

Dual Low Power JFET Input Operational Amplifier

General Description

The LF442 dual low power operational amplifier provides many of the same AC characteristics as the industry standard LM1458 while greatly improving the DC characteristics of the LM1458. The amplifier has the same bandwidth, slew rate, and gain (10 k Ω load) as the LM1458 and only draws one tenth the supply current of the LM1458. In addition the well matched high voltage JFET input devices of the LF442 reduce the input bias and offset currents by a factor of 10,000 over the LM1458. A combination of careful layout design and internal trimming guarantees very low input offset voltage and voltage drift. The LF442 also has a very low equivalent input noise voltage for a low power amplifier.

The LF442 is pin compatible with the LM1458 allowing an immediate 10 times reduction in power drain in many applications. The LF442 should be used where low power dissipation and good electrical characteristics are the major considerations.

Features

- 1/10 supply current of a LM1458: 400 µA (max)
- Low input bias current: 50 pA (Typ)
- Low input offset voltage: 1 mV (Typ)
- Low input offset voltage drift: 7 µV/°C (Typ)
- High gain bandwidth: 1 MHz (Typ)
- High slew rate: 1 V/µs (Typ)
- Low noise voltage for low power: $35 \text{ nV}/\sqrt{\text{Hz}}$ (Typ)
- Low input noise current: 0.01 pA/√Hz (Typ)
- High input impedance: 10¹²Ω

Ordering Information

NS Paart Number	SMD Part Number	NS Package Number	Package Description	
LF442MH/883	5962-9763301QGA	H08C	8LD T0–99 Metal Can	

Typical Connection



20149401

Connection Diagram



Pin 4 connected to case

Top View See NS Package Number H08C

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Simplified Schematic



Detailed Schematic



Absolute Maximum Ratings (Note 1)

Supply Voltage	+18V
Differential Input Voltage	±10V +30V
Input Voltage Bange (Note 3)	±30V
Output Short Circuit Duration (<i>Note 4</i>)	Continuous
Maximum Power Dissipation (<i>Note 2</i>)	900mW
Timax	150°C
Thermal Resistance	
θ _{JA}	
Still Air	161°C/W
500LF/Min Air flow	87°C/W
θ _{JC}	33°C/W
Operating Temperature Range	–55°C ≤ T _A ≤ 125°C
Storage Temperature Range	$-65^{\circ}C \le T_{A} \le 150^{\circ}C$
Lead Temperature (Soldering, 10 sec.)	260°C
ESD Tolerance (<i>Note 5</i>)	500V

Quality Conformance Inspection

Mil-Std-883, Method 5005 - Group A

Subgroup	Description	Temp (°C)
1	Static tests at	+25
2	Static tests at	+125
3	Static tests at	-55
4	Dynamic tests at	+25
5	Dynamic tests at	+125
6	Dynamic tests at	-55
7	Functional tests at	+25
8A	Functional tests at	+125
8B	Functional tests at	-55
9	Switching tests at	+25
10	Switching tests at	+125
11	Switching tests at	-55
12	Settling time at	+25
13	Settling time at	+125
14	Settling time at	-55

LF442 Electrical Characteristics

DC Parameters

The following conditions apply, unless otherwise specified. $V_S = \pm 15V$, $V_{CM} = 0V$, $R_S = 0\Omega$

Symbol	Parameter	Conditions	Notes	Min	Max	Unit	Sub- groups
I _{CC}	Supply Current				500	μA	1, 2, 3
V	Input Offset Voltage			-5.0	5.0	mV	1
♥ IO	input Onset Voltage	$n_{\rm S} = 10 R_{\rm S}^2$		-7.5	7.5	mV	2, 3
الد	Input Riss Current				0.1	nA	1
шВ	Input bias Current				20	nA	2
l				-0.05	0.05	nA	1
10				-10	10	nA	2
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 11V, R_{S} = 10K$		70		dB	1, 2, 3
PSRR	Dawar Currhy Dejection Detic	$V_{S}^{+} = +15V \text{ to } +6V,$ $V_{S}^{-} = -15V$		70		dB	1, 2, 3
	Power Supply Rejection Ratio	$V_{S}^{-} = -15V \text{ to } -6V,$ $V_{S}^{+} = +15V$		70		dB	1, 2, 3
. ^		$V_0 = 0V \text{ to } +10V,$		25		V/mV	4
+A _{VS}	Large Signal Voltage Gain	$R_L = 10K\Omega$	(Note 7)	15		V/mV	5, 6
^		$V_0 = 0V \text{ to } -10V,$		25		V/mV	4
-A _{VS}	Large Signal Voltage Gain	$R_L = 10K\Omega$	(Note 7)	15		V/mV	5, 6
V _O ⁺	Output Voltage Swing	$V_{I} = \pm 11V, R_{L} = 10K$		12		V	4, 5, 6
V _O -	Output Voltage Swing	$V_{I} = \pm 11V, R_{L} = 10K$			-12	V	4, 5, 6
V _{CM}	Input Common Mode Voltage Range		(<i>Note 6</i>)	11	-11	V	4, 5, 6

