



## TL5001

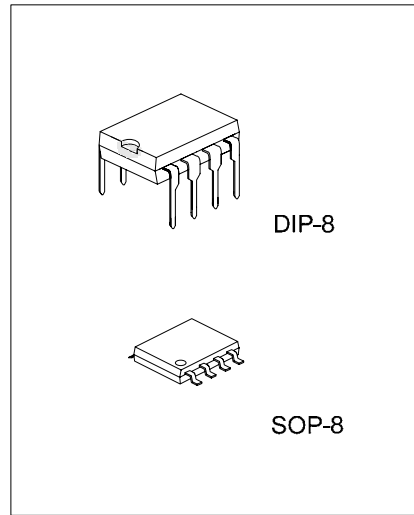
## LINEAR INTEGRATED CIRCUIT

### PULSE-WIDTH-MODULATION CONTROL CIRCUITS

#### DESCRIPTION

The UTC **TL5001** incorporates on a single monolithic chip all the functions required for a pulse width modulation (PWM) control circuit. Designed primarily for power-supply control, It contains an error amplifier, a regulator, an oscillator, a PWM comparator with a dead-time-control input, undervoltage lockout (UVLO), short-circuit protection (SCP), and an open-collector output transistor.

The error-amplifier common-mode voltage ranges from 0V to 1.5V. The noninverting input of the error amplifier is connected to a 1-V reference. Dead-time control (DTC) can be set to provide 0% to 100% dead time by connecting an external resistor between DTC and GND. The oscillator frequency is set by terminating RT with an external resistor to GND. During low Vcc conditions, the UVLO circuit turns the output off until Vcc recovers to its normal operating range.



\*Pb-free plating product number: TL5001L

#### FEATURES

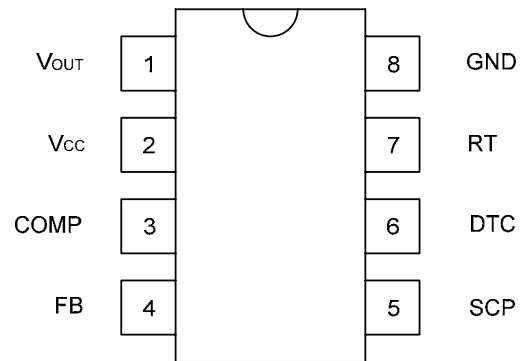
- \*Complete PWM power control
- \*3.6-V to 40-V operation
- \*Internal under voltage-lockout circuit
- \*Internal short-circuit protection
- \*Oscillator frequency : 20kHz to 500kHz
- \*Variable dead timer provides control over total range

#### ORDERING INFORMATION

Ordering Number		Package	Packing
Normal	Lead Free Plating		
TL5001-D08-T	TL5001L-D08-T	DIP-8	Tube
TL5001-S08-R	TL5001L-S08-R	SOP-8	Tape Reel
TL5001-S08-T	TL5001L-S08-T	SOP-8	Tube

<p>TL5001L-D08-R</p> <p>(1) Packing Type (2) Package Type (3) Lead Plating</p>	<p>(1) R: Tape Reel, T: Tube (2) D08: DIP-8, S08: SOP-8 (3) L: Lead Free Plating, Blank: Pb/Sn</p>
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### ■ PIN CONFIGURATION



### ■ ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	RATINGS	UNIT
Supply Voltage (Note 1)	$V_{CC}$	41	V
Amplifier Input Voltage	$V_{I(FB)}$	20	V
Output Voltage	$V_{OUT}$	51	V
Output Current	$I_{OUT}$	21	mA
Output Peak Current	$I_{O(PEAK)}$	100	mA
Continuous Total Power Dissipation	See dissipation rating table		
Operating Ambient Temperature Range	$T_{OPR}$	-20 ~ +85	°C
Storage Temperature Range	$T_{STG}$	-65 ~ +150	°C

Note: 1. All voltage values are with respect to the network ground terminal.

2. Absolute maximum ratings are those values beyond which the device could be permanently damaged.

Absolute maximum ratings are stress ratings only and functional device operation is not implied.

### ■ DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 70^\circ\text{C}$ POWER RATING
DIP-8	1000mW	8.0mW/°C	640mW	520mW	200mW
SOP-8	725mW	5.8mW/°C	464mW	377mW	145mW

### ■ RECOMMENDED OPERATING CONDITIONS

PARAMETER	SYMBOL	MIN	MAX	UNIT
Supply Voltage	$V_{CC}$	3.6	40	V
Amplifier Input Voltage	$V_{I(FB)}$	0	1.5	V
Output Voltage	$V_{OUT}$		50	V
Output Current	$I_{OUT}$		20	mA
COMP Source Current			45	$\mu\text{A}$
COMP dc Load Resistance		100		k
Oscillator Timing resistor	$R_t$	15	250	k
Oscillator Frequency	$f_{osc}$	20	500	kHz
Operating Ambient Temperature Range	$T_A$	-20	85	°C

