## 3-Channel 2-LD Driver for Optical Disc Drive

## Description

The CXA2640ER is a laser driver IC capable of driving two high output lasers (CD/DVD) for writeable optical discs.

## Features

- CD maximum drive current: 300 mA

DVD maximum drive current: 250 mA

- Capable of generating three-value write waveform through control of one read channel and two write channels
- Rise/Fall times $=1 \mathrm{~ns}$
- Read Channel: $\times 100$
- Write Channel: $\times 840$ (CD), $\times 400$ (DVD)
- Read channel has extensive low-noise design $1.5 \mathrm{nA} / \sqrt{\mathrm{Hz}}$ (@20MHz, ILD = 35mA, IMOD = 40mAp-p)
- Internal high frequency modulator circuit Frequency variable range: 200 to 600 MHz Maximum modulator current amplitude: $100 \mathrm{mAp}-\mathrm{p}$ Can be set separately for CD and DVD.
- Timing input for generating write waveform can be adapted to both differential input (LVDS/LVPECL) and single end input (3.3V CMOS/TTL).
- Single 5V power supply


## Applications

CD-R, CD-RW, DVD-R, DVD+RW, DVD-RW and DVD-RAM for high-speed writeable optical disc drives

## Structure

Bipolar silicon monolithic IC


Absolute Maximum Ratings ( $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| - Supply voltage | Vcc | 5.5 | V |
| :--- | :--- | :---: | :---: |
| - Storage temperature | Tstg | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |

## Operating Conditions

| - Supply voltage | Vcc | 4.5 to 5.5 | V |
| :--- | :--- | :---: | ---: |
| - Operating temperature | Topr | -10 to +75 | ${ }^{\circ} \mathrm{C}$ |

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Block Diagram and Pin Configuration


Pin Description

| Pin <br> No. | Symbol | I/O | Pin voltage |  | Equivalent circuit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | DC | AC |  |  |
| 1 | OUTEN2 | I | - | - |  | IIN2 set current control signal input. (positive logic) |
| 2 | xOUTEN2 | 1 |  |  |  | IIN2 set current control signal input. (negative logic) |
| 3 | OUTEN3 | I |  |  |  | IIN3 set current control signal input. (positive logic) |
| 4 | xOUTEN3 | I |  |  |  | IIN3 set current control signal input. (negative logic) |
| 5 | OUTENREF | 0 | 1.65 V | - |  | Reference voltage output for current control signal. Connects decoupling capacitance to ground. |
| 6 | VBG | O | 1.26 V | - |  | Internal reference voltage decoupling. |
| 7 | IIN1 | 1 | - | - |  | Current setting 1. <br> The set current $\times 100$ is output when LDEN1 or LDEN2 = high. |
| 8 | IIN2 | 1 | - | - |  | Current setting pin 2. The set current $\times 400$ is output through LDOUT1 when LDEN1 = high. The set current $\times 840$ is output through LDOUT2 when LDEN2 = high. |
| 9 | IIN3 | 1 |  |  |  | Current setting pin 3. The set current $\times 400$ is output through LDOUT1 when LDEN1 = high. The set current $\times 840$ is output through LDOUT2 when LDEN2 = high. |
| 10 | Vcc1 | I | - | - | - | Supply voltage for control system and modulator system. |
| 11 | Vcc2 | I | - | - | - | Supply voltage for timing system and current switch. |


| Pin No. | Symbol | I/O | Pin voltage |  | Equivalent circuit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | DC | AC |  |  |
| 12 | LDOUT2 | 0 | - | - |  | CD laser drive current output. Enabled when LDEN2 = high. |
| 13 | R FREQ1 | 0 |  |  | $300$ | Modulator frequency setting 1. Enabled when LDEN1 = high. Connects resistance to ground. |
| 14 | R FREQ2 | 0 |  |  | (14) | Modulator frequency setting 2. Enabled when LDEN2 = high. Connects resistance to ground. |
| 15 | R FREQ COMP | O | - | - |  | Modulator frequency variation adjustment. <br> Connects resistance to ground. |
| 16 | Vcc_LD | - | - | - | - | Output stage supply voltage. |
| 17 | RAMP1 | 0 |  |  |  | Modulator amplitude setting 1. Enabled when LDEN1 = high. Connects resistance to ground. |
| 18 | RAMP2 | 0 |  |  | (18) | Modulator amplitude setting 2. Enabled when LDEN2 = high. Connects resistance to ground. |
| 19 | LDOUT1 | 0 |  |  |  | DVD laser drive current output. <br> Enabled when LDEN1 = high. |
| 20 | GND2 | - | - | - | - | Ground. |


