

# LC72136N, 72136NM

# PLL Frequency Synthesizer for Electronic Tuning



# **Overview**

The LC72136N and LC72136NM are PLL frequency synthesizers for use in radio/cassette players. They allow high-performance AM/FM tuners to be implemented easily.

# **Features**

- High-speed programmable frequency divider
  - FMIN: 10 to 160 MHz.....Pulse swallower (divide-by-two prescaler built in)
  - AMIN: 2 to 40 MHz ......Pulse swallower
     0.5 to 10 MHz .....Direct division
- IF counter IFIN: 0.4 to 12 MHz.....For use as an AM/FM IF

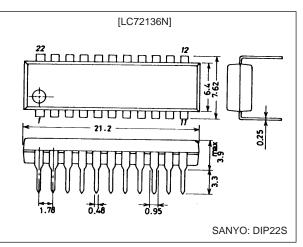
counter

- Reference frequency
  - Selectable from one of eight frequencies (crystal oscillator: 75 kHz)
    - 1, 3, 5, 3.125, 6.25, 12.5, 15, and 25 kHz
- · Phase comparator
  - Supports dead zone control
  - Built-in unlock detection circuit
  - Built-in deadlock clear circuit
- Built-in MOS transistor for forming an active low-pass filter
- I/O ports
  - Dedicated output ports: 6
  - I/O ports: 2
  - Supports clock time base output
- Serial Data I/O
  - Supports CCB format communication with the system controller.
- Operating ranges
  - Supply voltage: 4.5 to 5.5 V
  - Operating temperature: -20 to +70°C
- Packages
  - —DIP22S/MFP24S
  - CCB is a trademark of SANYO ELECTRIC CO., LTD.
  - CCB is SANYO's original bus format and all the bus addresses are controlled by SANYO.

# **Package Dimensions**

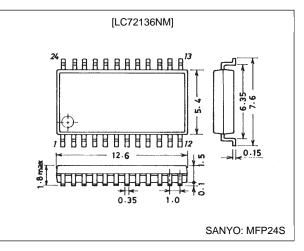
unit: mm

3059-DIP22S



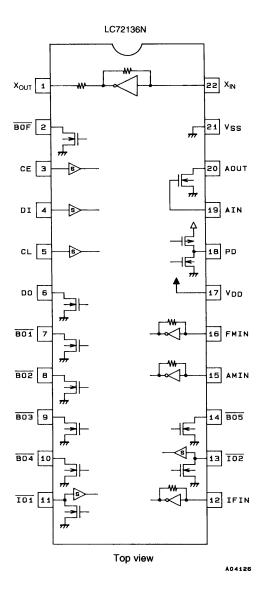
unit: mm

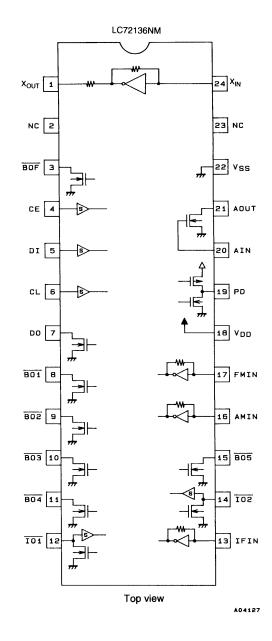
# 3112-MFP24S



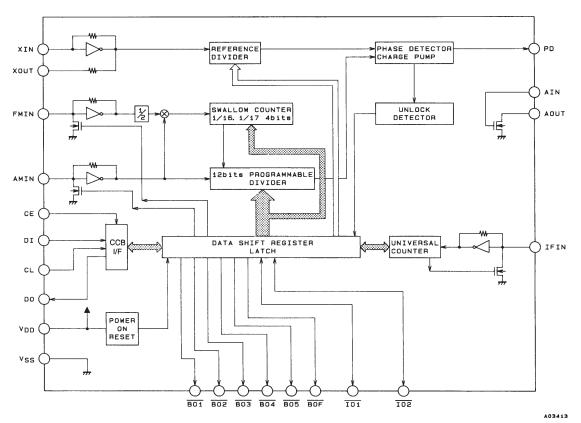
SANYO Electric Co., Ltd. Semiconductor Bussiness Headquarters TOKYO OFFICE Tokyo Bldg., 1-10, 1 Chome, Ueno, Taito-ku, TOKYO, 110 JAPAN

# **Pin Assignments**





# Block Diagram



# **Specifications**

# Absolute Maximum Ratings at $Ta=25^{\circ}C,\,V_{SS}$ = 0 V

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V <sub>DD</sub> max	V <sub>DD</sub>	-0.3 to +7.0	V
	V <sub>IN</sub> 1 max	CE, CL, DI, AIN	-0.3 to +7.0	V
Maximum input voltage	V <sub>IN</sub> 2 max	XIN, FMIN, AMIN, IFIN	-0.3 to V <sub>DD</sub> + 0.3	V
	V <sub>IN</sub> 3 max	<u>IO1, IO2</u>	-0.3 to +15	V
	V <sub>O</sub> 1 max	DO	-0.3 to +7.0	V
Maximum output voltage	V <sub>O</sub> 2 max	XOUT, PD	-0.3 to V <sub>DD</sub> + 0.3	V
	V <sub>O</sub> 3 max	BO1 to BO5, BOF, IO1, IO2, AOUT	-0.3 to +15	V
	I <sub>O</sub> 1 max	BO1	0 to 3.0	mA
Maximum output current	I <sub>O</sub> 2 max	AOUT, DO	0 to 6.0	mA
	I <sub>O</sub> 3 max	BO2 to BO5, BOF, IO1, IO2	0 to 10.0	mA
	Pd max	Ta ≤ 70°C: LC72136N (DIP22S)	350	mW
Allowable power dissipation	Polmax	Ta ≤ 70°C: LC72136NM (MFP24S)	200	mW
Operating temperature	Topr		-20 to +70	°C
Storage temperature	Tstg		-40 to +125	°C

