



For brush motors

H-bridge drivers (7V max.) BD6210, BD6211, BD6212, BD6215, BD6216, BD6217



RoHS Directive Compliance

Overview

These H-bridge drivers are full bridge drivers for brush motor applications. Each IC can operate at a wide range of power-supply voltages (from 3V to 36V), supporting output currents of up to 2A. MOS transistors in the output stage allow for PWM signal control, while the integrated VREF voltage control function of previous models offers direct replacement of deprecated motor driver ICs. These highly efficient H-bridge driver ICs facilitate low-power consumption design.

Features

- 1) Built-in, selectable one channel or two channels configuration
- 2) Low standby current
- 3) Supports PWM control signal input (20kHz to 100kHz)
- 4) VREF voltage setting pin enables PWM duty control
- 5) Cross-conduction prevention circuit
- 6) Four protection circuits provided: OCP, OVP, TSD and UVLO

Applications

VCR; CD/DVD players; audio-visual equipment; optical disc drives; PC peripherals; car audios; car navigation systems; OA equipments

Line up matrix

Poting voltage	Channels	Maximum output current				
Rating voltage	Channels	0.5A	1.0A	2.0A		
7V	1ch	BD6210 HFP / F	BD6211 HFP / F	BD6212 HFP / FP		
7.V	2ch	BD6215 FV / FP	BD6216 FP / FM	BD6217 FM		
18V	1ch	BD6220 HFP / F	BD6221 HFP / F	BD6222 HFP / FP		
180	2ch	BD6225 FV / FP	BD6226 FP / FM	BD6227 FM		
36V	1ch	BD6230 HFP / F	BD6231 HFP / F	BD6232 HFP / FP		
	2ch	BD6235 FV / FP	BD6236 FP / FM	BD6237 FM		

* Packages; F:SOP8, FV: SSOPB24, HFP:HRP7, FP:HSOP25, FM:HSOPM28

Absolute maximum ratings (Ta=25°C)

Parameter	Symbol	Ratings	Unit
Supply voltage	VCC	7	V
Output current	Іомах	0.5 ^{*1} / 1.0 ^{*2} / 2.0 ^{*3}	А
All other input pins	Vin	-0.3 ~ VCC	V
Operating temperature	Topr	-40 ~ +85	°C
Storage temperature	Тѕтс	-55 ~ +150	°C
Power dissipation	Pd	0.687 * ⁴ / 0.98 * ⁵ / 1.4 * ⁶ / 1.45 * ⁷ / 2.2 * ⁸	W
Junction temperature	Tjmax	150	°C

*** Notes: All voltages are with respect to ground.

*1 BD6210 / BD6215. Do not, exceed Pd or ASO.

*2 BD6211 / BD6216. Do not, exceed Pd or ASO.

*3 BD6212 / BD6217. Do not, exceed Pd or ASO.

*4 SOP8 package. Mounted on a 70mm x 70mm x 1.6mm FR4 glass-epoxy board with less than 3% copper foil. Derated at 5.5mW/°C above 25°C.

*5 SSOPB24 package. Mounted on a 70mm x 70mm x 1.6mm FR4 glass-epoxy board with less than 3% copper foil. Derated at 7.8mW/°C above 25°C.

*6 HRP7 package. Mounted on a 70mm x 70mm x 1.6mm FR4 glass-epoxy board with less than 3% copper foil. Derated at 11.2mW/°C above 25°C.

*7 HSOP25 package. Mounted on a 70mm x 70mm x 1.6mm FR4 glass-epoxy board with less than 3% copper foil. Derated at 11.6mW/°C above 25°C.

*8 HSOPM28 package. Mounted on a 70mm x 70mm x 1.6mm FR4 glass-epoxy board with less than 3% copper foil. Derated at 17.6mW/°C above 25°C.

