## AS5140H

10 BIT $360^{\circ}$ PROGRAMMABLE MAGNETIC ROTARY ENCODER FOR HIGH AMBIENT TEMPERATUES

## 1 General Description

The AS5140H is a contactless magnetic rotary encoder for accurate angular measurement over a full turn of $360^{\circ}$ and over an extended ambient temperature range of $-40^{\circ} \mathrm{C} \ldots+150^{\circ} \mathrm{C}$.
It is a system-on-chip, combining integrated Hall elements, analog front end and digital signal processing in a single device.
To measure the angle, only a simple two-pole magnet, rotating over the center of the chip, is required. The magnet may be placed above or below the IC.

The absolute angle measurement provides instant indication of the magnet's angular position with a resolution of $0.35^{\circ}=1024$ positions per revolution. This digital data is available as a serial bit stream and as a PWM signal. Furthermore, a user-programmable incremental output is available.

An internal voltage regulator allows the AS5140H to operate at either 3.3 V or 5 V supplies.

The AS5140H is pin-compatible to the AS5040; however it uses low-voltage OTP programming cells with additional programming options.


Figure 1: Typical arrangement of AS5140H and magnet

### 1.1 Applications

Automotive applications:

- Engine compartment sensors
- Transmission gearbox encoder
- Throttle Valve position control

Industrial applications

- rotary sensors in high temperature environment


### 1.2 Benefits

- Complete system-on-chip
- Flexible system solution provides absolute, PWM and incremental outputs simultaneously
- Ideal for applications in harsh environments due to contactless position sensing
- Tolerant to magnet misalignment and airgap variations
- Tolerant to external magnetic fields
- Operates up to $+150^{\circ} \mathrm{C}$ ambient temperature
- No temperature compensation necessary
- No calibration required


### 1.3 Key Features

- Contactless high resolution rotational position encoding over a full turn of 360 degrees
- Two digital 10bit absolute outputs:
- Serial interface and
- Pulse width modulated (PWM) output
- Three incremental output modes:
- Quadrature A/B and Index output signal
- Step / Direction and Index output signal
- 3-phase commutation for brushless DC motors
- 10, 9, 8 or 7-bit user programmable resolution
- User programmable zero / index position
- Failure detection mode for magnet placement monitoring and loss of power supply
- Rotational speeds up to $10,000 \mathrm{rpm}$
- Push button functionality detects movement of magnet in Z -axis
- Serial read-out of multiple interconnected AS5140H devices using Daisy Chain mode
- Fully automotive qualified to AEC-Q100, grade 0
- Wide ambient temperature range: $-40^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
- $\quad$ Small Pb-free package: SSOP 16 ( $5.3 \mathrm{~mm} \times 6.2 \mathrm{~mm}$ )


## 2 Pin Configuration



Figure 2: Pin configuration SSOP16

### 2.1 Pin Description

Table 2 shows the description of each pin of the standard SSOP16 package (Shrink Small Outline Package, 16 leads, body size: $5.3 \mathrm{~mm} \times 6.2 \mathrm{mmm}$; see Figure 2).

Pins 7,15 and 16 are supply pins, pins 5,13 and 14 are for internal use and must not be connected.

Pins 1 and 2 are the magnetic field change indicators, MagINCn and MagDECn (magnetic field strength increase or decrease through variation of the distance between the magnet and the device). These outputs can be used to detect the valid magnetic field range. Furthermore those indicators can also be used for contact-less push-button functionality.
Pins 3, 4 and 6 are the incremental pulse output pins. The functionality of these pins can be configured through programming the one-time programmable (OTP) register:

| Output Mode | Pin 3 | Pin 4 | Pin 6 | Pin 12 |
| :--- | :--- | :--- | :--- | :--- |
| 1.x: quadrature | A | B | Index | PWM |
| 2.x:step/direction | LSB | Direction | Index | PWM |
| 3.x: commutation | U | V | W | LSB |

Table 1: Pin assignment for the different incremental output modes

## Mode 1.x: Quadrature A/B Output:

Represents the default quadrature $A / B$ signal mode.

## Mode 2.x: Step / Direction Output:

Configures pin 3 to deliver up to 512 pulses (up to 1024 state changes) per revolution. It is equivalent to the LSB (least significant bit) of the absolute position value. Pin 4 provides the information of the rotational direction.

Both modes (mode 1.x and mode 2.x) provide an index signal (1 pulse/revolution) with an adjustable width of one LSB or three LSB's.

| Pin | Symbol | Type | Description |
| :--- | :--- | :--- | :--- |
| 1 | MagINCn | DO_OD | Magnet Field Magnitude INCrease; <br> active low, indicates a distance <br> reduction between the magnet and the <br> device surface. |
| 2 | MagDECn | DO_OD | Magnet Field Magnitude DECrease; <br> active low, indicates a distance increase <br> between the device and the magnet. |
| 3 | A_LSB_U | DO | Mode1.x: Quadrature A channel <br> Mode2.x: Least Significant Bit <br> Mode3.x: U signal (phase1) |
| 4 | B_Dir_V | DO | Mode1.x: Quadrature B channel <br> quarter period shift to channel A. <br> Mode2.x: Direction of Rotation <br> Mode3.x: V signal (phase2) |
| 5 | NC | - | Must be left unconnected |

Table 2: Pin description SSOP16

| DO_OD | digital output open drain | S | supply pin |
| :--- | :--- | :--- | :--- |
| DO | digital output | DI | digital input |
| DI_PD | digital input pull-down | DO_T | digital output /tri-state |
| DI_PU | digital input pull-up | ST | Schmitt-Trigger input |

## Mode 3.x: Brushless DC Motor Commutation Mode:

In addition to the absolute encoder output over the SSI interface, this mode provides commutation signals for brushless DC motors with either one pole pair or two pole pair rotors. The commutation signals are usually provided by 3 discrete Hall switches, which are no longer required, as the AS5140H can fulfill two tasks in parallel: absolute encoder + BLDC motor commutation.

In this mode, pin 12 provides the LSB output instead of the PWM (Pulse-Width-Modulation) signal.

Pin 8 (Prog) is also used to program the different incremental interface modes, the incremental resolution and the zero position into the OTP (see page 11).

This pin is also used as digital input to shift serial data through the device in Daisy Chain configuration, (see page 5).

Pin 11 Chip Select (CSn; active low) selects a device within a network of AS5140H encoders and initiates serial data transfer. A logic high at CSn puts the data output pin (DO) to tri-state and terminates serial data transfer. This pin is also used for alignment mode (page 12 ) and programming mode (page 9).

Pin 12 allows a single wire output of the 10 -bit absolute position value. The value is encoded into a pulse width modulated signal with $1 \mu$ s pulse width per step ( $1 \mu$ s to $1024 \mu$ s over a full turn). By using an external low pass filter, the digital PWM signal is converted into an analog voltage, allowing a direct replacement of potentiometers.

## 3 Functional Description

The AS5140H is manufactured in a CMOS standard process and uses a spinning current Hall technology for sensing the magnetic field distribution across the surface of the chip.

The integrated Hall elements are placed around the center of the device and deliver a voltage representation of the magnetic field at the surface of the IC.

Through Sigma-Delta Analog / Digital Conversion and Digital Signal-Processing (DSP) algorithms, the AS5140H provides accurate high-resolution absolute angular
position information. For this purpose a Coordinate Rotation Digital Computer (CORDIC) calculates the angle and the magnitude of the Hall array signals.

The DSP is also used to provide digital information at the outputs MagINCn and MagDECn that indicate movements of the used magnet towards or away from the device's surface.

A small low cost diametrically magnetized (two-pole) standard magnet provides the angular position information (see Figure 17).
The AS5140H senses the orientation of the magnetic field and calculates a 10 -bit binary code. This code can be accessed via a Synchronous Serial Interface (SSI). In addition, an absolute angular representation is given by a Pulse Width Modulated signal at pin 12 (PWM).

Besides the absolute angular position information the device simultaneously provides incremental output signals. The various incremental output modes can be selected by programming the OTP mode register bits (see page 11). As long as no programming voltage is applied to pin Prog, the new setting may be overwritten at any time and will be reset to default when power is turned off. To make the setting permanent, the OTP register must be programmed (see page 9 ff .). The default setting is a quadrature $A / B$ mode including the Index signal with a pulse width of 1 LSB. The Index signal is logic high at the user programmable zero position.

The AS5140H is tolerant to magnet misalignment and magnetic stray fields due to differential measurement technique and Hall sensor conditioning circuitry.


Figure 3: AS5140H block diagram

## 4 10-bit Absolute Angular Position Output

### 4.1 Synchronous Serial Interface (SSI)



Figure 4: Synchronous serial interface with absolute angular position data

If CSn changes to logic low, Data Out (DO) will change from high impedance (tri-state) to logic high and the read-out will be initiated.