| Pin <br> No. | Symbol | I/O | Pin voltage |  | Equivalent circuit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | DC | AC |  |  |
| 21 | LDEN1 | 1 | - | - |  | DVD output control. (positive logic) <br> When LDEN1 = high, the current set at IIN1 is output through LDOUT1. |
| 22 | LDEN2 | 1 | - | - | (22) | CD output control. (positive logic) <br> When LDEN2 = high, the current set at IIN1 is output through LDOUT2. |
| 23 | OSCEN | 1 | - | - |  | Modulator control. (positive logic) <br> Outputs modulator waveform when OSCEN = high. |
| 24 | GND1 | - | - | - | - | Ground. |

## Electrical Characteristics

$\left(\mathrm{Vcc}=5 \mathrm{~V}, \mathrm{Ta}=25^{\circ} \mathrm{C}\right)$

| No. | Measurement item | Symbol | Min. | Typ. | Max. | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | Current consumption 1 | Icc1 | 10 | 16 | 22 | mA | LDEN1, 2 = L |
| 2 | Current consumption 2 | Icc2 | 86 | 123 | 160 | mA | $\begin{aligned} & \text { LDEN1 }(2)=\mathrm{H}, \\ & \text { IOUT1 }=60 \mathrm{~mA}, \text { OSCEN }=H, \\ & \text { AMP }=40 \mathrm{mAp}-\mathrm{p} \end{aligned}$ |
| 3 | Current consumption 3 | Icc3 | 128 | 183 | 238 | mA | LDEN $=\mathrm{H}$, IOUT1 $=60 \mathrm{~mA}$, <br> IOUT2 $=120 \mathrm{~mA}$ (Duty $=25 \%$ ), <br> IOUT3 $=60 \mathrm{~mA}$ (Duty $=50 \%$ ), <br> IOUT = IOUT1 + IOUT2 + <br> IOUT3 |
| 4 | Current consumption 3-1 | Icc3_1 | 100 | 144 | 188 | mA | LDEN $=\mathrm{H}$, IOUT1 $=30 \mathrm{~mA}$, IOUT2 $=120 \mathrm{~mA}$ (Duty $=25 \%$ ), IOUT3 $=60 \mathrm{~mA}$ (Duty $=50 \%$ ), IOUT = IOUT1 + IOUT2 + IOUT3 |

<Logic input block: During single end transfer>

| 5 | Input voltage high level | VSH | 2 | - | Vcc | V |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| 6 | Input voltage low level | VSL | GND | - | 1.3 | V |