•Operating conditions (Ta=25°C)

Parameter	Symbol	Conditions	Unit
Supply voltage	VCC	3.0 ~ 5.5	V
VREF voltage	VREF	1.5 ~ 5.5	V

• Electrical characteristics (Unless otherwise specified, Ta=25°C and VCC=VREF=5V)

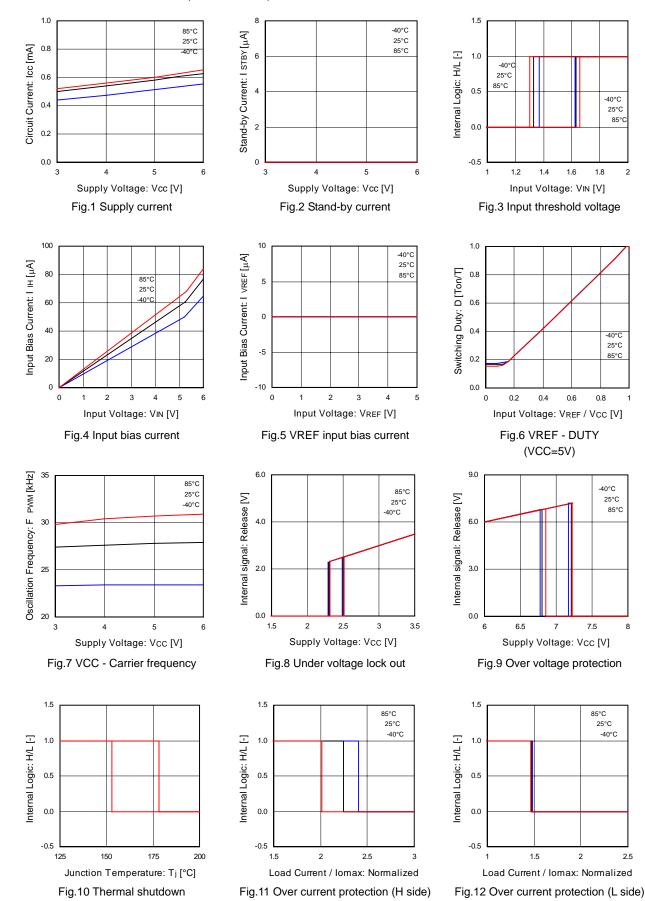
Deremeter	Current of	Limits		Linit	Conditions	
Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions
Supply current	lcc	0.4	0.7	1.5	mA	Forward / Reverse / Brake
Stand-by current	ISTBY	-	0	10	μA	Stand-by
Input high voltage	Vін	2.0	-	-	V	
Input low voltage	VIL	-	-	0.8	V	
Input bias current	Іін	30	50	100	μA	VIN=5.0V
Output ON resistance *1	Ron	0.5	1.0	1.5	Ω	Io=0.25A, vertically total
Output ON resistance *2	Ron	0.5	1.0	1.5	Ω	Io=0.5A, vertically total
Output ON resistance *3	Ron	0.2	0.5	1.0	Ω	Io=1.0A, vertically total
VREF bias current	IVREF	-10	0	10	μA	VREF=VCC
Carrier frequency	FPWM	20	25	35	kHz	VREF=3.75V
Input frequency range	Fмах	20	-	100	kHz	FIN / RIN

*1 BD6210 / BD6215

*2 BD6211 / BD6216

*3 BD6212 / BD6217

• Electrical characteristic curves (Reference data)



• Electrical characteristic curves (Reference data) - Continued

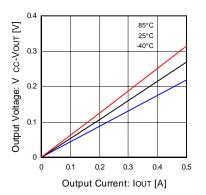


Fig.13 Output high voltage (0.5A class)

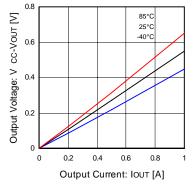


Fig.14 Output high voltage (1A class)

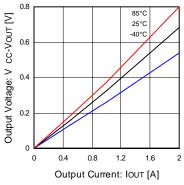


Fig.15 Output high voltage (2A class)

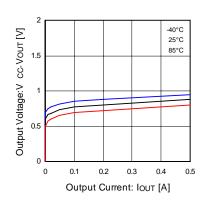


Fig.16 High side body diode (0.5A class)

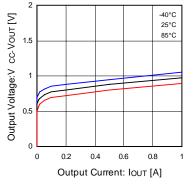


Fig.17High side body diode (1A class)

85°C

25°C

-40°C

0.8

0.6

0.4

0.2

0

0

0.2

0.4

Output Current: IOUT [A]

Fig.20 Output low voltage (1A class)

0.6

0.8

Output Voltage: V OUT [V]

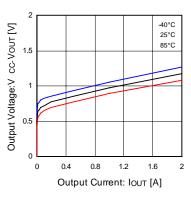


Fig.18 High side body diode (2A class)

