

SPT9712

12-BIT, 100 MWPS ECL D/A CONVERTER

TECHNICAL DATA

FEBRUARY 15, 2001

FEATURES

- 12-Bit, 100 MWPS digital-to-analog converter
- ECL compatibility
- Low power: 600 mW
- 1/2 LSB DNL
- 40 MHz multiplying bandwidth
- Industrial temperature range
- Superior performance over AD9712
 - Improved settling time of 13 ns
 - Improved glitch energy 15 pV-s
 - Master-slave latches

APPLICATIONS

- · Fast frequency hopping spread spectrum radios
- Direct sequence spread spectrum radios
- · Microwave and satellite modems
- Test & measurement instrumentation

GENERAL DESCRIPTION

The SPT9712 is a 12-bit, 100 MWPS digital-to-analog converter designed for direct digital synthesis, high resolution imaging, and arbitrary waveform generation applications.

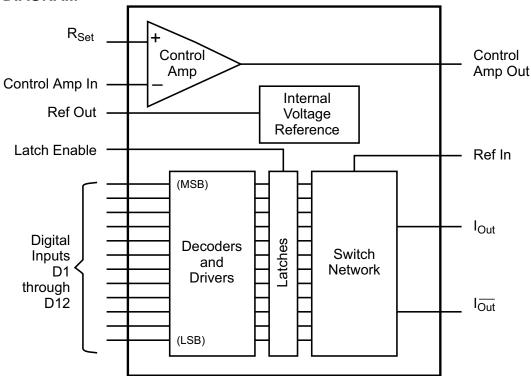
This device is pin-for-pin compatible with the AD9712 with significantly improved performance. The only difference between the SPT9712 and the AD9712 is that the Latch Enable (LE, pin 26) for the SPT9712 is rising-edge trig-

gered (see figure 1), whereas the Latch Enable (LE, pin 26) for the AD9712 functions in the transparent mode.

The SPT9712 is an ECL-compatible device. It features a fast settling time of 13 ns and low glitch impulse energy of 15 pV-s, which results in excellent spurious-free dynamic range characteristics.

The SPT9712 is available in a 28-lead PLCC package in the industrial temperature range (-40 to +85 °C).

BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS (Beyond which damage may occur) 25 °C

Output	Currents
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Internal Reference	ce Output Current	500 μΑ
Control Amplifier	Output Current	±2.5 mA

Temperature

Operating Temperature	40 to +85 °C
Junction Temperature	+150 ℃
Lead, Soldering (10 seconds)	+300 ℃
Storage	65 to +150 °C

Note: 1. Operation at any Absolute Maximum Rating is not implied. See Electrical Specifications for proper nominal applied conditions in typical applications.

ELECTRICAL SPECIFICATIONS

 $T_A = T_{MIN} - T_{MAX}$, $V_{EE} = -5.2$ V, $R_{Set} = 7.5$ k Ω , Control Amp In = Ref Out, $V_{OUT} = 0$ V, unless otherwise specified.

	TEST	ST TEST SPT9712A SPT9712B		SPT9712A		2B			
PARAMETERS	CONDITIONS	LEVEL	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
DC Performance Resolution Differential Linearity Differential Linearity Integral Linearity Integral Linearity Output Capacitance Gain Error1 Gain Error Tempco Zero-Scale Offset Error Offset Drift Coefficient Output Compliance Voltage Equivalent Output Resistance	Max at Full Temp. Best Fit Max at Full Temp. +25 °C +25 °C Full Temp. Full Temp. +25 °C Full Temp. Full Temp. Full Temp. Full Temp. Full Temp. +25 °C +25 °C +25 °C		-1.2 0.8	12 ±0.5 ±0.75 10 1.0 150 0.5 0.01	±0.75 ±1.5 ±1.0 ±1.75 5.0 8.0 2.5 5.0 +2.0 1.2	-1.2 0.8	12 ±1.0 ±1.0 10 1.0 150 0.5 0.01	±1.25 ±2.0 ±1.5 ±2.0 5.0 8.0 2.5 5.0 +2.0	Bits LSB LSB LSB LSB pF % FS % FS PPM/°C μA μA/°C V kΩ
Dynamic Performance Conversion Rate Settling Time t _{ST} ² Output Propagation Delay t _D ³ Glitch Energy ⁴ Full Scale Output Current ⁵ Spurious-Free Dynamic Range ⁶ 1.23 MHz; 10 MWPS 5.055 MHz; 20 MWPS 10.1 MHz; 50 MWPS 16 MHz; 40 MWPS Rise Time / Fall Time	+25 °C +25 °C +25 °C +25 °C +25 °C +25 °C 2 MHz Span 2 MHz Span 2 MHz Span 10 MHz Span R _L = 50 Ω	IV V V V V V	100	13 1 15 20.48 70 68 68 68 2		100	13 1 15 20.48 70 68 68 68 68 2		MWPS ns ns pV-s mA dBc dBc dBc dBc dBc ns
Power Supply Requirements Negative Supply Voltage Negative Supply Current (-5.2 V) Nominal Power Dissipation Power Supply Rejection Ratio	+25 ℃ Full Temp ±5% of V _{EE} External Ref, +25 ℃	IV I VI V	-5.46	-5.2 115 600 30	-4.94 140 148 100	-5.46	-5.2 115 600 30	-4.94 140 148 100	V mA mA mW µA/V