- ELECTRICAL CHARACTERISTICS OVER RECOMMENDED OPERATING FREE-AIR TEMPERATURE RANGE (V<sub>CC</sub>=6V, f<sub>osc</sub>=100kHz, all typical values at T<sub>A</sub>=25°C, unless otherwise noted)

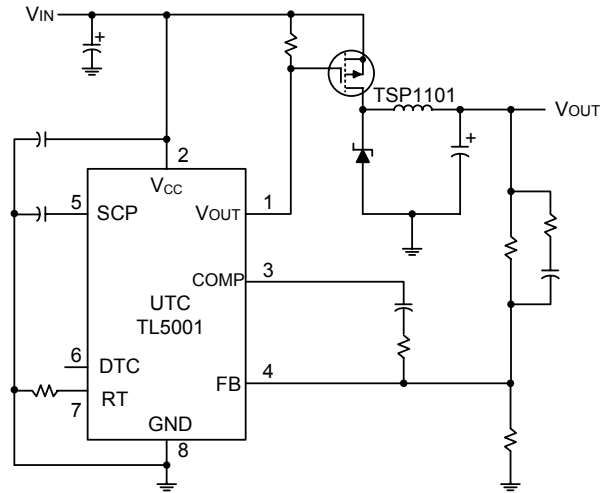
PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>Reference</b>						
Output Voltage	V <sub>OUT</sub>	COMP Connected to FB	0.95	1	1.05	V
Input Regulation		V <sub>CC</sub> =3.6V ~ 40V		2	12.5	mV
Output Voltage Change with Temperature		T <sub>A</sub> = -20°C ~ 25°C	-10	-1	10	mV/V
		T <sub>A</sub> = 25°C ~ 85°C	-10	2	10	
<b>Undervoltage Lockout</b>						
Threshold Voltage	Upper	V <sub>THR</sub>	T <sub>A</sub> = 25°C		3	V
	Lower		T <sub>A</sub> = 25°C		2.8	V
	Reset		T <sub>A</sub> = 25°C	2.1	2.55	V
Hysteresis		T <sub>A</sub> = 25°C	100	200		mV
<b>Short Circuit Protection</b>						
SCP Threshold voltage		T <sub>A</sub> = 25°C	0.95	1.00	1.05	V
SCP Voltage, Latched		No pullup	140	185	230	mV
SCP Voltage, UVLO Standby		No pullup		60	120	mV
Input Source Current		T <sub>A</sub> = 25°C	-10	-15	-20	μA
SCP Comparator 1 Threshold Voltage				1.5		V
<b>Oscillator</b>						
Frequency	f	R <sub>t</sub> =100kΩ		100		kHz
Standard Deviation of Frequency				15		kHz
Frequency Change with Voltage		V <sub>CC</sub> =3.6V ~ 40V		1		kHz
Frequency Change with Temperature		T <sub>A</sub> = -40°C ~ 25°C	-4	-0.4	4	kHz
		T <sub>A</sub> = -20°C ~ 25°C	-4	-0.4	4	kHz
		T <sub>A</sub> = 25°C ~ 85°C	-4	-0.2	4	kHz
Voltage at RT	V <sub>RT</sub>			1		V
<b>Dead-time Control</b>						
Output (source) Current	I <sub>O(SOURCE)</sub>	V <sub>(DT)</sub> =1.5V	0.9*I <sub>RT</sub> (NOTE)		1.1*I <sub>RT</sub> (NOTE)	μA
Input Threshold Voltage	V <sub>I(THR)</sub>	Duty cycle=0%	0.5	0.7		V
		Duty cycle=100%		1.3	1.5	
<b>Error Amplifier</b>						
Input Voltage	V <sub>IN</sub>	V <sub>CC</sub> =3.6V ~ 40V	0		1.5	V
Input Bias Current	I <sub>I(BIAS)</sub>			-160	-500	nA
Output Voltage Swing	Positive	V <sub>O(SW)</sub>		1.5	2.3	V
	Negative				0.3	0.4
Open-Loop Voltage Amplification	G <sub>V</sub>			80		dB
Unity-Gain Bandwidth				1.5		MHz
Output (sink) Current	I <sub>O(SINK)</sub>	V <sub>I(FB)</sub> =1.2V, COMP=1V	100	600		μA
Output (source) Current	I <sub>O(SOURCE)</sub>	V <sub>I(FB)</sub> =0.8V, COMP=1V	-45	-70		μA
<b>Output</b>						
Output Saturation Voltage	V <sub>O(SAT)</sub>	I <sub>OUT</sub> =10mA		1.5	2	V
Off-State Current	I <sub>OFF</sub>	V <sub>OUT</sub> =50V, V <sub>CC</sub> =0			10	μA
		V <sub>OUT</sub> =50V			10	
Short-Circuit Output Current	I <sub>O(SC)</sub>	V <sub>OUT</sub> =6V		40		mA
<b>Total Device</b>						
Standby Supply Current (Off state)	I <sub>STN-BY</sub>			1	1.5	mA
Average Supply Current		R <sub>t</sub> =100kΩ		1.4	2.1	mA