Parameter	Symbol	Conditions	min	typ	max	Unit
Supply voltage	V <sub>DD</sub>	V <sub>DD</sub>	4.5		5.5	V
	V <sub>IH</sub> 1	CE, CL, DI	0.7 V <sub>DD</sub>		6.5	V
Input high-level voltage	V <sub>IH</sub> 2	<u>101, 102</u>	0.7 V <sub>DD</sub>		13	V
Input low-level voltage	VIL	CE, CL, DI, 101, 102	0		0.3 V <sub>DD</sub>	V
Output valtage	V <sub>O</sub> 1	DO	0		6.5	V
Output voltage	V <sub>O</sub> 2	BO1 to BO5, BOF, IO1, IO2, AOUT	0		13	V
	f <sub>IN</sub> 1	XIN: V <sub>IN</sub> 1		75		kHz
Input frequency	f <sub>IN</sub> 2	FMIN: V <sub>IN</sub> 2	10		160	MHz
	f <sub>IN</sub> 3	AMIN: V <sub>IN</sub> 3, SNS = 1	2		40	MHz
	f <sub>IN</sub> 4	AMIN: $V_{IN}4$ , SNS = 0	0.5		10	MHz
	f <sub>IN</sub> 5	IFIN: V <sub>IN</sub> 5	0.4		12	MHz
	V <sub>IN</sub> 1	XIN: f <sub>IN</sub> 1	400		1500	mVrms
	V <sub>IN</sub> 2-1	FMIN: f = 10 to 130 MHz	40		1500	mVrms
	V <sub>IN</sub> 2-2	FMIN: f = 130 to 160 MHz	70		1500	mVrms
Input amplitude	V <sub>IN</sub> 3	AMIN: f <sub>IN</sub> 3, SNS = 1	40		1500	mVrms
	V <sub>IN</sub> 4	AMIN: $f_{IN}4$ , SNS = 0	40		1500	mVrms
	V <sub>IN</sub> 5-1	IFIN: f <sub>IN</sub> 5, IFS = 1	40		1500	mVrms
	V <sub>IN</sub> 5-2	IFIN: f <sub>IN</sub> 6, IFS = 0	70		1500	mVrms
Guaranteed crystal oscillator frequency	Xtal	XIN, XOUT*		75		kHz

# Allowable Operating Ranges at Ta = -20 to +70°C, $V_{SS}$ = 0 V

Note: \* Crystal oscillator recommended CI value

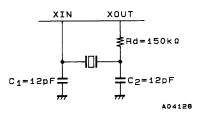
 $Cl \le 35 \text{ k}\Omega$  (for a 75 kHz crystal)

The circuit constants for the crystal oscillator circuit depend on the crystal used, the printed circuit board pattern, and other items. Therefore we recommend consulting with the manufacturer of the crystal for evaluation and reliability.

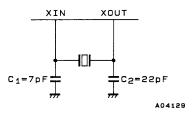
The extremely high input impedance of the XIN pins means that applications must take the possibility of leakage into account.

### Sample Oscillator Circuits

1. Seiko-Epson C-2-75kHz ( $C_L = 11 \text{ pF}$ )



2. Kyocera Corporation KF-38R5-09P0300 ( $C_L = 9 \text{ pF}$ )



# Electrical Characteristics at Ta = –20 to +70 $^{\circ}C,$ $V_{SS}$ = 0 V

Parameter	Symbol	Conditions	min	typ	max	Unit
	Rf1	XIN		8.0		MΩ
Internal feedback registers	Rf2	FMIN		500		kΩ
Internal feedback resistors	Rf3	AMIN		500		kΩ
	Rf4	IFIN		250		kΩ
laternal mult device realistance	Rpd1	FMIN		200		kΩ
Internal pull-down resistors	Rpd2	AMIN		200		kΩ
Internal output resistor	Rd	XOUT		250		kΩ
Hysteresis	V <sub>HIS</sub>	CE, CL, DI, IO1, IO2		0.1 V <sub>DD</sub>		V
Output high-level voltage	V <sub>OH</sub> 1	PD: $I_0 = -1 \text{ mA}$	V <sub>DD</sub> – 1.0			V
	V <sub>OL</sub> 1	PD: I <sub>O</sub> = 1 mA			1.0	V
	N/ 0	$\overline{\text{BO1}}$ : I <sub>O</sub> = 0.5 mA			0.5	V
	V <sub>OL</sub> 2	$\overline{\text{BO1}}$ : I <sub>O</sub> = 1 mA			1.0	V
	N 0	DO: I <sub>O</sub> = 1 mA			0.2	V
Output low-level voltage	V <sub>OL</sub> 3	DO: I <sub>O</sub> = 5 mA			1.0	V
		$\overline{\text{BO2}}$ to $\overline{\text{BO5}}$ , $\overline{\text{BOF}}$ , $\overline{\text{IO1}}$ , $\overline{\text{IO2}}$ : I <sub>O</sub> = 1 mA			0.2	V
	V <sub>OL</sub> 4	$\overline{\text{BO2}}$ to $\overline{\text{BO5}}$ , $\overline{\text{BOF}}$ , $\overline{\text{IO1}}$ , $\overline{\text{IO2}}$ : $I_{\text{O}} = 5 \text{ mA}$			1.0	V
	-	$\overline{\text{BO2}}$ to $\overline{\text{BO5}}$ , $\overline{\text{BOF}}$ , $\overline{\text{IO1}}$ , $\overline{\text{IO2}}$ : I <sub>O</sub> = 8 mA			1.6	V
	V <sub>OL</sub> 5	AOUT: I <sub>O</sub> = 1 mA, AIN = 1.3 V			0.5	V
	I <sub>IH</sub> 1	CE, CL, DI: V <sub>I</sub> = 6.5 V			5.0	μA
	I <sub>IH</sub> 2	$\overline{101}, \overline{102}: V_1 = 13 V$			5.0	μA
have the base of the second	I <sub>IH</sub> 3	$XIN: V_I = V_{DD}$	0.3	0.6	1.4	μA
Input high-level voltage	I <sub>IH</sub> 4	FMIN, AMIN: V <sub>I</sub> = V <sub>DD</sub>	4.0		22	μA
	I <sub>IH</sub> 5	IFIN: $V_I = V_{DD}$	8.0		44	μA
	I <sub>IH</sub> 6	AIN: V <sub>I</sub> = 6.5 V			200	nA
	I <sub>IL</sub> 1	CE, CL, DI: V <sub>I</sub> = 0 V			5.0	μA
	I <sub>IL</sub> 2	$\overline{101}, \overline{102}: V_1 = 0 V$			5.0	μA
	I <sub>IL</sub> 3	$XIN: V_I = 0 V$	0.3	0.6	1.4	μA
Input low-level current	I <sub>IL</sub> 4	FMIN, AMIN: $V_{I} = 0 V$	4.0		22	μA
	I <sub>IL</sub> 5	IFIN: $V_1 = 0 V$	8.0		44	μA
	I <sub>IL</sub> 6	AIN: $V_1 = 0 V$			200	nA
	I <sub>OFF</sub> 1	$\overline{BO1}$ to $\overline{BO5}$ , $\overline{BOF}$ , AOUT, $\overline{IO1}$ , $\overline{IO2}$ : V <sub>O</sub> = 13 V			5.0	μA
Output off leakage current	I <sub>OFF</sub> 2	DO: V <sub>O</sub> = 6.5 V			5.0	μA
High-level tree-state off leakage current	IOFFH	PD: $V_0 = V_{DD}$		0.01	200	nA
Low-level tree-state off leakage current	I <sub>OFFL</sub>	PD: V <sub>O</sub> = 0 V		0.01	200	nA
Input capacitance	CIN	FMIN		6		pF
	I <sub>DD</sub> 1	V <sub>DD</sub> : Xtal = 75 kHz, f <sub>IN</sub> 2 = 130 MHz, V <sub>IN</sub> 2 = 40 mVrms		5	10	mA
Current drain	I <sub>DD</sub> 2	V <sub>DD</sub> : PLL block stopped (PLL inhibit), Xtal oscillator operating (Xtal = 75 kHz)		0.1		mA
	I <sub>DD</sub> 3	V <sub>DD</sub> : PLL block stopped, Xtal oscillator stopped			10	μA