¹Gain is measured as a ratio of the full-scale current to I_{Set}. The ratio is nominally 128.

SPT9712

²Measured as voltage at mid-scale transition to $\pm 0.024\%$; R_L=50 Ω .

³Measured from the rising edge of Latch Enable to where the output signal has left a 1 LSB error band.

⁴Glitch is measured as the largest single transient.

⁵Calculated using I_{FS} = 128 x (Control Amp In / R_{Set})

⁶SFDR is defined as the difference in signal energy between the fundamental and worst case spurious frequencies in the output spectrum window, which is centered at the fundamental frequency and covers the indicated span.

ELECTRICAL SPECIFICATIONS

 $T_{A} = T_{MIN} - T_{MAX}, V_{EE} = -5.2 \text{ V}, R_{SET} = 7.5 \text{ k}\Omega, Control Amp In = Ref Out, } V_{OUT} = 0 \text{ V}, unless otherwise specified.}$

PARAMETERS	TEST CONDITIONS	TEST LEVEL	MIN	PT9712 TYP	A MAX	S MIN	PT9712 TYP	2B MAX	UNITS
Voltage Input and Control Reference Input Impedance Ref. Multiplying Bandwidth Internal Reference Voltage Internal Reference Voltage Drift Amplifier Input Impedance Amplifier Input Bandwidth	+25 ℃ +25 ℃ +25 ℃ +25 ℃	V V VI V V	-1.15	3 40 -1.20 50 3 1	-1.25	-1.15	3 40 -1.20 50 3 1	-1.25	kΩ MHz V ppm/°C MΩ MHz
Digital Inputs Logic 1 Voltage Logic 0 Voltage Logic 1 Current Logic 0 Current Input Capacitance Input Setup Time – ts Input Setup Time – ts Input Hold Time – t _H Input Hold Time – t _H Latch Pulse Width – t _{PWL} , t _{PWH}	Full Temp. Full Temp. Full Temp. Full Temp. +25 °C +25 °C Full Temp. +25 °C Full Temp. +25 °C Full Temp. +25 °C	VI VI VI V IV IV IV	-1.0 3 3.5 0.5 0.5 5.0	-0.8 -1.7 3 2 0 4.0	-1.5 20 10	-1.0 3 3.5 0.5 0.5 5.0	-0.8 -1.7 3 2 0 4.0	-1.5 20 10	V V µA µA pF ns ns ns ns

TEST LEVEL CODES

All electrical characteristics are subject to the following conditions:

All parameters having min/max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality Assurance inspection. Any blank section in the data column indicates that the specification is not tested at the specified condition.

LEVEL TEST PROCEDURE

IV

I 100% production tested at the specified temperature.

II 100% production tested at T_A = +25 °C , and sample tested at the specified temperatures.

III QA sample tested only at the specified temperatures.

Parameter is guaranteed (but not tested) by design and characterization data.

V Parameter is a typical value for information purposes only.

VI 100% production tested at $T_A = +25$ °C. Parameter is guaranteed over specified temperature range.