<Logic input block: During differential input>

| 7 | Input voltage high level | VDH | 0.8 |  | 3 | V |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8 | Input voltage amplitude | VDL | 0.2 |  | 3 | V |  |
| <LD driver block: DC> |  |  |  |  |  |  |  |
| 9 | LD drive current 1 | IOUTR | 100 |  | - | mA |  |
| 10 | LD drive current 2, 3 | IOUTW | 250 |  | - | mA |  |
| 11 | Total LD drive current 1 (DVD) | IOUT1 | 250 | 300 | - | mA | $\mathrm{Vcc}=4.5 \mathrm{~V}, \mathrm{VOP}=3 \mathrm{~V}$ |
| 12 | Total LD drive current 2 (CD) | IOUT2 | 300 | 350 | - | mA | $\mathrm{Vcc}=4.5 \mathrm{~V}, \mathrm{VOP}=2.5 \mathrm{~V}$ |
| 13 | Minimum LD drive current 1 (DVD) | OFFSET1 | - | - | 4 | mA | $\mathrm{IIN} 1=\mathrm{IIN} 2=\mathrm{IIN} 3=0 \mu \mathrm{~A}$, LDEN1 = OUTEN2 = OUTEN3 = H |
| 14 | Minimum LD drive current 2 (CD) | OFFSET2 | - | - | 4 | mA | $\operatorname{IIN} 1=\operatorname{IIN} 2=\operatorname{IIN} 3=0 \mu \mathrm{~A}$, LDEN2 = OUTEN2 = OUTEN3 = H |
| 15 | Output current noise 1 (DVD) | NOISE1 | - | 1.5 | - | $n \mathrm{~A} / \sqrt{\mathrm{Hz}}$ | $\begin{aligned} & \mathrm{f}=400 \mathrm{MHz}, \mathrm{ILD}=35 \mathrm{~mA} \\ & \mathrm{Imod}=40 \mathrm{mAp}-\mathrm{p} \\ & (20 \mathrm{MHz}: \text { NOISE }) \end{aligned}$ |
| 16 | Output current noise 2 (CD) | NOISE2 | - | 1.5 | - | $\mathrm{nA} / \sqrt{\mathrm{Hz}}$ | $\begin{aligned} & \mathrm{f}=400 \mathrm{MHz}, \mathrm{ILD}=35 \mathrm{~mA}, \\ & \mathrm{Imod}=20 \mathrm{mAp}-\mathrm{p} \\ & (20 \mathrm{MHz}: \text { NOISE }) \end{aligned}$ |
| <LD driver block: Pulse driving> |  |  |  |  |  |  |  |
| 17 | Propagation delay | DELAY | - | 3 | - | ns |  |
| 18 | Rise time (Tr) | TR | - | 1.5 | - | ns | ILD $=50$ to 100 mA pulse Settling 10\% to 90\% (resistance load) |
| 19 | Fall time (Tf) | TF | - | 1.5 | - | ns | ILD = 100 to 50 mA pulse Settling 10\% to 90\% (resistance load) |


| No. | Measurement item | Symbol | Min. | Typ. | Max. | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| <ILD control block> |  |  |  |  |  |  |  |
| 20 | Input resistance 1 (Pin 7) | ZIINR | 0.56 | 0.8 | 1.04 | k $\Omega$ |  |
| 21 | Input resistance 2 (Pins 8, 9) | ZIINW | 1.05 | 1.5 | 1.95 | k $\Omega$ |  |
| 22 | Input/output gain 1 | GAINR | 95 | 105 | 115 | - |  |
| 23 | Input/output gain 2, 3 (DVD) | GAINW1 | 360 | 400 | 440 | - |  |
| 24 | Input/output gain 2, 3 (CD) | GAINW2 | 765 | 840 | 935 | - |  |
| 25 | ILD control linearity 1 (DVD) | LINEA1 | -3.5 | - | 2.5 | \% | Based on linearity when ILD $=50$ to 150 mA $\mathrm{Vcc}=4.5 \mathrm{~V}, \mathrm{VI}=1.75 \mathrm{~V},$ $\mathrm{RL}=5 \Omega, \mathrm{ILD}=250 \mathrm{~mA}$ |
| 26 | ILD control linearity 2 (CD) | LINEA2 | -3.5 | - | 2.5 | \% | Based on linearity when ILD $=50$ to 150 mA $\mathrm{Vcc}=4.5 \mathrm{~V}, \mathrm{~V} 2=1 \mathrm{~V}$, $\mathrm{RL}=5 \Omega$, ILD $=300 \mathrm{~mA}$ |
| 27 | Input/output gain relative precision | GACCU | -5 | - | 5 | \% |  |
| 28 | Input/output transmission band | FBAND | 7 | - | - | MHz | Frequency for input/output gain of -3 dB |
| <High frequency modulator> |  |  |  |  |  |  |  |
| 29 | Frequency variable range | VARIF | 200 | - | 600 | MHz |  |
| 30 | Amplitude variable range | VARIAMP | - | - | 100 | mA | $\mathrm{fmod}=400 \mathrm{MHz}$ |
| 31 | Frequency variation | FREQ | -10 | - | 10 | \% | $\mathrm{fmod}=400 \mathrm{MHz}$ |
| 32 | Frequency temperature characteristic | TFREQ | - | -116 | - | ppm $/{ }^{\circ} \mathrm{C}$ | $\mathrm{fmod}=300 \mathrm{MHz}$ |
| 33 | Amplitude variation | AMP | 0 | 31 | 42 | mAp-p | $\mathrm{fmod}=400 \mathrm{MHz}$, |
| 34 | Amplitude temperature characteristic | TAMP | - | -319 | - | ppm $/{ }^{\circ} \mathrm{C}$ | $\begin{aligned} & \text { fmod }=300 \mathrm{MHz}, \\ & \text { RAMP }=10 \mathrm{k} \Omega \end{aligned}$ |
| 35 | OSCEN response time (ON) | OSCRES1 | - | 5 | - | ns |  |
| 36 | OSCEN response time (OFF) | OSCRES2 | - | 5 | - | ns |  |
| <LDEN control> |  |  |  |  |  |  |  |
| 37 | LDEN response time 1 (ON) | RLDRES 1 | - | - | 700 | ns | Time to reach $90 \%$ of Read set current (same condition as current consumption 2) |
| 38 | LDEN response time 1 (OFF) | RLDRES2 | - | - | 10 | ns | Time to reach 10\% of Read set current (same condition as current consumption 2) |
| 39 | LDEN response time 2 (ON) | WLDRES1 | - | - | 700 | ns | Time to reach $90 \%$ of Write set current (same condition as current consumption 3) |
| 40 | LDEN response time 2 (OFF) | WLDRES2 | - | - | 10 | ns | Time to reach $10 \%$ of Write set current (same condition as current consumption 3) |

