

## High Frequency LED Driver with Integrated Schottky and Simple Interconnect

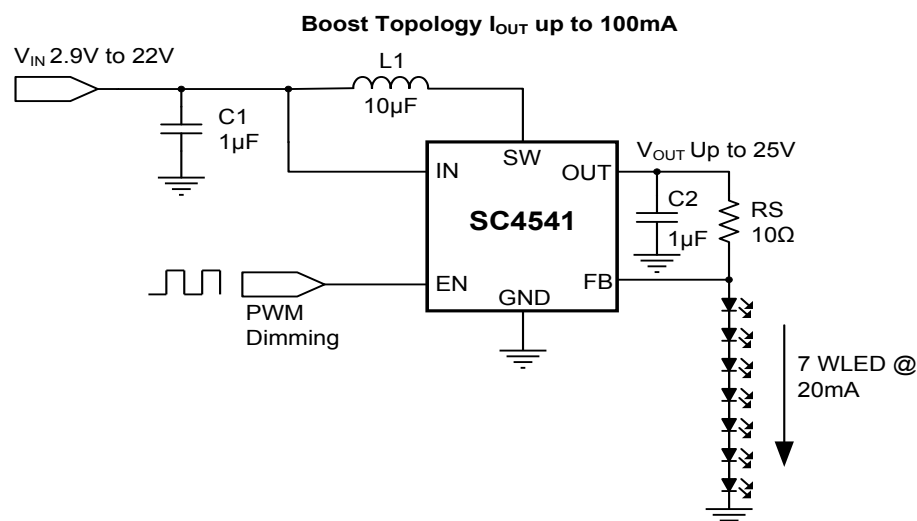
### POWER MANAGEMENT Features

- $V_{IN}$  Range — 2.9V to 22V
- $V_{OUT}$  Range — Up to 25V
- $I_{OUT}$  Range:
  - Up to 100mA (Boost)
  - Up to 200mA (Buck)
- Drives Up to 7 Series WLEDs (Boost)
- Drives Up to 5 Series WLEDs (Buck)
- Integrated Schottky Rectifier
- Single-Wire LED Connect
- Current Mode Control
  - Frequency — 2.0MHz
  - No External Compensation
- Direct PWM Dimming
  - 100:1 Range at 500Hz
- Open-LED Protection — 27V
- Low Shutdown Current < 1 $\mu$ A
- Package Options:
  - Low Profile MLPD-UT, 6-pin 2x2 x0.6(mm)
  - SOT23, 6-pin 2.9x2.8 x1.45(mm)
- Lead-free, Halogen-Free and WeEE/RoHS Compliant

### Applications

- Keypad Illumination
- DSC and DSLR Display Backlight
- Cellular Handsets
- Automotive Lighting

### Typical Application Circuit



### Description

The SC4541 is a fully integrated, high voltage boost (step-up) and buck (step-down) LED driver. The input voltage range is 2.9V to 22V with output voltage up to 25V. The device is capable of driving up to 7 white LEDs. The high side sense enables a single-wire LED connection by taking advantage of the available system ground. This reduces connector cost and complexity.

The device includes a current mode PWM controller, 260mA power switch, and high performance Schottky rectifier. The LED driver eliminates external compensation components for minimum solution size.

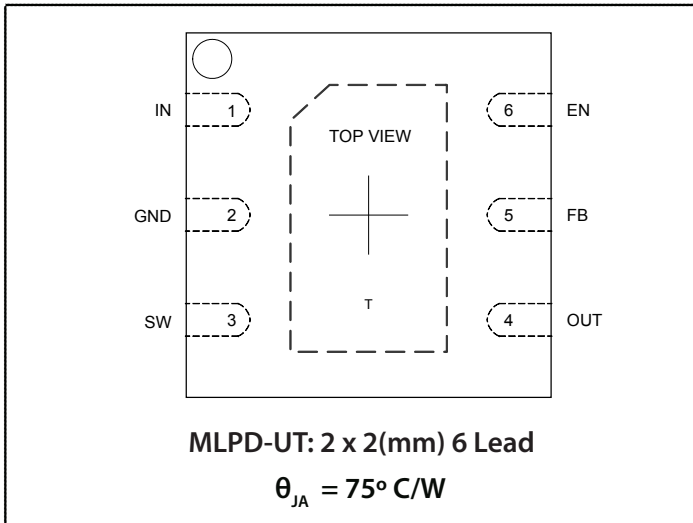
The wide input voltage range and low standby current are ideal for portable devices. The high input voltage allows direct connection to multi-cell battery packs which increases operating efficiency and extends battery life.

Output current is programmed with an external sense resistor up to 100mA (Boost) or 200mA (Buck). The device provides  $\pm 5\%$  constant current accuracy across the entire operating temperature range.

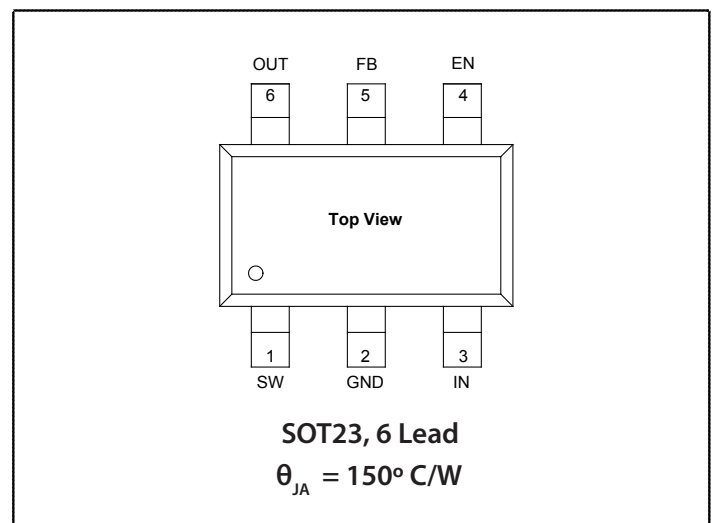
A logic level PWM dimming signal may be applied directly to the enable (EN) pin - for direct PWM dimming with a wide dimming range. Output OVP (over voltage protection) protects the SC4541 from open circuit faults in the LED string or connector. External OVP programming resistors are not required.

The SC4541 is available in low profile MLPD-UT and SOT23,6-pin packages. The device is rated across the temperature range -40°C to +85°C.