# **Pin Functions**

Symbol	Pin No. (MFP pin numbers are in parentheses.)	Туре	Functions	Circuit configuration
XIN XOUT	22 (24) 1 (1)	Xtal	<ul> <li>Crystal oscillator connections (75 kHz)</li> <li>The extremely high input impedance of the XIN pins means that applications must take the possibility of leakage into account.</li> </ul>	
FMIN	16 (17)	Local oscillator signal input	<ul> <li>FMIN is selected when the serial data input DVS bit is set to 1.</li> <li>The input frequency range is from 10 to 160 MHz.</li> <li>The input signal passes through the internal divide-by-two prescaler and is input to the swallow counter.</li> <li>The divisor can be in the range 272 to 65535. However, since the signal has passed through the divide-by-two prescaler, the actual divisor is twice the set value.</li> </ul>	A02599
AMIN	15 (16)	Local oscillator signal input	<ul> <li>AMIN is selected when the serial data input DVS bit is set to 0.</li> <li>When the serial data input SNS bit is set to 1: <ul> <li>The input frequency range is 2 to 40 MHz.</li> <li>The signal is directly input to the swallow counter.</li> <li>The divisor can be in the range 272 to 65535, and the divisor used will be the value set.</li> </ul> </li> <li>When the serial data input SNS bit is set to 0: <ul> <li>The input frequency range is 0.5 to 10 MHz.</li> <li>The signal is directly input to a 12-bit programmable divider.</li> <li>The divisor can be in the range 4 to 4095, and the divisor used will be the value set.</li> </ul> </li> </ul>	
CE	3 (4)	Chip enable	<ul> <li>Set this pin high when inputting (DI) or outputting (DO) serial data.</li> </ul>	S> A02500
CL	5 (6)	Clock	Used as the synchronization clock when inputting (DI) or outputting (DO) serial data.	□S ∧02500
DI	4 (5)	Input data	Inputs serial data transferred from the controller to the LC72136N.	□S A02500
DO	6 (7)	Output data	Outputs serial data transferred from the LC72136N to the controller. The data output is determined by the DOC0 to DOC2 bits in the serial data.	
V <sub>DD</sub>	17 (18)	Power supply	<ul> <li>The LC72136N power supply pin. (V<sub>DD</sub> = 4.5 to 5.5 V)</li> <li>The power on reset circuit operates when power is first applied.</li> </ul>	
V <sub>SS</sub>	21 (22)	Ground	The LC72136N ground	

Continued on next page.

Continued from preceding page.

Symbol	Pin No. (MFP pin numbers are in parentheses.)	Туре	Functions	Circuit configuration
B01 B02 B03 B04 B05 B0F	7 (8) 8 (9) 9 (10) 10 (11) 14 (15) 2 (3)	Output ports	<ul> <li>Dedicated outputs</li> <li>The output states are determined by the BO1 to BO5 bits in the serial data. Data: 0 = open, 1 = low</li> <li>A time base signal (8 Hz) can be output from the BO1 pin. (When the serial data TBC bit is set to 1.)</li> <li>Care is required when using the BO1 pin, since it has a higher on impedance that the other output ports (pins BO2 to BO5).</li> <li>The output state of the BOF pin is determined by the serial data DVS bit. Thus this pin can be used as an FM band selection switch. (Note that it should not be used as an AM band selection switch since it is susceptible to noise from the crystal oscillator.)</li> <li>DVS data: 0 = open, 1 = low</li> <li>All output ports are set to the open state following a power on reset.</li> </ul>	
101 102	11 (12) 13 (14)	Input or output ports	<ul> <li>I/O dual-use pins</li> <li>The direction (input or output) is determined by bits IOC1 and IOC2 in the serial data. Data: 0 = input port, 1 = output port</li> <li>When specified for use as input ports: The state of the input pin is transmitted to the controller over the DO pin. Input state: low = 0 data value high = 1 data value</li> <li>When specified for use as output ports: The output states are determined by the IO1 and IO2 bits in the serial data. Data: 0 = open, 1 = low</li> <li>These pins function as input pins following a power on reset.</li> </ul>	
PD	18 (19)	Charge pump output	<ul> <li>PLL charge pump output When the frequency generated by dividing the local oscillator signal frequency by N is higher than the reference frequency, a high level is output from the PD pin. Similarly, when that frequency is lower, a low level is output. The PD pin goes to the high-impedance state when the frequencies match.</li> </ul>	
AIN AOUT	19 (20) 20 (21)	LPF amplifier transistor connections	The n-channel MOS transistor used for the PLL active low-pass filter.	A02504
IFIN	12 (13)	IF counter	<ul> <li>Accepts an input in the frequency range 0.4 to 12 MHz.</li> <li>The input signal is directly transmitted to the IF counter.</li> <li>The result is output starting the MSB of the IF counter using the DO pin.</li> <li>Four measurement periods are supported: 4, 8, 32, and 64 ms.</li> </ul>	A02599