Note: Output source current at RT

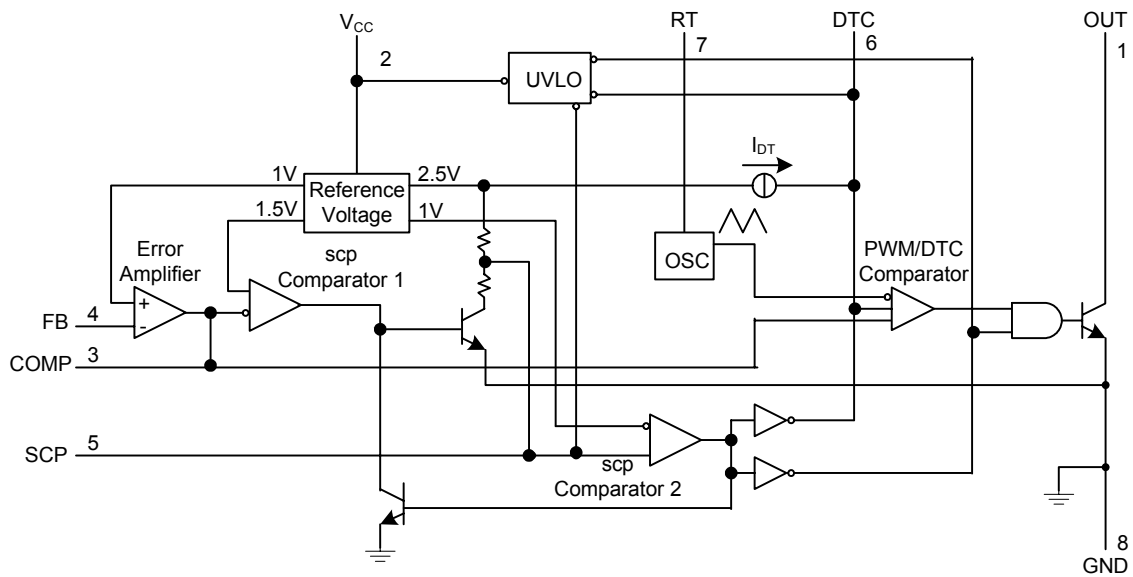
# TL5001

## LINEAR INTEGRATED CIRCUIT

### ■ SCHEMATIC FOR TYPICAL APPLICATION



### ■ FUNCTIONAL BLOCK DIAGRAM



## ■ DETAILED DESCRIPTION

### VOLTAGE REFERENCE

A 2.5-V regulator operating from Vcc is used to power the internal circuitry of the TL5001 and as a reference for the error amplifier and SCP circuit. A resistive divider provides a 1-V reference for the error amplifier non-inverting input which typically is within 2% of nominal over the operating temperature range.

### ERROR AMPLIFIER

The error amplifier compares a sample of the dc-to-dc converter output voltage to the 1-V reference and generates an error signal for the PWM comparator. The dc-to-dc converter output voltage is set by selecting the error –amplifier gain (see Figure 1), using the following expression.

$$V_{OUT} = (1 + R1/R2) (1V)$$

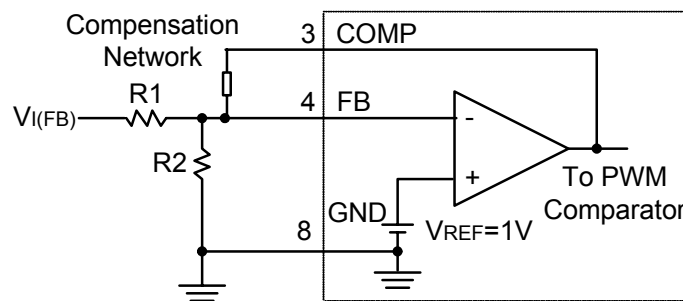


Figure 1. Error-Amplifier Gain Setting

The error-amplifier output is brought out as COMP for use in compensating the dc-to-dc converter control loop for stability. Because the amplifier can only source 45µA, the total dc load resistance should be 100kΩ or more.

### OSCILLATOR/PWM

The oscillator frequency (fosc) can be set between 20kHz and 500kHz by connecting a resistor between RT and GND. Acceptable resistor values range from 15 kΩ to 250 kΩ. The oscillator frequency can be determined by using the graph shown in Figure 5.

The oscillator output is a triangular wave with a minimum value of approximately 0.7V and a maximum value of approximately 1.3V. The PWM comparator compares the error-amplifier output voltage and the DTC input voltage to the triangular wave and turns the output transistor off whenever the triangular wave is greater than the lesser of the two inputs.

■ DETAILED DESCRIPTION(Cont.)

