1.0 A Low-Dropout Positive **Fixed and Adjustable Voltage Regulators**

The NCP1117LP is the low power version of the popular NCP1117 family of low dropout voltage regulators, with reduced quiescent current. It is intended primarily for high volume consumer applications over the 0 to 125 degree temperature range. Capable of providing an output current in excess of 1 A, with a dropout voltage of 1.3 V at 1 A full current load, the series consists of an adjustable and five fixed voltage versions of 1.5 V, 1.8 V, 2.5 V, 3.3 V and 5.0 V.

Internal protection features consist of output current limiting and built-in thermal shutdown. The NCP1117LP series can operate up to 18 V max input voltage. The device is available in the popular SOT-223 package.

Features

- Output Current in Excess of 1.0 A
- 1.4 V Maximum Dropout Voltage at 1 A
- Quiescent Current over 10 times Lower than Traditional 1117
- Fixed Output Voltages of 1.5 V, 1.8 V, 2.5 V, 3.3 V and 5.0 V
- Adjustable Output Voltage Option
- No Minimum Load Requirement for Fixed Voltage Output Devices
- Good Noise Rejection
- Current Limit and Thermal Shutdown Protection
- Operation up to 18 V Input
- These are Pb-Free Devices

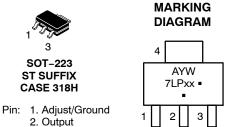
Applications

- TV and Monitors
- Set Top Boxes and Entertainment Devices
- Switching Power Supply Post Regulation
- Game Consoles and Consumer Applications
- Hard Drive Controllers



ON Semiconductor®

http://onsemi.com



3. Input

Heatsink tab is connected to Pin 2.

= 15, 18, 25, 33, 50, AD Α = Assembly Location

= Year W = Work Week

= Pb-Free Package

(Note: Microdot may be in either loca-

ORDERING INFORMATION

Device	Package	Shipping [†]
NCP1117LPST15T3G		
NCP1117LPST18T3G		
NCP1117LPST25T3G	SOT-223 (Pb-Free)	4000/Tape &
NCP1117LPST33T3G		Reel
NCP1117LPST50T3G		
NCP1117LPSTADT3G		

[†]For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

TYPICAL APPLICATIONS

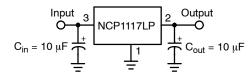


Figure 1. Fixed **Output Regulator**

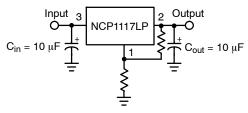


Figure 2. Adjustable **Output Regulator**

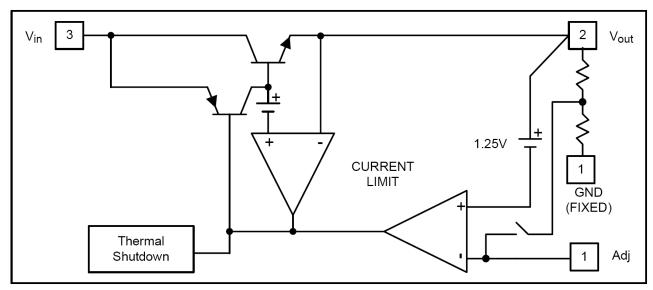


Figure 3. Block Diagram

Table 1. PIN FUNCTION DESCRIPTION

Pin No.	Pin Name	Description		
1	Adj (GND)	A resistor divider from this pin to the Vout pin and ground sets the output voltage (Ground only for Fixed-Mode).		
2	Vout	The output of the regulator. A minimum of 10 μ F capacitor (20 m Ω \leq ESR \leq 20 Ω) must be connected from this pin to ground to insure stability.		
from this pin to ground to insure that the input voltage does not sag below the minimu		The input pin of regulator. Typically a large storage capacitor ($20 \text{ m}\Omega \leq \text{ESR} \leq 20 \Omega$) is connected from this pin to ground to insure that the input voltage does not sag below the minimum dropout voltage during the load transient response. This pin must always be 1.3 V (typ.) higher than Vout in order for the device to regulate properly.		