## Electrical Characteristics Measurement Circuit



## Description of Operation

## (1) LD Drive Current Value Setting

The current controlled by the current setting pins IIN1, IIN2 and IIN3 is output from the LDOUT1 and LDOUT2 pins. IIN1, IIN 2 and IIN3 can be set respectively by LDEN1, LDEN2, OUTEN and xOUTEN for the output drive current from the LDOUT pin.

## (2) Differential Input and Single-end Input

External processing is required for the differential input and single-end input switching.
For the single-end input, if the device is used at the active Low, the OUTENREF pin and the OUTEN pin should be shorted externally; if it is used at the active High, the OUTENREF pin and the xOUTEN pin should be shorted externally. Leave the OUTENREF pin open for the differential input.

## (3) Modulator Circuit

The modulator ON/OFF is controlled by the OSCEN pin.
For the DVD side, the modulator frequency is varied by the external resistor connected to the RFREQ1 pin and the modulator amplitude can be varied by the external resistor value connected to the RAMP1 pin.
For the CD side, the modulator frequency is varied by the external resistor connected to the RFREQ2 pin and the modulator amplitude can be varied by the external resistor value connected to the RAMP2 pin.

## (4) RFREQ COMP Pin

The current depending on the internal resistor is generated using the RFREQ COMP pin external resistor to suppress the dispersion of the modulator frequency depending on the internal resistor. The RFREQ COMP pin external resistor is recommended to be fixed to $22 \mathrm{k} \Omega$.

## (5) Modulator Level Adjustment

The modulator level adjustment can be performed by varying the IIN1 input current value.


## Description of Functions

## 1. Logic table

Output control

| LDEN1 | LDEN2 | xOUTEN2 | xOUTEN3 | OSCEN | LDOUT1 (DVD) | LDOUT2 (CD) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| L | L | X | X | X | OFF | OFF |
| H | L | H | H | L | $100 \times$ IIN1 | OFF |
| H | L | L | H | L | $100 \times \mathrm{IIN} 1+400 \times \mathrm{IIN} 2$ | OFF |
| H | L | H | L | L | $100 \times \mathrm{IIN} 1+400 \times \mathrm{IIN} 3$ | OFF |
| H | L | L | L | L | $\begin{aligned} & 100 \times \text { IIN } 1+400 \times \mathrm{IIN} 2 \\ & +400 \times \text { IIN3 } \end{aligned}$ | OFF |
| L | H | H | H | L | OFF | $100 \times$ IIN1 |
| L | H | L | H | L | OFF | $100 \times \mathrm{IIN} 1+840 \times \mathrm{IIN} 2$ |
| L | H | H | L | L | OFF | $100 \times \mathrm{IIN} 1+840 \times \mathrm{IIN} 3$ |
| L | H | L | L | L | OFF | $\begin{aligned} & 100 \times \mathrm{IIN} 1+840 \times \mathrm{IIN} 2 \\ & +800 \times \mathrm{IIN} 3 \end{aligned}$ |
| H | H | X | X | X | OFF (inhibit) | OFF (inhibit) |