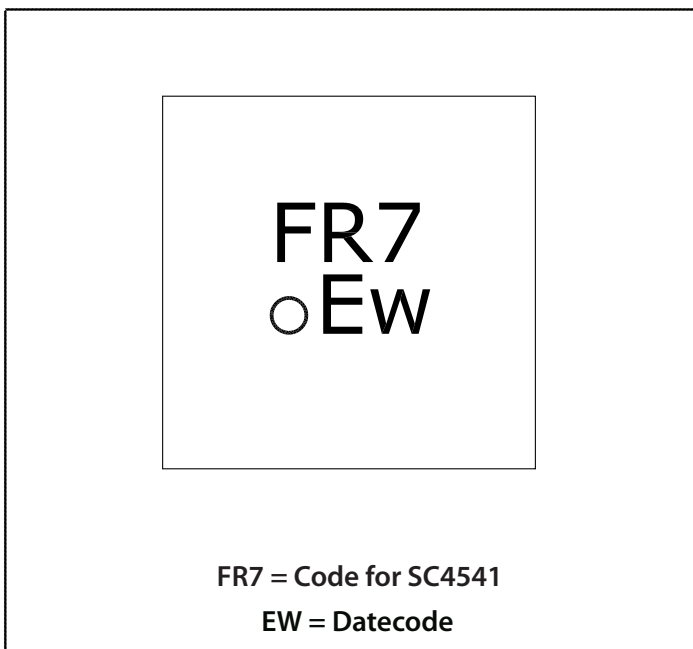
### Pin Configuration, MLPD-UT6



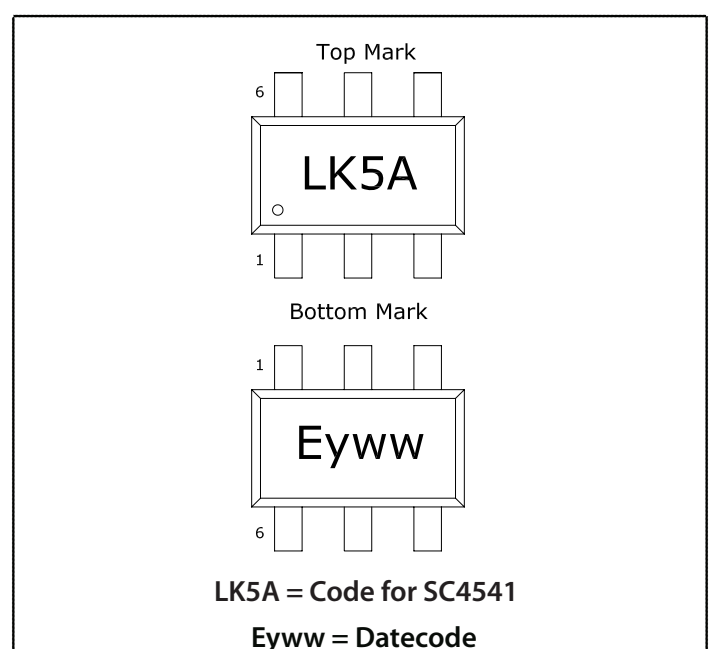
### Pin Configuration, SOT23-6



### Marking Information, MLPD-UT6



### Marking Information, SOT23-6



### Ordering Information

Device	Package
SC4541SKTRT <sup>(1,2)</sup>	SOT23-6
SC4541ULTRT <sup>(1,2)</sup>	2 x 2 x 0.6(mm) MLPD-UT6
SC4541EVB-1	Evaluation Board, SOT23-6
SC4541EVB-2	Evaluation Board, MLPD-UT6

**Notes:**

- (1) Available in tape and reel only. A reel contains 3,000 devices.
- (2) Available in lead-free package only. Device is WEEE/RoHS compliant and halogen-free.

## Absolute Maximum Ratings

Supply IN Voltage(V) .....	-0.3 to 26
SW Voltage, OUT Voltage(V) .....	-0.3 to 28
FB, Enable Voltage(V) .....	-0.3 to VIN+0.3
ESD Protection Level <sup>(1)</sup> (kV) .....	4

## Recommended Operating Conditions

Supply IN Voltage (V) .....	2.9 to 22
Maximum Output Voltage (V) .....	25
Ambient Temperature Range (°C) .....	-40 to +85

## Thermal Information

Thermal Resistance, Junction to Ambient<sup>(2)</sup>

MLPD-UT6(°C/W) .....	75
SOT23-6(°C/W).....	150
Maximum Junction Temperature (°C) .....	+125
Storage Temperature Range (°C) .....	-65 to +150
Peak IR Reflow Temperature (10s to 30s) (°C) .....	+300

Exceeding the above specifications may result in permanent damage to the device or device malfunction. Operation outside of the parameters specified in the Electrical Characteristics section is not recommended.

NOTES:

(1) Tested according to JEDEC standard JESD22-A114-B.

(2) Calculated from package in still air, mounted to 3 x 4.5(in), 4 layer FR4 PCB with thermal vias under the exposed pad per JESD51 standards.

## Electrical Characteristics

$V_{IN}=5V$ ,  $C_{IN}=1\mu F$ ,  $C_{OUT}=1\mu F$ ,  $L=10\mu H$ ,  $T_A=-40^{\circ}C$  to  $85^{\circ}C$ , unless specified otherwise.

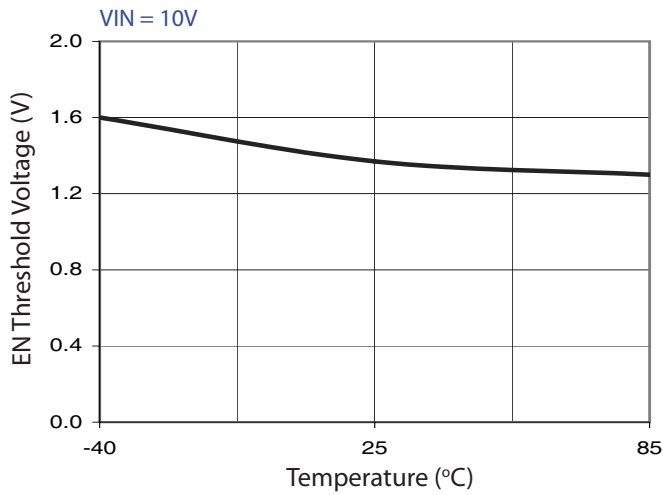
Parameter	Symbol	Conditions	Min	Typ	Max	Units
Input Voltage	$V_{IN}$		2.9		22	V
Supply Current	$I_{IN(Q)}$	$V_{IN}=22V, V_{OUT}=18V, V_{FB}=17V, EN=3V$		1.5	3	mA
Shutdown Supply Current	$I_{IN(SHDN)}$	$EN=0V$			1	$\mu A$
EN Logic High Voltage	$V_{EN(H)}$		2			V
EN Logic Low Voltage	$V_{EN(L)}$				0.4	V
EN Logic Bias Current	$I_{EN(L)}, I_{EN(H)}$				1	$\mu A$
<b>Boost Converter Characteristics</b>						
Switching Frequency	$f_{SW}$		1.6	2.0	2.4	MHz
Maximum Duty Cycle	$D_{MAX}$		84			%
Switch Current limit	$I_{LIMIT}$		260	350	450	mA
Switch Leakage Current	$I_{SW}$	$V_{SW} = 18V$		0.1	5	$\mu A$
Switch Saturation Voltage	$V_{SAT}$	$I_{SW} = 0.2A$		200		mV
OUT Over-Voltage Protection	$V_{OUT(OVP)}$	$T_A=25^{\circ}C$	25.5		28.0	V
Schottky Forward Drop	$V_{F(D)}$	$I_{OUT} = 100mA, T_A=25^{\circ}C$		0.7		V

**Electrical Characteristics (continued)**

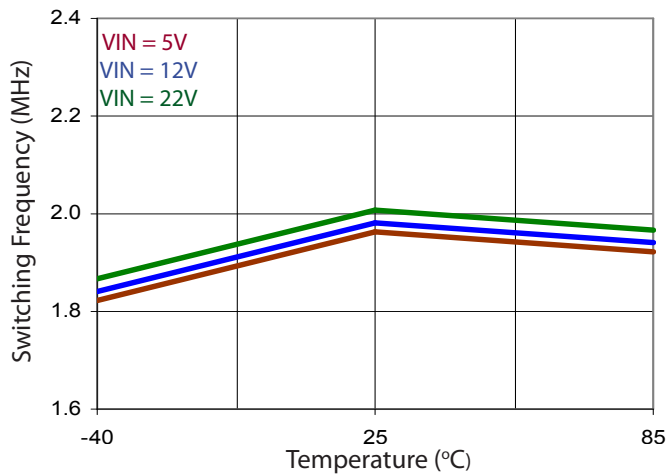
Parameter	Symbol	Conditions	Min	Typ	Max	Units
<b>Boost Converter Characteristics (continued)</b>						
Schottky Leakage Current	$I_{LEAK(D)}$	$V_{OUT} = 20V$			4	$\mu A$
<b>Current Sink Characteristics</b>						
LED Current Sense Voltage ( $V_{OUT} - V_{FB}$ )	$V_{SENSE}$	$V_{OUT} = 18V, T_A = 25^\circ C$	192	200	208	mV
		$V_{OUT} = 18V, T_A = -40^\circ C \text{ to } 85^\circ C$	188		212	mV
OUT Pin Bias Current	$I_{OUT(BIAS)}$	$V_{OUT} = 18V$		55		$\mu A$
FB Pin Bias Current	$I_{FB(BIAS)}$	$V_{FB} = 18V$		12		$\mu A$
$V_{OUT}, V_{FB}$ Common Mode Voltage	$V_{CM}$		2.9			V

