

CMOS LDO Regulator Series for Portable Equipments CMOS LDO Regulators with Auto Power Saving Function



BHDDPB1WHFV Series

No.09020EAT05

Description

The BH $\Box\Box$ PB1WHFV regulator series can respond to changes in output current by switching to a state in which regulator characteristics are ideal. The regulators cut power consumption by lowering their own current consumption to approximately 2 μ A when the application is operating in the standby state. During normal-current operation it will automatically switch to high-speed operating mode. The IC's soft start function reduce the rush current that flows to the output capacitors during startup. The HVSOF5 package, which features excellent heat dissipation, contributes to space-saving application designs.

Features

- 1) Automatic switching between low-consumption and high-speed modes
- 2) Built-in rush current prevention circuit
- 3) Low-voltage 1.7 V operation
- 4) High accuracy output voltage: ± 1%
- 5) Circuit current during low-consumption operation: 2 μ A
- 6) Stable with a ceramic capacitor (0.47 μ F)
- 7) Built-in temperature and overcurrent protection circuits
- 8) Built-in output discharge during standby operation function
- 9) Ultra-small HVSOF5 power package

Applications

Battery-driven portable devices, etc.

Product lineup

■150 mA BH□□PB1WHFV Series

		51163								
Product name	1.2	1.5	1.8	2.5	2.8	2.9	3.0	3.1	3.3	Package
BHDDPB1WHFV	\checkmark	HVSOF5								

Model name: BH

а

b

Symbol	Description							
		Output voltag	ge specific	ation				
		Output voltage (V)	Output voltage (V)					
	12	1.2 V (Typ.)	29	2.9 V (Typ.)				
а	15 1.5 V (Typ.)		30	3.0 V (Typ.)				
	18	1.8 V (Typ.)	31	3.1 V (Typ.)				
	25	2.5 V (Typ.)	33	3.3 V (Typ.)				
	28	2.8 V (Typ.)						
b		Package HFV: HVSOF5						

●Absolute maximum ratings (Ta = 25°C)

U (
Parameter	Symbol	Limits	Unit
Power supply voltage	VMAX	-0.3 to +6.5	V
Power dissipation	Pd	410 ^{*1}	mW
Operating temperature range	Topr	-40 to +85	°C
Storage temperature range	Tslg	−55 to +125	°C
Junction temperature	Tjmax	125	°C

*1: Reduced by 4.1 mW/°C over 25°C, when mounted on a glass epoxy board (70 mm × 70 mm × 1.6 mm)

Recommended operating ranges (not to exceed Pd)

Parameter	Symbol	Limits	Unit
Power supply voltage	Vin	1.7 to 5.5	V
Output MAX current	IMAX	0 to 150	mA

Recommended operating conditions

Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions
Input capacitor	CIN	0.33 *2	0.47	-	μF	The use of ceramic capacitors is recommended.
Output capacitor	Co	0.33 *2	0.47	I	μF	The use of ceramic capacitors is recommended.

*2: Make sure that the output capacitor value is not kept lower than this specified level across a variety of temperature, DC bias characteristic. And also make sure that the capacitor value can not change as time progresses.

•Electrical characteristics

(Unless otherwise specified, Ta = 25°C, VIN = VOUT + 1.0 V, STBY = 1.5 V, SEL = 0 V, CIN = 0.47 µF, Co = 0.47 µF)