Module control

| LDEN1 | LDEN2 | xOUTEN2 | xOUTEN3 | OSCEN | LDOUT1 | LDOUT2 |
| :---: | :---: | :---: | :---: | :---: | :--- | :--- |
| L | L | X | X | X | OFF | OFF |
| H | L | X | X | L | MODOFF | OFF |
| H | L | X | X | H | MODON <br> (Rfreq1, Ramp1) | OFF |
| L | H | X | X | L | OFF | MODOFF |
| L | H | X | X | H | OFF | MODON <br> (Rfreq2, Ramp2) |
| H | H | X | X | X | OFF (inhibit) | OFF (inhibit) |

Note: Module control does not depend on a data timing signals.
2. Timing Chart
(DVD side)

(CD side)


## Notes on Operation

- Arrange the external resistors connected to the IIN1, IIN2 and IIN3 pins near the IC package to reduce the affects from other signal lines.
- Wiring between the output LDOUT pin and the laser diode, and wiring between the Vcc_LD pin and external decoupling capacitors should be the shortest. Making the distance for wiring long increases output waveform overshoots and undershoots caused by the affect of wiring inductance.
- The Vcc_LD pin's external decoupling capacity ground can be grounded to the GND grounding the load from the LDOUT pin. This reverses the phase of the drive waveform at the LDOUT and Vcc_LD and moves in the direction that suppresses overshoots and undershoots.
- Temperature guarantee

Thermal resistance ( $\theta j-\mathrm{a}$ ) when the CXA2640ER is mounted on PWB varies according to the set (PWB) and because it is difficult to predict along with the tendency for higher power for power consumption (Po), the following points should be considered when using.
Use in a range that does not exceed a junction temperature of $150^{\circ} \mathrm{C}$. Also, power consumption (Po) should be below allowable power dissipation (Pd). Use with the thermal resistance ( $\theta \mathrm{j}-\mathrm{a}$ ) of the PWB mounting lowered so that it can be operated normally at a maximum operating temperature of $75^{\circ} \mathrm{C}$. To lower $\theta \mathrm{j}$-a, radiating measures on the set, such as widening the GND region with the set PWB are needed. Also, the diepad on the CXA2640ER 24-pin VQFN package is exposed on the backside, so thermal transmission from the IC backside to the PWB is excellent. For that reason, it is possible to release heat from the PBC to the set chassis thereby lowering the thermal resistance of the PWB mount.
Find the thermal resistance ( $\theta \mathrm{j}-\mathrm{a}$ ) when mounted on PWB and power consumption ( Po ) using the following method.

$$
\begin{aligned}
\text { Po }= & (\text { Icc } \times V \mathrm{Vcc})-(\text { lop } \times \text { Vop }) \\
& \text { Icc: IC current consumption when operating (Including Iop) } \\
& \text { Iop: Output drive current flowed from the LDOUT pin to the Laser Diode } \\
& \text { Vop: Operating voltage of the laser diode }
\end{aligned}
$$

It is also possible for Po when a modulator is $\mathrm{ON}(\operatorname{lmod}=40 \mathrm{mAp}-\mathrm{p})$, although the precision will decrease.
On the DVD side: $\mathrm{Po}=(50 \mathrm{~mA}+\mathrm{IIN} 1 \times 2.6+(\mathrm{IIN} 2+\mathrm{IIN} 3) \times 10) \times \mathrm{Vcc}+(\mathrm{Iop} \times(\mathrm{Vcc}-\mathrm{Vop}))$
On the CD side: $P o=(50 \mathrm{~mA}+\mathrm{IIN} 1 \times 2.6+(\mathrm{IIN} 2+\mathrm{IIN} 3) \times 21) \times \mathrm{Vcc}+(\mathrm{Iop} \times(\mathrm{Vcc}-\mathrm{Vop}))$