**DEAD TIME CONTORL (DTC)**

DTC provides a means of limiting the output-switch duty cycle to a value less than 100%, which is critical for boost and flyback converters. A current source generates a reference current (IDT) at DTC that is nominally equal to the current at the oscillator timing terminal, RT. Connecting a resistor between DTC and GND generates a dead-time reference voltage (VDT), which the PWM/DTC comparator compares to the oscillator triangle wave as described in the previous section. Nominally, the maximum duty cycle is 0% when VDT is 0.7V or less and 100% when VDT is 1.3V or greater. Because the triangle wave amplitude is a function of frequency and the source impedance of RT is relatively high(1250Ω),choosing RDT for a specific maximum duty cycle, D, is accomplished using the following equation and the voltage limits for the frequency in question as found in Figure 11(Voscmax and Voscmin are the maximum and minimum oscillator levels):

$$R_{DT}=(R_t + 1250)[D(V_{osc \ max}-V_{osc \ min})+V_{osc \ min}]$$

Where

R<sub>DT</sub> and R<sub>t</sub> are in ohms, D in decimal

Soft start can be implemented by paralleling the DTC resistor with a capacitor (C<sub>DT</sub>) as shown in Figure 2. During soft start, the voltage at DTC is derived by the following equation:

$$V_{DT} \approx I_{DT} R_{DT} (1 - e^{-t/R_{DT} C_{DT}})$$

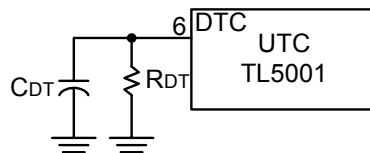


Figure 2. Soft- Start Circuit

If the dc-to-dc converter must be in regulation within a specified period of time, the time constant, R<sub>DT</sub>C<sub>DT</sub>, should be t<sub>0</sub>/3 to t<sub>0</sub>/5. The UTC **TL5001** remains off until V<sub>DT</sub> ≈ 0.7V, the minimum ramp value. C<sub>DT</sub> is discharged every time UVLO or SCP becomes active.

**UNDERVOLTAGE-LOCKOUT (UVLO) PROTECTION**

The undervoltage-lockout circuit turns the output transistor off and resets the SCP latch whenever the supply voltage drops too low (approximately 3V at 25 ) for proper operation. A hysteresis voltage of 200mV eliminates false triggering on noise and chattering.

## ■ DETAILED DESCRIPTION(Cont.)

### SHORT-CIRCUIT PROTECTION (SCP)

The UTC **TL5001** includes short-circuit protection (see Figure 3), which turns the power switch off to prevent damage when the converter output is shorted. When activated, The SCP prevents the switch from being turned on until the internal latching circuit is reset. The circuit is reset by reducing the input voltage until UVLO becomes active or until the SCP terminal is pulled to ground externally.

When a short circuit occurs, the error-amplifier output at COMP rises to increase the power-switch duty cycle in an attempt to maintain the output voltage. SCP comparator 1 starts an RC timing circuit when COMP exceeds 1.5V. If the short is removed and the error-amplifier output drops below 1.5V before time out, normal converter operation continues. If the fault is still present at the end of the time-out period, the time sets the latching circuit and turns off the UTC **TL5001** output transistor.