- After a minimum time tclk fe, data is latched into the output shift register with the first falling edge of CLK.
- Each subsequent rising CLK edge shifts out one bit of data.
- The serial word contains 16 bits, the first 10 bits are the angular information $D[9: 0]$, the subsequent 6 bits contain system information, about the validity of data such as OCF, COF, LIN, Parity and Magnetic Field status (increase/decrease).
- A subsequent measurement is initiated by a log. "high" pulse at CSn with a minimum duration of tcsn.


## Data Content:

D9:D0 absolute angular position data (MSB is clocked out first)

OCF (Offset Compensation Finished), logic high indicates the finished Offset Compensation Algorithm. For fast startup, this bit may be polled by the external microcontroller. As soon as this bit is set, the AS5140H has completed the startup and the data is valid (see Table 4)

COF (Cordic Overflow), logic high indicates an out of range error in the CORDIC part. When this bit is set, the data at D9:DO is invalid. The absolute output maintains the last valid angular value.

This alarm may be resolved by bringing the magnet within the $X-Y-Z$ tolerance limits.

LIN (Linearity Alarm), logic high indicates that the input field generates a critical output linearity. When this bit is set, the data at D9:D0 may still be
used, but can contain invalid data. This warning may be resolved by bringing the magnet within the $X-Y-Z$ tolerance limits.

MagINCn, (Magnitude Increase) becomes HIGH, when the magnet is pushed towards the IC, thus the magnetic field strength is increasing.

MagDECn, (Magnitude Decrease) becomes HIGH, when the magnet is pulled away from the IC, thus the magnetic field strength is decreasing.

Both signals HIGH indicate a magnetic field that is out of the allowed range (see Table 3).

| Mag <br> INCn | Mag <br> DECn | Description |
| :---: | :---: | :--- |
| 0 | 0 | No distance change <br> Magnetic Input Field OK (in range) |
| 0 | 1 | Distance increase: Pull-function. This state <br> is dynamic, it is only active while the magnet <br> is moving away from the chip in Z-axis |
| 1 | 0 | Distance decrease: Push- function. This <br> state is dynamic, it is only active while the <br> magnet is moving towards the chip in Z.- <br> axis. |
| 1 | 1 | Magnetic Input Field invalid - out of range: <br> too large, too small (missing magnet) |

Table 3: Magnetic magnitude variation indicator
Note: Pins 1 and 2 (MagINCn, MagDECn) are open drain outputs and require external pull-up resistors. If the magnetic field is in range, both outputs are turned off.
The two pins may also be combined with a single pull-up resistor. In this case, the signal is high when the magnetic field is in range. It is low in all other cases (see Table 3).

Even Parity bit for transmission error detection of bits 1... 15 (D9...D0, OCF, COF, LIN, MagINCn, MagDECn)

The absolute angular output is always set to a resolution of 10 bit. Placing the magnet above the chip, angular values increase in clockwise direction by default.

Data $\mathrm{D} 9: \mathrm{DO}$ is valid, when the status bits have the following configurations:

| OCF | COF | LIN | Mag INCn | Mag DECn | Parity |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 0 | 0 | 0 | 0 | even checksum of bits 1:15 |
|  |  |  | 0 | 1 |  |
|  |  |  | 1 | 0 |  |

Table 4: Status bit outputs
The absolute angular position is sampled at a rate of $10 \mathrm{kHz}(0.1 \mathrm{~ms})$. This allows reading of all 1024 positions per 360 degrees within 0.1 seconds $=9.76 \mathrm{~Hz}(\sim 10 \mathrm{~Hz})$ without skipping any position. Multiplying 10 Hz by 60 , results the corresponding maximum rotational speed of 600 rpm .

Readout of every second angular position allows for rotational speeds of up to 1200 rpm .

Consequently, increasing the rotational speed reduces the number of absolute angular positions per revolution (see Table 8). Regardless of the rotational speed or the number of positions to be read out, the absolute angular value is always given at the highest resolution of 10 bit.

The incremental outputs are not affected by rotational speed restrictions due to the implemented interpolator. The incremental output signals may be used for highspeed applications with rotational speeds of up to 10,000 rpm without missing pulses.

### 4.2 Daisy Chain Mode

The Daisy Chain mode allows connection of several AS5140H's in series, while still keeping just one digital input for data transfer (see "Data $\operatorname{IN}$ " in Figure 5 below). This mode is accomplished by connecting the data output (DO; pin 9) to the data input (Prog; pin 8) of the subsequent device. The serial data of all connected devices is read from the DO pin of the first device in the chain. The Prog pin of the last device in the chain should be connected to VSS. The length of the serial bit stream increases with every connected device, it is
n * (16+1) bits:
e.g. 34 bit for two devices, 51 bit for three devices, etc...

The last data bit of the first device (Parity) is followed by a logic low bit and the first data bit of the second device (D9), etc... (see Figure 6).

### 4.2.1 Programming Daisy Chained Devices

In Daisy Chain mode, the Prog pin is connected directly to the DO pin of the subsequent device in the chain (see Figure 5). During programming (see section 9), a programming voltage of 7.5 V must be applied to pin Prog. This voltage level exceeds the limits for pin DO, so one of the following precautions must be made during programming:

- open the connection DO $\rightarrow$ Prog during programming or
- add a Schottky diode between DO and Prog (Anode = DO, Cathode $=$ Prog)
Due to the parallel connection of CLK and CSn, all connected devices may be programmed simultaneously.


Figure 5: Daisy Chain hardware configuration


Figure 6: Daisy Chain mode data transfer

## 5 Incremental Outputs

Three different incremental output modes are possible with quadrature $A / B$ being the default mode.
Figure 7 shows the two-channel quadrature as well as the step / direction incremental signal (LSB) and the direction bit in clockwise (CW) and counter-clockwise (CCW) direction.

### 5.1 Quadrature A/B Output (Quad A/B Mode)

The phase shift between channel $A$ and $B$ indicates the direction of the magnet movement. Channel A leads channel B at a clockwise rotation of the magnet (top view) by 90 electrical degrees. Channel B leads channel $A$ at a counter-clockwise rotation.

### 5.2 LSB Output (Step / Direction Mode)

Output LSB reflects the LSB (least significant bit) of the programmed incremental resolution (OTP Register Bit Div0, Div1). Output Dir provides information about the rotational direction of the magnet, which may be placed above or below the device ( $1=$ clockwise; $0=$ counter clockwise; top view). Dir is updated with every LSB change.
In both modes (quad $A / B$, step/direction) the resolution and the index output are user programmable. The index pulse indicates the zero position and is by default one angular step (1LSB) wide. However, it can be set to three LSBs by programming the Index-bit of the OTP register accordingly (see Table 7).


Figure 7: Incremental output modes

### 5.2.1 Incremental Power-up Lock Option

After power-up, the incremental outputs can optionally be locked or unlocked, depending on the status of the CSn pin:

CSn = low at power-up:
CSn has an internal pull-up resistor and must be externally pulled low ( $R_{\text {ext }} \leq 5 k \Omega$ ). If $C$ sn is low at power-up, the incremental outputs (A, B, Index) will be high until the internal offset compensation is finished.
This unique state $(A=B=$ Index $=$ high $)$ may be used as an indicator for the external controller to shorten the waiting time at power-up. Instead of waiting for the specified maximum power up-time (0), the controller can start requesting data from the A 5140 H as soon as the state ( $\mathrm{A}=\mathrm{B}=$ Index $=$ high ) is cleared.

CSn = high or open at power-up:
In this mode, the incremental outputs (A, B, Index) will remain at logic high state, until CSn goes low or a low pulse is applied at CSn. This mode allows intentional disabling of the incremental outputs until for example the system microcontroller is ready to receive data.

### 5.3 Incremental Output Hysteresis

To avoid flickering incremental outputs at a stationary magnet position, a hysteresis is introduced.

In case of a rotational direction change, the incremental outputs have a hysteresis of 2 LSB.
Regardless of the programmed incremental resolution, the hysteresis of 2 LSB always corresponds to the highest resolution of 10 bit. In absolute terms, the hysteresis is set to 0.704 degrees for all resolutions.

For constant rotational directions, every magnet position change is indicated at the incremental outputs (see Figure 8). If for example the magnet turns clockwise from position " $x+3$ " to " $x+4$ ", the incremental output would also indicate this position accordingly.

A change of the magnet's rotational direction back to position " $x+3$ " means, that the incremental output still remains unchanged for the duration of 2 LSB, until position " $x+2$ " is reached. Following this direction, the incremental outputs will again be updated with every change of the magnet position.