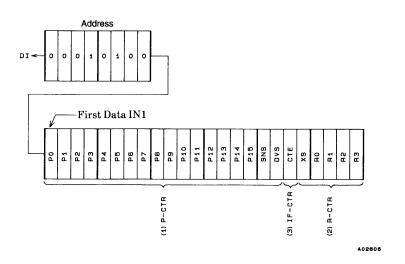
# Serial Data I/O Procedures

The LC72136N inputs and outputs data using the Sanyo CCB (computer control bus) audio LSI serial bus format. This LSI adopts an 8-bit address format CCB.

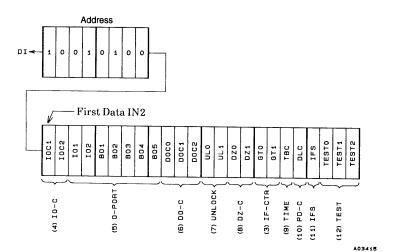
	I/O mada				Add	ress				Function			
	I/O mode	B0	B1	B2	B3	A0	A1	A2	A3	Function			
1	IN1 (82)	0	0	0	1	0	1	0	0	<ul> <li>Control data input mode (serial data input)</li> <li>24 data bits are input.</li> <li>See the "DI Control Data (serial data input) Structure" item for details on the meaning of the input data.</li> </ul>			
2	IN2 (92)	1	0	0	1	0	1	0	0	<ul> <li>Control data input mode (serial data input)</li> <li>24 data bits are input.</li> <li>See the "DI Control Data (serial data input) Structure" item for details on the meaning of the input data.</li> </ul>			
3	OUT (A2)	0	1	0	1	0	1	0	0	<ul> <li>Data output mode (serial data output)</li> <li>The number of bits output is equal to the number of clock cycles.</li> <li>See the "DO Output Data (Serial Data Output) Structure" item for details on the meaning of the output data.</li> </ul>			
	$CE$ $CL \begin{cases} (1) \\ (2) \\ (2) \\ (3) \\ (3) \\ (4) \\ (4) \\ (5) \\ (5) \\ (5) \\ (6) \\ $	CL: norr	nal high	A	ј	 	A1	<u>}</u>		<pre>//O mode determined /// // // // // // // // // // // // //</pre>			

## DI Control Data (serial data input) Structure

1. IN1 Mode



2. IN2 Mode



# **DI Control Data Functions**

No.	Control block/data				De	scription			Related data			
	Programmable divider data	Data that	sets the p	rogrammat	le divider							
	P0 to P15	A binary v	alue in wh	ich P15 is	the MSB.	The LSB char	iges depending	on DVS and SNS.				
		DVS	SNS	LSB	Divisor	setting (N)	Actu	al divisor				
		1	*	P0		0 65535		e of the setting				
		0	1	P0		o 65535	The value of the					
		0	0	P4		to 4095	The value of the					
~				ignored wh				ie setting				
(1)	DVS, SNS			•			orammable divid	der, switches the				
	2.00, 0.10						tate. (*: Don't c					
		DVS	SNS	Input pir	n	Input frequen	cy range	BOF pin				
		1	*	FMIN		10 to 160	MHz	Low				
		0	1	AMIN		2 to 40 M	ЛНz	Open				
		0	0	AMIN		0.5 to 10	MHz	Open				
		Note: See	e the "Proo	grammable	Divider" it	tem for details	-	· · · · · · · · · · · · · · · · · · ·				
	Reference divider data	Reference	e frequenc	y (fref) sele	ection data	à						
	R0 to R3	R3	R2	R1	R0	Re	ference frequen	cy (kHz)				
		0	0	0	0		25					
		0	0	0	1		25					
		0	0	1	0 1		25 25					
		0	1	0	0		12.5					
		0	1	1	0 1		3.125 3.125					
		1	0	0	0		5					
		1	0	0	1	5						
(2)		1	0	1	0		5					
(-)		1	0	1	1 0		1 3					
		1	1	0	1		15					
		1	1	1	0	PLL I	NHIBIT + Xtal C	SC STOP				
		1	1	1	1		PLL INHIBI	Т				
		Note: PLI										
							are stopped, the	e FMIN, AMIN, output pin goes to				
				dance state		notato, and an	o onargo pamp c	diput pin good to				
	XS	<ul> <li>Oscillator</li> </ul>										
				mode" The	oscillator	margin is redu	uced and the cry	stal radiation				
		IS XS = 1: N	reduced.	le.								
					ving a pow	ver-on reset.						
	IF counter control data	IF counter			specification	on						
	CTE	CTE = 1: CTE = 0:										
	GT0 GT1				dotormin-	tion						
	GT0, GT1	IF counter	measure									
(3)		GT1	GT0	Meas	urement t	ime (ms)	Wait	time (ms)	IFS			
(0)		0	0		4			3 to 4				
		0	1		8			3 to 4				
		1	0		32		7	7 to 8				
		1	1		64		1	7 to 8				
						m for details.						
(4)	I/O port specification data					I/O dual-use	pins					
	IOC1, IOC2			e, 1 = outp		loto						
(5)	Output port data BO1 to BO5, IO1, IO2	<ul> <li>BO1 to B0</li> <li>Data: 0 =</li> </ul>			out state d	เสเส			IOC1			
(0)			Open" is						IOC2			

Continued on next page.

Continued from preceding page.