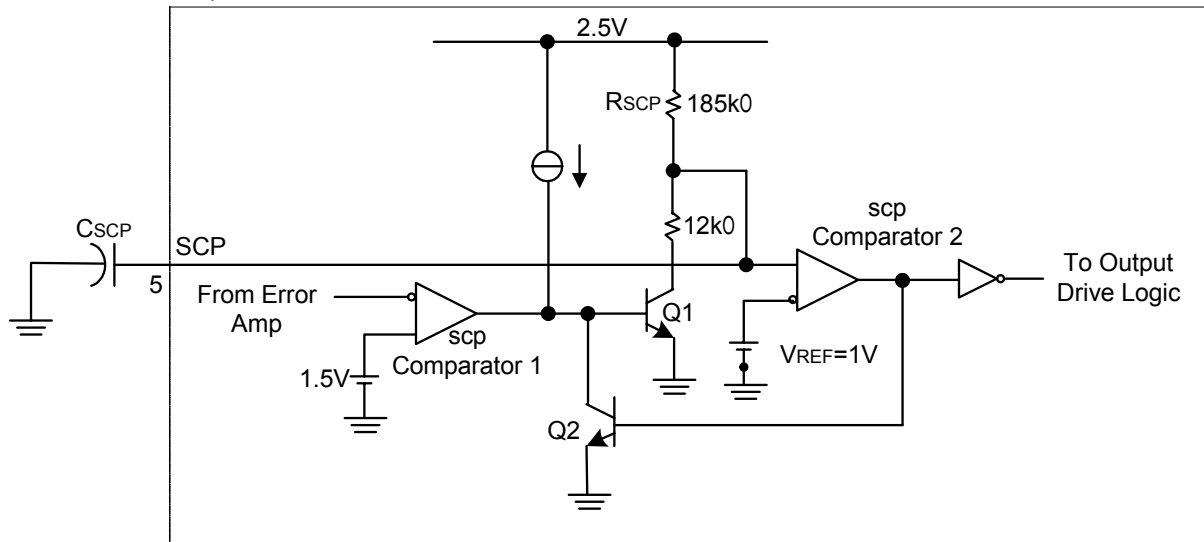


Figure 3.SCP Circuit

The timer operates by charging an external capacitor ( $C_{scp}$ ), connected between the SCP terminal and ground, towards 2.5V through a  $185k\Omega$  resistor ( $R_{scp}$ ). The circuit begins charging from an initial voltage of approximately 185mV and times out when capacitor voltage reaches 1V. The output of SCP comparator 2 then goes high, turns on Q2, and latches the timer circuit. The expression for setting the SCP time period is derived from the following equation:

$$V_{scp} = (2.5 - 0.185)(1 - e^{-t/\tau}) + 0.185$$

Where

$$\tau = R_{scp} C_{scp}$$

The end of the time-out period,  $t_{scp}$ , occurs when  $V_{scp} = 1V$ . Solving for  $C_{scp}$  yields:

$$C_{scp} = 12.46 * t_{scp}$$

Where

$$t \text{ is in seconds, } C \text{ in } \mu F$$

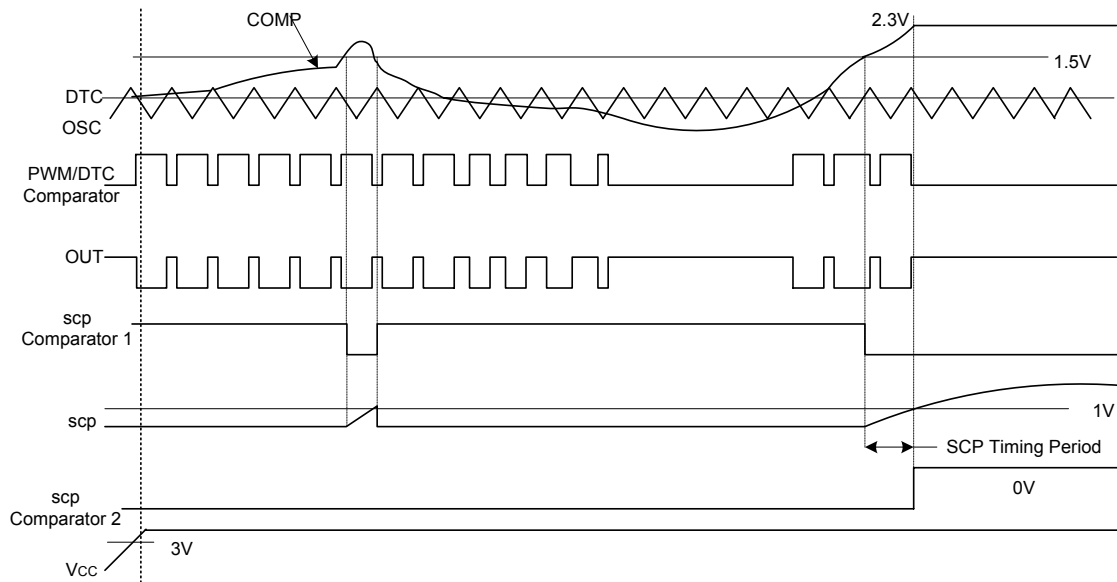
$t_{scp}$  must be much longer (generally 10 to 15 times) than the converter start-up period or the converter will not start.

### OUTPUT TRANSISTOR

The output of the UTC **TL5001** is an open-collector transistor with a maximum collector current rating of 21mA and a voltage rating of 51V. The output is turned on under the following conditions: the oscillator triangle wave is lower than both the DTC voltage and the error-amplifier output voltage, the UVLO circuit is inactive, and the short-circuit protection circuit is inactive.