Figure 8: Hysteresis window for incremental outputs

## 6 Pulse Width Modulation (PWM) Output

The AS5140H provides a pulse width modulated output (PWM), whose duty cycle is proportional to the measured angle:
Position $=\frac{t_{\text {on }} \cdot 1025}{\left(t_{\text {on }}+t_{\text {off }}\right)}-1$
The PWM frequency is internally trimmed to an accuracy of $\pm 5 \% ~(~ \pm 10 \%$ over full temperature range). This tolerance can be cancelled by measuring the complete duty cycle as shown above.


Figure 9: PWM output signal

| Parameter | Symbol | Typ | Unit | Note |
| :---: | :---: | :---: | :---: | :--- |
| PWM <br> frequency | fPwM | 0.9756 | kHz | Signal period: $1025 \mu \mathrm{~s}$ |
| MIN pulse <br> width | PWMIN | 1 | $\mu \mathrm{~s}$ | - Position 0d <br> - Angle 0 deg |
| MAX pulse <br> width | PWMAX | 1024 | $\mu \mathrm{~s}$ | - Position 1023d <br> - Angle 359,65 deg |

Table 5: PWM signal parameters

## 7 Analog Output

An analog output may be generated by averaging the PWM signal, using an external active or passive lowpass filter. The analog output voltage is proportional to the angle: $0^{\circ}=0 \mathrm{~V} ; 360^{\circ}=\mathrm{VDD5V}$.
Using this method, the AS5140H can be used as direct replacement of potentiometers.


Figure 10: Simple passive $\mathbf{2 d}^{\text {nd }}$ order lowpass filter

## $\mathrm{R} 1, \mathrm{R} 2 \geq 4 \mathrm{k} 7 \quad \mathrm{C} 1, \mathrm{C} 2 \geq 1 \mu \mathrm{~F} / 6 \mathrm{~V}$

R1 should be $\geq 4 k 7$ to avoid loading of the PWM output. Larger values of $R x$ and $C x$ will provide better filtering and less ripple, but will also slow down the response time.

## 8 Brushless DC Motor Commutation Mode

Brushless DC motors require angular information for stator commutation. The AS5140H provides U-V-W commutation signals for one and two pole pair motors. In addition to the three-phase output signals, the step (LSB) output at pin 12 allows high accuracy speed measurement. Two resolutions (9 or 10 bit) can be selected by programming Div0 according to Table 7.
Mode 3.0 (3.1) is used for brush-less DC motors with one-pole pair rotors. The three phases (U, V, W) are 120 degrees apart, each phase is 180 degrees on and 180 degrees off.

Mode 3.2 (3.3) is used for motors with two pole pairs requiring a higher pulse count to ensure a proper current commutation. In this case the pulse width is 256 positions, equal to 90 degrees.

The precise physical angle at which the $\mathrm{U}, \mathrm{V}$ and W signals change state ("Angle" in Figure 11 and Figure 12) is calculated by multiplying each transition position by the angular value of 1 count:
Angle [deg] = Position $\times$ (360 degree / 1024)


Figure 11: U, V and V-signals for BLDC motor commutation (Div1=0, Div0=0)


Figure 12: U, V and W-signals for 2pole BLDC motor commutation (Div1=1; Div0=0)

## 9 Programming the AS5140H

Note: A detailed description of the austriamicrosystems low voltage polyfuse OTP programming method is given in Application Note AN5000-30, which can be downloaded from the austriamicrosystems website. The OTP programming description in this datasheet is for general information only.

After power-on, programming the AS5140H is enabled with the rising edge of CSn with Prog $=$ high and CLK $=$ low.

The AS5140H programming is a one-time-programming (OTP) method, based on polysilicon fuses. The advantage of this method is that a programming voltage of only 3.3 V is required for programming.
The OTP consists of 52 bits, of which 21 bits are available for user programming. The remaining 31 bits contain factory settings and a unique chip identifier (Chip-ID).

A single OTP cell can be programmed only once. Per default, the cell is " 0 "; a programmed cell will contain a " 1 ". While it is not possible to reset a programmed bit from " 1 " to " 0 ", multiple OTP writes are possible, as long as only unprogrammed " 0 "-bits are programmed to " 1 ".
Independent of the OTP programming, it is possible to overwrite the OTP register temporarily with an OTP write command at any time. This setting will be cleared and overwritten with the hard programmed OTP settings at each power-up sequence or by a LOAD operation.

The OTP memory can be accessed in several ways:
The OTP memory can be accessed in several ways:

- Load Operation

The Load operation reads the OTP fuses and loads the contents into the OTP register. Note: a Load operation is automatically executed after each power-on-reset.

- Write Operation

The Write operation allows a temporary modification of the OTP register. It does not program the OTP. This operation can be invoked multiple times and will remain set while the chip is supplied with power and while the OTP register is not modified with another Write or Load operation.

- Read Operation

The Read operation reads the contents of the OTP register, for example to verify a Write command or to read the OTP memory after a Load command

- Program Operation

The Program operation writes the contents of the OTP register permanently into the OTP ROM.

- Analog Readback Operation

The Analog Readback operation allows a quantifiable verification of the programming. For each programmed or unprogrammed bit, there is a representative analog value (in essence, a resistor value) that is read to verify whether a bit has been successfully programmed or not

### 9.1 OTP Memory Assignment

| Bit | Symbol | Function |  |
| :---: | :---: | :---: | :---: |
|  | mbit1 | Factory Bit |  |
| 51 | Md0 | Incremental Output Mode Selection | 흠00$\vdots$$\vdots$00000 |
| 50 | Md1 |  |  |
| 49 | Div0 |  |  |
| 48 | Div1 |  |  |
| 47 | Index |  |  |
| 46 | Z0 | 10 bit Zero Position |  |
| : | : |  |  |
| 37 | Z9 |  |  |
| 36 | CCW | Direction |  |
| 35 | RAO | Redundancy Address |  |
| : | : |  |  |
| 31 | RA4 |  |  |
| 30 | FS 0 | Factory Bit |  |
| : | FS 1 | Factory Bit |  |
| 18 | FS 12 | Factory Bit |  |
| 17 | ChipID0 | 18 bit Chip ID | $\begin{aligned} & \text { 든 } \\ & \text { 心 } \\ & \text { 응 } \end{aligned}$ |
| 16 | ChipID1 |  |  |
| : | : |  |  |
| 0 | ChipID17 |  |  |
|  | mbit0 | Factory Bit |  |

Table 6: OTP Bit Assignment

### 9.2 User selectable settings

The AS5140H allows programming of the following user selectable options:

| Md1, Md0 | Incremental Output Mode Selection |
| :--- | :--- |
| Div1,Div0 | Divider Setting of Incremental Output |
| Index | Index Pulse Width Selection: 1LSB / 3LSB |
| Z [9:0] | Programmable Zero / Index Position |
| CCW | Counter Clockwise Bit |

$\mathrm{ccw}=0$ - angular value increases in clockwise direction $\mathrm{ccw}=1$ - angular value increases in counterclockwise direction
RA [4:0] Redundant Address: an OTP bit location addressed by this address is always set to " 1 " independent of the corresponding original OTP bit setting

### 9.3 OTP Default Setting

The AS5140H can also be operated without programming. The default, un-programmed setting is:

Md0, MD1: 00 = incremental mode = quadrature
Div0,Div1 : $00=$ incremental resolution $=10$ bit
Index: $0 \quad=$ Index bit width $=1 \mathrm{LSB}$
Z9 to ZO: 00 = no programmed zero position
CCW: $0 \quad$ = clockwise operation
RA4 to RA0:0 = no OTP bit is selected

### 9.4 Redundant programming option

In addition to the regular programming, a redundant programming option is available. This option allows that one selectable OTP bit can be set to "1" (programmed state) by writing the location of that bit into a 5 -bit address decoder. This address can be stored in bits RA5..0 in the OTP user settings.

Example: setting RA5... 0 to " 00001 " will select bit $51=$ MD0, " 00010 " selects bit $50=$ MD1, " 10000 " selects bit $36=$ CCW, etc..

### 9.5 OTP register entry and exit condition



Figure 13:OTP access timing diagram

To avoid accidental modification of the OTP during normal operation, each OTP access (Load, Write, Read, Program) requires a defined entry and exit procedure, using the CSn, PROG and CLK signals as shown in Figure 13.

### 9.6 Incremental Mode Programming

Three different incremental output modes are available.
Mode: $\operatorname{Md1}=0 / \mathrm{Md} 0=1$ sets the AS 5140 H in quadrature mode.

Mode: Md1=1 / MdO=0 sets the AS5140H in step / direction mode (see Table 1)
In both modes, the incremental resolution may be reduced from 10 bit down to 9,8 or 7 bit using the divider OTP bits Div1 and Div0. (see Table 7 below).