## Typical Characteristics

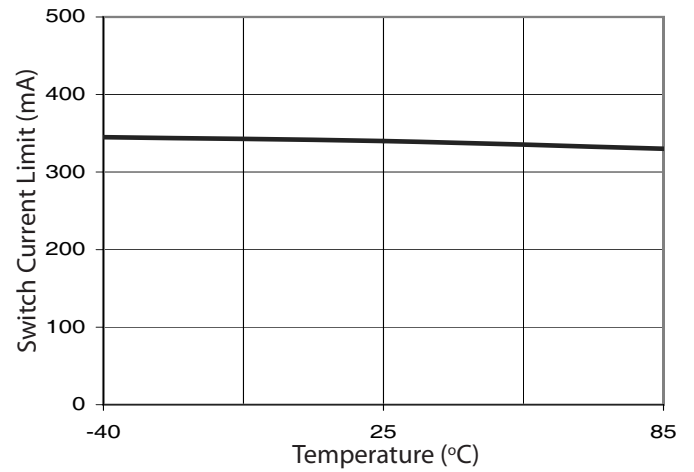
### EN Threshold



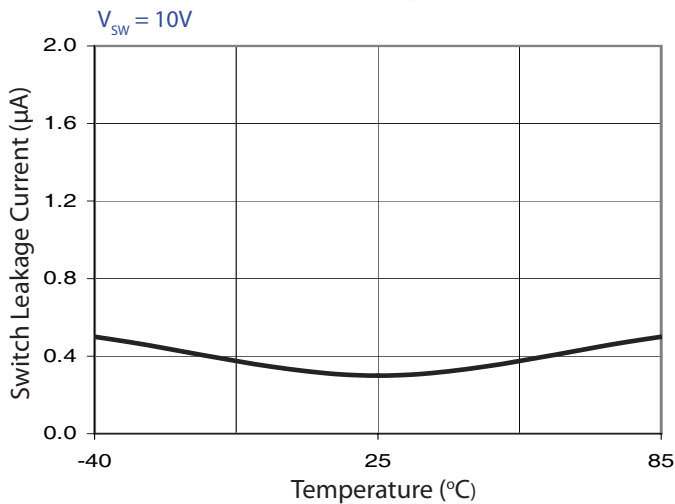
### Switching Frequency



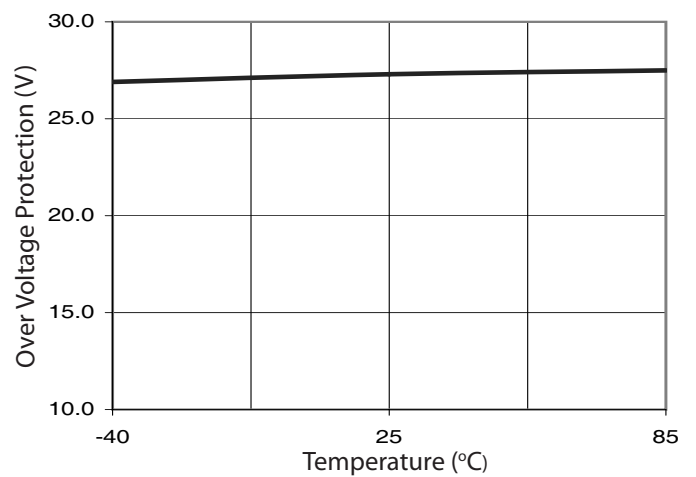
### Switch Current Limit

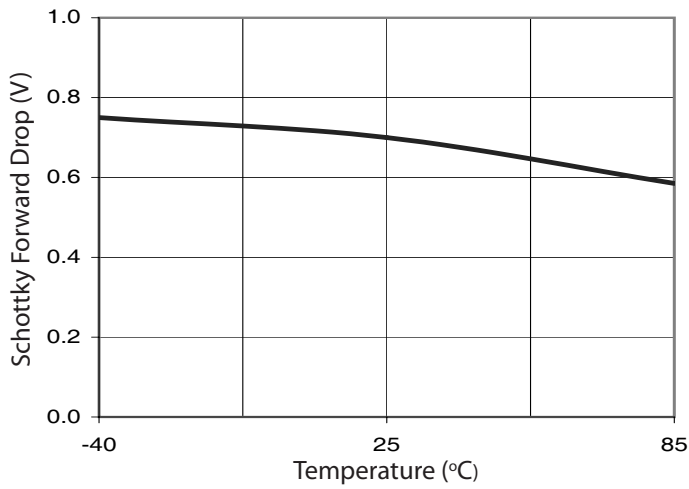
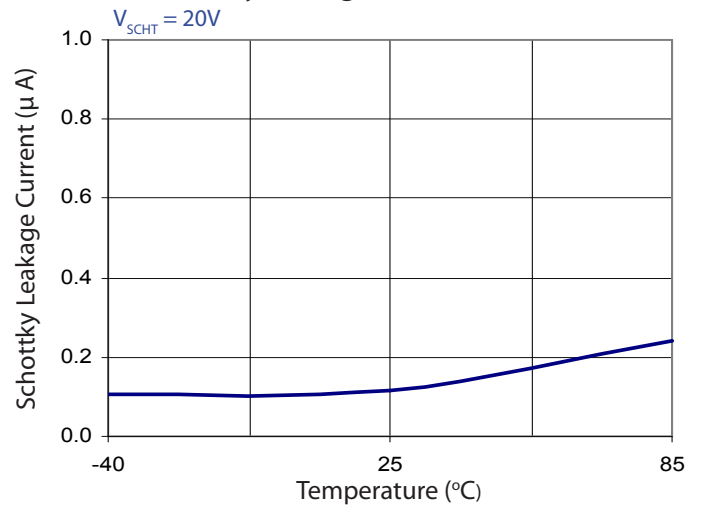
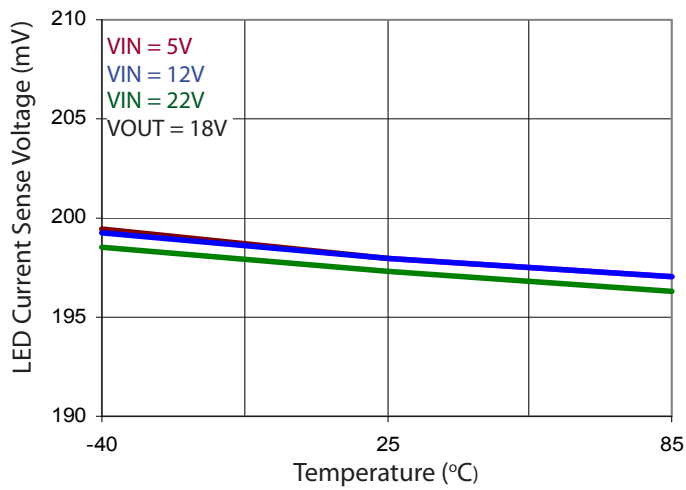


### Switch Leakage Current



### Output Over Voltage Protection

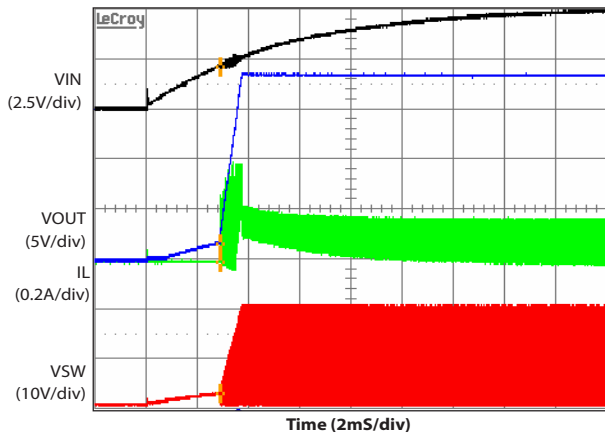


**Typical Characteristics (continued)**
**Schottky Forward Drop**

**Schottky Leakage Current**

**LED Current Sense Voltage(FB)**


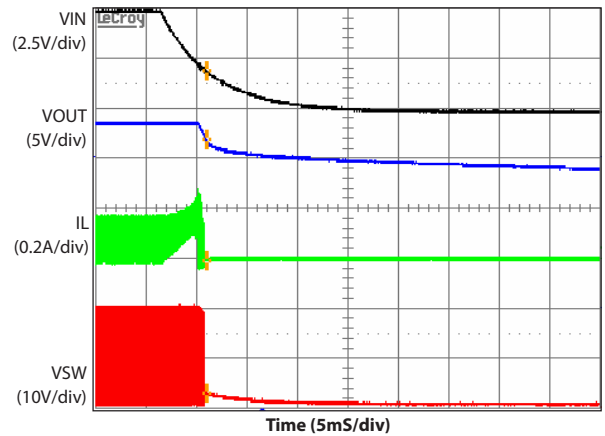
## Typical Characteristics (continued)