No.	Control block/data				Description		Related data
	DO pin control data	Data that	determine	es DO pin o	output		
	DOC0, DOC1, DOC2	DOC2	DOC1	DOC0		DO pin state	
		0	0	0	Open	F	
		0	0	1	Low when the unlock	state is detected	
		0	1	0	end-UC*1		
		0	1	1	Open		
			0	0	Open The IO1 pin state <sup>*2</sup>		
		1	1	0	The $\overline{IO2}$ pin state <sup>*2</sup>		
			1	1	Open		
					lowing a power-on rese measurement completi		
		1000.11					UL0, UL1,
(6)		DO p	oin		))	<u>۲</u>	CTE, IOC1, IOC2
			-	Count star		Count end ③ CE: High	
					set and an IF count is s les to the open state.	started (CTE = $0 \rightarrow 1$ ), the DO pin	
						etes, the DO pin goes low and	
					etion check operation is		
						to serial data I/O (CE: high). set to be an output port.	
					•	ing the data input period (during the	
		pe	eriod wher	CE is high	n in mode IN1 or IN2), re	egardless of the values of the DO pin	
						n outputs the content of the internal	
				,		bin signal during the data output period mode) regardless of the values of	
					a (DOC0 to DOC2).		
	Unlock detection data			( )	detection range for PLL		
	UL0, UL1				nan the specified range Don't care.)	occurs, the LC72136N determines	
(7)		UL1	UL0	ø	E detection width	Detector output	DOC0, DOC1,
(.,		0	0	Stopped		Open	DOC2
		0	1	0		øE is output directly	
		1	*	±6.67 µs		øE is extended by 1 to 2 ms	
						serial data output UL bit is 0.	
	Phase comparator control data	<ul> <li>Phase cor</li> </ul>	mparator	dead zone	control data		
	DZ0, DZ1	DZ1	DZ0		Dead z	zone mode	
		0	0	DZA			
(8)		0	1	DZB			
		1	0	DZC			
		1	1	DZD			
		Dead zon	e width: D	ZA < DZB	< DZC < DZD		
(9)	Clock time base TBC	• An 8 Hz 4 (The BO1	,		0	but from $\overline{BO1}$ by setting TBC to 1.	BO1
	Charge pump control data	Data that	forcibly co	ontrols the	charge pump output		
	DLC	D	LC		Charge	oump output	
			0	Normal	operation		
			1	Forced lo			
(10)					-	ing from deadlock by setting Vtune to	
						en the circuit is deadlocked due to the ol voltage (Vtune) being 0 V.	
		The crysta	al oscillato	r circuit mu	· · · · ·	llowing a power on reset. y before this data is changed to	

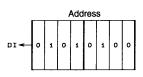
Continued on next page.

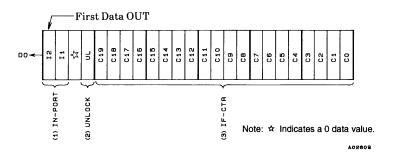
Continued from preceding page.

No.	Control block/data	Description	Related data
(11)	IF counter control data IFS	This data should be set to 1 in normal operation. Setting this data to 0 switches the LC72136N to a reduced input sensitivity mode in which the sensitivity is reduced by 10 to 30 mVrms.     * See the "IF Counter Operation" item for details.	
(12)	LSI test data TEST 0 to TEST3	LSI test data     TEST0     TEST1     All three bits must be set to 0.     TEST2     All the test data is set to 0 following a power-on reset.	

## DO Output Data (Serial Data Output) Structure

### 3. OUT mode



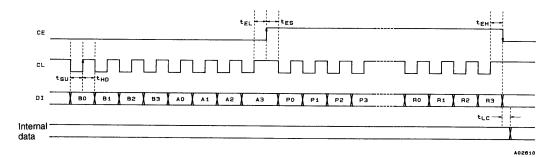


# **DO Output Data**

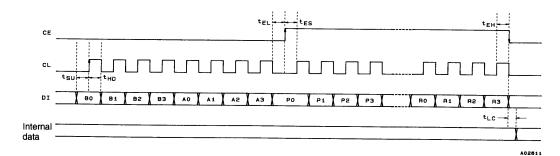
No.	Control block/data	Description	Related data
(1)	I/O port data I2, I1	Data latched from the states of the I/O ports, pins IO1 and IO2. This data reflects the pin states, regardless of whether they are in input or output mode. The data is latched when OUT mode is selected. I1 ← IO1 pin state ) High: 1 I2 ← IO2 pin state ) Low: 0	IOC1, IOC2
(2)	PLL unlock data UL	• Data latched from the state of the unlock detection circuit UL $\leftarrow$ 0: Unlocked UL $\leftarrow$ 1: Locked or in detection stopped mode	ULO, UL1
(3)	IF counter binary data C19 to C0	Data latched from the state of the IF counter, which is a 20-bit binary counter.     C19 ← Binary counter MSB     C0 ← Binary counter LSB	CTE, GT0, GT1

# Serial Data Input (IN1/IN2) $t_{SU},\,t_{HD},\,t_{EL},\,t_{ES},\,t_{EH},\geq 0.75~\mu s~t_{LC}$ < 0.75 $\mu s$

1. CL: Normal high

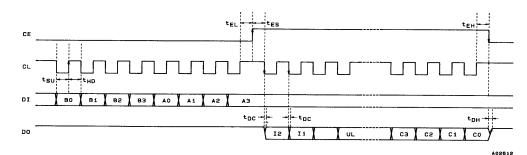


2. CL: Normal low

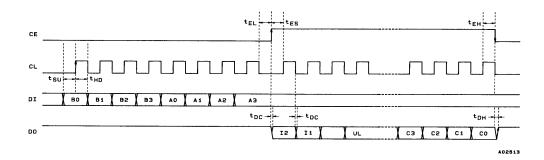


# Serial Data Output (OUT) $t_{SU},\,t_{HD},\,t_{EL},\,t_{ES},\,t_{EH},\geq$ 0.75 $\mu s$ $t_{DC},\,t_{DH}$ < 0.35 $\mu s$

1. CL: Normal high

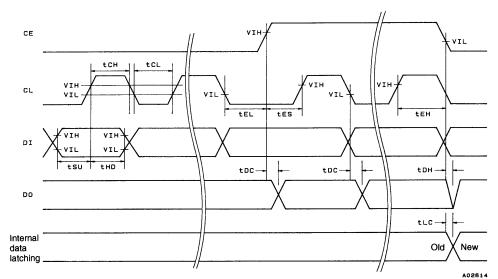


2. CL: Normal low

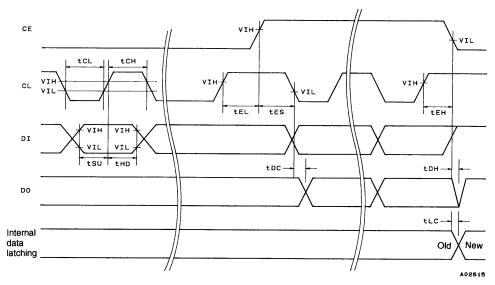


Note: Since the DO pin is an n-channel open drain circuit, the times for the data to change (t<sub>DC</sub> and t<sub>DH</sub>) will differ depending on the value of the pull-up resistor, printed circuit board capacitance.