Mode: Md1=1 / Md0=1 sets the AS5140H in brushless DC motor commutation mode with an additional LSB incremental signal at pin 12 (PWM_LSB).
To allow programming of all bits, the default factory setting is all bits $=0$. This mode is equal to mode 1:0 (quadrature A/B, 1LSB index width, 256 ppr ).

The absolute angular output value, by default, increases with clockwise rotation of the magnet (top view). Setting the CCW-bit (see Table 6) allows for reversing the
indicated direction, e.g. when the magnet is placed underneath the IC:

CCW $=0$ - angular value increases clockwise;
CCW = 1 - angular value increases counterclockwise.
By default, the zero / index position pulse is one LSB wide. It can be increased to a three LSB wide pulse by setting the Index-bit of the OTP register.

Further programming options (commutation modes) are available for brushless DC motor-control.
$\mathrm{Md} 1=\mathrm{Md} 0=1$ changes the incremental output pins 3, 4 and 6 to a 3 -phase commutation signal. Div1 defines the number of pulses per revolution for either a two-pole (Div1=0) or four-pole (Div1=1) rotor.
In addition, the LSB is available at pin 12 (the LSB signal replaces the PWM-signal), which allows for high rotational speed measurement of up to $10,000 \mathrm{rpm}$.

|  | OTP-Mode-Register-Bit |  |  |  |  | Pin \# |  |  |  | Pulses per Revolution ppr | Incremental Resolution bit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mode | Md1 | MdO | Div1 | Div0 | Index | 3 | 4 | 6 | 12 |  |  |
| Default (Mode0.0) | 0 | 0 | 0* | 0* | 0* | A | B | 1LSB | PWM 10 bit | 2x256 | 10 |
| quadAB-Mode1.0 | 0 | 1 | 0 | 0 | 0 |  |  | 1LSB |  |  |  |
| quadAB-Mode1.1 | 0 | 1 | 0 | 0 | 1 |  |  | 3LSBs |  |  |  |
| quadAB-Mode1.2 | 0 | 1 | 0 | 1 | 0 |  |  | 1LSB |  | 2x128 | 9 |
| quadAB-Mode1.3 | 0 | 1 | 0 | 1 | 1 |  |  | 3LSBs |  |  |  |
| quadAB-Mode1.4 | 0 | 1 | 1 | 0 | 0 |  |  | 1LSB |  | $2 \times 64$ | 8 |
| quadAB-Mode1.5 | 0 | 1 | 1 | 0 | 1 |  |  | 3LSBs |  |  |  |
| quadAB-Mode1.6 | 0 | 1 | 1 | 1 | 0 |  |  | 1LSB |  | 2x32 | 7 |
| quadAB-Mode1.7 | 0 | 1 | 1 | 1 | 1 |  |  | 3LSBs |  |  |  |
| Step/Dir-Mode2.0 | 1 | 0 | 0 | 0 | 0 | LSB | Dir | 1LSB | PWM$10 \text { bit }$ | 512 | 10 |
| Step/Dir-Mode2.1 | 1 | 0 | 0 | 0 | 1 |  |  | 3LSBs |  |  |  |
| Step/Dir -Mode2.2 | 1 | 0 | 0 | 1 | 0 |  |  | 1LSB |  | 256 | 9 |
| Step/Dir -Mode2.3 | 1 | 0 | 0 | 1 | 1 |  |  | 3LSBs |  |  |  |
| Step/Dir -Mode2.4 | 1 | 0 | 1 | 0 | 0 |  |  | 1LSB |  | 128 | 8 |
| Step/Dir -Mode2.5 | 1 | 0 | 1 | 0 | 1 |  |  | 3LSBs |  |  |  |
| Step/Dir -Mode2.6 | 1 | 0 | 1 | 1 | 0 |  |  | 1LSB |  | 64 | 7 |
| Step/Dir -Mode2.7 | 1 | 0 | 1 | 1 | 1 |  |  | 3LSBs |  |  |  |
| Commutation-Mode3.0 | 1 | 1 | 0 | 0 | 0 | $\mathrm{U}\left(0^{\circ}\right)$ | $V\left(120^{\circ}\right)$ | W(240 ${ }^{\circ}$ | LSB | $3 \times 1$ | 10 |
| Commutation-Mode3.1 | 1 | 1 | 0 | 1 | 0 |  |  |  |  |  | 9 |
| Commutation-Mode3.2 | 1 | 1 | 1 | 0 | 0 | $\begin{gathered} U^{\prime} \\ \left(0^{\circ}, 180^{\circ}\right) \end{gathered}$ | $\begin{gathered} V^{\prime} \\ \left(60^{\circ}, 240^{\circ}\right) \end{gathered}$ | $\begin{array}{\|c\|} \hline \text { W' } \\ \left(120^{\circ}, 300^{\circ}\right) \end{array}$ | LSB | $2 \times 3$ | 10 |
| Commutation-Mode3.3 | 1 | 1 | 1 | 1 | 0 |  |  |  |  |  | 9 |

Table 7: One Time Programmable (OTP) register options
*Note: Div1, Div0 and Index cannot be programmed in Mode 0:0

### 9.7 Zero Position Programming

Zero position programming is an OTP option that simplifies assembly of a system, as the magnet does not need to be manually adjusted to the mechanical zero position. Once the assembly is completed, the mechanical and electrical zero positions can be matched by software. Any position within a full turn can be defined as the permanent new zero/index position.

For zero position programming, the magnet is turned to the mechanical zero position (e.g. the "off"-position of a rotary switch) and the actual angular value is read.

## 10 Alignment Mode

The alignment mode simplifies centering the magnet over the center of the chip to gain maximum accuracy.

Alignment mode can be enabled with the falling edge of CSn while Prog = logic high (Figure 14). The Data bits D9-D0 of the SSI change to a 10 -bit displacement amplitude output. A high value indicates large X or Y displacement, but also higher absolute magnetic field strength. The magnet is properly aligned, when the difference between highest and lowest value over one full turn is at a minimum.

Under normal conditions, a properly aligned magnet will result in a reading of less than 128 over a full turn. The MagINCn and MagDECn indicators will be $=1$ when the alignment mode reading is < 128. At the same time, both hardware pins MagINCn (\#1) and MagDECn (\#2) will be pulled to VSS. A properly aligned magnet will therefore produce a MagINCn $=\mathrm{MagDECn}=1$ signal throughout a full $360^{\circ}$ turn of the magnet.
Stronger magnets or short gaps between magnet and IC may show values larger than 128. These magnets are still properly aligned as long as the difference between highest and lowest value over one full turn is at a minimum.

This value is written into the OTP register bits Z9:Z0 (see Figure 13) and programmed as described in section 9.

This new absolute zero position is also the new Index pulse position for incremental output modes.

Note: The zero position value may also be modified before programming, e.g. to program an electrical zero position that is $180^{\circ}$ (half turn) from the mechanical zero position, just add 512 to the value read at the mechanical zero position and program the new value into the OTP register.

The Alignment mode can be reset to normal operation by a power-on-reset (disconnect / re-connect power supply) or by a falling edge on CSn with Prog $=$ low.


Figure 14: Enabling the alignment mode


Figure 15: Exiting alignment mode

## $113.3 \mathrm{~V} / 5 \mathrm{~V}$ Operation

The AS5140H operates either at $3.3 \mathrm{~V} \pm 10 \%$ or at 5 V $\pm 10 \%$. This is made possible by an internal 3.3 V LowDropout (LDO) voltage regulator. The internal supply voltage is always taken from the output of the LDO, meaning that the internal blocks are always operating at 3.3V.

For 3.3V operation, the LDO must be bypassed by connecting VDD3V3 with VDD5V (see Figure 16).

For 5 V operation, the 5 V supply is connected to pin VDD5V, while VDD3V3 (LDO output) must be buffered by a $2,2 . .10 \mu \mathrm{~F}$ capacitor, which is supposed to be placed close to the supply pin (see Figure 16).

The VDD3V3 output is intended for internal use only It must not be loaded with an external load.

The output voltage of the digital interface I/O's corresponds to the voltage at pin VDD5V, as the I/O buffers are supplied from this pin (see Figure 16).

A buffer capacitor of 100 nF is recommended in both cases close to pin VDD5V. Note that pin VDD3V3 must always be buffered by a capacitor. It must not be left floating, as this may cause an instable internal 3.3 V supply voltage which may lead to larger than normal jitter of the measured angle.


### 3.3V Operation



Figure 16: Connections for 5V / 3.3V supply voltages

## 12 Choosing the Proper Magnet

Typically the magnet should be 6 mm in diameter and $\geq 2.5 \mathrm{~mm}$ in height. Magnetic materials such as rare earth AINiCo, SmCo5 or NdFeB are recommended.

The magnet's field strength perpendicular to the die surface should be verified using a gauss-meter. The magnetic field $B_{v}$ at a given distance, along a concentric circle with a radius of 1.1 mm (R1), should be in the range of $\pm 45 \mathrm{mT} \ldots \pm 75 \mathrm{mT}$. (see Figure 17).