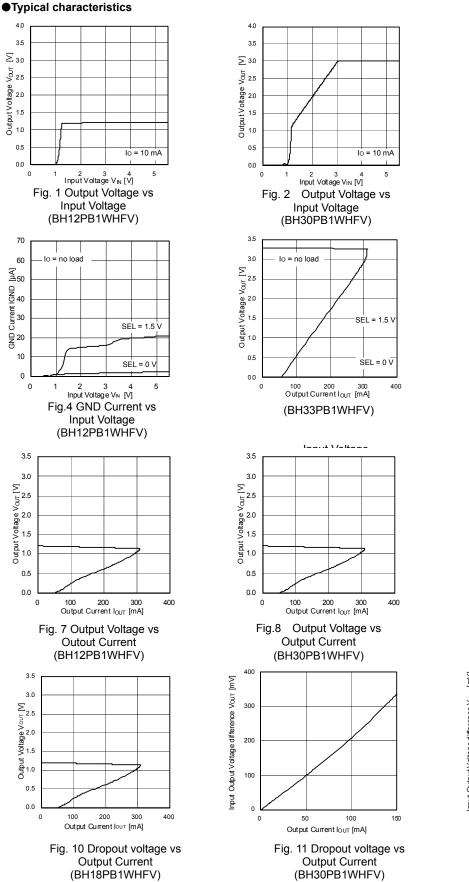
			Limit					
PARAMETER		Symbol	MIN.	TYP.	MAX.	Unit	Conditions	
[Regulator]								
Output voltage		Vout	VOUT1×0.99	-	VOUT1×1.01	V	Vout≧2.5V,Iout=0.1mA,SEL=1.5V	
(high-speed mode)		Vout1	VOUT1-0.025	-	VOUT1+0.025	V	Vout≦1.8V,Iout=0.1mA,SEL=1.5V	
Output voltage		Vour2	VOUT2×0.97	-	VOUT2×1.038	V	Vout≧2.5V,Iout=0.1mA,SEL=0V	
(low-consumption mode)		Vout2	VOUT2×0.967	-	VOUT2×1.043	V	Vout≦1.8V,Iout=0.1mA,SEL=0V	
Circuit current (high-speed mode)		ICC1	-	20	40	μA	IOUT=0mA, VIN pin monitor,SEL=1.5V	
Circuit current (low-consumption mode)		ICC2	-	2	4	μA	IOUT=0mA, VIN pin monitor, SEL=0V	
Circuit current (STBY)		ISTBY	-	-	1.0	μA	STBY=0V	
Ripple rejection ratio (high-speed mode)		RR1	42	60	-	dB	VRR=-20dBv, fRR=1kHz, Iout=10mA, SEL=1.5V	
Dropout voltage 1 *1		VSAT1	-	100	200	mV	VIN=VOUT×0.98,IOUT=50mA	
Dropout voltage 2 *1		VSAT2	-	210	400	mV	VIN=VOUT×0.98,IOUT=100mA	
Dropout voltage 3 *1		VSAT3	-	315	600	mV	VIN=VOUT×0.98,IOUT=150mA	
Line regulation 1 (high-speed mode)		VDL1	-	2	20	mV	VIN=VOUT+1V to 5.5V,IouT=10mA	
Line regulation 2 (low-consumption mode)		VDL2	-	2	20	mV	Vιν=Vout+1V to 5.5V,Ιουτ=100μΑ	
Load regulation		VDLO	-	10	40	mV	IOUT=10mA to 100mA	
[Mode switch]								
Current threshold (low-consumption mode)		ITH1	0.09	0.3	-	mA	SEL=0V louт=3mA⇒0mA sweep	
Current threshold (high-speed mode)		ITH2	-	1.2	2.2	mA	SEL=0V louт=0mA⇒3mA sweep	
Over Current Protection	า 1】	[1			
Limit Current		ILMAX	160	300	500	mA	Vo=Vout×0.90	
Short current		ISHORT	20	50	100	mA	Vo=0V	
[Stand-by block]		I						
STBY pin sink current		ISTB	-	2	4	μA	STBY=1.5V	
STBY control voltage	ON	VSTBH	1.5	-	Vin	V		
J	OFF	VSTBL	-0.3	-	0.3	V		
Discharge resistance at standby		RDCG	1.5	2.2	3.0	kΩ	STBY=0V	
[SEL PIN]		[1		1			
Pull-down resistance of S	EL pin	RSEL	0.5	1.0	2.0	MΩ		
SEL control voltage	ON	VSELH	1.5	-	Vin	V	Fixed high speed mode	
CEL CONTO VOILAGO	OFF	VSELL	-0.3	-	0.3	V	Automatic switch mode	

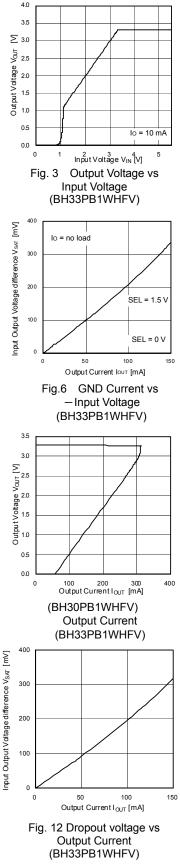
* Note: This IC is not designed to be radiation-resistant. *3: Except at VOUT \leq 1.5 V.