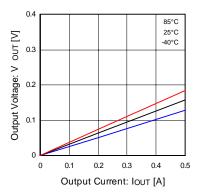
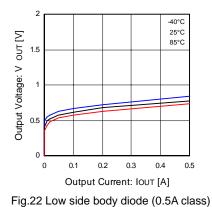


Fig.19 Output low voltage (0.5A class)



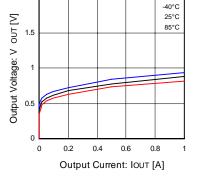


Fig.23 Low side body diode (1A class)

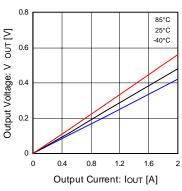
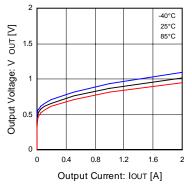
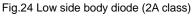


Fig.21 Output low voltage (2A class)





• Block diagram and pin configuration

BD6210F / BD6211F

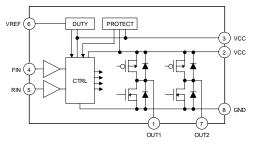


Fig.25 BD6210F / BD6211F

OUT1 📼	ത്	GND
vcc 🗖	\sim	
vcc 🎞		D VREF
		🗖 RIN
•		

Fig.26 SOP8

Table 1 BD6210/11F Pin Function Name OUT1 Driver output 1 VCC 2 Power supply 3 VCC Power supply 4 FIN Control input (forward) 5 RIN Control input (reverse) 6 VREF Duty setting pin 7 OUT2 Driver output 8 GND Ground

BD6210HFP / BD6211HFP / BD6212HFP

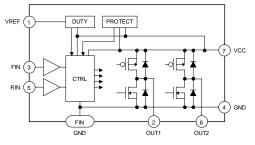


Fig.27 BD6210/11/12HFP

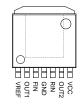


Fig.28 HRP7

Pin	Name	Function
1	VREF	Duty setting pin
2		Driver output

Table 2 BD6210/11/12HFP

2	OUT1	Driver output
3	FIN	Control input (forward)
4	GND	Ground
5	RIN	Control input (reverse)
6	OUT2	Driver output
7	VCC	Power supply
FIN	GND	Ground

• Block diagram and pin configuration – Continued

BD6212FP

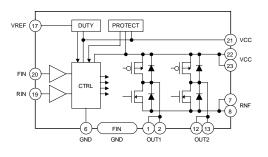


Fig.29 BD6212FP

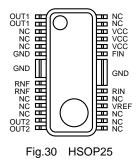


	Table 3	BD6212FP
Pin	Name	Function
1,2	OUT1	Driver output
6	GND	Small signal ground
7,8	RNF	Power stage ground
12,13	OUT2	Driver output
17	VREF	Duty setting pin
19	RIN	Control input (reverse)
20	FIN	Control input (forward)
21	VCC	Power supply
22,23	VCC	Power supply
FIN	GND	Ground

Note: All pins not described above are NC pins.

BD6215FV

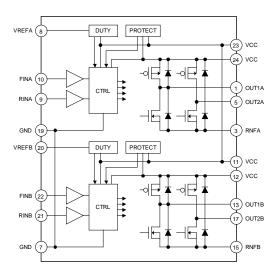


Fig. 31 BD6215FV

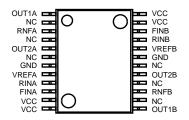


Fig. 32 SSOPB24

	Table 4	BD6215FV
Pin	Name	Function
1	OUT1A	Driver output
3	RNFA	Power stage ground
5	OUT2A	Driver output
7	GND	Small signal ground
8	VREFA	Duty setting pin
9	RINA	Control input (reverse)
10	FINA	Control input (forward)
11	VCC	Power supply
12	VCC	Power supply
13	OUT1B	Driver output
15	RNFB	Power stage ground
17	OUT2B	Driver output
19	GND	Small signal ground
20	VREFB	Duty setting pin
21	RINB	Control input (reverse)
22	FINB	Control input (forward)
23	VCC	Power supply
24	VCC	Power supply

Note: All pins not described above are NC pins.

