## **ZXCT1020** Low offset current output current monitor

### Description

The ZXCT1020 is a precision high-side current sense monitor. Using this type of device eliminates the need to disrupt the ground plane when sensing a load current.

The ZXCT1020 uses two external resistors to set the overall voltage gain for applications where improved accuracy at small sense voltages is required. For fixed gain variants Zetex offers the ZXCT1021 (G=10) and ZXCT1022 (G=100).

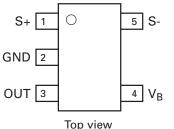
The ZXCT1020 footprint follows that of the ZXCT1021/2 with only 2 additional resistors required:

One resistor between pins 1 and 4 for setting transconductance, and the other between pins 3 and 2 for setting overall gain.

#### Features

- Accurate high-side current sensing
- Versatile current output scaling
- 2.5V 20V operating range
- 25µA quiescent current
- 1% typical accuracy
- SOT23-5 package

### **Pinout information**



Current output enables the user to set the gain via these external resistors. Using two external resistors to set the gain ensures optimal versatility as the transconductance can be varied to meet the output impedance requirements of the load that the ZXCT1020 has to drive.

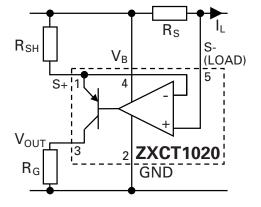
The very low offset voltage enables a typical accuracy of 3% for sense voltages of only 10mV, giving better tolerances for small sense resistors necessary at higher currents.

The wide input voltage range of 20V down to as low as 2.5V make it suitable for a range of applications. With a minimum operating current of just 25µA, combined with its SOT23-5 package make it suitable for portable battery equipment too.

#### Applications

- Battery chargers
- Over-current monitor
- Motherboard power supply current measurement
- Level translating
- Programmable current source

### **Typical application circuit**



### **Ordering information**

Order reference	Package	Device marking	Status	Reel size (inches)	Quantity per reel	Tape width (mm)
ZXCT1020E5TA	SOT23-5	1020	Preview	7	3000	8

### Absolute maximum ratings

Voltage on V <sub>B</sub> with respect to GND pin	-0.5V to 20V
Voltage on S+ <sup>(a)</sup> , S- <sup>(b)</sup> , OUT with respect to GND pin	-0.5V to V <sub>B</sub> +0.5V
V <sub>SENSE</sub> <sup>(c)</sup>	-0.5V to +2.5V <sup>(d)</sup>
Junction temperature	-40°C to125°C
Storage temperature	-55°C to 150°C
Package power dissipation (T <sub>amb</sub> = 25°C) SOT23-5	300mW

#### NOTES:

(a) Subject to  $V_{\mbox{SENSE}}\mbox{+}$  never going 6V below  $V_{\mbox{B}}\mbox{-}$ 

(b) Subject to absolute maximum V<sub>SENSE</sub> not being exceeded.

(c) V<sub>SENSE</sub> is defined as the voltage difference across the sense resistor. and is the voltage across resistor R<sub>SH</sub> plus the voltage between S+ and S-.

(d) V<sub>SENSE</sub> might need to be reduced when used with smaller values of R<sub>SH</sub> and at larger rails due to increased power dissipation.

#### **Pin out information**

Pin	Name	Pin function
1	S+	Positive sense input. Should be tied to positive side of sense resistor via resistance (R <sub>SH</sub> ) of the order of 150 $\Omega$ to 1.5k $\Omega$ .
2	GND	Ground and substrate connection of device.
3	OUT	Current output. A gain setting resistor (R <sub>G</sub> ) referenced to GND should be connected to this pin to set overall voltage gain of: Gain = R <sub>G</sub> /R <sub>SH</sub> The resistance, R <sub>G</sub> , placed on out will set the ZXCT1020 output impedance equal to R <sub>G</sub> . When driving low impedance loads both R <sub>G</sub> and R <sub>SH</sub> should be reduced.
4	V <sub>B</sub>	Input voltage pin. Provides bias to current monitor and should be tied to the rail whose current is being monitored.
5	S-	High impedance negative sense voltage input

## **Recommended operating conditions**

	Parameter	Min.	Max.	Units
$V_{SENSE+}$	Common-mode sense input range	2.5	20	V
V <sub>B</sub>	Bias pin input voltage range <sup>(*)</sup>	2.5	20	V
V <sub>SENSE</sub>	Differential sense Input voltage range	0	1.5	V
V <sub>OUT</sub>	Output voltage range	0	V <sub>SENSE-</sub> - 1	V
R <sub>SH</sub>	Shunt resistor value	120	2000	Ω
T <sub>A</sub>	Ambient temperature range	-40	85	°C

NOTES:

(\*) For best performance  $V_B$  and  $V_{SENSE+}$  should be referred to the rail whose current is being measured.