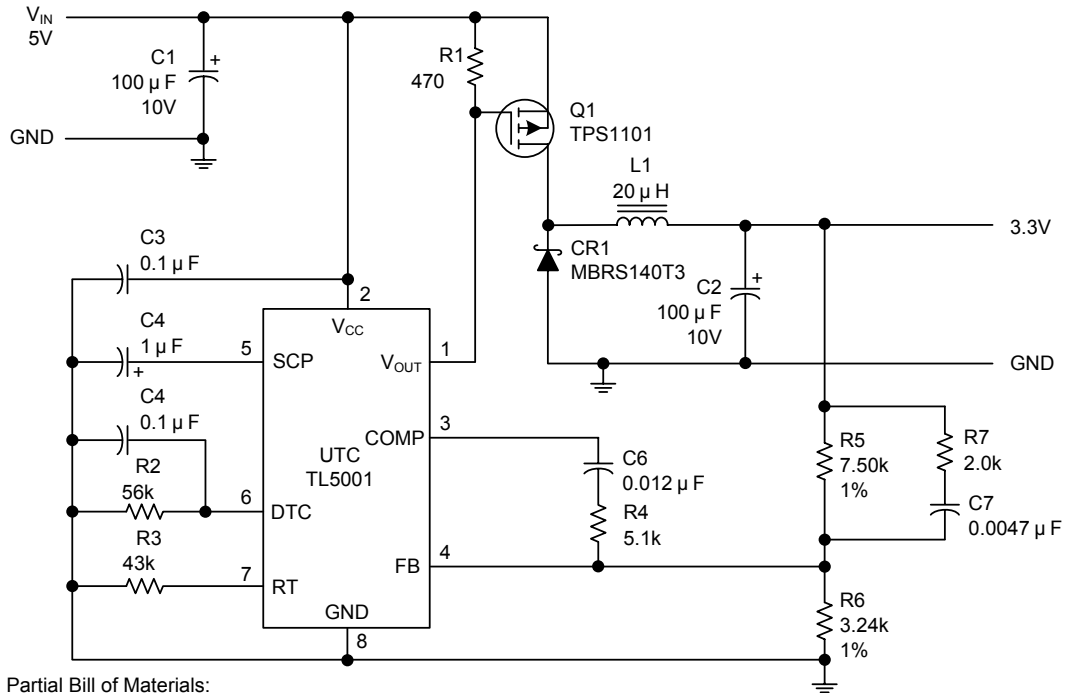
■ PARAMETER MEASUREMENT INFORMATION



NOTE A: The waveforms show timing characteristics for an intermittent short circuit and a longer short circuit that is sufficient to activate SCP.

Figure 4. PWM Timing Diagram

### APPLICATION INFORMATION



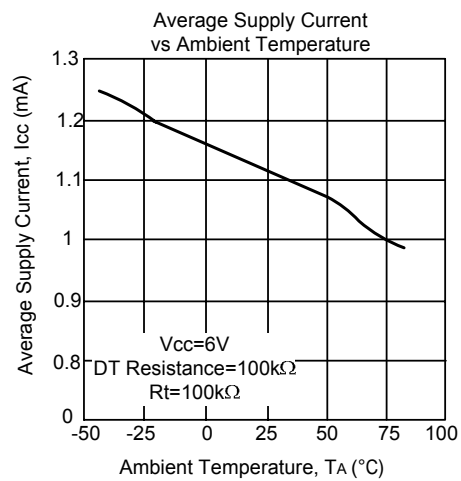
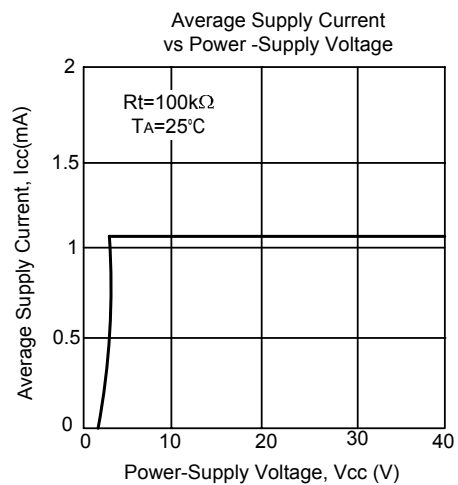
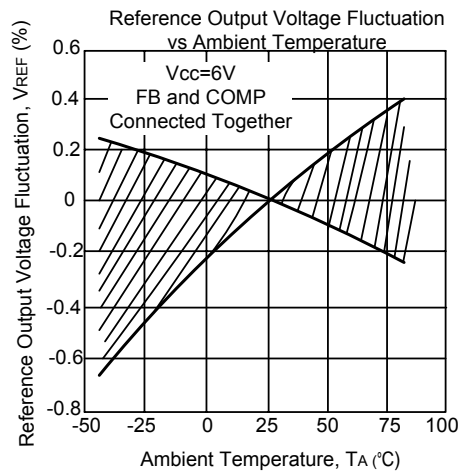
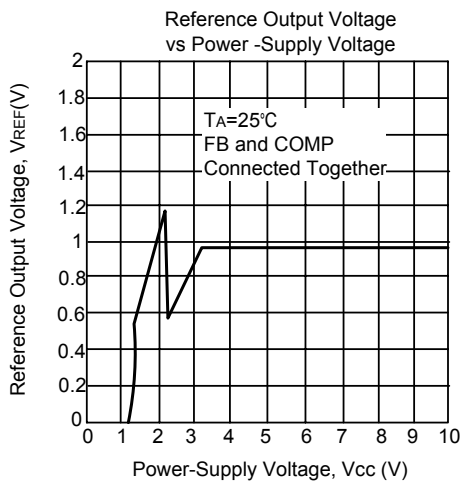
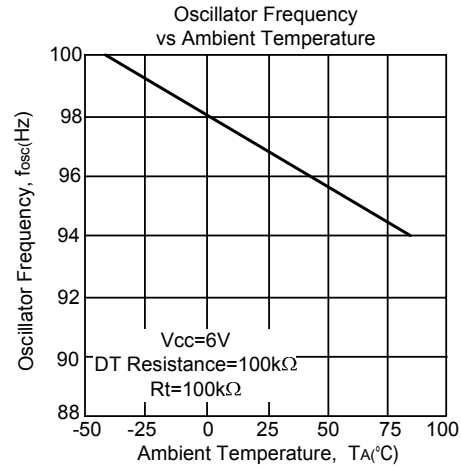
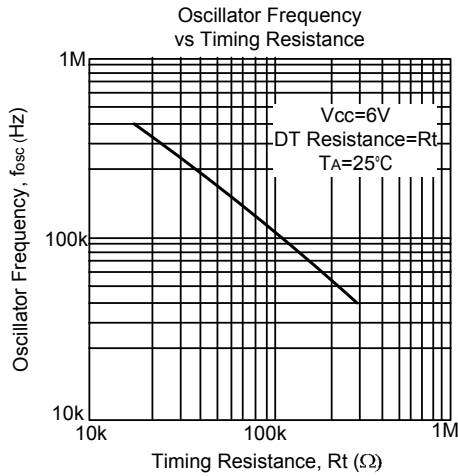
Partial Bill of Materials:

