

HT82P604B DC-DC Converter

Features

- Operating voltage: 5V~12V
- Drives two N-Channel MOSFETs
- Simple single-loop control design
 Voltage-mode PWM control
- Fast transient response
 - High-bandwidth error amplifier
 - Full 0% to 100% duty ratio
- Excellent output voltage regulation
 - ±1% over line voltage and temperature
- 5 bit digital-to-analog output voltage selection
 - Wide range -1.3V to 3.5V
 - 0.1V binary steps from 2.1V to 3.5V
 - 0.05V binary steps from 1.3V to 2.05V
- Power-good output voltage monitor
- Over-voltage and over-current sensing element, uses MOSFET's $r_{DS\,(ON)}$
- Small converter size
 - Constant frequency operation
 - 200kHz free-running oscillator programmable from 50kHz to over 1MHz

Applications

Power supply for PentiumTM, PentiumTM
 Pro, PowerPCTM and AlphaTM Microprocessors

- High-power5Vto3.xVDC-DCregulators
- Low-voltage distributed power supplies

General Description

The HT82P604B provides complete control and protection for a DC-DC converter optimized for high performance microprocessor applications. It is designed to drive two N-Channel MOSFETs in a synchronous-rectified buck topology. The HT82P604B integrates all of the control, output adjustment, monitoring and protection functions into a single package.

The output voltage of the converter is easily adjusted and precisely regulated. The HT82P604B includes a 5-input digital-to-analog converter (DAC) that adjusts the output voltage from 2.1V to 3.5V in 0.1V increments and from 1.3V to 2.05V in 0.05V steps. The precision reference and voltage-mode regulator hold the selected output voltage to within $\pm 1\%$ over temperature and line voltage variations.

The HT82P604B provides simple, single feedback loop, voltage-mode control with fast tran-

sient response. It includes a 200kHz free-running triangle-wave oscillator that is adjustable from below 50kHz to over 1MHz. The error amplifier features a 15MHz gain-bandwidth product and 6V/ms slew rate which enables high converter bandwidth for fast transient performance. The resulting PWM duty ratio ranges from 0% to 100%.

The HT82P604B monitors the output voltage with a window comparator that tracks the DAC output and issues a Power Good signal when the output is within $\pm 10\%$. The HT82P604B protects against over-current conditions by inhibiting PWM operation. Built-in over-voltage protection triggers an external SCR to crowbar the input supply. The HT82P604B monitors the current by using the $r_{DS\ (ON)}$ of the upper MOSFET which eliminates the need for a current sensing resistor.

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Pin Assignment

VSEN [1	20	RT				
OCSET [2	19	OVP				
SS [3	18	□ vcc				
VID0 □	4	17	LGATE				
VID1 □	5	16	PGND				
VID2 □	6	15	🗆 воот				
VID3 □	7	14	UGATE				
VID4 □	8	13	PHASE				
COMP [9	12	PGOOD				
FB 🗆	10	11	GND				
HT82P604B							
- 20 SOP							

Pin Description

Pin No.	Pin Name	I/O	Description
1	VSEN	I	This pin is connected to the converters output voltage. The PGOOD and OVP comparator circuits use this signal to report output voltage status and for overvoltage protection.
2	OCSET	I	Connect a resistor (R _{OCSET}) from this pin to the drain of the upper MOSFET. R _{OCSET} , an internal 200mA current source (I _{OCS}), and the upper MOSFET on-resistance (r _{DS (ON)}) set the converter over-current (OC) trip point. $I_{PEAK} = \frac{I_{OCS} \times R_{OCSET}}{r_{DS(ON)}}$ An over-current trip cycles the soft-start function.
3	SS	0	Connect a capacitor from this pin to ground. This capacitor, along with an internal $10\mu A$ current source, sets the soft-start interval of the converter.
4~8	VID0~4	I	VID0-4 are the input pins to the 5-bit DAC. The states of these five pins program the internal voltage reference (DACOUT). The level of DACOUT sets the converter output voltage. It also sets the PGOOD and OVP thresholds. Table 1 specifies DACOUT for the 32 combinations of DAC inputs.
9, 10	COMP, FB	I	COMP and FB are the available external pins of the error amplifier. The FB pin is the inverting input of the error amplifier and the COMP pin is the error amplifier output. These pins are used to compensate the voltage-control feedback loop of the converter.
11	GND	P	Signal ground for the IC. All voltage levels are measured with respect to this pin.