Figure 17: Typical magnet and magnetic field distribution

### 12.1 Physical Placement of the Magnet

The best linearity can be achieved by placing the center of the magnet exactly over the defined center of the IC package as shown in Figure 18:


Figure 18: Defined IC center and magnet displacement radius

## Magnet Placement:

The magnet's center axis should be aligned within a displacement radius $\mathrm{R}_{\mathrm{d}}$ of 0.25 mm from the defined center of the IC with reference to the edge of pin \#1 (see Figure 18). This radius includes the placement tolerance of the chip within the SSOP-16 package ( $+/-0.235 \mathrm{~mm}$ ). The displacement radius $R_{d}$ is 0.485 mm with reference to the center of the chip (see section 10:Alignment Mode:).

The vertical distance should be chosen such that the magnetic field on the die surface is within the specified limits (see Figure 17). The typical distance "z" between the magnet and the package surface is 0.5 mm to 1.8 mm with the recommended magnet ( $6 \mathrm{~mm} \times 2.5 \mathrm{~mm}$ ). Larger gaps are possible, as long as the required magnetic field strength stays within the defined limits.

A magnetic field outside the specified range may still produce usable results, but the out-of-range condition will be indicated by MagINCn (pin 1) and MagDECn (pin $2)$, see Table 3.


Figure 19: Vertical placement of the magnet

## 13 Simulation Modelling



Figure 20: Arrangement of Hall sensor array on chip (principle)
With reference to Figure 20, a diametrically magnetized permanent magnet is placed above or below the surface of the AS5140H. The chip uses an array of Hall sensors to sample the vertical vector of a magnetic field distributed across the device package surface. The area of magnetic sensitivity is a circular locus of 1.1 mm radius with respect to the center of the die. The Hall sensors in the area of magnetic sensitivity are grouped and configured such that orthogonally related components of the magnetic fields are sampled differentially.
The differential signal $\mathrm{Y} 1-\mathrm{Y} 2$ will give a sine vector of the magnetic field. The differential signal X1-X2 will give an orthogonally related cosine vector of the magnetic field.

The angular displacement $(\Theta)$ of the magnetic source with reference to the Hall sensor array may then be modelled by:
$\Theta=\arctan \frac{(Y 1-Y 2)}{(X 1-X 2)} \pm 0.5^{\circ}$
The $\pm 0.5^{\circ}$ angular error assumes a magnet optimally aligned over the center of the die and is a result of gain mismatch errors of the AS5140H. Placement tolerances of the die within the package are $\pm 0.235 \mathrm{~mm}$ in X and Y direction, using a reference point of the edge of pin \#1 (Figure 20).

In order to neglect the influence of external disturbing magnetic fields, a robust differential sampling and ratiometric calculation algorithm has been implemented. The differential sampling of the sine and cosine vectors
removes any common mode error due to DC components introduced by the magnetic source itself or external disturbing magnetic fields. A ratiometric division of the sine and cosine vectors removes the need for an accurate absolute magnitude of the magnetic field and thus accurate $Z$-axis alignment of the magnetic source.

The recommended differential input range of the magnetic field strength $\left(\mathrm{B}_{\left(\mathrm{X}_{1}-\mathrm{X}_{2}\right)}, \mathrm{B}_{\left(\mathrm{Y} 1-\mathrm{Y}_{2}\right)}\right)$ is $\pm 75 \mathrm{mT}$ at the surface of the die. In addition to this range, an additional offset of $\pm 5 \mathrm{mT}$, caused by unwanted external stray fields is allowed.

The chip will continue to operate, but with degraded output linearity, if the signal field strength is outside the recommended range. Too strong magnetic fields will introduce errors due to saturation effects in the internal preamplifiers. Too weak magnetic fields will introduce errors due to noise becoming more dominant.

## 14 Failure Diagnostics

The AS5140H also offers several diagnostic and failure detection features:

### 14.1 Magnetic Field Strength Diagnosis

By software: the MagINCn and MagDECn status bits will both be high when the magnetic field is out of range.

By hardware: Pins \#1 (MagINCn) and \#2 (MagDECn) are open-drain outputs and will both be turned on (= low with external pull-up resistor) when the magnetic field is out of range. If only one of the outputs is low, the magnet is either moving towards the chip (MagINCn) or away from the chip (MagDECn).

### 14.2 Power Supply Failure Detection

By software: If the power supply to the AS5140H is interrupted, the digital data read by the SSI will be all "0"s. Data is only valid, when bit OCF is high, hence a data stream with all "0"s is invalid. To ensure adequate low levels in the failure case, a pull-down resistor ( $\sim 10 \mathrm{k} \Omega$ ) should be added between pin DO and VSS at the receiving side.

By hardware: The MagINCn and MagDECn pins are open drain outputs and require external pull-up resistors. In normal operation, these pins are high ohmic and the outputs are high (see Table 3). In a failure case, either when the magnetic field is out of range or the power supply is missing, these outputs will become low. To ensure adequate low levels in case of a broken power
supply to the A 5140 H , the pull-up resistors ( $>10 \mathrm{k} \Omega$ ) from each pin must be connected to the positive supply at pin 16 (VDD5V).

By hardware: PWM output: The PWM output is a constant stream of pulses with 1 kHz repetition frequency. In case of power loss, these pulses are missing.

By hardware: Incremental outputs: In normal operation, pins A(\#3), B(\#4) and Index (\#6) will never be high at the same time, as Index is only high when $A=B=l o w$. However, after a power-on-reset, if VDD is powered up or restarts after a power supply interruption, all three outputs will remain in high state until pin CSn is pulled low. If CSn is already tied to VSS during power-up, the incremental outputs will all be high until the internal offset compensation is finished (within tpwrup).

## 15 Angular Output Tolerances

### 15.1 Accuracy

Accuracy is defined as the error between measured angle and actual angle. It is influenced by several factors:

- the non-linearity of the analog-digital converters,
- internal gain and mismatch errors,
- non-linearity due to misalignment of the magnet

As a sum of all these errors, the accuracy with centered magnet $=\left(E_{r r}{ }_{m a x}-E r r_{\text {min }}\right) / 2$ is specified as better than $\pm 0.5$ degrees @ $25^{\circ} \mathrm{C}$ (see Figure 22).

Misalignment of the magnet further reduces the accuracy. Figure 21 shows an example of a 3D-graph displaying non-linearity over XY-misalignment. The center of the square XY-area corresponds to a centered magnet (see dot in the center of the graph). The $X$ - and $Y$ - axis extends to a misalignment of $\pm 1 \mathrm{~mm}$ in both directions. The total misalignment area of the graph
covers a square of $2 \times 2 \mathrm{~mm}$ ( $79 \times 79 \mathrm{mil}$ ) with a step size of $100 \mu \mathrm{~m}$.

For each misalignment step, the measurement as shown in Figure 22 is repeated and the accuracy
$\left(\right.$ Errmax $_{\text {- }} \mathrm{Errmin}_{\text {n }} / 2$ (e.g. $0.25^{\circ}$ in Figure 22) is entered as the $Z$-axis in the 3D-graph.


Figure 21: Example of linearity error over XY misalignment
The maximum non-linearity error on this example is better than $\pm 1$ degree (inner circle) over a misalignment radius of $\sim 0.7 \mathrm{~mm}$. For volume production, the placement tolerance of the IC within the package ( $\pm 0.235 \mathrm{~mm}$ ) must also be taken into account.

The total nonlinearity error over process tolerances, temperature and a misalignment circle radius of 0.25 mm is specified better than $\pm 1.4$ degrees.

The magnet used for this measurement was a cylindrical $N d F e B$ (Bomatec® BMN-35H) magnet with 6 mm diameter and 2.5 mm in height.


Figure 22: Example of linearity error over $360^{\circ}$

### 15.2 Transition Noise

Transition noise is defined as the jitter in the transition between two steps.

Due to the nature of the measurement principle (Hall sensors + Preamplifier + ADC), there is always a certain degree of noise involved.

This transition noise voltage results in an angular transition noise at the outputs. It is specified as 0.06 degrees rms (1 sigma) ${ }^{* 1}$.

This is the repeatability of an indicated angle at a given mechanical position.

The transition noise has different implications on the type of output that is used:

- Absolute output; SSI interface:

The transition noise of the absolute output can be reduced by the user by applying an averaging of readings. An averaging of 4 readings will reduce the transition noise by $50 \%=0.03^{\circ} \mathrm{rms}(1$ sigma).

- PWM interface:

If the PWM interface is used as an analog output by adding a low pass filter, the transition noise can be reduced by lowering the cutoff frequency of the filter.
If the PWM interface is used as a digital interface with a counter at the receiving side, the transition noise may again be reduced by averaging of readings.