THEORY OF OPERATION

The SPT9712 uses a segmented architecture incorporating most significant bit (MSB) decoding. The four MSBs (D1–D4) are decoded to thermometer code lines to drive 15 discrete current sinks. For the eight least significant bits (LSBs), D5 and D6 are binary weighted and D7–D12 are applied to the R-2R network. The 12-bit decoded data is input to internal master/slave latches. The latched data is input to the switching network and is presented on the output pins as complementary current outputs.

TYPICAL INTERFACE CIRCUIT

The SPT9712 requires few external components to achieve the stated operation and performance. Figure 2 shows the typical interface requirements when using the SPT9712 in normal circuit operation. The following sections provide descriptions of the pin functions and outline critical performance criteria to consider for achieving optimal device performance.

POWER SUPPLIES AND GROUNDING

The SPT9712 requires the use of a single -5.2 V supply. All supplies should be treated as analog supply sources. This means the ground returns of the device should be connected to the analog ground plane. All supply pins should be bypassed with .01 μ F and 10 μ F decoupling capacitors as close to the device as possible.

The two grounds available on the SPT9712 are DGND and AGND. These grounds are not tied together internal to the device. The use of ground planes is recommended to achieve the best performance of the SPT9712. All ground, reference and analog output pins should be tied directly to the DAC ground plane. The DAC and system ground planes should be separate from each other and only connected at a single point through a ferrite bead to reduce ground noise pickup.

DIGITAL INPUTS AND TIMING

The SPT9712 uses single-ended, 10K ECL-compatible inputs for data inputs D1–D12 and Latch Enable. It also employs master/slave latches to simplify digital interface timing requirements and reduce glitch energy by synchronizing the current switches. This is an improvement over the AD9712, which typically requires external latches for digital input synchronization.

Referring to figure 1, data is latched into the DAC on the rising edge of the latch enable clock with the associated setup and hold times. The output transition occurs after a typical 1 ns propagation delay and settles to within ±1 LSB in typically 13 ns. Because of the SPT9712's rising-edge triggering, no timing changes are required when replacing an AD9712 operating in the transparent mode.

VOLTAGE REFERENCE

When using the internal reference, Ref Out should be connected to Control Amp In and decoupled with a 0.1 μ F capacitor. Control Amp Out should be connected to Ref In and decoupled to the analog supply. (See figure 2.)

Full-scale output current is determined by Control Amp In and R_{Set} using the following formula:

 I_{Out} (FS) = (Control Amp In / R_{Set}) x 128 (Current Out is a constant 128 factor of the reference current)

The internal reference is typically -1.20 V with a tolerance of ± 0.05 V and a typical drift of 50 ppm/°C. If greater accuracy or temperature stability is required, an external reference can be utilized.

OUTPUTS

The output of the SPT9712 is comprised of complementary current sinks, I_{Out} and $I_{\overline{Out}}$. The output current levels at either I_{Out} or $I_{\overline{Out}}$ are based upon the digital input code. The sum of the two is always equal to the full-scale output current minus one LSB.

By terminating the output current through a resistive load to ground, an associated voltage develops. The effective resistive load (R_{Eff}) is the output resistance of the device (R_{Out}) in parallel with the resistive load (R_{L}). The voltage which develops can be determined using the following formulas:

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Control Amp Out = -1.2 V, and R<sub>Set</sub> = 7.5 k\Omega I<sub>Out</sub> (FS) = (-1.2 V / 7.5 k\Omega) x 128 = -20.48 mA R<sub>L</sub> = 51 \Omega R<sub>Out</sub> = 1.0 k\Omega R<sub>Eff</sub> = 51 \Omega || 1.0 k\Omega = 48.52 \Omega V<sub>Out</sub> = R<sub>Eff</sub> x I<sub>Out</sub> (FS) = 48.52 \Omega x -20.48 mA = -0.994 V
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The resistive load of the SPT9712 can be modified to incorporate a wide variety of signal levels. However, optimal device performance is achieved when the outputs are equivalently loaded.