Conditions:  $5V_{IN}$ , 20mA, 6 LEDs in series,  $C_{IN}/C_{OUT}=1\mu F$ ,  $L=10\mu H$

Start-up — by  $V_{IN}$

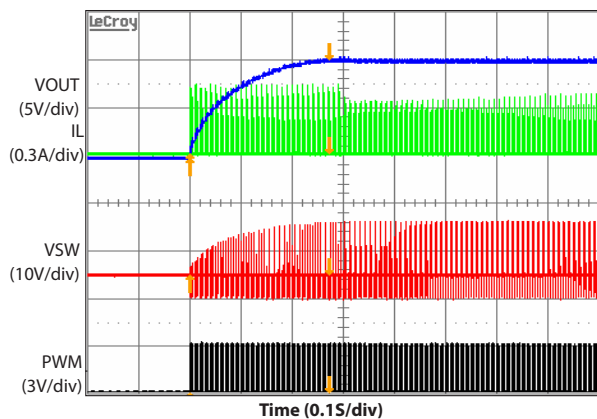


Shut Down

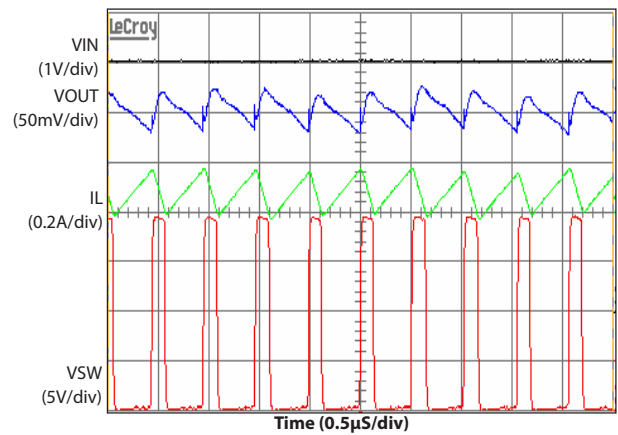


Start-up — by PWM

300Hz PWM Dimming, 0.1% Duty Cycle

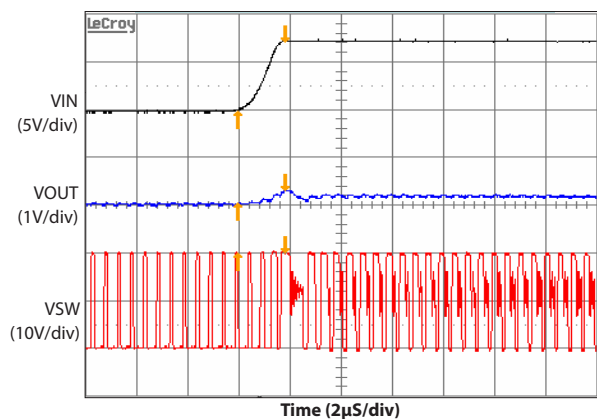


Main Power Switching



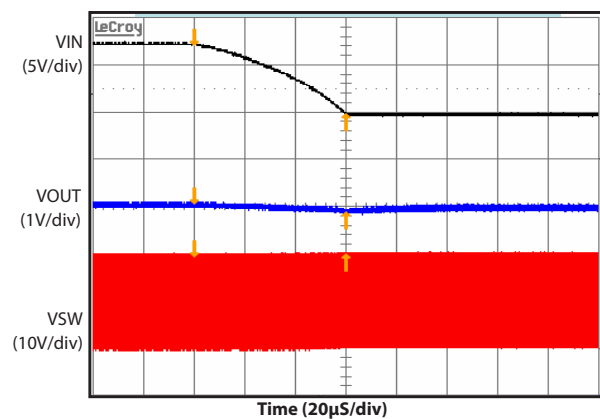
VIN Fluctuation — Rising

5VIN to 12VIN



VIN Fluctuation — Falling

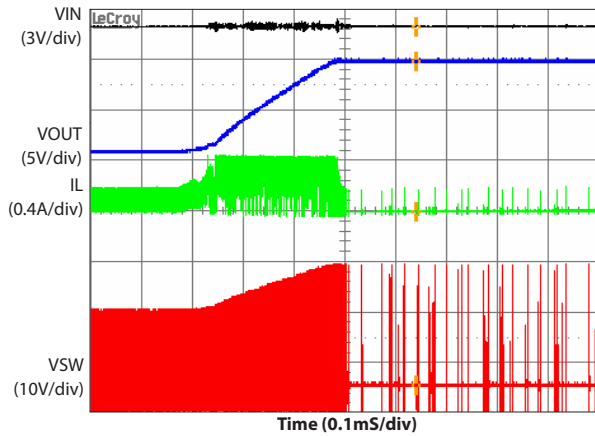
12VIN to 5VIN



## Typical Characteristics (continued)

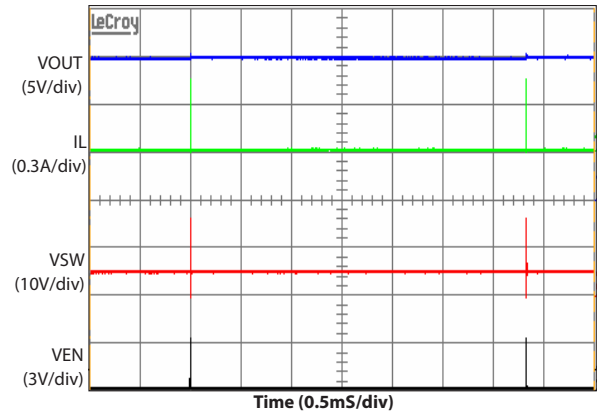
Conditions:  $5V_{IN}$ ,  $20mA$ , 6 LEDs in series,  $C_{IN}/C_{OUT}=1\mu F$ ,  $L=10\mu H$

### Over Voltage Protection



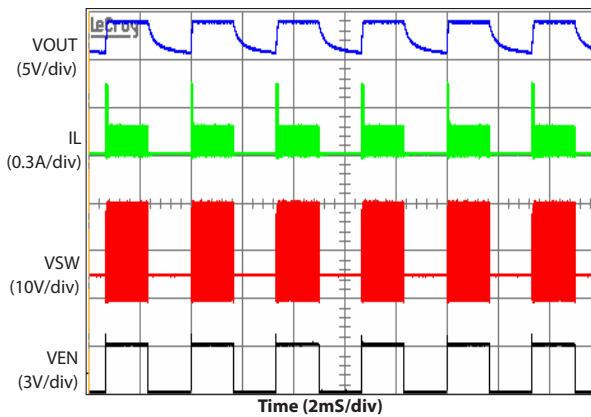
### PWM Dimming (0.1% duty cycle)

300Hz PWM Dimming, 0.1% Duty Cycle



### PWM Dimming (50% duty cycle)

300Hz PWM Dimming, 50% Duty Cycle







## Applications Information

### Operation

The SC4541 is a fixed frequency step-up current-mode switching regulator with an integrated Schottky diode and power transistor (see the Block Diagram for more information).

When EN is low, the device is in shutdown mode and draws less than 1µA of current. The internal circuitry is enabled and turns on when EN is logic high. The voltage drop across OUT and FB is reflected to ground and fed to the negative input of the error amplifier and a 200mV reference signal is fed to the positive input of the error amplifier. The output of the error amplifier is one input to the PWM comparator. The device is designed to be stable without additional external compensation components.