## Serial Data Timing



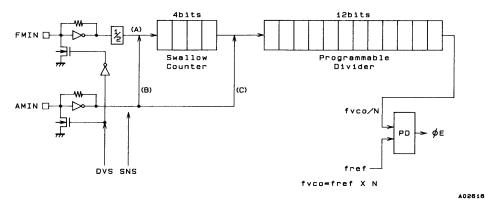
CL Stopped at the Low Level



CL Stopped at the High Level

Parameter	Symbol	Pins	Conditions	min	typ	max	Unit
Data setup time	t <sub>SU</sub>	DI, CL		0.75			μs
Data hold time	t <sub>HD</sub>	DI, CL		0.75			μs
Clock low-level time	t <sub>CL</sub>	CL		0.75			μs
Clock high-level time	t <sub>CH</sub>	CL		0.75			μs
CE wait time	t <sub>EL</sub>	CE, CL		0.75			μs
CE setup time	t <sub>ES</sub>	CE, CL		0.75			μs
CE hold time	t <sub>EH</sub>	CE, CL		0.75			μs
Data latch change time	t <sub>LC</sub>					0.75	μs
Data output time	t <sub>DC</sub>	DO, CL	These times depend on the pull-up resistance			0.35	μs
	t <sub>DH</sub>	DO, CE	and the printed circuit board capacitances.			0.35	μs

### **Programmable Divider Structure**



	DVS	SNS	Input pin	Set divisor	Actual divisor: N	Input frequency range (MHz)
A	1	*	FMIN	272 to 65535	Twice the set value	10 to 160
В	0	1	AMIN	272 to 65535	The set value	2 to 40
С	0	0	AMIN	4 to 4095	The set value	0.5 to 10

Note: \* Don't care.

### Sample Programmable Divider Divisor Calculations

1. For a 50 kHz FM step size (DVS = 1, SNS = \*: FMIN selected)

• FM RF = 90.0 MHz (IF = +10.7 MHz)

FM VCO = 100.7 MHz

PLL fref = 25 kHz (R0 to R1 = 1, R2 to R3 = 0)

100.7 MHz (FM VCO)  $\div$  25 kHz (fref)  $\div$  2 (FMIN: divide-by-two prescaler) = 2014  $\rightarrow$  07DE (HEX)

~		Ę	_			<u> </u>	_		<u>ز</u>		_		_	<u> </u>									
0	1	1	1	1	0	1	1	1	1	1	0	0	0	0	0	*	1			1	1	0	0
04	đ	ЪЗ	БЧ	4 4	5 d	9 d	P7	8 d	6 d	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	xs	ВО	В1	ВЗ	E8

A02817

2. For a 5 kHz SW step size (DVS = 0, SNS = 1: AMIN high-speed side selected)

• SW RF = 21.75 MHz (IF = +450 kHz)

SW VCO = 22.20 MHz

PLL fref = 5 kHz (R0 = R2 = 0, R1 = R3 = 1)

22.2 MHz (SW VCO)  $\div$  5 kHz (fref) = 4440  $\rightarrow$  1158 (HEX)

_		<u> </u>				5		_	ز_	L		_		<u> </u>									
0	0	0	1	1	0	1	0	1	0	0	0	1	0	0	0	1	0			0	1	0	1
0	P1	ЪЗ	БЧ	P.4	52	90	Р7	8 d	6 d	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	SX	В	н1	58	R3

A02618

3. For a 9 kHz MW step size (DVS = 0, SNS = 0: AMIN low-speed side selected)

• MW RF = 1008 kHz (IF = +450 kHz)

MW VCO = 1458 kHz

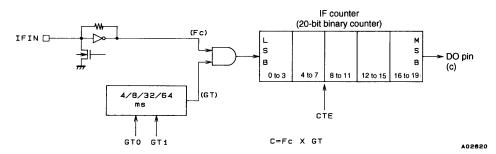
PLL fref = 3 kHz (R0 to R1 = 0, R2 to R3 = 1): using a 3 kHz reference frequency 1458 kHz (MW VCO)  $\div$  3 kHz (fref) = 486  $\rightarrow$  1E6 (HEX)

				_	_	<u> </u>		_	_5		_		ز_	L									
*	*	*	*	0	1	1	0	0	1	1	1	1	0	0	0	0	0			0	0	1	1
0	1 d	Ъ2	Еd	P.4	Ъ.	96	P7	8d	6d	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	sx	ВO	R1	В2	ВЭ

A03418

### **IF Counter Structure**

The LC72136N IF counter is a 20-bit binary counter. The result of the count can be read out serially, MSB first, from the DO pin.



GT1	GT0	Measurement time									
GII	GIU	Measurement period (GT) (ms)	Wait time (t <sub>WU</sub> ) (ms)								
0	0	4	3 to 4								
0	1	8	3 to 4								
1	0	32	7 to 8								
1	1	64	7 to 8								

IF frequency (Fc) measurement consists of determining how many pulses enter the IF counter in a specified measurement time (GT).