AC Parameters

The following conditions apply, unless otherwise specified. $V_S = \pm 15V$, $V_{CM} = 0V$, $R_S = 0\Omega$

Symbol	Parameter	Conditions	Notes	Min	Max	Unit	Sub- groups
SR+	Slew Rate	$V_{O} = -5V$ to +5V, $A_{V} = 1$, $R_{L} = 2K\Omega$, $C_{L} - 100pF$		0.6		V/µS	7
SR-	Slew Rate	$V_{O} = +5V$ to -5V, $A_{V} = 1$, $R_{L} = 2K\Omega$, $C_{L} - 100pF$		0.6		V/µS	7
GBW	Gain Band Width	$V_1 = 50 \text{mV}, f = 20 \text{KHz}$		0.6		MHz	7

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.

Note 2: The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{Jmax} (maximum junction temperature), θ_{JA} (package junction to ambient thermal resistance), and T_A (ambient temperature). The maximum allowable power dissipation at any temperature is $P_{Dmax} = (T_{Jmax} - T_A)/\theta_{JA}$ or the number given in the Absolute Maximum Ratings, whichever is lower.

Note 3: Unless otherwise specified the absolute maximum negative input voltage is equal to the negative power supply voltage.

Note 4: Any of the amplifier outputs can be shorted to ground indefinitely, however, more than one should not be simultaneously shorted as the maximum junction temperature will be exceeded.

Note 5: Human Body Model, 100pF discharged through 1.5K $\!\Omega$

Note 6: Parameter tested go-no-go only, guaranteed by CMRR test..

Note 7: V/mV in units column is equivalent to K in datalog.

Typical Performance Characteristics









Positive Current Limit



Supply Current













Gain Bandwidth



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 $\begin{array}{l} V_S=\pm\,15V\\ T_A=25^\circ C \end{array}$

1M

20149436

100k

Av = 1

1k

10k

FREQUENCY (Hz)

Pulse Response

 $R_{L} = 10 \text{ k}\Omega, C_{L} = 10 \text{ pF}$







20149409





20149408

Large Signal Non-Inverting



LF442QML

Application Hints

This device is a dual low power op amp with internally trimmed input offset voltages and JFET input devices (BI-FET II). These JFETs have large reverse breakdown voltages from gate to source and drain eliminating the need for clamps across the inputs. Therefore, large differential input voltages can easily be accommodated without a large increase in input current. The maximum differential input voltage is independent of the supply voltages. However, neither of the input voltages should be allowed to exceed the negative supply as this will cause large currents to flow which can result in a destroyed unit.

Exceeding the negative common-mode limit on either input will force the output to a high state, potentially causing a reversal of phase to the output. Exceeding the negative common-mode limit on both inputs will force the amplifier output to a high state. In neither case does a latch occur since raising the input back within the common-mode range again puts the input stage and thus the amplifier in a normal operating mode. Exceeding the positive common-mode limit on a single input will not change the phase of the output; however, if both inputs exceed the limit, the output of the amplifier will be forced to a high state.

The amplifiers will operate with a common-mode input voltage equal to the positive supply; however, the gain bandwidth and slew rate may be decreased in this condition. When the negative common-mode voltage swings to within 3V of the negative supply, an increase in input offset voltage may occur.

Each amplifier is individually biased to allow normal circuit operation with power supplies of $\pm 3.0V$. Supply voltages less

than these may degrade the common-mode rejection and restrict the output voltage swing.

The amplifiers will drive a 10 k load resistance to \pm 10V over the full temperature range.

Precautions should be taken to ensure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a socket as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

As with most amplifiers, care should be taken with lead dress, component placement and supply decoupling in order to ensure stability. For example, resistors from the output to an input should be placed with the body close to the input to minimize "pick-up" and maximize the frequency of the feedback pole by minimizing the capacitance from the input to ground.

A feedback pole is created when the feedback around any amplifier is resistive. The parallel resistance and capacitance from the input of the device (usually the inverting input) to AC ground set the frequency of the pole. In many instances the frequency of this pole is much greater than the expected 3 dB frequency of the closed loop gain and consequenty there is negligible effect on stability margin. However, if the feedback pole is less than approximately 6 times the expected 3 dB frequency a lead capacitor should be placed from the output to the input of the op amp. The value of the added capacitor should be such that the RC time constant of this capacitor and the resistance it parallels is greater than or equal to the original feedback pole time constant.



Runs from 9v batteries (±9V supplies)

Fully settable gain and time constant

Battery powered supply allows direct plug-in interface to strip chart recorder without common-mode problems



Trim 1M pot for 1 kHz full-scale output 15 mW power drain No integrator reset FET required Mount D1 and D2 in close proximity 1% linearity to 1 kHz





Revision History

Date Released	Revision	Section	Changes
12/16/2010	A	New release to corporate format	1 MDS datasheet converted into one corporate datasheet format. MNLF442M-X Rev 0A1 will be archived.



Notes

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Switching Regulators	www.national.com/switchers	Distributors	www.national.com/contacts		
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