Pin No.	Pin Name	I/O	Description
12	PGOOD	O	PGOOD is an open collector output used to indicate the status of the converter output voltage. This pin is pulled low when the converter output is not within $\pm 10\%$ of the DACOUT reference voltage. Exception to this behavivor is the '11111' VID pin combination which disables the converter; in this case, PGOOD asserts a high level.
13	PHASE	О	Connect the PHASE pin to the upper MOSFET source. This pin is used to monitor the voltage drop across the MOSFET for over-current protection. This pin also provides the return path for the upper gate drive.
14	UGATE	О	Connect UGATE to the upper MOSFET gate. This pin provides the gate drive for the upper MOSFET.
15	воот	I	This pin provides bias voltage to the upper MOSFET driver. A bootstrap circuit may be used to create a BOOT voltage suitable to drive a standard N-Channel MOSFET.
16	PGND	P	This is the power ground connection. Tie the lower MOSFET source to this pin.
17	LGATE	О	Connect LGATE to the lower MOSFET gate. This pin provides the gate drive for the lower MOSFET.
18	VCC	P	Provide a 12V bias supply for the chip to this pin.
19	OVP	О	The OVP pin can be used to drive an external SCR in the event of an overvoltage condition.
20	RT	I	This pin provides oscillator switching frequency adjustment. By placing a resistor (R_T) from this pin to GND, the nominal 200kHz switching frequency is increased. $F_S \approx 200 \text{kHz} + \frac{5 \times 10^6}{R_T \ (k\Omega)} \ (R_T \ \text{to GND})$ Conversely, connecting a pull-up resistor (R_T) from this pin to VCC reduces the switching frequency. $F_S \approx 200 \text{kHz} + \frac{4 \times 10^7}{R_T \ (k\Omega)} \ (R_T \ \text{to } 12\text{V})$

Absolute Maximum Ratings

Supply Voltage	+15V	Storage Temperature	–50°C to 125°C
Innut Voltage	GND_0 3 to Vgg+0 3V	Operating Temperature	-20°C to 75°C

Note: These are stress ratings only. Stresses exceeding the range specified under "Absolute Maximum Ratings" may cause substantial damage to the device. Functional operation of this device at other conditions beyond those listed in the specification is not implied and prolonged exposure to extreme conditions may affect device reliability.



Electrical Characteristics

Sumbal	Donomoton		Test Conditions	ъл.			TT
Symbol	$oxed{f V_{CC}} oxed{f Parameter} oxed{f V_{CC}} oxed{f Condition}$		Conditions	Min.	Тур.	Max.	Unit
V _{CC} Sup	ply Current						
I_{CC}	Nominal Supply	12V	Ugate and Lgate Open	_	5	_	mA
Power-C	On Reset						
	Rising V _{CC} Threshold	12V	V _{OCSET} =4.5V	_	_	10.4	V
	Falling V _{CC} Threshold	12V	V _{OCSET} =4.5V	8.2	_	_	V
	Rising V _{OCSET} Threshold	12V		_	1.26	_	V
Oscillate	or	•				'	
	Free Running Frequency	12V	RT=Open	185	200	215	kHz
	Total Variation	12V	6kΩ <rt gnd<200kω<="" td="" to=""><td>-15</td><td>_</td><td>+15</td><td>%</td></rt>	-15	_	+15	%
$\Delta V_{ m OSC}$	Ramp Amplitude	12V	RT=Open	_	1.9	_	V _{P-P}
Referen	ce and DAC	•				'	
	DACOUT Voltage Accuracy	12V		-1.0	_	+1.0	%
ERROR	Amplifier	•				'	
	DC Gain	12V		_	88	_	dB
GBW	Gain-Bandwidth Product	12V		_	15	_	MHz
SR	Slew Rate	12V	COMP=10pF	_	6	_	V/µs
Gate Dri	vers	ļ					
I_{UGATE}	Upper Gate Source	12V	$egin{array}{l} V_{BOOT}\!-\!V_{PHASE}\!=\!12V, \ V_{UGATE}\!=\!6V \end{array}$	350	500	_	mA
R _{UGATE}	Upper Gate Sink	12V	I _{LGATE} =0.3A	_	5.5	10	Ω
I_{LGATE}	Lower Gate Source	12V	$V_{\rm CC}$ =12V, $V_{\rm LGATE}$ =6V	300	450	_	mA
R _{LGATE}	Lower Gate Sink	12V	I _{LGATE} =0.3V	_	3.5	6.5	Ω
Protecti	on	•		•	•		•
	Over-Voltage Trip (V _{SEN} /DACOUT)	12V			115	120	%
I _{OCSET}	OCSET Current Source	12V	$V_{\rm OCSET}$ =4.5 $V_{\rm DC}$	170	200	230	μΑ
I _{OVP}	OVP Sourcing Current	12V	V_{SEN} =5.5, V_{OVP} =0V	60		_	mA
$\overline{\mathrm{I}_{\mathrm{SS}}}$	Soft Start Current	12V		_	10	_	μΑ
Power G	Good	•					
	Upper Threshold (V _{SEN} /DACOUT)	12V	$ m V_{SEN}$ Rising	106	_	111	%



