PACKAGE OUTLINE DIMENSIONS

## TSSOP-16



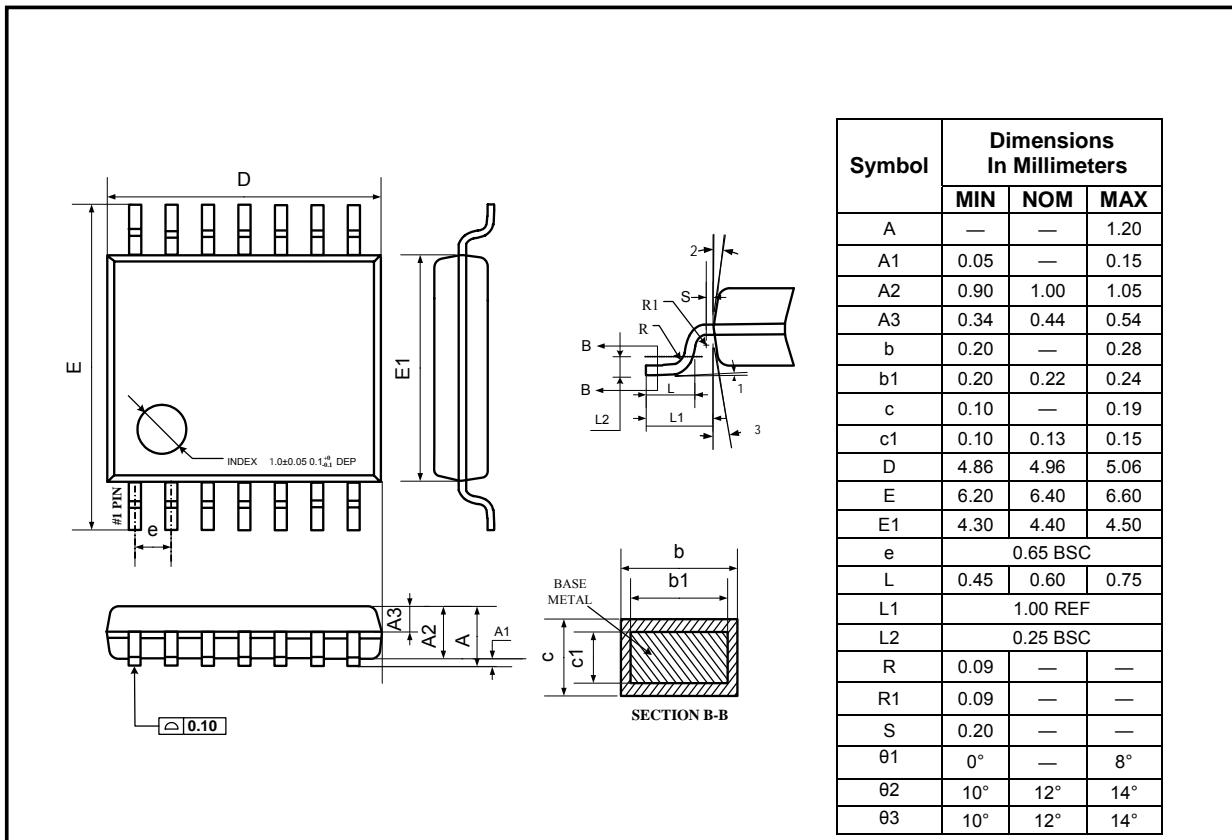
# PACKAGE OUTLINE DIMENSIONS

**SO-14**



# PACKAGE OUTLINE DIMENSIONS

## TSSOP-14



## REVISION HISTORY

Location	Page
<b>10/2007— Data Sheet REV.A</b>	
<b>02/2008— Data Sheet changed from REV. A to REV. B</b>	
Added SO-14 and TSSOP-14 Packages .....	1, 3, 11, 12

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