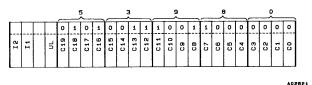
$$Fc = \frac{C}{GT}$$
 (C = Fc × GT)

C: count value (number of pulses)

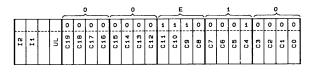
402622

### Sample IF Counter Frequency Calculations

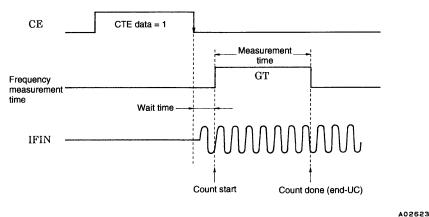
1. For a measurement time (GT) of 32 ms and a count value (C) of 53980 (hexadecimal), which is 342,400 (decimal) IF frequency (Fc) = 342,400 ÷ 32 ms = 10.7 MHz



2. For a measurement time (GT) of 8 ms and a count value (C) of E10 (hexadecimal), which is 3600 (decimal) IF frequency (Fc) = 3600 ÷ 8 ms = 450 kHz



### **IF Counter Operation**



Before starting the IF count, the IF counter must be reset in advance by setting CTE in the serial data to 0. The IF count is started by changing the CTE bit in the serial data from 0 to 1. The serial data is latched by the LC72136N when the CE pin is dropped from high to low. The IF signal must be supplied to the IFIN pin in the period between the point the CE pin goes low and the end of the wait time at the latest. Next, the value of the IF count at the end of the measurement period must be read out during the period CTE is 1. This is because the IF counter is reset when CTE is set to 0.

Note: When operating the IF counter, the control microprocessor must first check the state of the IF-IC SD (station detect) signal and only after determining that the SD signal is present turn on IF buffer output and execute an IF count operation. Auto-search techniques that use only the IF counter are not recommended, since it is possible for IF buffer leakage output to cause incorrect stops at points where there is no station.

			f (MHz)
IFS	$0.4 \le f < 0.5$	0.5 ≤ f < 8	$8 \le f \le 12$
1: Normal mode	40 mVrms (0.1 to 3 mVrms)	40 mVrms	40 mVrms (1 to 10 mVrms)
0: Degradation mode	70 mVrms (10 to 15 mVrms)	70 mVrms	70 mVrms (30 to 40 mVrms)

#### **IFIN Minimum Sensitivity Ratings**

Note: Values in parentheses are actual performance values presented as reference data.

#### **Unlock Detection Timing**

1. Unlock Detection Determination Timing

Unlock detection is performed in the reference frequency (fref) period (interval). Therefore, in principle, unlock determination requires a time longer than the period of the reference frequency. However, immediately after changing the divisor N (frequency) unlock detection must be performed after waiting at least two periods of the reference frequency.

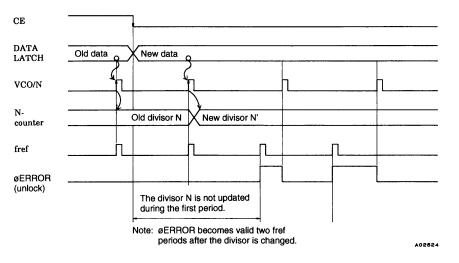
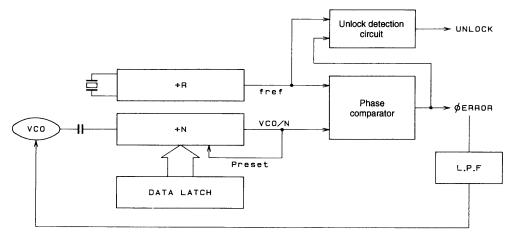


Figure 1 Unlock Detection Timing

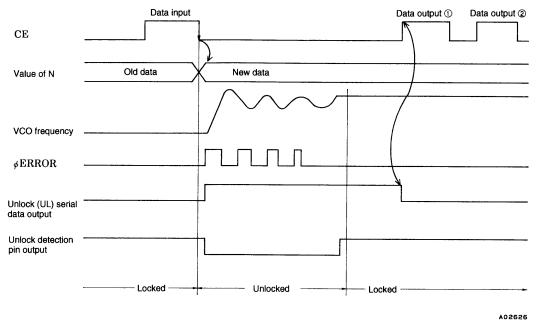
For example, if fref is 1 kHz (and thus the period is 1 ms), after changing the divisor N, the system must wait at least 2 ms before checking for the unlocked state.



A02625

Figure 2 Circuit Structure

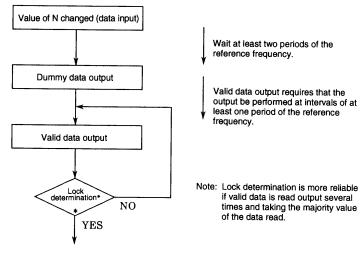
### 2. Unlock Detection Software



## Figure 3

3. When Outputting Unlock Data Using Serial Data Output:

Once the LC72136N detects an unlocked state, it does not reset the unlock data (UL) until the next data output (or data input) operation is performed. At the data output ① point in Figure 3, although the VCO frequency is stable (locked), the unlock data remains set to the unlocked state since no data output has been performed since the value of N was changed. Thus, even though the frequency became stable (locked), from the point of view of the data, the circuit is in the unlocked state. Therefore, the data output ① immediately following a change to the value of N should be seen as a dummy data, and the data from the second data output (data output ②) and later outputs should be seen as valid data.



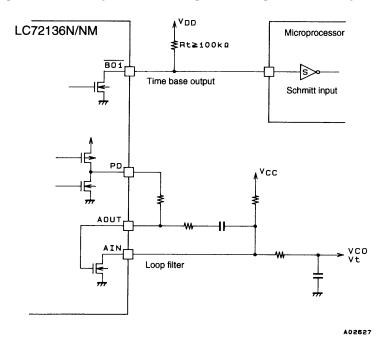


### When directly outputting data from the DO pin (set up by the DO pin control data)

Since the DO pin outputs the unlocked state (locked: high, unlocked: low) the timing considerations in the technique described in the previous section are not necessary. After changing the value of N, the locked state can be determined after waiting at least two periods of the reference frequency.

### Notes on Clock Time Base Usage

When the clock time base output is used, the value of the pull-up resistor for the output pin ( $\overline{BO1}$ ) must be at least 100 k $\Omega$ . This is to avoid degradation of the VCO C/N characteristics when using the built-in low-pass filter transistor to form the loop filter. Since the clock time base output pin and the low-pass filter transistor ground are the same node in the IC, the time base output pin current fluctuations must be suppressed to limit the influence on the low-pass filter. We recommend the use of a Schmitt input on the receiving controller (microprocessor) to prevent chattering.