### **Recommended resistor gain setting combinations**

Gain	R <sub>SHUNT</sub>	R <sub>GAIN</sub>
10	1.5k $\Omega$	15k $\Omega$
20	750Ω	15k $\Omega$
50	300Ω	15k $\Omega$
100	150Ω	15k $\Omega$

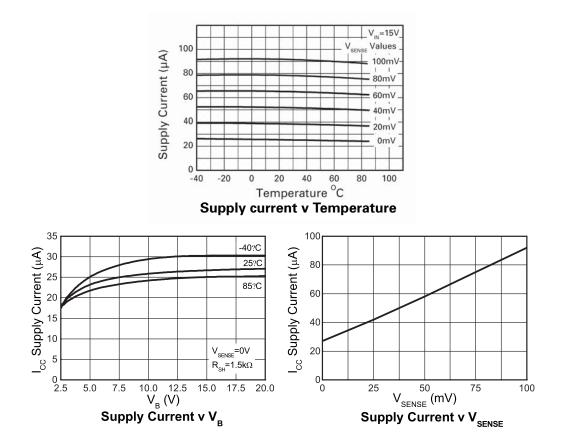
### **Electrical characteristics**

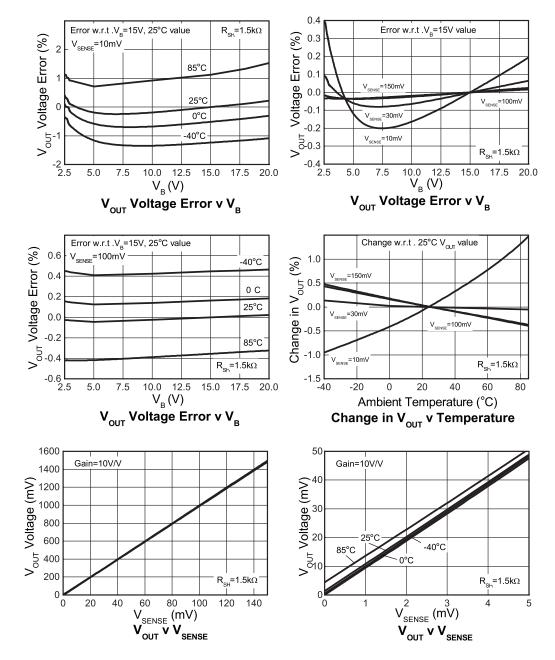
 $T_{amb} = 25^{\circ}C, V_{SENSE+} = V_B = 15V, V_{SENSE} = 100 \text{mV}, R_G = 15 \text{k}\Omega, R_{SH} = 1.5 \text{k}\Omega \text{ unless otherwise stated}.$ 

Symbol	Parameter	Conditions	Limits			Unit
			Min.	Тур.	Max.	
V <sub>OUT</sub>	Output voltage	V <sub>SENSE</sub> = 0mV		3	15	mV
		V <sub>SENSE</sub> = 30mV	291	300	309	mV
		V <sub>SENSE</sub> = 100mV	0.98	1	1.02	V
		V <sub>SENSE</sub> = 150mV	1.47	1.5	1.53	V
TC[1]	Output voltage temperature coefficient			50	300	ppm
۱ <sub>0</sub>	Ground pin current	V <sub>SENSE</sub> = 0V		25	35	μA
I <sub>S-</sub>	S- input current	V <sub>SENSE</sub> = 0V		20	100	nA
I <sub>S</sub> +	S+ input current	V <sub>SENSE</sub> = 0V		100		nA
Acc	Accuracy	V <sub>SENSE</sub> = 100mV	-2		2	%
Gain	V <sub>OUT</sub> /V <sub>SENSE</sub>	V <sub>SENSE</sub> = 100mV		10		V/V
R <sub>OUT</sub>	Output resistance	R <sub>G</sub> not connected		370		MΩ
BW	Bandwidth	V <sub>SENSE</sub> (DC) = 10mV		300		kHz
		V <sub>SENSE</sub> (DC) = 100mV		2		MHz
PSRR	Power supply rejection ratio	$V_{\text{SENSE+}} = V_{\text{B}} = 2.5 \text{ to } 20 \text{V}$	70	80		dB