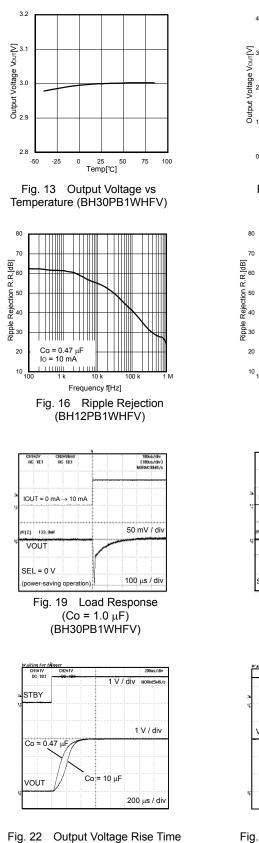
•Electrical characteristics of each output voltage

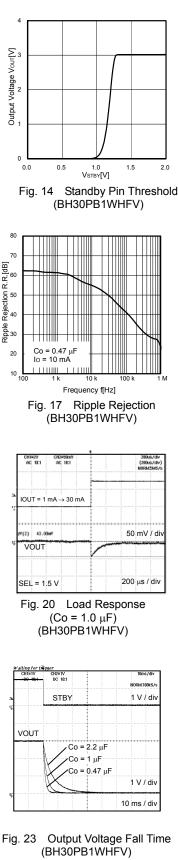
Output Voltage	Parameter	Min.	Тур.	Max.	Unit	Conditions
1.2 V		70	120	-		VCC = 1.7 V
1.2 V	Max. output current	150	-	_	mA	VCC = 2.0 V
1.5 V		50	100	_		VCC = 1.8 V
1.5 V		150	-	_		VCC = 2.2 V
1.8 V ≤ Vout		75	143	_		VCC = Vout + 0.3 V
1.0 V S VOUT		150	_	_		VCC = Vout + 0.6 V

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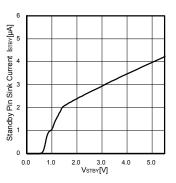


Fig. 15 Standby Pin Sink Current (BH30PB1WHFV)

	CH1=1V DC 10:1	CH2=50mV	L = 0 V	⇔ 1.5	10ms/div NORM:100k5/s 1 V / div
÷					
ţ	VOUT			5	50 mV / div
	lo = no	load	+++++++++++++++++++++++++++++++++++++++	ſ	10 ms / div

Fig. 18 Output Voltage Waveform During SEL Switching (BH30PB1WHFV)

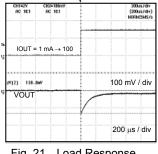
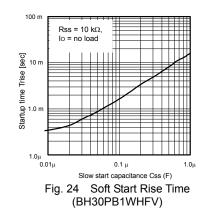


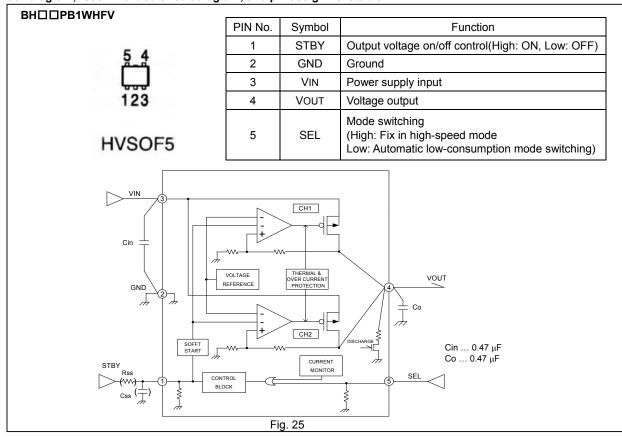
Fig. 21 Load Response (Co = 1.0 μ F) (BH30PB1WHFV)



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(BH30PB1WHFV)

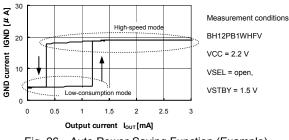
•Block diagram, recommended circuit diagram, and pin assignment table

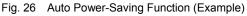


Auto Power-saving Function

The IC incorporates a built-in auto power-saving function that continuously monitors the output current and switches automatically between a low current consumption regulator and a high-speed operation regulator. This function reduces the regulator's own current consumption to approximately 1/10 or lower of normal levels when the output current falls below approximately 300 µA.