### **Other Items**

1. Notes on the Phase Comparator Dead Zone

DZ1	DZ0	Dead-zone mode	Charge pump	Dead zone
0	0	DZA	ON/ON	0 sec
0	1	DZB	ON/ON	-0 sec
1	0	DZC	OFF/OFF	+0 sec
1	1	DZD	OFF/OFF	+ +0 sec

Since correction pulses are output from the charge pump even if the PLL is locked when the charge pump is in the ON/ON state, the loop can easily become unstable. This point requires special care when designing application circuits.

The following problems may occur in the ON/ON state.

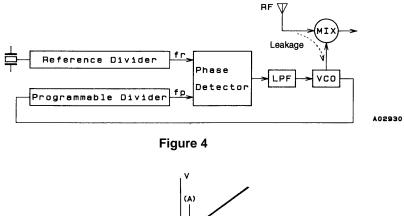
- · Side band generation due to reference frequency leakage
- Side band generation due to both the correction pulse envelope and low frequency leakage

Schemes in which a dead zone is present (OFF/OFF) have good loop stability, but have the problem that acquiring a high C/N ratio can be difficult. On the other hand, although it is easy to acquire a high C/N ratio with schemes in which there is no dead zone, it is difficult to achieve high loop stability. Therefore, it can be effective to select DZA or DZB, which have no dead zone, in applications which require an FM S/R ratio in excess of 90 to 100 dB, or in which an increased AM stereo pilot margin is desired. On the other hand, we recommend selecting DZC or DZD, which provide a dead zone, for applications which do not require such a high FM signal-to-noise ratio and in which either AM stereo is not used or an adequate AM stereo pilot margin can be achieved.

### Dead Zone

The phase comparator compares fp to a reference frequency (fr) as shown in Figure 4. Although the characteristics of this circuit (see Figure 5) are such that the output voltage is proportional to the phase difference  $\emptyset$  (line A), a region (the dead zone) in which it is not possible to compare small phase differences occurs in actual ICs due to internal circuit delays and other factors (line B). A dead zone as small as possible is desirable for products that must provide a high S/N ratio.

However, since a larger dead zone makes this circuit easier to use, a larger dead zone is appropriate for popularlypriced products. This is because it is possible for RF signals to leak from the mixer to the VCO and modulate the VCO in popularly-priced products in the presence of strong RF inputs. When the dead zone is narrow, the circuit outputs correction pulses and this output can further modulate the VCO and generate beat frequencies with the RF signal.



(B)



- Coupling capacitors must be placed as close as possible to their respective pin. A capacitance of about 100 pF is desirable. In particular, if a capacitance of 1000 pF or over is used for the IF pin, the time to reach the bias level will increase and incorrect counting may occur due to the relationship with the wait time.
- 3. Notes on IF Counting → SD must be used in conjunction with the IF counting time When using IF counting, always implement IF counting by having the microprocessor determine the presence of the IF-IC SD (station detect) signal and turn on the IF counter buffer only if the SD signal is present. Schemes in which auto-searches are performed with only IF counting are not recommended, since they can stop at points where there is no signal due to leakage output from the IF counter buffer.

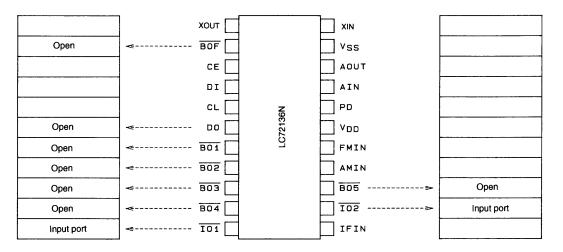
### 4. DO Pin Usage Techniques

In addition to data output mode times, the DO pin can also be used to check for IF counter count completion and for unlock detection output. Also, an input pin state can be output unchanged through the DO pin and input to the controller.

### 5. Power Supply Pins

A capacitor of at least 2000 pF must be inserted between the power supply  $V_{DD}$  and  $V_{SS}$  pins for noise exclusion. This capacitor must be placed as close as possible to the  $V_{DD}$  and  $V_{SS}$  pins.

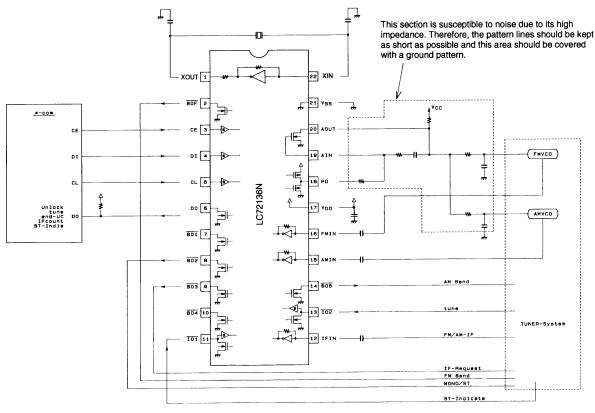
## Pin States Following a Power-On Reset



A03417

#### Sample Application System

(Using the DIP22S package)



A03418

- No products described or contained herein are intended for use in surgical implants, life-support systems, aerospace equipment, nuclear power control systems, vehicles, disaster/crime-prevention equipment and the like, the failure of which may directly or indirectly cause injury, death or property loss.
- Anyone purchasing any products described or contained herein for an above-mentioned use shall:
  - ① Accept full responsibility and indemnify and defend SANYO ELECTRIC CO., LTD., its affiliates, subsidiaries and distributors and all their officers and employees, jointly and severally, against any and all claims and litigation and all damages, cost and expenses associated with such use:
- Intersection 20 Not impose any responsibility for any fault or negligence which may be cited in any such claim or litigation on SANYO ELECTRIC CO., LTD., its affiliates, subsidiaries and distributors or any of their officers and employees jointly or severally.
- Information (including circuit diagrams and circuit parameters) herein is for example only; it is not guaranteed for volume production. SANYO believes information herein is accurate and reliable, but no guarantees are made or implied regarding its use or any infringements of intellectual property rights or other rights of third parties.

This catalog provides information as of November, 1996. Specifications and information herein are subject to change without notice.