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• Block diagram and pin configuration – Continued

PROTECT DUTY VREFA (24) vcc (25) vcc -dÐ **本** ⊸⊡≰ FINA (1) OUT1A CTRL RINA OUT2A 6 Ż ╢╋┓ + GND (20 RNFA VREFB (21 DUTY PROTECT (12) vcc 13) vcc -017 FINB (23 (14) OUT1B CTRL RINB (22 (19) OUT2B 4 ╢╋┑ ۱th RNFB GND 16) (8 FIN GND

BD6215FP, BD6216FP/FM



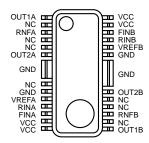


Fig. 34 HSOP25

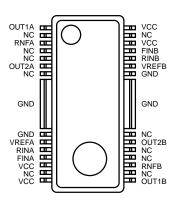


Fig. 35 HSOPM28

Pin	Name	Function		
1	OUT1A	Driver output		
3	RNFA	Power stage ground		
6	OUT2A	Driver output		
8	GND	Small signal ground		
9	VREFA	Duty setting pin		
10	RINA	Control input (reverse)		
11	FINA	Control input (forward)		
12	VCC	Power supply		
13	VCC	Power supply		
14	OUT1B	Driver output		
16	RNFB	Power stage ground		
19	OUT2B	Driver output		
20	GND	Small signal ground		
21	VREFB	Duty setting pin		
22	RINB	Control input (reverse)		
23	FINB	Control input (forward)		
24	VCC	Power supply		
25	VCC	Power supply		
FIN	GND	Ground		

Table 5 BD6215FP / BD6216FP

Note: All pins not described above are NC pins.

Table 6 BD6216FM

Pin	Name	Function
1	OUT1A	Driver output
3	RNFA	Power stage ground
6	OUT2A	Driver output
8	GND	Small signal ground
9	VREFA	Duty setting pin
10	RINA	Control input (reverse)
11	FINA	Control input (forward)
12	VCC	Power supply
14	VCC	Power supply
15	OUT1B	Driver output
17	RNFB	Power stage ground
20	OUT2B	Driver output
22	GND	Small signal ground
23	VREFB	Duty setting pin
24	RINB	Control input (reverse)
25	FINB	Control input (forward)
26	VCC	Power supply
28	VCC	Power supply
FIN	GND	Ground

Note: All pins not described above are NC pins.

• Block diagram and pin configuration – Continued

BD6217FM

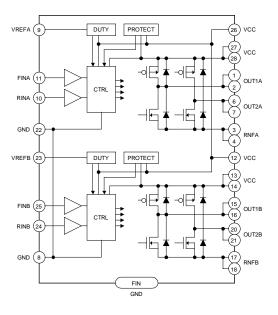


Fig. 36 BD6217FM

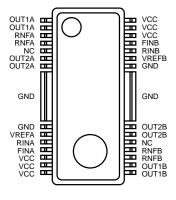


Fig. 37 HSOPM28

	Table 7	BD6217FM	
Pin	Name	Function	
1,2	OUT1A	Driver output	
3,4	RNF A	Power stage ground	
6,7	OUT2A	Driver output	
8	GND	Small signal ground	
9	VREFA	Duty setting pin	
10	RINA	Control input (reverse)	
11	FINA	Control input (forward)	
12	VCC	Power supply	
13,14	VCC	Power supply	
15,16	OUT1B	Driver output	
17,18	RNFB	Power stage ground	
20,21	OUT2B	Driver output	
22	GND	Small signal ground	
23	VREFB	Duty setting pin	
24	RINB	Control input (reverse)	
25	FINB	Control input (forward)	
26	VCC	Power supply	
27,28	VCC	Power supply	
FIN	GND	Ground	
NUMBER OF STREET	Note: All nins not described above are NC nins		

Note: All pins not described above are NC pins.

Functional descriptions

1) Operation modes

\square	FIN	RIN	VREF	OUT1	OUT2	Operation
а	L	L	Х	Hi-Z*	Hi-Z*	Stand-by (idling)
b	Н	L	VCC	Н	L	Forward (OUT1 > OUT2)
С	L	Н	VCC	L	Н	Reverse (OUT1 < OUT2)
d	Н	Н	Х	L	L	Brake (stop)
е	PWM	L	VCC	Н	PWM	Forward (PWM control)
f	L	PWM	VCC	PWM	Н	Reverse (PWM control)
g	Н	L	Option	Н	PWM	Forward (VREF control)
h	L	Н	Option	PWM	Н	Reverse (VREF control)

Table 8 Logic table

* Hi-Z is the off state of all output transistors. Please note that this is the state of the connected diodes, which differs from that of the mechanical relay.