- Incremental mode:

In incremental mode, the transition noise influences the period, width and phase shift of the output signals A, B and Index. However, the algorithm used to generate the incremental outputs guarantees no missing or additional pulses even at high speeds (up to $10,000 \mathrm{rpm}$ and higher)
${ }^{*}$ : statistically, 1 sigma represents $68.27 \%$ of readings,
3 sigma represents $99.73 \%$ of readings.

### 15.3 High Speed Operation

### 15.3.1 Sampling Rate

The AS5140H samples the angular value at a rate of 10.42k samples per second. Consequently, the incremental, as well as the absolute outputs are updated each $96 \mu \mathrm{~s}$.
At a stationary position of the magnet, this sampling rate creates no additional error.

## Absolute Mode:

With the given sampling rate of 10.4 kHz , the number of samples ( $n$ ) per turn for a magnet rotating at high speed can be calculated by:

$$
n=\frac{60}{r p m \cdot 96 \mu s}
$$

In practice, there is no upper speed limit. The only restriction is that there will be fewer samples per revolution as the speed increases.
Regardless of the rotational speed, the absolute angular value is always sampled at the highest resolution of 10 bit.
Likewise, for a given number of samples per revolution ( $n$ ), the maximum speed can be calculated by:

$$
r p m=\frac{60}{n \cdot 96 \mu \mathrm{~s}}
$$

In absolute mode (serial interface and PWM output), 610 rpm is the maximum speed, where 1024 readings per revolution can be obtained.

In incremental mode, the maximum error caused by the sampling rate of the ADCs is $0 /+96 \mu$ s. It has a peak of $1 \mathrm{LSB}=0.35^{\circ}$ at 610 rpm .
At higher speeds this error is reduced again due to interpolation and the output delay remains at $192 \mu \mathrm{~s}$ as the DSP requires two sampling periods $(2 \times 96 \mu s)$ to synthesize and redistribute any missing pulses.

## Incremental Mode:

Incremental encoders are usually required to produce no missing pulses up to several thousand rpm's.

Therefore, the AS5140H has a built-in interpolator, which ensures that there are no missing pulses at the incremental outputs for rotational speeds of up to 10,000 rpm, even at the highest resolution of 10 bits ( 512 pulses per revolution).

| Absolute Output Mode | Incremental Output <br> Mode |
| :---: | :---: |
| $610 \mathrm{rpm}=1024$ samples / turn | no missing pulses <br> $@ 10$ bit resolution <br> $(512 \mathrm{ppr}):$ |
| $122 \mathrm{rpm}=512$ samples / turn | max. speed $=10,000 \mathrm{rpm}$ |
| $2441 \mathrm{rpm}=256$ samples / turn |  |
| etc... |  |

Table 8: Speed performance

### 15.4 Propagation Delays

The propagation delay is the delay between the time that a sample is taken until it is converted and available as angular data. This delay is $48 \mu \mathrm{~s}$ for the absolute interface and $192 \mu$ s for the incremental interface.

Using the SSI interface for absolute data transmission, an additional delay must be considered, caused by the asynchronous sampling ( $\mathrm{t}=0 \ldots 1 / \mathrm{f}_{\mathrm{s}}$ ) and the time it takes the external control unit to read and process the data.

### 15.4.1 Angular Error Caused by Propagation Delay

A rotating magnet will therefore cause an angular error caused by the output delay.
This error increases linearly with speed:
$e_{\text {sampling }}=r p m * 6 *$ prop.delay
where: esampling $=$ angular error [ ${ }^{\circ}$ ]
rpm = rotating speed [rpm]
prop.delay $=$ propagation delay [seconds]

Note: since the propagation delay is known, it can be automatically compensated by the control unit that is processing the data from the AS5140H, thus reducing the angular error caused by speed.

### 15.5 Internal Timing Tolerance

The AS5140H does not require an external ceramic resonator or quartz. All internal clock timings for the AS5140H are generated by an on-chip RC oscillator. This oscillator is factory trimmed to $\pm 5 \%$ accuracy at room temperature ( $\pm 10 \%$ over full temperature range). This tolerance influences the ADC sampling rate and the pulse width of the PWM output:

- Absolute output; SSI interface:

A new angular value is updated every $100 \mu$ s (typ.)

- Incremental outputs:
the incremental outputs are updated every
$100 \mu \mathrm{~s}$ (typ.)
- PWM output:

A new angular value is updated every $100 \mu$ s (typ.).
The PWM pulse timings $T_{\text {on }}$ and $T_{\text {off }}$ also have the
same tolerance as the internal oscillator.
If only the PWM pulse width Ton is used to measure
the angle, the resulting value also has this timing
tolerance.
However, this tolerance can be cancelled by measuring both $T_{\text {on }}$ and $T_{\text {off }}$ and calculating the angle from the duty cycle (see section 6 ):

$$
\text { Position }=\frac{t_{\text {on }} \cdot 1025}{\left(t_{\text {on }}+t_{\text {off }}\right)}-1
$$

### 15.6 Temperature

### 15.6.1 Magnetic Temperature Coefficient

One of the major benefits of the AS5140H compared to linear Hall sensors is that it is much less sensitive to temperature. While linear Hall sensors require a compensation of the magnet's temperature coefficient, the AS 5140 H automatically compensates for the varying magnetic field strength over temperature. The magnet's temperature drift does not need to be considered, as the AS5140H operates with magnetic field strengths from $\pm 45 \ldots \pm 75 \mathrm{mT}$.

## Example:

A NdFeB magnet has a field strength of $75 \mathrm{mT} @-40^{\circ} \mathrm{C}$ and a temperature coefficient of $-0.12 \%$ per Kelvin. The temperature change is from $-40^{\circ}$ to $+150^{\circ}=190 \mathrm{~K}$.
The magnetic field change is: $190 \times-0.12 \%=-22.8 \%$, which corresponds to
75 mT at $-40^{\circ} \mathrm{C}$ and 57.9 mT at $150^{\circ} \mathrm{C}$.
The AS5140H can compensate for this temperature related field strength change automatically, no user adjustment is required.

### 15.6.2 Accuracy over Temperature

The influence of temperature in the absolute accuracy is very low. While the accuracy is $\leq \pm 0.5^{\circ}$ at room temperature, it may increase to $\leq \pm 0.9^{\circ}$ due to increasing noise at high temperatures.

### 15.6.3 Timing Tolerance over Temperature

The internal RC oscillator is factory trimmed to $\pm 5 \%$. Over temperature, this tolerance may increase to $\pm 10 \%$. Generally, the timing tolerance has no influence in the accuracy or resolution of the system, as it is used mainly for internal clock generation.
The only concern to the user is the width of the PWM output pulse, which relates directly to the timing tolerance of the internal oscillator. This influence however can be cancelled by measuring the complete PWM duty cycle (see 15.5).

## 16 Electrical Characteristics

### 16.1 AS5140H Differences to AS5040

The AS5140H and the AS5040 differ in the following features:

| Parameter | AS5140H | AS5040 |
| :---: | :---: | :---: |
| Pin - assignment | Pin - compatible |  |
| Ambient temperature range | $-40^{\circ} \mathrm{C} \ldots . .+150^{\circ} \mathrm{C}$ | $-40^{\circ} \mathrm{C} \ldots . .+125^{\circ} \mathrm{C}$ |
| Alignment mode | Exit alignment mode by power-on-reset , <br> Exit alignment mode by POR or with PROG= low @ falling edge of CSn | Exit alignment mode by power-on-reset only |
| OTP programming voltage | 3,0 to $3,6 \mathrm{~V}$ | 7,3 to 7.5V |
| OTP programming options | Incremental modes (quad AB, step/dir, BLDC) Incremental resolution Incremental Index bit width 10-bit Zero position direction bit (cw/ccw) Redundancy address (1 of 16) 18-bit Chip-Identifier | Incremental modes (quad AB, step/dir, BLDC) Incremental resolution Incremental Index bit width 10-bit Zero position direction bit (cw/ccw) |
| OTP Programming protocol | CSn, PROG and CLK; 52-bit serial data protocol | $\begin{aligned} & \text { CSn, PROG and CLK; } \\ & \text { 16-bit (32-bit) serial data protocol } \end{aligned}$ |