Switch current is sensed with an integrated sense resistor and is summed with the slope-compensating ramp which is then fed as the modulating ramp input to the PWM comparator. The latch is reset and Q1 is turned off when the modulating ramp intersects the error amplifier output. The latch will also be reset if the current in Q1 exceeds the current limit threshold, typically 350mA. The latch is set and Q1 turns on at the beginning of every oscillator clock cycle.

The SC4541 has a built-in Schottky diode. When input voltage is applied an inrush current flows through the inductor and the Schottky diode and charges up the output capacitor connected to OUT. The Schottky diode inside the SC4541 can sustain a maximum current of 1A.

### Programming the LED Current

The circuit will maintain the necessary output voltage to ensure 200mV across OUT and FB pins. Refer to the Typical Application Circuit on page 1. The LED current is programmed by connecting a resistor ( $R_{SENSE}$ ) between the OUT pin and FB pin. The LED current, in mA, is set by using the following relationship.

$$I_{LED} = 200/R_{SENSE}$$

where  $R_{SENSE}$  value is in Ohms.  
 $I_{LED}$  value is in mA.

The feedback voltage has a direct effect on the converter efficiency. Because the voltage drop across the feedback resistor does not contribute to the output power (LED brightness), the lower the feedback voltage, the higher

the efficiency. The internal oscillator sets a high switching frequency of 2.0MHz which further reduces the total solution size by minimizing external L/C filter size.

### Over-Voltage Protection (OVP)

An integrated over-voltage protection circuit prevents the OUT voltage from exceeding the maximum switch voltage rating (28V). When EN is high, a resistor divider from OUT to GND is switched in to monitor OUT for over-voltage protection. If OUT rises higher than the OVP threshold, typically 27V, the OVP comparator will trip and reset the latch to turn Q1 off. When OUT drops back below the OVP threshold, Q1 will turn back on in the next oscillator cycle. This sequence will repeat for as long as the over-voltage condition persists.

### Enable and PWM Dimming

The EN pin could be toggled for direct PWM dimming. In a typical application, a microcontroller sets a register which varies the pulse width on a GPIO which is tied to EN pin. Direct PWM dimming can be used to control LED brightness when a PWM signal (up to 1kHz) with wide duty ratio is applied.

### Inductor Selection

The inductance value affects the converter's steady state operation, transient response, and its loop stability. Attention must be given to inductor parameters; including inductance, DC resistance, and saturation current. The inductance and operating conditions determine the inductor ripple current. The converter can operate in continuous conduction mode (CCM) or discontinuous conduction mode (DCM). The inductor DC current or input current can be calculated using the following equation.

$$I_{IN} = \frac{V_{OUT} \times I_{OUT}}{V_{IN} \times \eta}$$

$I_{IN}$  - Input current;

$I_{OUT}$  - Output current;

$V_{OUT}$  - Boost output voltage;

$V_{IN}$  - Input voltage;

$\eta$  - Efficiency of the boost converter.

## Applications Information ( continued)

The converter has higher efficiency under CCM and the inductor peak current is calculated by the following equation.

$$I_{L\_PEAK} = I_{IN} + \frac{V_{IN} \times D}{2 \times F_S \times L}$$

Assuming CCM, the duty ratio is calculated using the following equation.

$$D = \frac{V_{OUT} - V_{IN} + V_D}{V_{OUT} + V_D}$$

$V_D$  is the forward voltage drop of the internal Schottky rectifier.

The inductor ripple current is further shown by the following equation.

$$I_{L\_RIPPLE} = \frac{V_{IN} \times D}{F_S \times L}$$

The recommended inductance shown in Table 1 is based on operation with regulated 5V input, dual Li-ion or three Li-ion battery packs. Increased inductance is required for higher input and output voltage configurations.

**Table 1 — Inductance Requirements, Driving 6 bWLED**

Input Type	Input Voltage Range	L1 Inductance
Regulated 5V	4.5 to 5.5V	10μH
Dual Lithium Ion Battery Pack	5.4 to 8.4V	22μH
Three Lithium Ion Battery Pack	8.1 to 12.6V	33μH

## Output Capacitor Selection

Ceramic capacitors are placed near IN and OUT pins to ground in order to minimize ripple voltage. The minimum capacitance needed for a given ripple can be estimated using the following equation.

$$C_{OUT} = \frac{(V_{OUT} - V_{IN}) \times I_{OUT}}{V_{OUT} \times F_S \times V_{RIPPLE}}$$

$V_{RIPPLE}$  = Peak to peak output voltage ripple

$I_{OUT}$  = Output current

$V_{OUT}$  = Boost output voltage

$V_{IN}$  = Input voltage

$F_S$  = Switching frequency

A 1μF ceramic input and output capacitor should be adequate for most applications. Larger capacitance reduces voltage ripple, overshoot and undershoot during load transient as well as improving loop stability margin. Recommended ceramic capacitor manufacturers are listed in Table 1.

**Table 2 — Recommended Ceramic Capacitor Manufacturers**

Vendor	Phone	Website
Kemet	408-986-0424	www.kemet.com
Murata	814-237-1431	www.murata.com
Taiyo Yuden	408-573-4150	www.t-yuden.com

## Applications Information ( continued)

### PCB Layout Considerations

Poor layout can degrade the performance of the DC-DC converter and can be a contributory factor in EMI problems, ground bounce, thermal issues, and resistive voltage losses. Poor regulation and instability can result. A typical application schematic is shown in Figure 1. A typical PCB layout is shown in Figure 4.

The following design rules are recommended:

- Place the inductor and filter capacitors as close to the device as possible and use short, wide traces between the power components.
- Use a ground plane to further reduce noise interference on sensitive circuit nodes.

