

# **FIP** Technologies

# Dual Low Noise LDO in 3mm×3mm MLF<sup>™</sup>

## Description

The HIP6211 is a dual CMOS low dropout linear regulator with ultra-low-noise output, very low dropout voltage and very low ground current.

The HIP6211 operates from a 2.5V to 5.5V input voltage range and delivers up to 300mA, with low dropout of 150 mV at 300mA. The other features of HIP6211 include short-circuit protection and thermal-shutdown protection.

Its dual channel and low noise feature is ideal for microcontroller and DSP based handheld applications.

Other key application areas for HIP6211 also include palmtop computers, PCMCIA cards and WLAN cards.

The HIP6211 has a special feature that if both EN pin is enabled simultaneously the output of LDO2 delays 20us from output of LDO1 which helps to minimize inrush startup current.

The HIP6211 is available in tiny 10pin 3mm x 3mm MLF (10L-TDLMF) package with fixed output voltage versions.

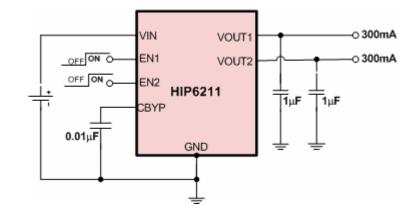
### Features

- 300mA+300mA dual channel high accuracy LDO
- Ultra low output noise:  $140 \mu V_{RMS}$
- Low ground current: 200μA
- Very low dropout: 150mV @300mA
- Zero shutdown supply current
- TTL-logic-controlled independent enable
  input
- Thermal and current limit protections
- Ultra low droop load transient response
- Ultra fast line transient response
- Tiny 10pin 3mm x 3mm MLF (10L-TDLMF) package
- Fixed options 1.5V, 1.8V, 2.5V, 2.8V, 3.0V and 3.3V

## Applications

- Dual supply handheld products
- Cellular and cordless phones
- Wireless LAN cards
- Digital camera
- MP3 / MP4 / CD player
- Pen drives
- USB Hubs and USB 2.0
- Mini PCI & PCI express cards

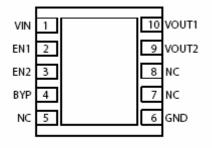
## **Typical Application**



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## **Pin Configuration**



10-Pin 3mm × 3mm MLF<sup>™</sup> (ML) (Top View)

## **Ordering Information**

Output Voltage	Marking	Package
1.5	DBAMW	10L-TDLMF
1.5	DBBMW	10L-TDLMF
1.8	DBCMW	10L-TDLMF
1.8	DBDMW	10L-TDLMF
2.5	DBEMW	10L-TDLMF
2.5	DBFMW	10L-TDLMF
2.7	DBGMW	10L-TDLMF
2.7	DBHMW	10L-TDLMF
2.8	DBIMW	10L-TDLMF
2.8	DBJMW	10L-TDLMF
3.0	DBKMW	10L-TDLMF
3.0	DBLMW	10L-TDLMF
3.3	DBMMW	10L-TDLMF
3.3	DBNMW	10L-TDLMF
	1.5      1.5      1.8      1.8      2.5      2.5      2.7      2.7      2.8      3.0      3.0      3.3	1.5      DBAMW        1.5      DBBMW        1.8      DBCMW        1.8      DBDMW        2.5      DBEMW        2.5      DBFMW        2.7      DBGMW        2.7      DBHMW        2.8      DBIMW        3.0      DBKMW        3.3      DBMMW

Note 1: Contact the factory for other output voltages that are not in the above table

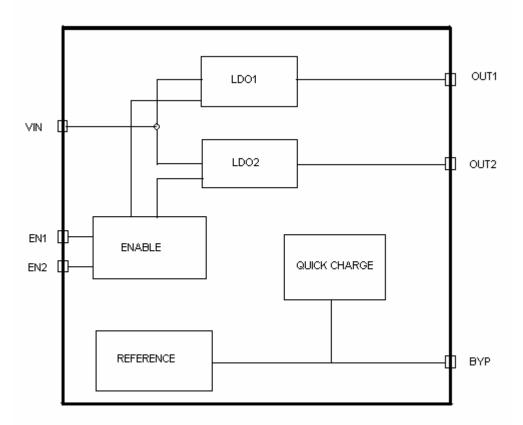


## **Pin Descriptions**

Name	Pin No.	Туре	Function	
IN	1	Supply	Supply voltage. 2.5V ~ 5.5V.	
EN1	2	Logic input 1	Enable/Shutdown. CMOS compatible input. Logic 'H' : enable, logic 'L' : shutdown.	
EN2	3	Logic input 2	Enable/Shutdown. CMOS compatible input. Logic 'H' : enable, logic 'L' : shutdown.	
BYP	4	Bypass	Reference voltage bypass pin. Connect $0.01 \text{uF} \le C_{BYP} \le 0.1 \text{uF}$ to GNDreduce output noise. May be left open.	
GND	6	Ground	Ground pin	
OUT2	9	Analog output2	Regulator 2 Output.	
OUT1	10	Analog output1	Regulator 1 Output.	