### 16.2 Absolute Maximum Ratings (non operating)

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated under "Operating Conditions" is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

| Parameter | Symbol | Min | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :--- |
| DC supply voltage at pin VDD5V | VDD5V | -0.3 | 7 | V |  |
| DC supply voltage at pin VDD3V3 | VDD3V3 | -0.3 | 5 | V |  |
| Input pin voltage | $\mathrm{V}_{\text {in }}$ | -0.3 | 7 | V | Pins Prog, MagIncn, MagDecn, CLK, CSn, |
| Input current (latchup immunity) | $\mathrm{I}_{\text {scr }}$ | -100 | 100 | mA | Norm: JEDEC 78 |
| Electrostatic discharge | ESD |  | $\pm 2$ | kV | Norm: MLL 883 E method 3015 |
| Storage temperature | $\mathrm{T}_{\text {strg }}$ | -55 | +150 | ${ }^{\circ} \mathrm{C}$ |  |
| Body temperature (Lead-free <br> package) | $\mathrm{T}_{\text {Body }}$ |  | 260 | ${ }^{\circ} \mathrm{C}$ | $\mathrm{t}=20$ to 40s, Norm: IPC/JEDEC J-Std-020C <br> Lead finish 100\% Sn "matte tin" |
| Humidity non-condensing | H | 5 | 85 | $\%$ |  |

### 16.3 Automotive Qualification

The AS5140H is fully automotive qualified according to AEC-Q100, grade 0

### 16.4 Operating Conditions

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Ambient temperature | Tamb | -40 |  | 150 | ${ }^{\circ} \mathrm{C}$ |  |
| Supply current | $\mathrm{I}_{\text {supp }}$ |  | 16 | 21 | mA |  |
| External supply voltage at pin VDD5V <br> Internal regulator output voltage at pin VDD3V3 | VDD5V VDD3V3 | $\begin{aligned} & 4.5 \\ & 3.0 \end{aligned}$ | $\begin{aligned} & 5.0 \\ & 3.3 \end{aligned}$ | $\begin{aligned} & 5.5 \\ & 3.6 \end{aligned}$ | $\begin{aligned} & \mathrm{V} \\ & \mathrm{~V} \end{aligned}$ | 5 V operation |
| External supply voltage at pin VDD5V , VDD3V3 | VDD5V <br> vDD3V3 |  | $\begin{aligned} & 3.3 \\ & 3.3 \end{aligned}$ | $\begin{aligned} & 3.6 \\ & 3.6 \end{aligned}$ | $\begin{aligned} & \mathrm{V} \\ & \mathrm{~V} \end{aligned}$ | 3.3V operation (pins VDD5V and VDD3V3 connected) |
| External VDD3V3 supply voltage rise time at power-up | $\mathrm{t}_{\text {pwrup }}$ | 1 |  | 150 | $\mu \mathrm{s}$ | $10 \%-90 \%$ level in 3.3 V mode (pins VDD5V and VDD3V3 connected) |

### 16.5 DC Characteristics for Digital Inputs and Outputs

16.5.1 CMOS Schmitt-Trigger Inputs: CLK, CSn (CSn = internal Pull-up)
(operating conditions: $T_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}$, VDD5V $=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: |
| High level input voltage | $\mathrm{V}_{1}$ | 0.7 * VDD5V |  | V | Normal operation |
| Low level input voltage | VIL |  | 0.3 * VDD5V | V |  |
| Schmitt Trigger hysteresis | $V_{\text {Ion }}$ - $\mathrm{V}_{\text {loff }}$ | 1 |  | V |  |
| Input leakage current <br> Pull-up low level input current | ILEAK | -1 | 1 | $\mu \mathrm{A}$ $\mu \mathrm{A}$ | CLK only |
|  |  | -30 | -100 |  | CSn only, VDD5V: 5.0V |

### 16.5.2 CMOS / Program Input: Prog

(operating conditions: $T_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :--- |
| High level input voltage | $\mathrm{V}_{\text {IH }}$ | $0.7^{*}$ VDD5V | 5 | V |  |
| High level input voltage | $\mathrm{V}_{\text {PROG }}$ | See "programming <br> conditions" |  | V | During programming |
| Low level input voltage | $\mathrm{V}_{\text {IL }}$ |  | $0.3^{*}$ <br> VDD5V | V |  |
| Pull-down high level input current | $\mathrm{I}_{\mathrm{IL}}$ |  | 100 | $\mu \mathrm{~A}$ | $\mathrm{VDD5V}: 5.5 \mathrm{~V}$ |

### 16.5.3 CMOS Output Open Drain: MagINCn, MagDECn

(operating conditions: $\mathrm{T}_{\mathrm{amb}}=-40$ to $+150^{\circ} \mathrm{C}$, VDD5V $=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Low level output voltage | VoL |  | VSS +0.4 | V |  |
| Output current | lo |  | 4 <br> 2 | mA | VDD5V: 4.5 V <br> VDD5V: 3 V |
| Open drain leakage current | loz |  | 1 | $\mu \mathrm{~A}$ |  |

### 16.5.4 CMOS Output: A, B, Index, PWM

(operating conditions: $\mathrm{T}_{\mathrm{amb}}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :---: |
| High level output voltage | VoH | VDD5V- 0.5 |  | V |  |
| Low level output voltage | VoL |  | $\mathrm{VSS}+0.4$ | V |  |
| Output current | lo |  | 4 | mA | VDD5V: 4.5 V |
|  |  |  | 2 | mA | VDD5V: 3 V |

### 16.5.5 Tristate CMOS Output: DO

(operating conditions: $\mathrm{T}_{\mathrm{amb}}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :---: |
| High level output voltage | VoH | VDD5V -0.5 |  | V |  |
| Low level output voltage | VOL |  | $\mathrm{VSS}+0.4$ | V |  |
| Output current | lo |  | 4 | mA | VDD5V: 4.5 V |
|  |  |  | 2 | mA | VDD5V: 3 V |
| Tri-state leakage current | loz |  | 1 | $\mu \mathrm{~A}$ |  |

### 16.6 Magnetic Input Specification

(operating conditions: $\mathrm{T}_{\mathrm{amb}}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted) Two-pole cylindrical diametrically magnetised source:

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Diameter | $\mathrm{dmag}^{\text {m }}$ | 4 | 6 |  | mm | Recommended magnet: $\varnothing 6 \mathrm{~mm} \times 2.5 \mathrm{~mm}$ for cylindrical magnets |
| Thickness | $\mathrm{t}_{\text {mag }}$ | 2.5 |  |  | mm |  |
| Magnetic input field amplitude | Bpk | 45 |  | 75 | mT | Required vertical component of the magnetic field strength on the die's surface, measured along a concentric circle with a radius of 1.1 mm |
| Magnetic offset | Boff |  |  | $\pm 10$ | mT | Constant magnetic stray field |
| Field non-linearity |  |  |  | 5 | \% | Including offset gradient |
| Input frequency <br> (rotational speed of magnet) | $\mathrm{f}_{\text {mag_abs }}$ |  |  | 10 | Hz | Absolute mode: 600 rpm @ readout of 1024 positions (see table 6) |
|  | fmag _nc |  |  | 166 | Hz | Incremental mode: no missing pulses at rotational speeds of up to $10,000 \mathrm{rpm}$ (see table 6) |
| Displacement radius | Disp |  |  | 0.25 | mm | Max. X-Y offset between defined IC package center and magnet axis (see Figure 18) |
|  |  |  |  | 0.485 |  | Max. X-Y offset between chip center and magnet axis. |
| Chip placement tolerance |  |  |  | $\pm 0.235$ | mm | Placement tolerance of chip within IC package (see Figure 20) |
| Recommended magnet material and temperature drift |  |  | -0.12 |  | \%/K | NdFeB (Neodymium Iron Boron) |
|  |  |  | -0.035 |  |  | SmCo (Samarium Cobalt) |