Table 2. MAXIMUM RATINGS

Rat	Symbol	Value	Unit	
DC Input Voltage		V _{in}	-0.3 to 18	V
Operating Junction Temperature Range		T _{OP}	0 to 125	°C
Operating Ambient Temperature Range		T _A	0 to 125	°C
Maximum Junction Temperature Range		T _{J(max)}	-55 to 150	°C
Power Dissipation and Thermal Characteristics - Power Dissipation (Note 1) - Thermal Resistance, Junction-to-Ambient (Note 2) - Thermal Resistance, Junction-to-Case		P _D R _{θJA} R _{θJC}	Internally Limited 108 15	W °C/W °C/W
Electrostatic Discharge	Human Body Model	ESD	2000	V
	Machine Model		200	
Storage Temperature Range		T _{STG}	-65 to 150	°C

NOTE: This device series contains ESD protection and exceeds the following tests:

ESD HBM tested per AEC-Q100-002 (EIA/JESD22-A114) ESD MM tested per AEC-Q100-003 (EIA/JESD22-A115)

Latch–up Current Maximum Rating: ≤ 150mA per JEDEC standard: JESD78

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

NOTE: All voltages are referenced to GND pin.

1. The maximum package power dissipation is:

$$P_{D} = \frac{T_{J(max)} - T_{A}}{R_{\theta JA}}$$

2. $R_{\theta JA}$ on a 100 x 100 mm PCB Cu thickness 1 oz; T_A = 25°C

Table 3. ELECTRICAL CHARACTERISTICS ($Cin = 10 \mu F$, $Cout = 10 \mu F$, for typical value TA = 25°C, for min and max values TA is the operating ambient temperature range that applies unless otherwise noted.)

Parameter		Conditions	Symbol	Min	Тур	Max	Unit
Reference Voltage, Adjustable Output Devices	NCP1117-ADJ	T _J = 25°C (Vin – Vout) = 1.5 V, Io = 10 mA	V _{ref}	1.225	1.250	1.275	V
Output Voltage, Fixed Output Devices	NCP1117-1.5	$T_J = 25^{\circ}C$ 3 V \le Vin \le 12 V, lo = 10 mA	V _{out}	1.470	1.5	1.530	V
	NCP1117-1.8	$T_J = 25^{\circ}C$ $3.3 \text{ V} \leq \text{Vin} \leq 12 \text{ V, lo} = 10 \text{ mA}$		1.760	1.8	1.840	٧
	NCP1117-2.5	$T_J = 25^{\circ}C$ 4 V \leq Vin \leq 12 V, lo = 10 mA		2.450	2.5	2.550	٧
	NCP1117-3.3	$T_J = 25^{\circ}C$ 4.8 V \leq Vin \leq 12 V, lo = 10 mA		3.235	3.3	3.365	V
	NCP1117-5.0	$T_J = 25^{\circ}C$ 6.5 V \leq Vin \leq 12 V, lo = 10 mA		4.900	5	5.100	٧
Line Regulation, Adjustable & Fixed (Note 3)	NCP1117-XXX	$T_J = 25^{\circ}C$ Vout + 1.5 V < Vin < 12 V, Io = 10 mA	Reg _{line}			0.2	%
Load Regulation (Note 3)	NCP1117-ADJ	T _J = 25°C 10 mA < lo < 1 A, Vin = 3.3 V	Reg _{load}			1	%
	NCP1117-1.5	T _J = 25°C 10 mA < lo < 1 A, Vin = 3 V			12	15	mV
	NCP1117-1.8	T _J = 25°C 10 mA < lo < 1 A, Vin = 3.3 V			15	18	mV
	NCP1117-2.5	T _J = 25°C 10 mA < lo < 1 A, Vin = 4 V			20	25	mV
	NCP1117-3.3	T _J = 25°C 10 mA < lo < 1 A, Vin = 4.7 V			26	33	mV
	NCP1117-5.0	T _J = 25°C 10 mA < lo < 1 A, Vin = 6.5 V			40	50	mV
Dropout Voltage (Vin – Vout), Adjustable & Fixed	NCP1117-XXX	lout = 1 A, T _A = 25°C ΔVout = Vout – 100 mV			1.3	1.4	V
Current Limit, Adjustable & Fixed	NCP1117-XXX	Vin = 7 V, T _A = 25°C	lout	1.1			Α
Minimum Load Current (Note 4)	NCP1117-XXX	0°C ≤ Tj ≤ 125°C	I _{Lmin}		1	5	mA
Quiescent Current	NCP1117-fixed	Vin = 12 V Io = 10 mA	I _{QFIX}		550	700	μΑ
	NCP1117-ADJ		I _{QADJ}		30	50	μΑ
Thermal Regulation (Note 5)		$T_A = 25$ °C, $T = 30$ ms pulse			0.008	0.04	%W
Ripple Rejection	NCP1117-XXX	F = 120 Hz, Cout = 25 μF tantalum, lout = 1 A, Vin = Vout + 3 V	RR		60		dB
Thermal Shutdown	NCP1117-XXX		T _{shdn}		165		°C
Thermal Hysteresis	NCP1117-XXX		T _{hyst}		10		°C