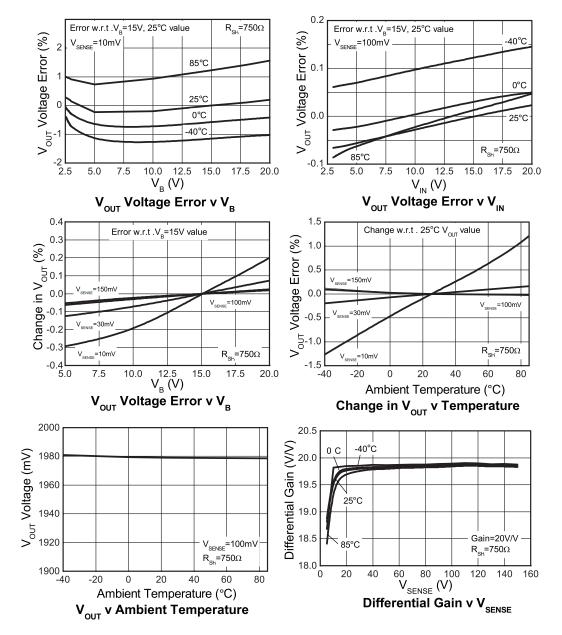
### **Typical characteristics**

Test conditions unless otherwise stated: T<sub>A</sub> = 25°C, V<sub>B</sub> = V<sub>SENSE+</sub> (via R<sub>SH</sub>) =15V, V<sub>SENSE</sub> = 100mV R<sub>SH</sub> = 1.5k $\Omega$ , R<sub>G</sub> = 15k $\Omega$ .

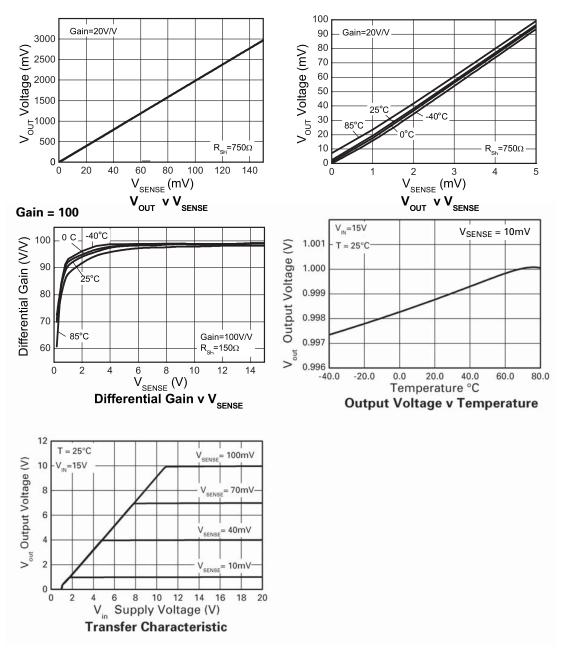




Test conditions unless otherwise stated:  $T_A = 25^{\circ}C$ ,  $V_B = V_{SENSE+}$  (via  $R_{SH}$ ) =15V,  $V_{SENSE}$  = 100mV Gain = 10,  $R_G = 15k\Omega$ .



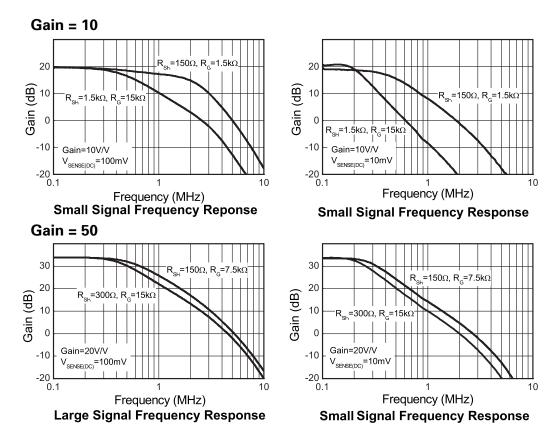
Test conditions unless otherwise stated:  $T_A = 25^{\circ}C$ ,  $V_B = V_{SENSE+}$  (via  $R_{SH}$ ) =15V,  $V_{SENSE}$  = 100mV Gain = 20,  $R_G = 15k\Omega$ .



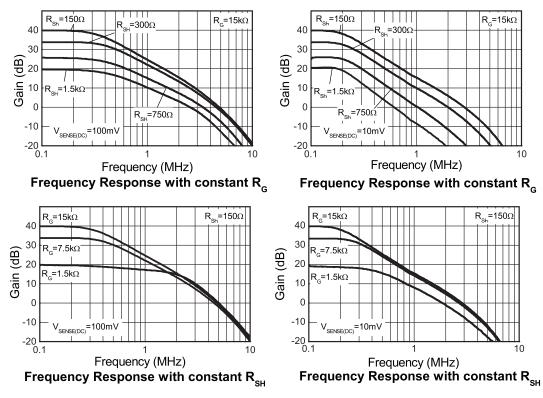
Test conditions unless otherwise stated: T<sub>A</sub> = 25°C, V<sub>B</sub> = V<sub>SENSE+</sub> (via R<sub>SH</sub>) =15V, V<sub>SENSE</sub> = 100mV R<sub>G</sub> = 15k $\Omega$ .