Q1	TPS1101	Texas Instruments
L1	CTX20-1 or 23 turns of #28 wire on Micrometals No. T50-26B core	Coiltronics
C1	TPSD 107M010R0100	AVX
C2	TPSD 107M010R0100	AVX
CR1	MBRS 140T3	Motorola

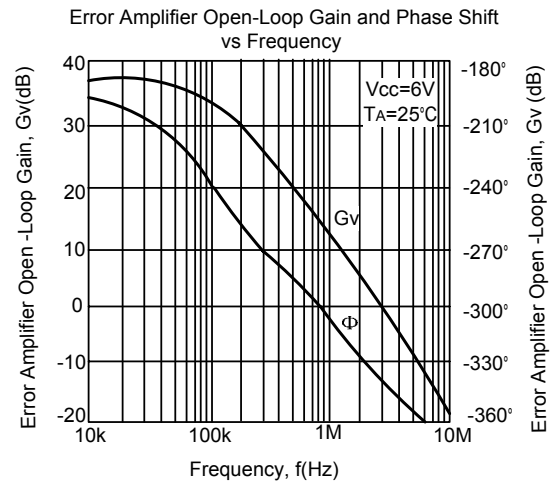
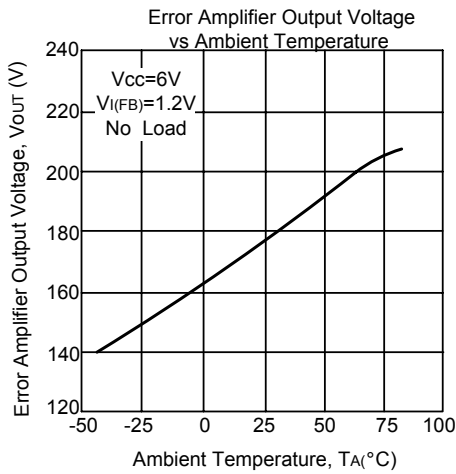
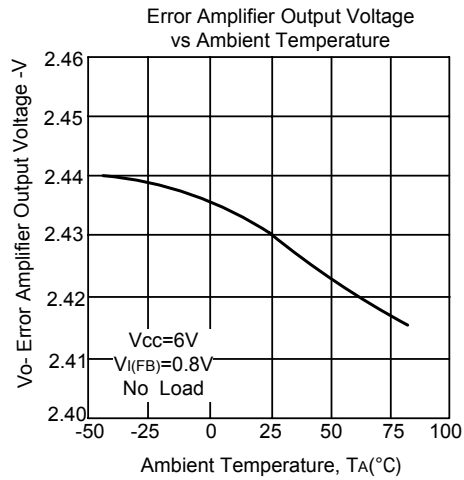
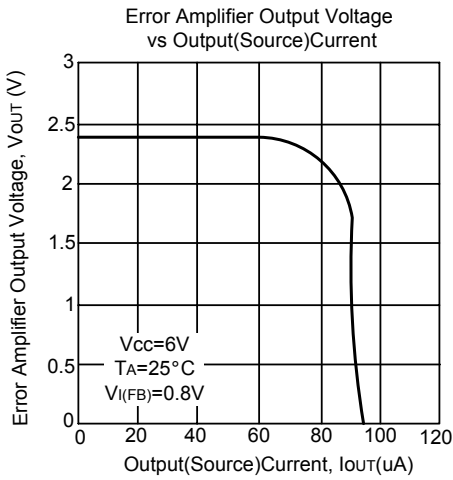
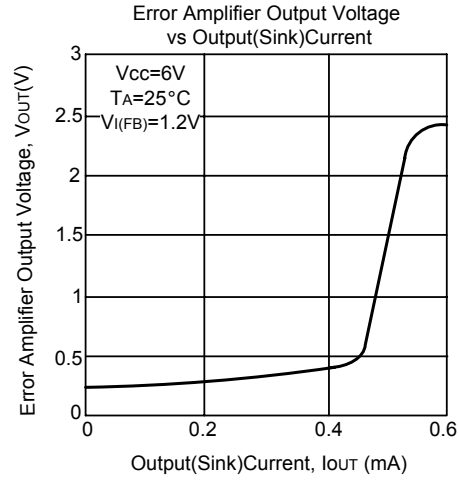
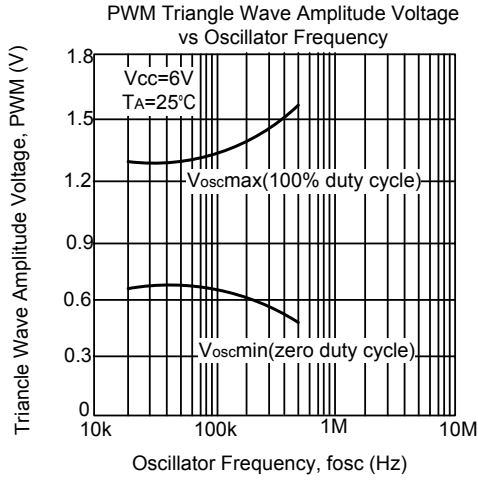
NOTES: A. Frequency = 200KHz  
 B. Duty cycle = 90% max  
 C. Soft-start time constant (TC) = 5.6ms  
 D. SCP TC = 70msA