X : Don't care

a) Stand-by mode

Stand-by operates independently of the VREF pin voltage. In stand-by mode, all internal circuits are turned off, including the output power transistors. Motor output goes to high impedance. If the motor is running at the switch to stand-by mode, the system enters an idling state because of the body diodes. However, when the system switches to stand-by from any other mode (except the brake mode), the control logic remains in the high state for at least 50µs before shutting down all circuits.

b) Forward mode

This operating mode is defined as the forward rotation of the motor when the OUT1 pin is high and OUT2 pin is low. When the motor is connected between the OUT1 and OUT2 pins, the current flows from OUT1 to OUT2. For operation in this mode, connect the VREF pin with VCC pin.

c) Reverse mode

This operating mode is defined as the reverse rotation of the motor when the OUT1 pin is low and OUT2 pin is high. When the motor is connected between the OUT1 and OUT2 pins, the current flows from OUT2 to OUT1. For operation in this mode, connect the VREF pin with VCC pin.

d) Brake mode

This operating mode is used to quickly stop the motor (short circuit brake). It differs from the stand-by mode because the internal control circuit is operating in the brake mode. Please switch to the stand-by mode (rather than the brake mode) to save power and reduce consumption.

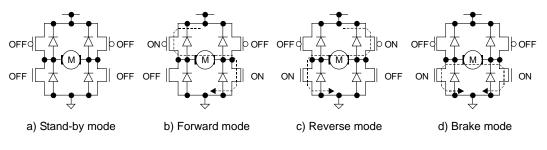
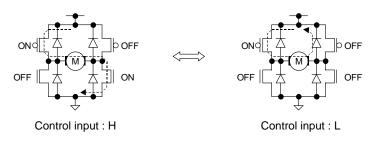
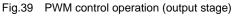


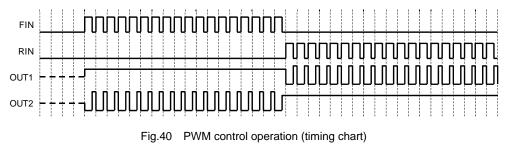
Fig.38 Four basic operations (output stage)

e) f) PWM control mode

The rotational speed of the motor can be controlled by the switching duty when the PWM signal is input to the FIN pin or the RIN pin. In this mode, the high side output is fixed and the low side output does the switching, corresponding to the input signal. The switching operates by the output state toggling between "L" and "Hi-Z". The PWM frequency can be input in the range between 20kHz and 100kHz. Note that control may not be attained by switching on duty at frequencies lower than 20kHz, since the operation functions via the stand-by mode. Also, circuit operation may not respond correctly when the input signal is higher than 100kHz. To operate in this mode, connect the VREF pin with VCC pin. In addition, establish a current path for the recovery current from the motor, by connecting a bypass capacitor (10 μ F or more is recommended) between VCC and ground.





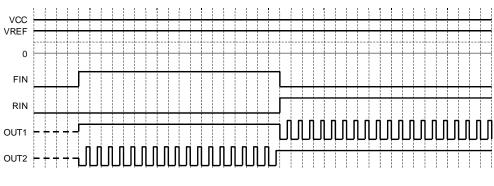


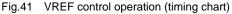
g) h) VREF control mode

The built-in VREF-switching on duty conversion circuit provides switching duty corresponding to the voltage of the VREF pin and the VCC voltage. The function offers the same level of control as the high voltage output setting function in previous models. The on duty is shown by the following equation.

DUTY ≈ VREF [V] / VCC [V]

For example, if VCC voltage is 5V and VREF pin voltage is 3.75V, the switching on duty is about 75 percent. However, please note that the switching on duty might be limited by the range of VREF pin voltage (Refer to the operating conditions, shown on page 2). The PWM carrier frequency in this mode is 25kHz (nominal), and the switching operation is the same as it is the PWM control modes. When operating in this mode, do not input the PWM signal to the FIN and RIN pins. In addition, establish a current path for the recovery current from the motor, by connecting a bypass capacitor (10µF or more is recommended) between VCC and ground.