To operate only the high-speed operation regulator without using the auto power-saving function, fix the SEL pin to high.





Power Dissipation (Pd)

1. Power Dissipation (Pd)

Power dissipation calculations include estimates of power dissipation characteristics and internal IC power consumption, and should be treated as guidelines. In the event that the IC is used in an environment where this power dissipation is exceeded, the attendant rise in the junction temperature will trigger the thermal shutdown circuit, reducing the current capacity and otherwise degrading the IC's design performance. Allow for sufficient margins so that this power dissipation is not exceeded during IC operation.

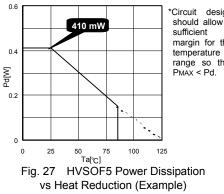
Calculating the maximum internal IC power consumption (PMAX)

 $PMAX = (VIN - VOUT) \times IOUT (MAX.)$

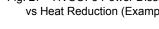
VOUT : Output voltage IOUT (MAX) : Max. output current

VIN : Input voltage

2. Power Dissipation/Heat Reduction (Pd) HVSOF5



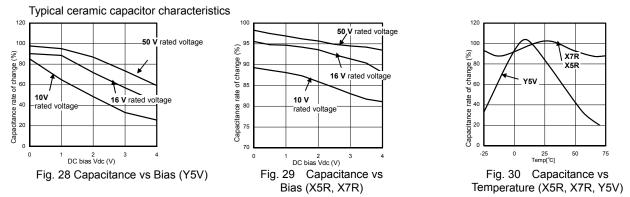
*Circuit design should allow a margin for the range so that PMAX < Pd.



Input Output capacitors

It is recommended to insert bypass capacitors between input and GND pins, positioning them as close to the pins as possible. These capacitors will be used when the power supply impedance increases or when long wiring paths are used, so they should be checked once the IC has been mounted.

Ceramic capacitors generally have temperature and DC bias characteristics. When selecting ceramic capacitors, use X5R or X7R, or better models that offer good temperature and DC bias characteristics and high tolerant voltages.



Output capacitors

Mounting input capacitor between input pin and GND(as close to pin as possible), and also output capacitor between output pin and GND(as close to pin as possible) is recommended.

The input capacitor reduces the output impedance of the voltage supply source connected to the VCC. The higher value the output capacitor goes, the more stable the whole operation becomes. This leads to high load transient response. Please confirm the whole operation on actual application board.

Generally, ceramic capacitor has wide range of tolerance, temperature coefficient, and DC bias characteristic. And also its value goes lower as time progresses. Please choose ceramic capacitors after obtaining more detailed data by asking capacitor makers.

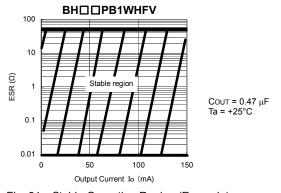


Fig. 31 Stable Operation Region (Example)

Operation Notes

- 1. Absolute maximum ratings
- An excess in the absolute maximum ratings, such as supply voltage, temperature range of operating conditions, etc., can break down the devices, thus making impossible to identify breaking mode, such as a short circuit or an open circuit. If any over rated values will expect to exceed the absolute maximum ratings, consider adding circuit protection devices, such as fuses.
- 2. Thermal design
- Use a thermal design that allows for a sufficient margin in light of the power dissipation (Pd) in actual operating conditions. 3. Inter-pin shorts and mounting errors

Use caution when positioning the IC for mounting on printed circuit boards. The IC may be damaged if there is any connection error or if pins are shorted together.