# **Block Diagram**

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## **Absolute Maximum Ratings**

Symbol	Parameter	Value	Unit
V <sub>IN</sub>	DC Supply Voltage at Pin 1	-0.3 to +6.0	V
V <sub>EN</sub>	Enable Input Voltage at Pin 2 and Pin 3	-0.3 to +6.0	V
P <sub>D</sub>	Continuous Power Dissipation	Internally limited	W
T <sub>STG</sub>	Storage Temperature Range	-65 to +150	°C
R <sub>eja</sub>	Thermal Resistance, Junction-To-Air	235	°c/W
T <sub>J,MAX</sub>	Operating Junction Temperature	-40 to +125	°c
TL	Lead Temperature (Soldering, 5sec)	260	°C
ESD	ESD Capability, HBM model	2	kV

### **Recommended Operating Conditions**

Symbol	Parameter	Value	Unit
V <sub>IN</sub>	DC Supply Voltage at Pin 1	+2.5 to 5.5	V
V <sub>EN</sub>	Enable Input Voltage at Pin 2 and Pin 3	0 to $V_{IN}$	V
T <sub>A</sub>	Operating Ambient Temperature	-40 to +85	°C

## **Electrical Characteristics**

 $(V_{IN}=V_{OUT(nominal)}+1V \text{ OR } 2.5V \text{ (WHICHEVER IS GREATER), } V_{EN}=V_{IN}, C_{IN}=C_{OUT}=1\mu\text{F}, I_{O}=1\text{MA}, T_{A}=25^{\circ}\text{C}, \text{ UNLESS OTHERWISE NOTED.}$ 

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V <sub>IN</sub>	Supply Voltage		2.5		5.5	V
ΔV <sub>OUT</sub>	Output Voltage Accuracy	$I_0 = 1 m A$ ( both regulator )	-1.0		1.0	%
$\Delta V_{LOAD}$	Load Regulation	$I_0=1$ mA to 300mA ( both regulator )			0.5	%
ΔVI <u>line</u>	Line Regulation dV <sub>OUT</sub> /( dV <sub>IN</sub> *V <sub>OUT(NOMINAL)</sub> )*100%	$V_{IN} = V_{OUT(NOMINAL)} + 0.1V$ (or 2.5V, whichever is greater) to 5.5V, $I_0=1mA$ ( both regulator )			0.05	%/V
VDP	Dropout Voltage (Note 1)	$I_{LOAD} = 300 \text{mA}$ ( both regulator )		150	220	mV
1		$I_{LOAD} = 100 \text{mA}$ ( both regulator )		80	100	mV
Ι <sub>Ο</sub>	Maximum Output Current	Continuous	300			mA <sub>RMS</sub>
ILIM	Current Limit/Output Current	$V_{IN}-V_{OUT} = 1.3V$ (regulator 1)	350	600		mA
l I		$V_{IN}-V_{OUT} = 1.3V$ (regulator 2)	350	600		
IQ	Standby Current	$V_{EN} = 0V$		0.02	1	μA
IG	Ground pin current	$I_{LOAD1} = I_{LOAD2} = 1mA$ ( both regulator active)		200	300	μA
PSRR		f = 100Hz, Cout= 1µF, Cbyp=10nF		75		dB
PSRR	Ripple Rejection, $I_{OUT} = 10$ mA,	f = 10KHz, Cout= 1µF, Cbyp =10nF		70		dB
	Output voltage noise	$C_{OUT} = 1\mu F, C_{BYP} = 10nF,$ F = 10Hz to 100K Hz (Vp-p/2/ $\sqrt{2}$ )		140		μV <sub>RMS</sub>
	Thermal Shutdown Temperature			165		°C
	Thermal Shutdown Hysteresis			20		°C
VIH	Logic Input High Voltage (EN)	( both regulator )	1.2			V
VIL	Logic Input Low Voltage (EN\)	( both regulator )			0.4	V
IEN	Logic Input Current (SHDN\)	( both regulator )	-1			