2) Cross-conduction protection circuit

In the full bridge output stage, when the upper and lower transistors are turned on at the same time, and this condition exists during the period of transition from high to low, or low to high, a rush current flows from the power supply to ground, resulting in a loss. This circuit protects against the rush current by providing a dead time (about 400ns, nominal) at the transition.

3) Output protection circuits

a) Under voltage lock out (UVLO) circuit

To secure the lowest power supply voltage necessary to operate the controller, and to prevent under voltage malfunctions, a UVLO circuit has been built into this driver. When the power supply voltage falls to 2.3V (nominal) or below, the controller forces all driver outputs to high impedance. When the voltage rises to 2.5V (nominal) or above, the UVLO circuit ends the lockout operation and returns the chip to normal operation.

b) Over voltage protection (OVP) circuit

When the power supply voltage exceeds 7.3V (nominal), the controller forces all driver outputs to high impedance. The OVP circuit is released and its operation ends when the voltage drops back to 6.8V (nominal) or below. This protection circuit does not work in the stand-by mode. Also, note that this circuit is supplementary, and thus if it is asserted, the absolute maximum rating will have been exceeded. Therefore, do not continue to use the IC after this circuit is activated, and do not operate the IC in an environment where activation of the circuit is assumed.

c) Thermal shutdown (TSD) circuit

The TSD circuit operates when the junction temperature of the driver exceeds the preset temperature (175°C nominal). At this time, the controller forces all driver outputs to high impedance. Since thermal hysteresis is provided in the TSD circuit, the chip returns to normal operation when the junction temperature falls below the preset temperature (150°C nominal). Thus, it is a self-returning type circuit.

The TSD circuit is designed only to shut the IC off to prevent thermal runaway. It is not designed to protect the IC or guarantee its operation in the presence of extreme heat. Do not continue to use the IC after the TSD circuit is activated, and do not operate the IC in an environment where activation of the circuit is assumed.

d) Over current protection (OCP) circuit

To protect this driver IC from ground faults, power supply line faults and load short circuits, the OCP circuit monitors the output current for the circuit's monitoring time (10µs, nominal). When the protection circuit detects an over current, the controller forces all driver outputs to high impedance during the off time (290µs, nominal). The IC returns to normal operation after the off time period has elapsed (self-returning type). At the two channels type, this circuit works independently for each channel.

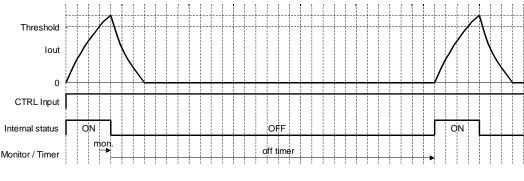
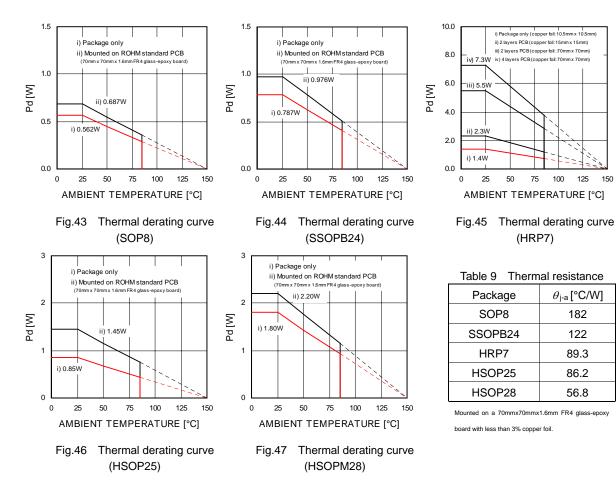


Fig.42 Over current protection (timing chart)



Thermal design needs to meet the following operating conditions.

In creating the thermal design, sufficient margin must be provided to guarantee the temperature conditions below.

- 1. The ambient temperature Ta must be 85°C or below
- 2. The junction temperature Tj must be 150°C or below

The junction temperature Tj can be determined using the following equation.

Tj ≈ Ta +θ_{j-a} x Pc [°C]

The power consumption Pc can be determined using the following equation. Refer to page 4 about V_{ON(H)} and V_{F(H)}.