4. Thermal shutdown circuit (TSD)

The IC incorporates a built-in thermal shutdown circuit (TSD circuit). The thermal shutdown circuit is designed only to shut the IC off to prevent runaway thermal operation. It is not designed to protect the IC or guarantee its operation. Do not continue to use the IC after operating this circuit or use the IC in an environment where the operation of this circuit is assumed.

5. Ground wiring patterns

The power supply and ground lines must be as short and thick as possible to reduce line impedance. Fluctuating voltage on the power ground line may damage the device.

6. Overcurrent protection circuit

The IC incorporates a built-in overcurrent protection circuit that operates according to the output current capacity. This circuit serves to protect the IC from damage when the load is shorted. The protection circuit is designed to limit current flow by not latching in the event of a large and instantaneous current flow originating from a large capacitor or other component. These protection circuits are effective in preventing damage due to sudden and unexpected accidents. However, the IC should not be used in applications characterized by the continuous operation or transitioning of the protection circuits. At the time of thermal designing, keep in mind that the current capability has negative characteristics to temperatures.

- 7. Actions in strong electromagnetic field
- Use caution when using the IC in the presence of a strong electromagnetic field as doing so may cause the IC to malfunction. 8. Back Current

In applications where the IC may be exposed to back current flow, it is recommended to create a path to dissipate this current by inserting a bypass diode between the VIN and VOUT pins.

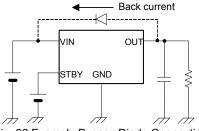


Fig. 32 Example Bypass Diode Connection

9. I/O voltage difference

Using the IC in automatic switching mode when the I/O voltage differential becomes saturated (VIN - VOUT < 150 mV) may result in a large output noise level. If the noise level becomes problematic, use the IC with the SEL pin in the high state when the voltage differential is saturated.

10.GND Voltage

The potential of GND pin must be minimum potential in all operating conditions.

11. Preventing Rush Current

By attaching the Rss and Css time constants to the STBY pin, sudden rises in the regulator output voltage can be prevented, dampening the flow of rush current to the output capacitors. The larger the time constant used, the greater the resulting reduction. However, large time constants also result in longer startup times, so the constant should be selected after considering the conditions in which the IC is to be used.

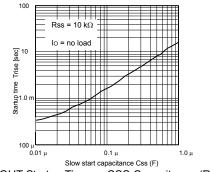


Fig. 33 VOUT Startup Time vs CSS Capacitance (Reference)

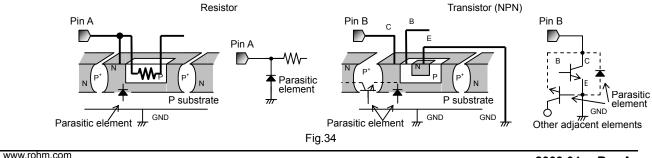
12. Regarding input Pin of the IC (Fig.34)

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of these P layers with the N layers of other elements, creating a parasitic diode or transistor. For example, the relation between each potential is as follows:

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

When GND > Pin B, the P-N junction operates as a parasitic transistor.

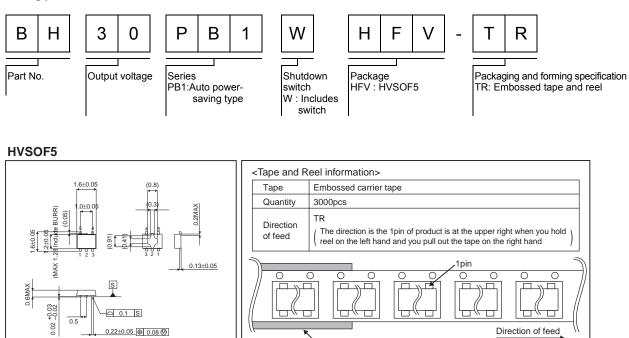
Parasitic diodes can occur inevitable in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Accordingly, methods by which parasitic diodes operate, such as applying a voltage that is lower than the GND (P substrate) voltage to an input pin, should not be used.



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BHDDPB1WHFV Series

Ordering part number



`Reel

 $\ast \, \text{Order}$ quantity needs to be multiple of the minimum quantity.

(Unit : mm)

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