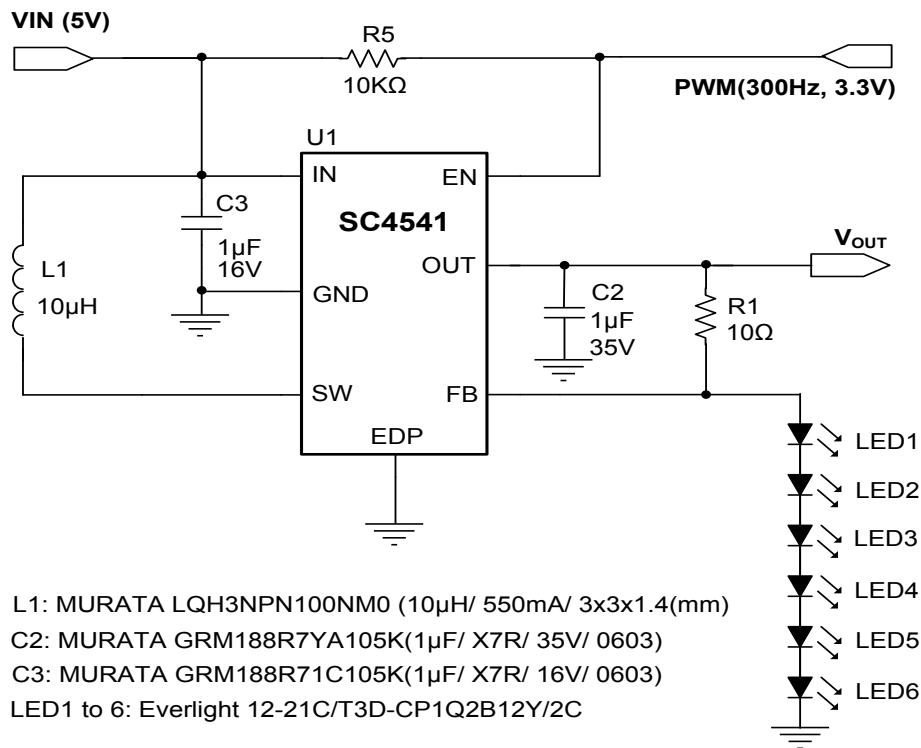
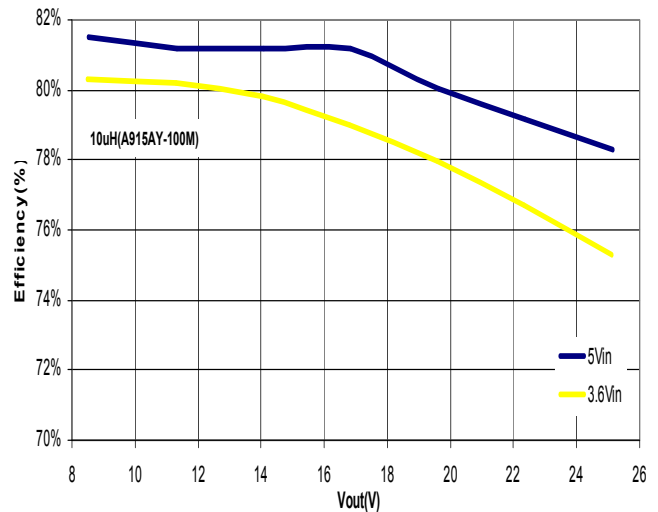
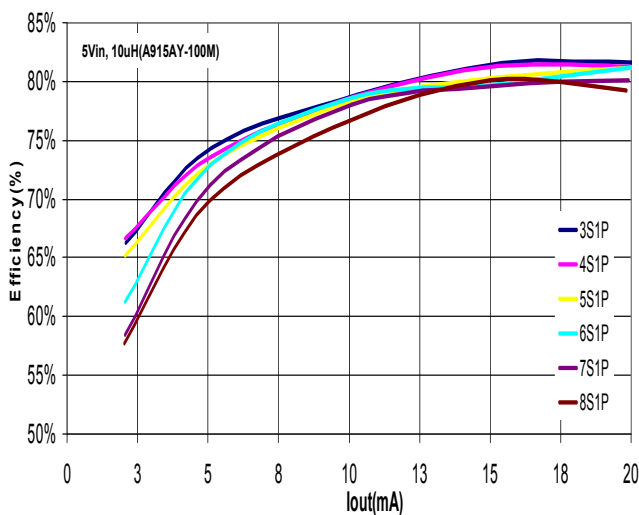
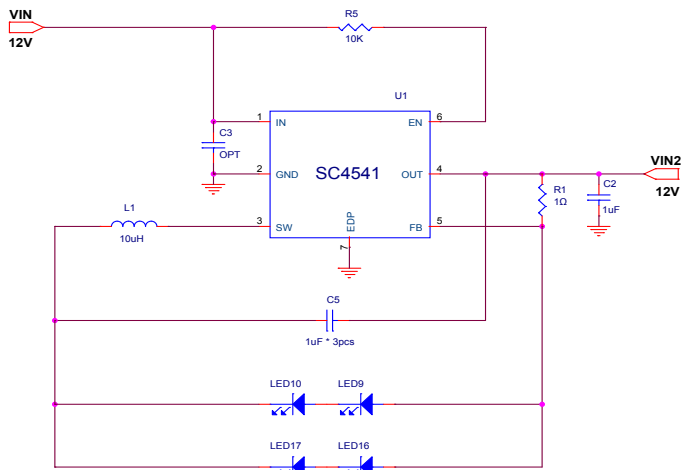


Figure 1 — Typical Application Schematic, Boost Topology



## Applications Information ( continued)



- (1) L1: Murata LQH3NPN100MM0L
- (2) C2, C5: Murata GRM188R7YA105K

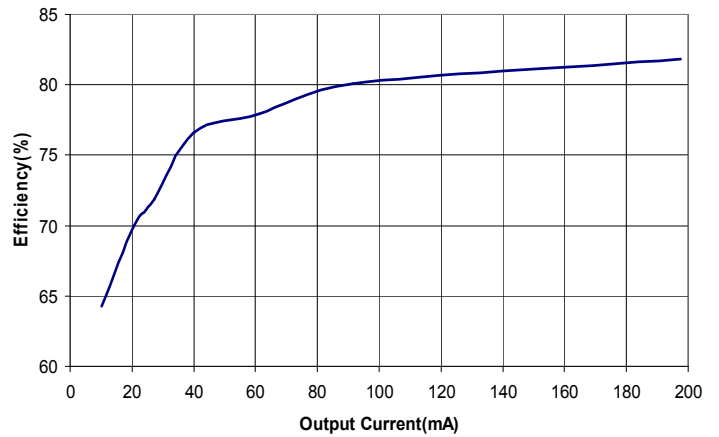
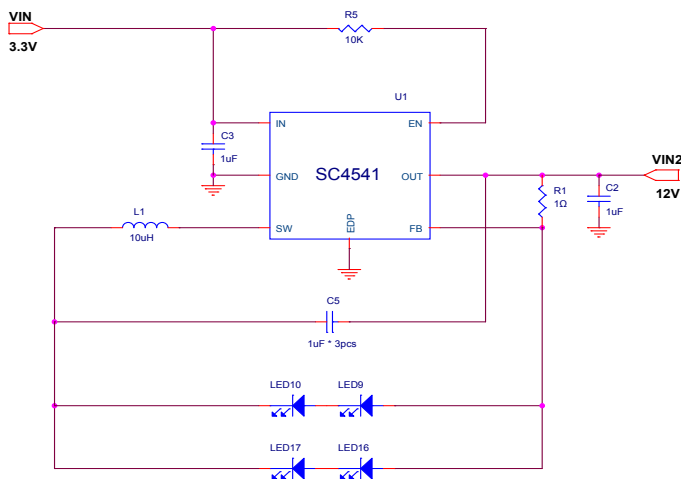


Figure 2 — Typical Application Schematic, Buck Topology



- (1) L1: Murata LQH3NPN100MM0L
- (2) C2, C3, C5: Murata GRM188R7YA105K

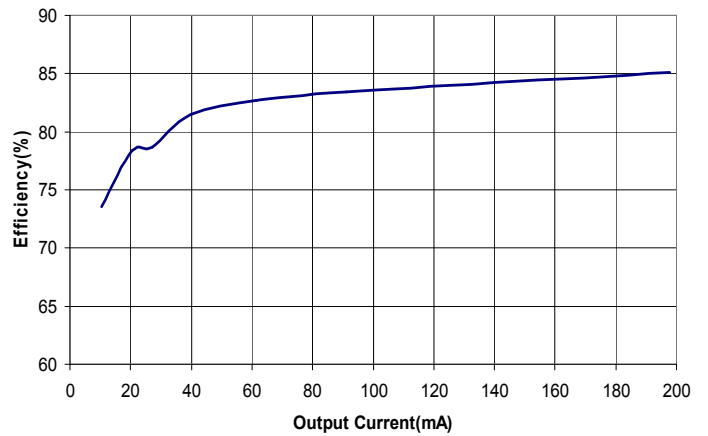


Figure 3 — Typical Application Schematic, High Efficiency of Buck Topology

Applications Information ( continued)

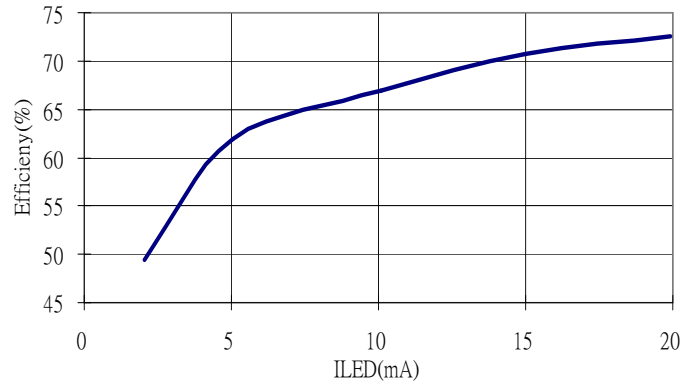
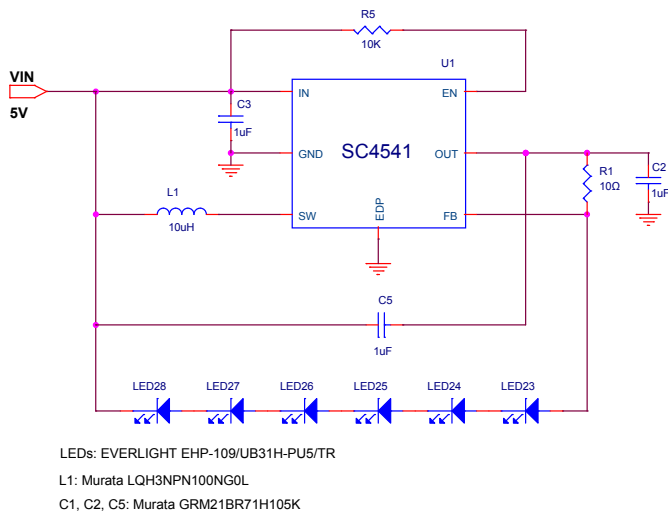