G 1 1	D 4		Test Conditions	3.4.	m	3.4	T7 *4
Symbol	Parameter	$\mathbf{v}_{\mathbf{c}\mathbf{c}}$	Conditions	Min. Typ		Max.	Unit
	Lower Threshold (V _{SEN} /DACOUT)	12V	$ m V_{SEN}$ Falling	89	_	94	%
	Hysteresis (V _{SEN} /DACOUT)	12V	Upper and Lower Threshold	_	2	_	%
V_{PGOOD}	PGOOD Voltage Low	12V	I _{PGOOD} =-5mA	_	0.5	_	V

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Functional Description

Output voltage program

The output voltage of a HT82P604B converter is programmed to discrete levels between 1.3V and 3.5V. The voltage identification (VID) pins program an internal voltage reference (DACOUT) with a 5-bit digital-to-analog converter (DAC). The level of DACOUT also sets the PGOOD and OVP thresholds. Table 1 specifies the DACOUT voltage for the 32 combinations of open or short connections on the VID pins. The output voltage should not be adjusted while the converter is delivering power. Remove input power before changing the output voltage. Adjusting the output voltage during operation could toggle the PGOOD signal and exercise the overvoltage protection.

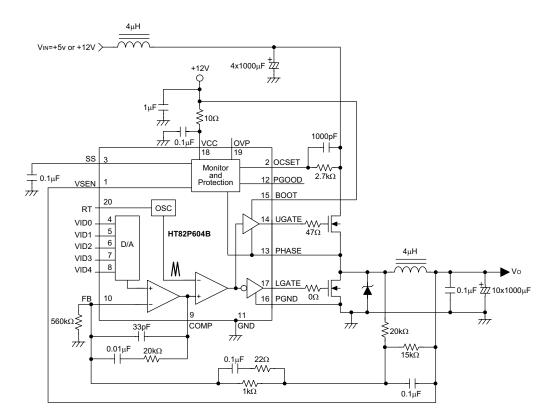
	Nominal				
VID4	VID3	VID2	VID1	VID0	Output Voltage Dacout
0	1	1	1	1	1.30
0	1	1	1	0	1.35
0	1	1	0	1	1.40
0	1	1	0	0	1.45
0	1	0	1	1	1.50
0	1	0	1	0	1.55
0	1	0	0	1	1.60
0	1	0	0	0	1.65
0	0	1	1	1	1.70
0	0	1	1	0	1.75
0	0	1	0	1	1.80

	Pin Name							
VID4	VID3	VID2	VID1	VID0	Output Voltage Dacout			
0	0	1	0	0	1.85			
0	0	0	1	1	1.90			
0	0	0	1	0	1.95			
0	0	0	0	1	2.00			
0	0	0	0	0	2.05			
1	1	1	1	1	0			
1	1	1	1	0	2.1			
1	1	1	0	1	2.2			
1	1	1	0	0	2.3			
1	1	0	1	1	2.4			
1	1	0	1	0	2.5			
1	1	0	0	1	2.6			
1	1	0	0	0	2.7			
1	0	1	1	1	2.8			
1	0	1	1	0	2.9			
1	0	1	0	1	3.0			
1	0	1	0	0	3.1			
1	0	0	1	1	3.2			
1	0	0	1	0	3.3			
1	0	0	0	1	3.4			
1	0	0	0	0	3.5			

Note: 0=connected to GND or VSS, 1=OPEN ${\it Table~1~Output~voltage~program}$



Application Circuits





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