 $Pc \approx (I_{OUT}^{2} x R_{ON}) x (V_{REF} / V_{CC}) + I_{OUT} x (V_{ON(H)} + V_{F(H)}) x (1 - V_{REF} / V_{CC}) + V_{CC} x I_{CC} [W]$

Example using the BD6211F

Conditions: Ta=50°C, VCC=VREF=5V, lout=0.1A.

The power consumption of the IC and the junction temperature are as follows:

 $Pc \approx 0.1^2 \text{ x } 1.0 + 5 \text{ x } 0.7 \text{m} = 13.5 \text{mW}$

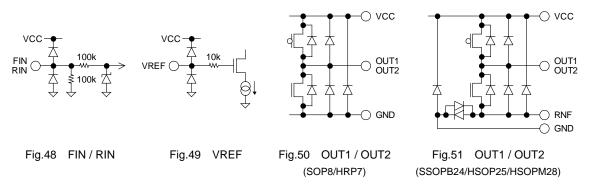
Tj ≈ 50 + 182 x 13.5m = 52.5 [°C]

Where the Tjmax parameter is 150°C and the derating is set to 80 percents, the maximum ambient temperature Tamax is determined as follows.

Ta ≤ Tjmax x 0.8 -θ_{j-a} x Pc ≈ 115 [°C]

In this example, thermal design can be considered satisfactory (meaning that there are no problems in thermal design), since the system meets the operating temperature conditions.

Interfaces



•NOTES FOR USE

1) Absolute maximum ratings

Devices may be destroyed when supply voltage or operating temperature exceeds the absolute maximum rating. Because the cause of this damage cannot be identified as, for example, a short circuit or an open circuit, it is important to consider circuit protection measures – such as adding fuses – if any value in excess of absolute maximum ratings is to be implemented.

2) Connecting the power supply connector backward

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply lines, such as adding an external direction diode.

3) Power supply lines

Return current generated by the motor's Back-EMF requires countermeasures, such as providing a return current path by inserting capacitors across the power supply and GND. In this case, it is important to conclusively confirm that none of the negative effects sometimes seen with electrolytic capacitors – including a capacitance drop at low temperatures - occurs. Also, the connected power supply must have sufficient current absorbing capability. Otherwise, the regenerated current will increase voltage on the power supply line, which may in turn cause problems with the product, including peripheral circuits exceeding the absolute maximum rating. To help protect against damage or degradation, physical safety measures should be taken, such as providing a voltage clamping diode across the power supply and GND.

4) Electrical potential at GND

Keep the GND terminal potential to the minimum potential under any operating condition. In addition, check to determine whether there is any terminal that provides voltage below GND, including the voltage during transient phenomena. When both a small signal GND and high current GND are present, single-point grounding (at the set's reference point) is recommended, in order to separate the small signal and high current GND, and to ensure that voltage changes due to the wiring resistance and high current do not affect the voltage at the small signal GND. In the same way, care must be taken to avoid changes in the GND wire pattern in any external connected component.

5) Thermal design

Use a thermal design that allows for a sufficient margin in light of the power dissipation (Pd) under actual operating conditions.

- 6) Inter-pin shorts and mounting errors Use caution when positioning the IC for mounting on printed circuit boards. The IC may be damaged if there is any connection error, or if pins are shorted together.
- Operation in strong electromagnetic fields
 Using this product in strong electromagnetic fields may cause IC malfunctions. Use extreme caution with electromagnetic fields.

8) ASO - Area of Safety Operation

When using the IC, set the output transistor so that it does not exceed absolute maximum ratings or ASO.

9) Built-in thermal shutdown (TSD) circuit

The TSD circuit is designed only to shut the IC off to prevent thermal runaway. It is not designed to protect the IC or guarantee its operation in the presence of extreme heat. Do not continue to use the IC after the TSD circuit is activated, and do not operate the IC in an environment where activation of the circuit is assumed.

10) Capacitor between output and GND

In the event a large capacitor is connected between the output and GND, if VCC and VIN are short-circuited with 0V or GND for any reason, the current charged in the capacitor flows into the output and may destroy the IC. Use a capacitor smaller than 1µF between output and GND.

11) Testing on application boards

When testing the IC on an application board, connecting a capacitor to a low impedance pin subjects the IC to stress. Therefore, always discharge capacitors after each process or step. Always turn the IC's power supply off before connecting it to or removing it from the test setup during the inspection process. Ground the IC during assembly steps as an antistatic measure. Use similar precaution when transporting or storing the IC.