### 16.7 Electrical System Specifications

(operating conditions: $\mathrm{T}_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution | RES |  |  | 10 | bit | 0.352 deg |
| 7 bit <br> 8 bit <br> 9 bit <br> 10 bit | LSB |  | $\begin{aligned} & 2.813 \\ & 1.406 \\ & 0.703 \\ & 0.352 \end{aligned}$ |  | deg | Adjustable resolution only available for incremental output modes; <br> Least significant bit, minimum step |
| Integral non-linearity (optimum) | INLopt |  |  | $\pm 0.5$ | deg | Maximum error with respect to the best line fit. Verified at optimum magnet placement, $\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$. |
| Integral non-linearity (optimum) | INLtemp |  |  | $\pm 0.9$ | deg | Maximum error with respect to the best line fit. Verified at optimum magnet placement, $T_{\text {amb }}=-40 \text { to }+150^{\circ} \mathrm{C}$ |
| Integral non-linearity | INL |  |  | $\pm 1.4$ | deg | Best line fit $=\left(\right.$ Errmax $_{\text {max }}-$ Errmin $_{\text {min }} / 2$ <br> Over displacement tolerance with 6 mm diameter magnet, $T_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}$ |
| Differential non-linearity | DNL |  |  | $\pm 0.176$ | deg | 10bit, no missing codes |
| Transition noise | TN |  |  | 0.12 | $\begin{aligned} & \text { Deg } \\ & \text { RMM } \end{aligned}$ | RMS equivalent to 1 sigma |
| Hysteresis | Hyst |  | 0.704 |  | deg | Incremental modes only |
| Power-on reset thresholds On voltage; 300 mV typ. hysteresis Off voltage; 300 mV typ. hysteresis | $\begin{aligned} & V_{\text {on }} \\ & V_{\text {off }} \end{aligned}$ | $\begin{aligned} & 1,37 \\ & 1.08 \end{aligned}$ | $\begin{aligned} & 2.2 \\ & 1.9 \end{aligned}$ | $\begin{aligned} & 2.9 \\ & 2.6 \end{aligned}$ | $\begin{aligned} & \text { V } \\ & \text { V } \end{aligned}$ | DC supply voltage 3.3 V (VDD3V3) DC supply voltage 3.3 V (VDD3V3) |
| Power-up time | $t_{\text {tpwup }}$ |  |  | 50 | ms | Until offset compensation finished |
| System propagation delay absolute output | $\mathrm{t}_{\text {delay }}$ |  |  | 48 | $\mu \mathrm{s}$ | Includes delay of ADC and DSP |
| System propagation delay incremental output |  |  |  | 192 | $\mu \mathrm{s}$ | Calculation over two samples |
| Sampling rate for absolute output | $\mathrm{fs}_{s}$ | 9.90 | 10.42 | 10.94 | kHz | Internal sampling rate, $\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$ |
|  |  | 9.38 | 10.42 | 11.46 |  | Internal sampling rate, $\mathrm{T}_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}$ |
| Read-out frequency | CLK |  |  | 1 | MHz | Max. clock frequency to read out serial data |



Figure 23: Integral and differential non-linearity example (exaggerated curve)

Integral Non-Linearity (INL) is the maximum deviation between actual position and indicated position.
Differential Non-Linearity (DNL) is the maximum deviation of the step length from one position to the next.
Transition Noise (TN) is the repeatability of an indicated position.

### 16.8 Timing Characteristics

## Synchronous Serial Interface (SSI)

(operating conditions: $\mathrm{T}_{\mathrm{amb}}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :---: | :--- |
| Data output activated (logic <br> high) | $\mathrm{t}_{\text {Do active }}$ |  |  | 100 | ns | Time between falling edge of CSn and data output <br> activated |
| First data shifted to output <br> register | tcLL $^{2}$ FE | 500 |  |  | ns | Time between falling edge of CSn and first falling <br> edge of CLK |
| Start of data output | T CLK/2 | 500 |  |  | ns | Rising edge of CLK shifts out one bit at a time |
| Data output valid | $\mathrm{t}_{\text {Do valid }}$ | 357 | 375 | 394 | ns | Time between rising edge of CLK and data output <br> valid |
| Data output tristate | $\mathrm{t}_{\text {Do tristae }}$ |  |  | 100 | ns | After the last bit DO changes back to "tristate" |
| Pulse width of CSn | $\mathrm{t}_{\text {CSn }}$ | 500 |  |  | ns | CSn = high; To initiate read-out of next angular <br> position |
| Read-out frequency | fcLK | $>0$ |  | 1 | MHz | Clock frequency to read out serial data |

Pulse Width Modulation Output
(operating conditions: $T_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PWM frequency | $\mathrm{f}_{\text {pwm }}$ | 0.927 | 0.976 | 1.024 | KHz | Signal period $=1025 \mu \mathrm{~s} \pm 5 \%$ at $\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$ |
|  |  | 0.878 | 0.976 | 1.074 |  | $=1025 \mu \mathrm{~s} \pm 10 \%$ at $\mathrm{T}_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}$ |
| Minimum pulse width | PW min | 0.95 | 1 | 1.05 | $\mu \mathrm{s}$ | Position 0d; angle 0 degree |
| Maximum pulse width | PW max | 973 | 1024 | 1075 | $\mu \mathrm{s}$ | Position 1023d; angle 359.65 degree |

## ncremental Outputs

(operating conditions: $\mathrm{T}_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}$, VDD5V $=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :---: | :--- |
| Incremental outputs valid after <br> power-up | t nocremental <br> outputs valid |  |  | 500 | ns | Time between first falling edge of CSn after power-up <br> and valid incremental outputs |
| Directional indication valid | t Dir valid |  |  | 500 | ns | Time between rising or falling edge of LSB output and <br> valid directional indication |

### 16.9 Programming Conditions

(operating conditions: $\mathrm{T}_{\text {amb }}=-40$ to $+150^{\circ} \mathrm{C}, \mathrm{VDD5V}=3.0-3.6 \mathrm{~V}$ ( 3 V operation) VDD5V $=4.5-5.5 \mathrm{~V}$ ( 5 V operation) unless otherwise noted)

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Programming voltage | $V_{\text {prog }}$ | 3.0 | 3.3 | 3.6 | V | Voltage applied during programming. |
| Programming voltage off level | $V$ Progoff | 0 |  | 1 | V | Line must be discharged to this level |
| Programming current | I prog |  |  | 100 | mA | Current during programming |
| Programmed fuse resistance (log 1) | $\mathrm{R}_{\text {programmed }}$ | 100k |  | $\infty$ | Ohm | $10 \mu \mathrm{~A}$ max. current @ 100mV |
| Unprogrammed fuse resistance ( $\log 0$ ) | Runprogrammed | 50 |  | 100 | Ohm | 2mA max. current @ 100mV |
| Programming time per bit | tprog | 10 |  | 20 | $\mu \mathrm{s}$ | Time to prog. a singe fuse bit |
| Refresh time per bit | tcharge | 1 |  |  | $\mu \mathrm{s}$ | Time to charge the cap after tprog |
| LOAD frequency | fload |  |  | 500 | kHz | Data can be loaded at $\mathrm{n}^{*} 2 \mu \mathrm{~s}$. |
| READ frequency | fread |  |  | 2.5 | MHz | Read the data from the latch. |
| WRITE frequency | $\mathrm{f}_{\text {WRITE }}$ |  |  | 2.5 | MHz | Write the data to the latch. |
|  |  |  |  |  |  |  |

## 17 Package Drawings and Markings

16-Lead Shrink Small Outline Package SSOP-16


| Dimensions |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Symbol | mm |  |  |  | inch |  |  |  |  |  |  |
|  | Min | Typ | Max | Min | Typ | Max |  |  |  |  |  |
| A | 1.73 | 1.86 | 1.99 | .068 | .073 | .078 |  |  |  |  |  |
| A1 | 0.05 | 0.13 | 0.21 | .002 | .005 | .008 |  |  |  |  |  |
| A2 | 1.68 | 1.73 | 1.78 | .066 | .068 | .070 |  |  |  |  |  |
| b | 0.25 | 0.315 | 0.38 | .010 | .012 | .015 |  |  |  |  |  |
| C | 0.09 | - | 0.20 | .004 | - | .008 |  |  |  |  |  |
| D | 6.07 | 6.20 | 6.33 | .239 | .244 | .249 |  |  |  |  |  |
| E | 7.65 | 7.8 | 7.9 | .301 | .307 | .311 |  |  |  |  |  |
| E1 | 5.2 | 5.3 | 5.38 | .205 | .209 | .212 |  |  |  |  |  |
| e | 0.65 |  |  |  |  |  |  |  |  |  | .0256 |
| K | $0^{\circ}$ | - | $8^{\circ}$ | $0^{\circ}$ | - | $8^{\circ}$ |  |  |  |  |  |
| L | 0.63 | 0.75 | 0.95 | .025 | .030 | .037 |  |  |  |  |  |

## 18 Packing Options

$$
\begin{array}{|ll}
\text { Delivery: } & \text { Tape and Reel ( } 1 \text { reel }=2000 \text { devices }) \\
& \text { Tubes (1 box }=100 \text { tubes à } 77 \text { devices })
\end{array}
$$



Marking: AYWWIZZ
A: Pb-free Identifier
Y: Last Digit of Manufacturing Year WW: Manufacturing Week

I: Plant Identifier
ZZ: Traceability Code

JEDEC Package Outline Standard: MO - 150 AC

Thermal Resistance $R_{\text {th }(j-a)}$ :
typ. 151 K/W in still air, soldered on PCB

IC's marked with a white dot or the letters "ES" denote Engineering Samples

## 19 Recommended PCB Footprint



| Recommended Footprint Data |  |  |
| :---: | :---: | :---: |
|  | mm | inch |
| A | 9.02 | 0.355 |
| B | 6.16 | 0.242 |
| C | 0.46 | 0.018 |
| D | 0.65 | 0.025 |
| E | 5.01 | 0.197 |

## 20 Revision History

| Revision | Date |  | Description |
| :--- | :---: | :--- | :---: |
| 1.0 | Oct. 3,2006 | Initial revision |  |

## 21 Contact

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