Note 1: The Dropout Voltage is defined as  $V_{IN}$  -  $V_{OUT}$ , when  $V_{OUT}$  is 2% below the value of  $V_{OUT}$  measured for  $V_{IN} = V_{OUT(nominal)} + 1V.$ 

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## **Operation Description**

The HIP6211 are ultra-low-noise, low-dropout, lowquiescent current linear regulators designed for space-restricted applications. These devices can supply loads up to 300mA. As shown in the Block Diagram, the HIP6211 consists of a highly accurate band gap core, noise bypass circuit, error amplifier, P-channel pass transistor and an internal feedback voltage divider. The 1.0V band gap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the reference voltage, the pass transistor gate is pulled low. This allows more current to pass to the output and increases the output voltage. If the feedback voltage is too high, the pass transistor gate is pulled high, allowing less current to pass to the output. The output voltage is feedback through an internal resistor voltage divider connected to the OUT pin. An external bypass capacitor connected to BYP reduces noise at the output. Additional blocks include a current limiter, over temperature protection, and shutdown logic.

#### Internal P-Channel Pass Transistor

The HIP6211 feature a  $1\Omega$  (typ) P-channel MOSFET pass transistor. This provides several advantages over similar designs using a PNP pass transistor, including longer battery life. The P-channel MOSFET requires no base drive, which considerably reduces quiescent current. PNP-based regulators waste considerable current in dropout when the pass transistor saturates. They also use high base-drive current under heavy loads. The HIP6211 does not suffer from these problems and consume only  $200\mu$ A of quiescent current in light load.

#### **Current Limit**

The HIP6211 includes a current limiter. It monitors the output current and controls the pass transistor's gate voltage to limit the output current under 600mA (typ). The output can be shorted to ground for an indefinite amount of time without damaging the part.

#### Enable Input

The HIP6211 features an active-high Enable input (EN) pin that allows on/off control of the regulator. The HIP6211 bias current reduces to less than microampere of leakage current when it is shutdown. The Enable input is TTL/CMOS compatible threshold for simple logic interfacing. When EN is 'H,' the output voltage startup rising time is 35us typically at 300mA output current. Connect EN pin to IN pin for normal operation

#### Under Voltage Lockout

When the input supply goes too low (below 2.0V) the HIP6211 produces an internal UVLO (under voltage lockout) signal that generates a fault signal and shuts down the chip. This mechanism protects the chip from producing false logic due to low input supply.

#### Quick Charging Mode

The HIP6211 has a quick charge block to get the reference up very quickly by charging the BYP capacitor with very high current when the chip comes out of shut down. This quick charge block stops charging the BYP capacitor when the reference reaches 95% of its nominal value and then the chip switches out of quick charging mode to normal operating mode.

#### **Over Temperature Protection**

Over temperature protection limits total power dissipation in the HIP6211. When the junction temperature exceeds  $T_j$ = +165°C, the thermal sensor signals the shutdown logic and turns off the pass transistor. The thermal sensor turns the pass transistor on again after the IC's junction temperature drops by 20°C, resulting in a pulsed output during continuous thermal-overload conditions.

Thermal-Overload protection is design to protect the HIP6211 in the event of a fault condition. For continual operation, do not exceed the absolute maximum junction temperature rating of Tj =  $+150^{\circ}$ C.

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Operating Region and Power Dissipation The HIP6211 maximum power dissipation depends on 1) the thermal resistance of the case and circuit board, 2) the temperature difference between the die junction and ambient, and 3) the rate of airflow. The power dissipation across the device is:

P = Iout (Vin - Vout)

The maximum power dissipation is:

 $Pmax = (Tj - Ta) / (\theta jc + \theta ca)$ 

Where (Tj - Ta) is the temperature difference between the HIP6211 die junction and the ambient air;  $\theta$ jc is the thermal resistance of the package; and  $\theta$ ca is the thermal resistance through the PC board, copper traces, and other materials to the surrounding air.

The GND pin of the HIP6211 performs the dual function of providing an electrical connection to ground and channeling heat away. Connect the GND pin to ground using a large pad or ground plane.

#### Noise Reduction

For the HIP6211, an external  $0.01\mu$ F bypass capacitor between BYP and GND with innovative noise bypass scheme reduces output noises dramatically, exhibiting  $100\mu$ Vrms of output voltage noise with Cbyp =  $0.01\mu$ F and Cout =  $1\mu$ F.

## **Application Information**

Capacitor Selection and Regulator Stability Use a  $1\mu$ F capacitor on the HIP6211 input and a  $1\mu$ F capacitor on the output. Large input capacitor values and lower ESRs provide better noise rejection and line-transient response.