## Thermal resistance ( $\theta \mathrm{j}-\mathrm{a}$ ) when mounted on PWB

- The thermal resistance $(\theta \mathrm{c}-\mathrm{a})$ is obtained by measuring the Package surface temperature using a thermo couple or a radiation thermometer.
In order to improve the precision of measurement, it is desired to calculate by the following formula.
$\Delta$ Package surface temperature when lop is variable $/ \Delta \mathrm{Po}$
Assume the thermal resistance $(\theta \mathrm{j}-\mathrm{c})$ to be approximately $2^{\circ} \mathrm{C} / \mathrm{W}$.
- Diode thermal coefficient $-2.27 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ and the positive protection diode thermal characteristics are used to find this.
The V2 voltage found in (2) below cancels the voltage decrease caused by the wiring resistance between the positive protection diode connection Vcc and the Vcc pins as reference and is measured to find the precise temperature characteristics of the positive protection diode.
(1) V1 to OSCEN pin voltage to Vcc pin voltage, Icc1 to current consumption when OV is applied to the IIN1, IIN2 and IIN3 pins.
(2) V 2 to OSCEN pin voltage to Vcc pin voltage immediately after applying the arbitrary voltage to the IINx pin.
(3) V3 to OSCEN pin voltage to Vcc pin voltage, Icc3 to consumption current when applying the arbitrary voltage to the IINx pin and heat reaches equilibrium.
$\Delta T j$ using the voltage drop ( V 1 to V 2 ) between the positive protection diode connection Vcc and the Vcc pins that are the reference, as described above are:

$$
\Delta \mathrm{Tj}=((\mathrm{V} 3+(\mathrm{V} 1-\mathrm{V} 2))-\mathrm{V} 1) /-2.27 \mathrm{mV} /{ }^{\circ} \mathrm{C}
$$

Thermal resistance $(\theta j-a)$ is:

$$
\theta \mathrm{j}-\mathrm{a}=\Delta \mathrm{Tj} /((\operatorname{lcc} 3-\mathrm{Icc} 1) \times \mathrm{Vcc}-\mathrm{Iop} \times \mathrm{Vop})
$$

- Allowable power dissipation (PD) $\geq \mathrm{Po}[\mathrm{W}]$
$\mathrm{PD}=\left(150^{\circ} \mathrm{C}-\right.$ Ambient temperature $) / \theta \mathrm{j}-\mathrm{a}$
- Maximum operating temperature
$\left(150^{\circ} \mathrm{C}-\Delta \mathrm{Tj}\right) \geq 75^{\circ} \mathrm{C}$



## Example of Representative Characteristics



Modulator frequency control characteristics
RAMP $=3.3 \mathrm{k} \Omega$ (approximately $40 \mathrm{mAp}-\mathrm{p}$ ) RFREQ_COMP $=22 k \Omega$, IIN $1=642 \mu A$


IIN2/3 input current vs. CD/DVD output current characteristics
$\mathrm{Vcc}=5 \mathrm{~V}$, resistance load


RAMP resistance value vs. modulator waveform peak current characteristics

RFREQ $=12 \mathrm{k} \Omega$ (approximately 390 MHz ) RFREQ_COMP $=22 \mathrm{k} \Omega$, IIN1 $=642 \mu \mathrm{~A}$


## Application circuit 1



Application circuits shown are typical examples illustrating the operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits or for any infringement of third party patent and other right due to same.

## Application circuit 2



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## 24PIN VQFN(PLASTIC)



PACKAGE STRUCTURE

| SONY CODE | VQFN-24P-04 |
| :--- | :---: |
| EIAJ CODE | - |
| JEDEC CODE | - |


| PACKAGE MATERIAL | EPOXY RESIN |
| :--- | :--- |
| LEAD TREATMENT | SOLDER PLATING |
| LEAD MATERIAL | COPPER ALLOY |
| PACKAGE MASS | 0.04 g |

LEAD PLATING SPECIFICATIONS

| ITEM | SPEC. |
| :--- | :--- |
| LEAD MATERIAL | COPPER ALLOY |
| SOLDER COMPOSITION | Sn-Bi Bi:1-4wt\% |
| PLATING THICKNESS | $5-18 \mu \mathrm{~m}$ |