Figure 4 — Typical Application Schematic, Floating Buck-Boost Topology

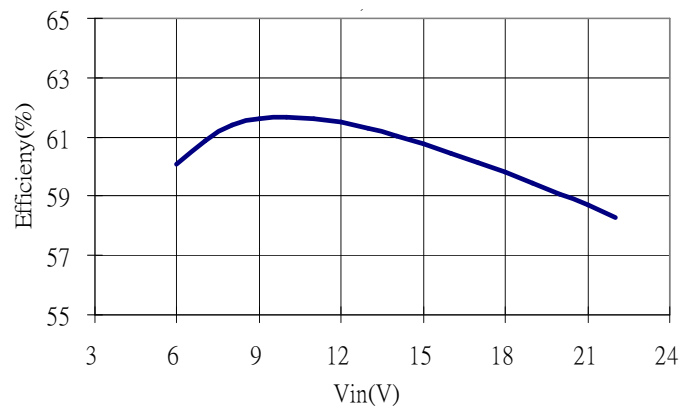
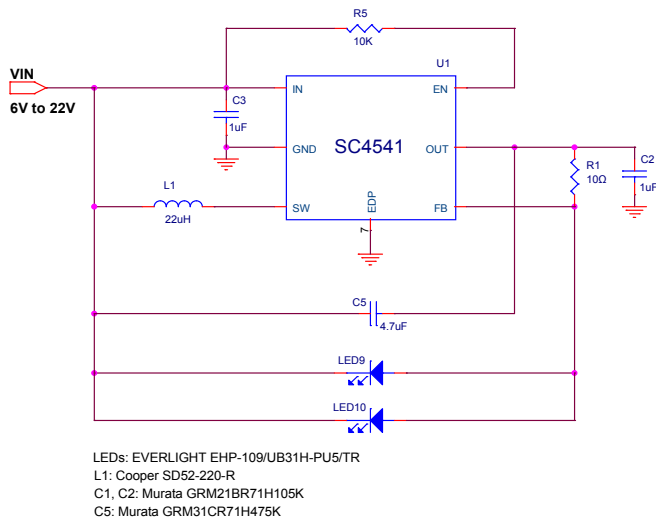
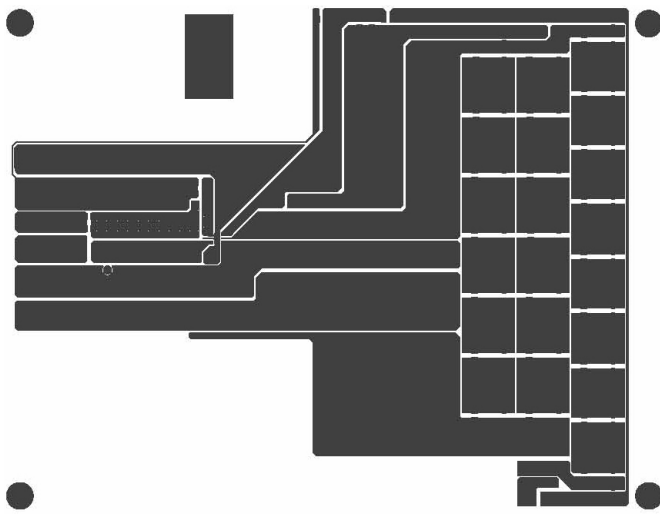
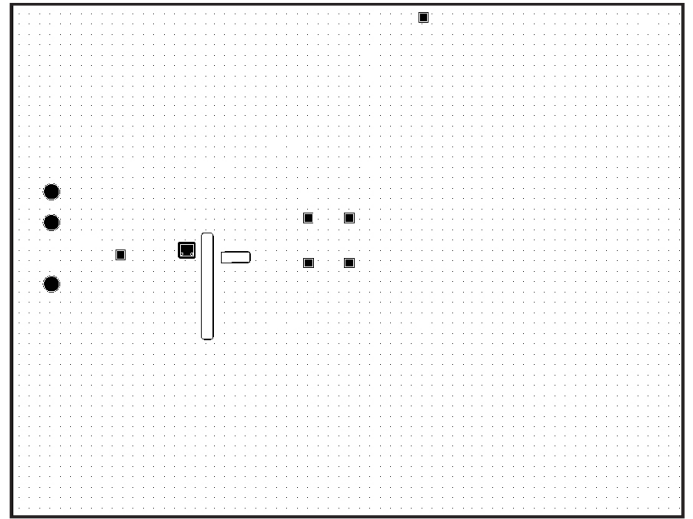


Figure 5 — Typical Application Schematic, Floating Buck-Boost Topology

**Applications Information ( continued)**

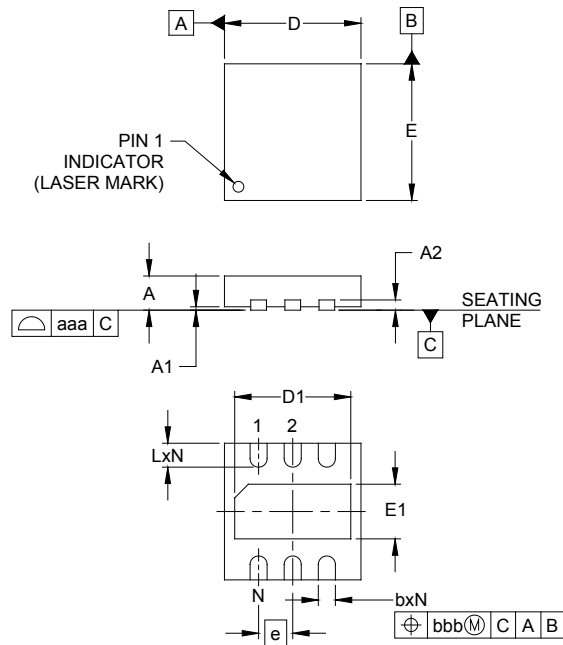


(TOP layer)



(Bottom layer)

**Figure 4 — Typical PCB Layout**

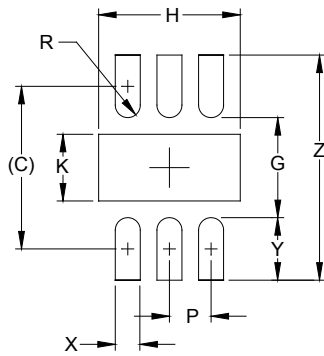
**Outline Drawing — 2x2 MLPD-UT6**


DIM	INCHES			MILLIMETERS		
	MIN	NOM	MAX	MIN	NOM	MAX
A	.020	-	.024	0.50	-	0.60
A1	.000	-	.002	0.00	-	0.05
A2	(.006)			(0.152)		
b	.007	.010	.012	0.18	0.25	0.30
D	.075	.079	.083	1.90	2.00	2.10
D1	.061	.067	.071	1.55	1.70	1.80
E	.075	.079	.083	1.90	2.00	2.10
E1	.026	.031	.035	0.65	0.80	0.90
e	.020 BSC			0.50 BSC		
L	.010	.014	.018	0.25	0.35	0.45
N	6			6		
aaa	.003			0.08		
bbb	.004			0.10		

**NOTES:**

1. CONTROLLING DIMENSIONS ARE IN MILLIMETERS (ANGLES IN DEGREES).
2. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS TERMINALS.