Reduce output noise and improve load-transient response, stability, and power-supply rejection by using large output capacitors. Note that some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. With dielectrics such as Z5U and Y5V, it may be necessary to use a  $1\mu$ F or larger output capacitor to ensure stability at

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temperatures below -10°C. With X7R or X5R

dielectrics,  $1\mu\text{F}$  is sufficient at all operating

temperatures. A graph of the region of stable Cout

ESR vs. load current is shown in the Typical

Characteristics.

Use a  $0.01\mu$ F bypass capacitor at BYP for lowoutput voltage noise. The leakage current going into the BYP pin should be less than 10nA.

Noise, PSRR, and Transient Response

The HIP6211 are designed to deliver ultra-low noise and high PSRR, as well as low dropout and low quiescent currents in battery-powered systems. The HIP6211 PSRR is 75dB at 100Hz and 70dB at 10kHz (see the Power-Supply Rejection Ratio vs. Frequency graph in the Typical Characteristic).

When operating from sources other than batteries, improved supply-noise rejection and transient response can be achieved by increasing the values of the input and output bypass capacitors, and through passive filtering techniques. The Typical Characteristics show the HIP6211 line and load transient responses.

### Dropout Voltage

A regulator's minimum dropout voltage determines the lowest usable supply voltage. In batterypowered systems, this determines the useful endof-life battery voltage. Because the HIP6211 use a P-channel MOSFET pass transistor, their dropout voltage is a function of drain-to-source on resistance (RDS (on)) multiplied by the load current (see the Typical Characteristics).

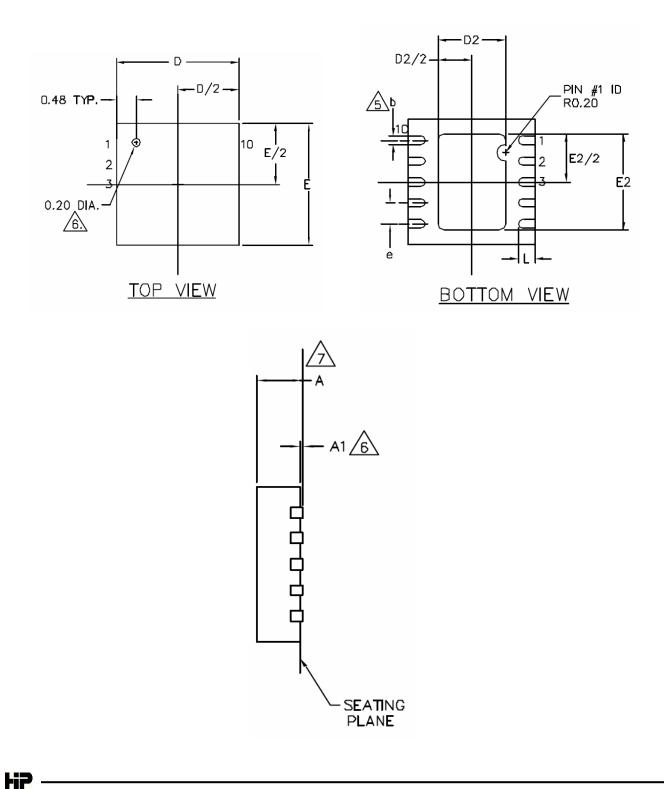
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# Package Information 10-Pin 3mm X 3mm Mlf Outline Dimension



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## **Dimension:**

	DIMENSION (mm)		
	MIN,	NOM,	MAX,
Α	0.80	0.85	1.00
A1	0.00	0.01	0.05
D		3.00 BSC	
D2	1.45	1.60	1.75
E		3.00 BSC	
<b>E</b> 2	2.15	2.30	2.45
е		0.50 BSC	
L	0.35	0.40	0.55
þ	0.18	0.23	0.30

#### NOTE:

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- ALL DIMENSIONS ARE IN MILLIMETERS. 1.
- MAX. PACKAGE WARPAGE IS 0.05 mm. 2.
- MAXIMUM ALLOWABE BURRS IS 0.076 mm IN ALL DIRECTIONS. з.
- **Å**:
- PIN #1 ID ON TOP WILL BE LASER/INK MARKED. DIMENSION & APPLIES TO METALIZED TERMINAL AND IS MEASURED BETWEEN 0.20 AND 0.25 mm FROM TERMINAL TIP.
- a A APPLIED ONLY FOR TERMINALS,
- APPLIED FOR EXPOSED PAD AND TERMINALS.





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