Figure 5. Step-Down Converter

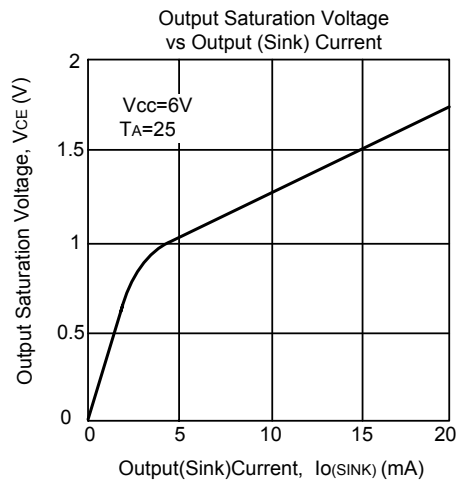
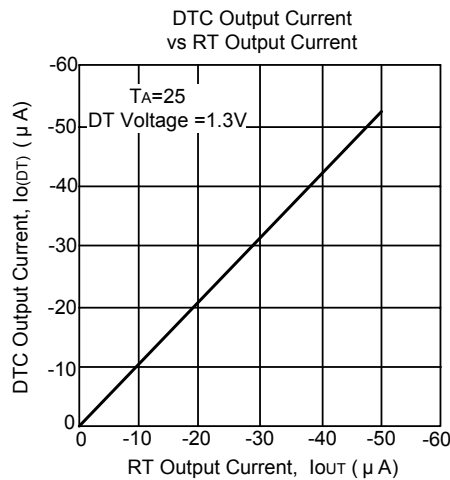
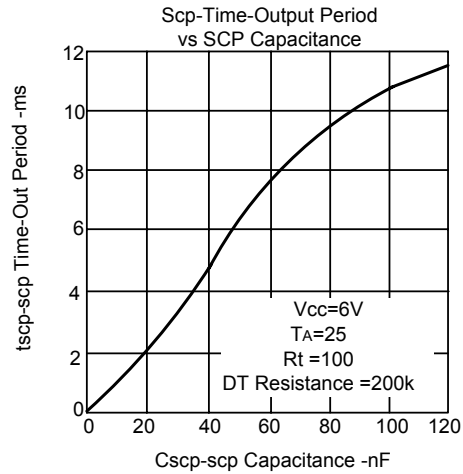
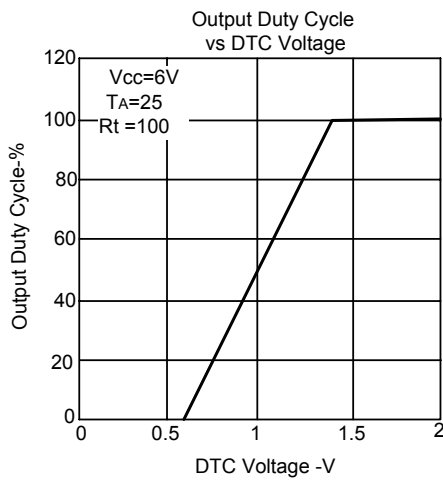
## ■ TYPICAL CHARACTERISTICS



■ TYPICAL CHARACTERISTICS(Cont.)



### ■ TYPICAL CHARACTERISTICS(Cont.)



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