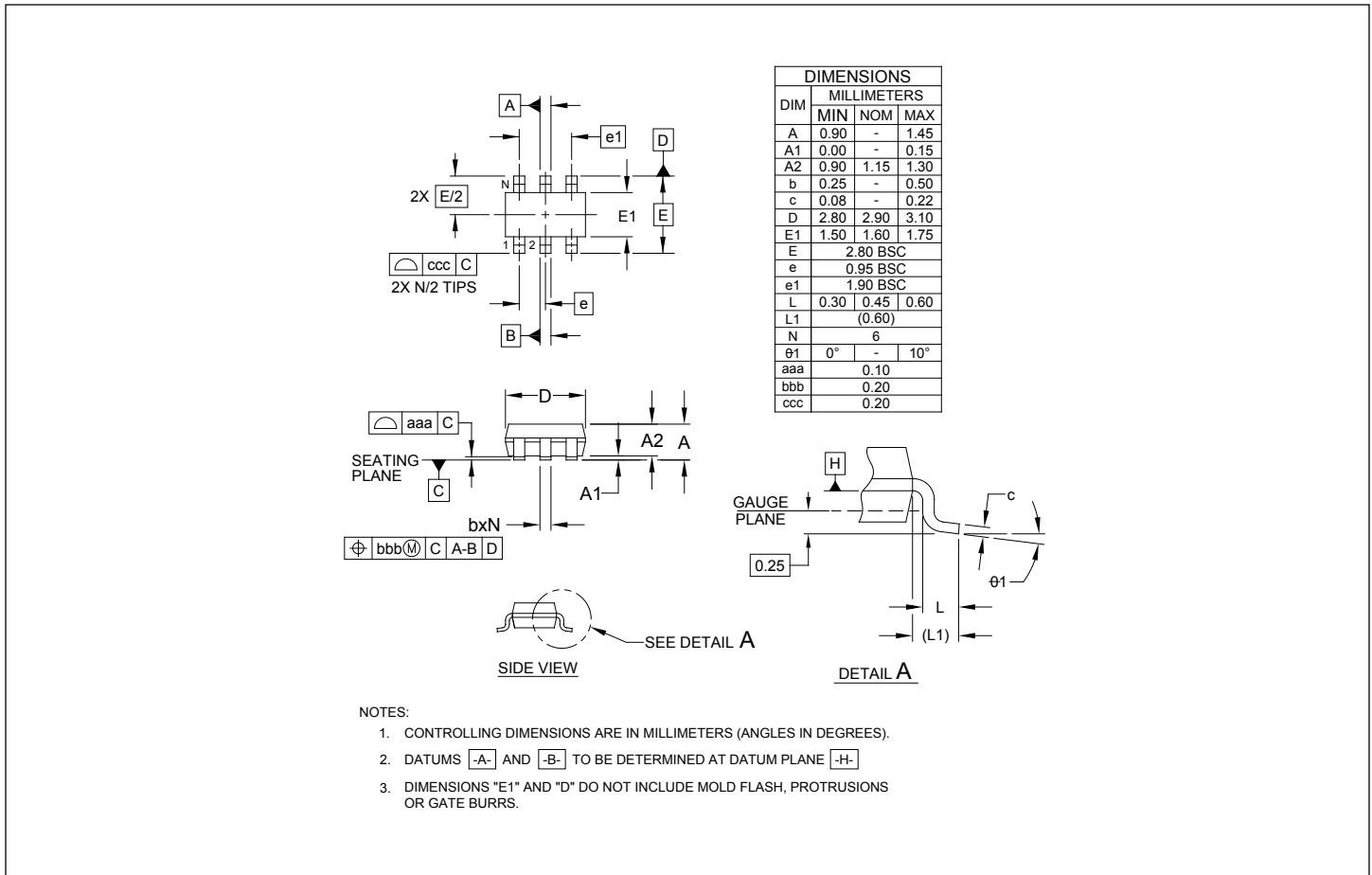
**Land Pattern — 2x2 MLPD-UT6**


DIMENSIONS		
DIM	INCHES	MILLIMETERS
C	(.077)	(1.95)
G	.047	1.20
H	.067	1.70
K	.031	0.80
P	.020	0.50
R	.006	0.15
X	.012	0.30
Y	.030	0.75
Z	.106	2.70

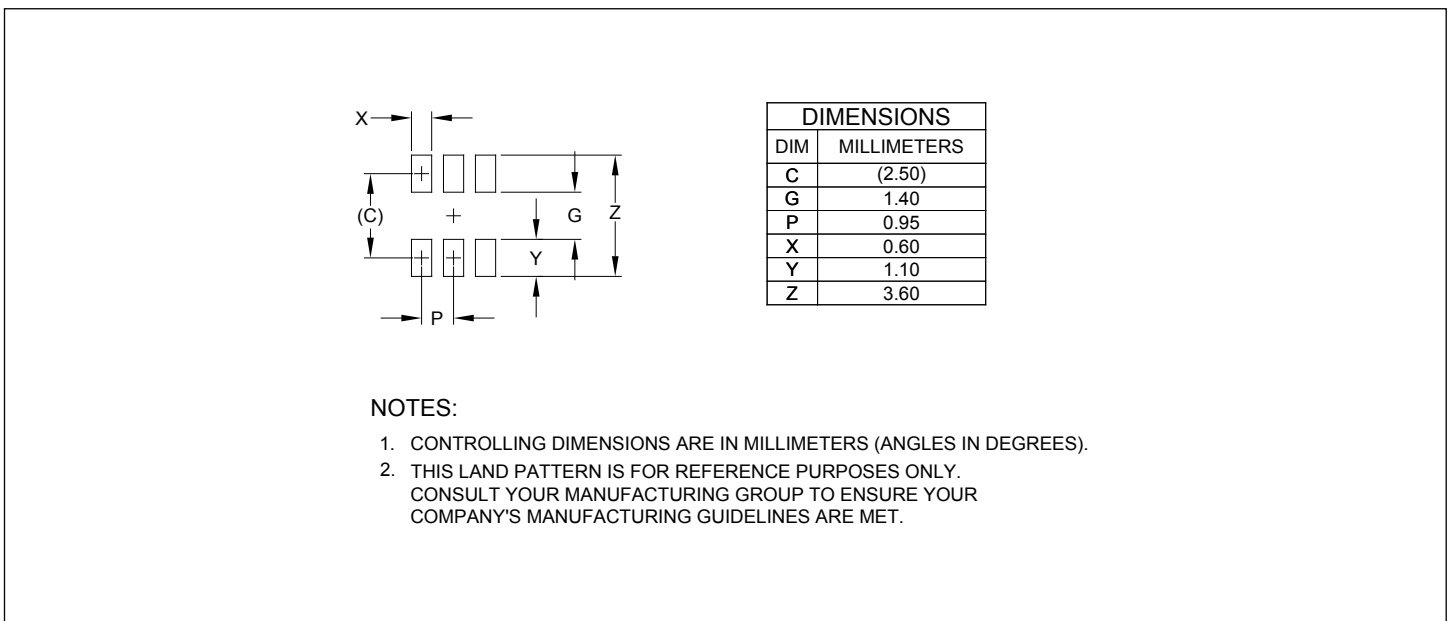
**NOTES:**

1. CONTROLLING DIMENSIONS ARE IN MILLIMETERS (ANGLES IN DEGREES).
2. THIS LAND PATTERN IS FOR REFERENCE PURPOSES ONLY. CONSULT YOUR MANUFACTURING GROUP TO ENSURE YOUR COMPANY'S MANUFACTURING GUIDELINES ARE MET.
3. THERMAL VIAS IN THE LAND PATTERN OF THE EXPOSED PAD SHALL BE CONNECTED TO A SYSTEM GROUND PLANE. FAILURE TO DO SO MAY COMPROMISE THE THERMAL AND/OR FUNCTIONAL PERFORMANCE OF THE DEVICE.

Outline Drawing — SOT23-6



Land Pattern — SOT23-6



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