12) Switching noise

When the operation mode is in PWM control or VREF control, PWM switching noise may effects to the control input pins and cause IC malfunctions. In this case, insert a pulled down resistor ($10k\Omega$ is recommended) between each control input pin and ground.

13) Regarding the input pin of the IC

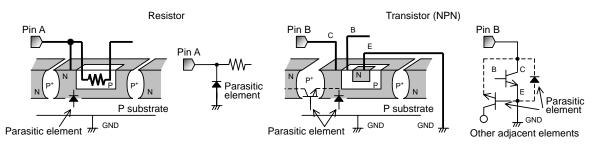
This monolithic IC contains P+ isolation and P substrate layers between adjacent elements, in order to keep them isolated.

P-N junctions are formed at the intersection of these P layers with the N layers of other elements, creating a parasitic diode or transistor. For example, the relation between each potential is as follows:

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

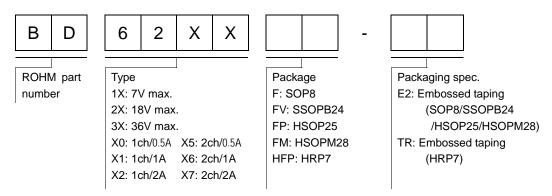
When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, as well as operating malfunctions and physical damage. Therefore, do not use methods by which parasitic diodes operate, such as applying a voltage lower than the GND (P substrate) voltage to an input pin.

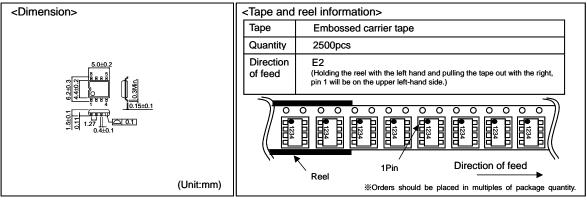


Appendix: Example of monolithic IC structure

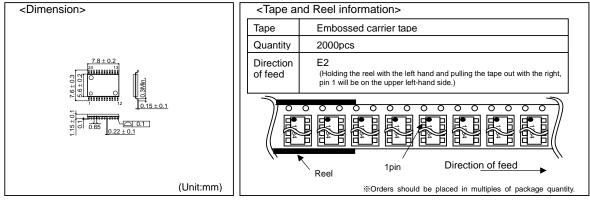
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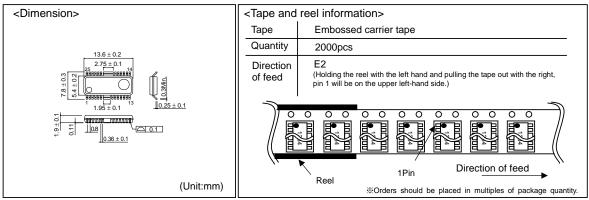
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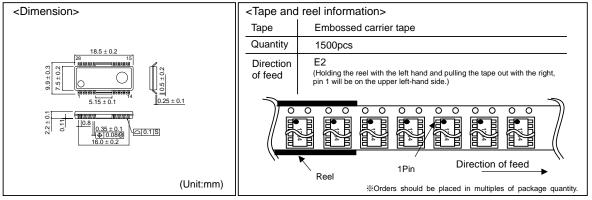
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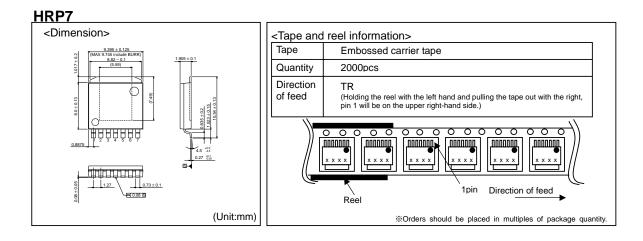


HSOP25



HSOP-M28





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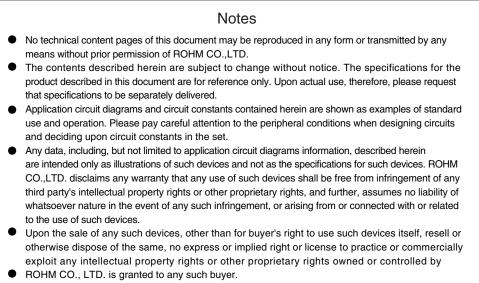
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