### **Typical AC characteristics**

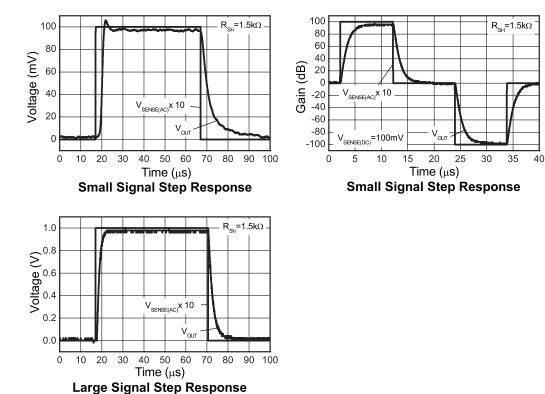
Test conditions unless otherwise stated:  $T_A = 25^{\circ}C$ ,  $V_B = V_{SENSE+}$  (via  $R_{SH}$ ) =15V,  $V_{SENSE}$  = 100mV,  $R_G = 15k\Omega$ .



Test conditions unless otherwise stated:  $T_A = 25^{\circ}C$ ,  $R_G=15k$ ,  $V_B = V_{SENSE+}$  (via  $R_{SH}$ ) =15V,  $V_{SENSE} = 100mV$  unless otherwise stated.



Various gains with constant  $\mathrm{R}_{\mathrm{G}}$ 

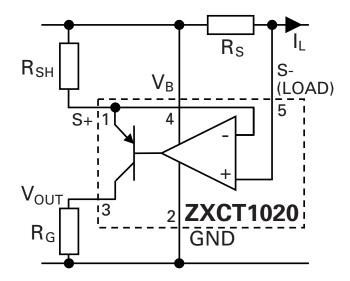


Test conditions unless otherwise stated: T<sub>A</sub> = 25°C, G=100, R<sub>G</sub> = 15k, V<sub>B</sub> = V<sub>SENSE+</sub> (via R<sub>SH</sub>), V<sub>SENSE</sub> = 100mV.

### **Application information**

The ZXCT1020 has a V<sub>B</sub> pin that is used to provide power to the current monitor. The maximum voltage applied to the ZXCT1020 must be applied to this pin. The S+ and S- pins are used to measure the current flowing to the load through the sense resistor. In normal use, the S+ is tied to V<sub>B</sub> via a shunt resistor, R<sub>H</sub> making the ZXCT1020 essentially line powered.

The ZXCT1020 has a programmable gain set by the ratio of two external resistors R<sub>G</sub> and R<sub>SH</sub>.



 ${\sf R}_{\sf SH}$  sets the transconductance whereas  ${\sf R}_{\sf G}$  set the gain and results in an output voltage defined as:

$$V_{OUT} = \frac{R_G}{R_{SH}} \times V_{SENSE}$$

Where  $V_{SENSE} = R_{SENSE} \times I_L$ 

The ZXCT1020 has been tested to the same conditions as the ZXCT1021 giving an overall voltage gain of 10. The gain of the ZXCT1020 can be adjusted simply by varying R<sub>G</sub>. So to achieve a gain of 50 R<sub>G</sub> is increased from 15k $\Omega$  to 75k $\Omega$ . An alternative is to decrease R<sub>SH</sub> from 1.5k $\Omega$  to 300 $\Omega$ .

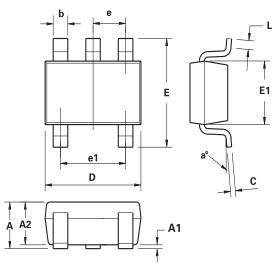
Decreasing  $R_{SH}$  increases the transconductance and, if for any given gain, reducing the  $R_{SH}$  will reduce the overall output impedance.

To achieve a gain of 100, for example, the following resistor values could be used:

 $R_{SH} = 150 R_{G} = 15k$ 

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## Package outline - SOT23-5



Dim.	Millimeters		Inc	hes
	Min.	Max.	Min.	Max.
A	0.90	1.45	0.0354	0.0570
A1	0.00	0.15	0.00	0.0059
A2	0.90	1.30	0.0354	0.0511
b	0.20	0.50	0.0078	0.0196
С	0.09	0.26	0.0035	0.0102
D	2.70	3.10	0.1062	0.1220
E	2.20	3.20	0.0866	0.1181
E1	1.30	1.80	0.0511	0.0708
е	0.95 REF		0.0374 REF	
e1	1.90 REF		0.0748 REF	
L	0.10	0.60	0.0039	0.0236
a°	0°	30°	0°	30°

Note: Controlling dimensions are in millimeters. Approximate dimensions are provided in inches

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