SPT9712

4 2/15/01

Figure 1 - Timing Diagram

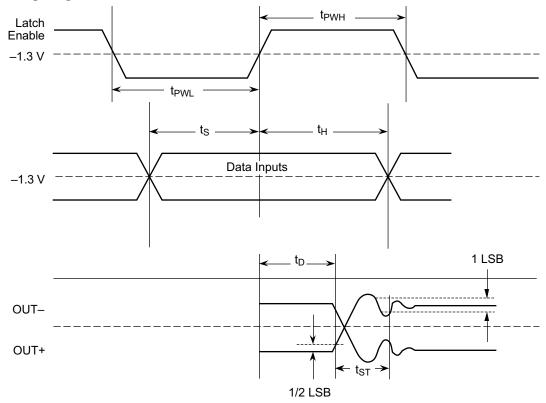
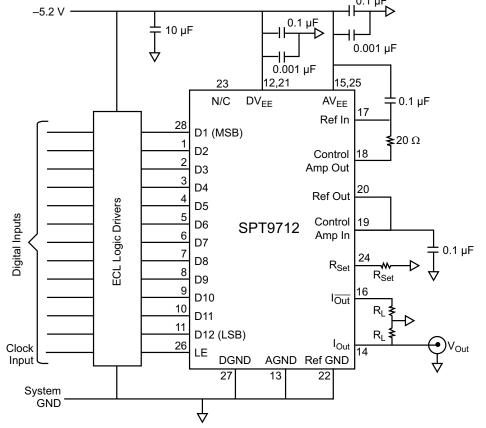
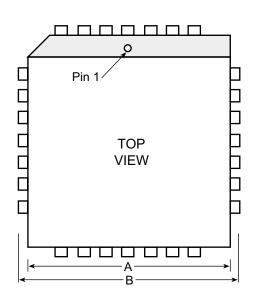


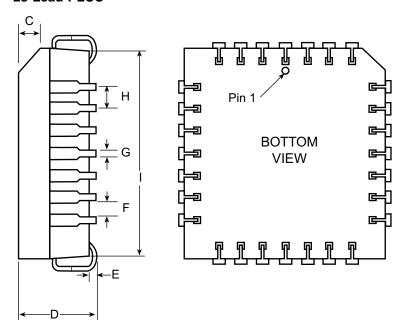
Figure 2 - Typical Interface Circuit



PACKAGE OUTLINE

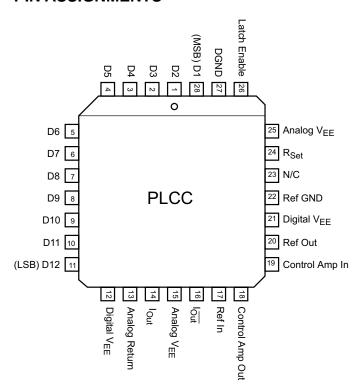
28-Lead PLCC





	INCH	IES	MILLIN	METERS
SYMBOL	MIN	MAX	MIN	MAX
Α	0.452	0.456	11.48	11.58
В	0.485	0.495	12.32	12.57
С	30		30)°
D	0.170	0.179	4.32	4.55
E	0.020	0.025	0.51	0.64
F	0.031	0.035	0.79	0.89
G	0.013	0.021	0.33	0.53
Н	0.048	0.052	1.22	1.32
	0.410	0.430	10.41	10.92

PIN ASSIGNMENTS



PIN FUNCTIONS

Name	Function
Out+	Analog Current Output
Out-	Complementary Analog Current Output
D1-D12	Digital Input Bits (D12 is the LSB)
Latch Enable	Latch Control Line
Ref In	Voltage Reference Input
Ref Out	Internal Voltage Reference Output Normally Connected to Control Amp In
Ref GND	Ground Return For Internal Voltage Reference and Amplifier
Control Amp In	Normally Connected to Ref Out If Not Connected to External Reference
Control Amp Out	Output of Internal Control Amplifier Normally Connected to Ref In
R _{Set} 1	Connection for External Resistance Reference When Using Internal Amplifier Nominally 7.5 $k\Omega$
Analog Return	Analog Return Ground
Analog V _{EE}	Analog Negative Supply (-5.2 V)
Digital V _{EE}	Digital Negative Supply (-5.2 V)
DGND	Digital Ground Return
N/C	Not Connected

¹Full-Scale Current Out = 128 (Control Amp In / R_{Set})

ORDERING INFORMATION

PART NUMBER	DNL/INL	TEMPERATURE RANGE	PACKAGE
SPT9712AIP	±0.75/±1.0	−40 to +85 °C	28L PLCC
SPT9712BIP	±1.25/±1.5	−40 to +85 °C	28L PLCC

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7

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SPT9712

2/15/01