^{3.} Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible.

^{4.} Guaranteed by design.

^{5.} Thermal Regulation is defined as the change in output voltage at a time after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to lo_{max} at VIN = VIN + 1.5 V for T = 30 msec. Guaranteed by characterization.

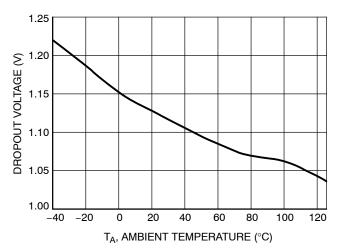


Figure 4. Dropout Voltage vs. Temperature $I_{load} = 10 \text{ mA}$

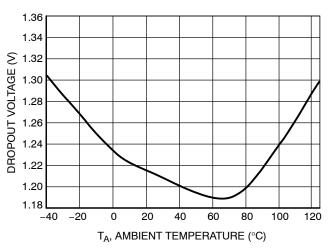


Figure 5. Dropout Voltage vs. Temperature $I_{load} = 1 A$

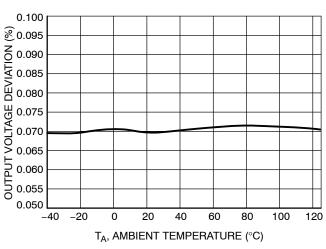


Figure 6. Line Regulation vs. Temperature $I_{load} = 10 \text{ mA}$

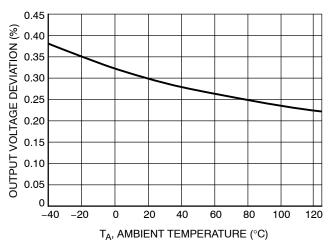


Figure 7. Load Regulation vs. Temperature $I_{load} = 1 A$

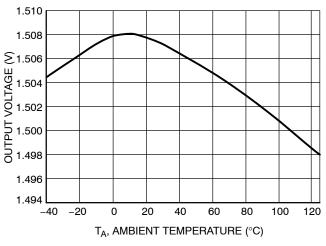


Figure 8. Output Voltage vs. Temperature $I_{load} = 10 \text{ mA}$

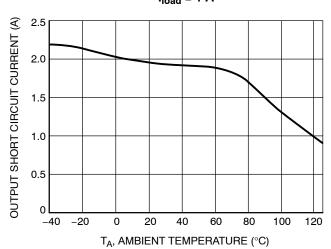


Figure 9. Output Short Circuit Current vs. Temperature

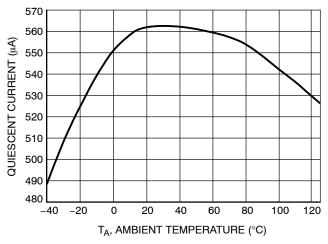


Figure 10. Quiescent Current vs. Temperature $I_{load} = 10 \text{ mA}$

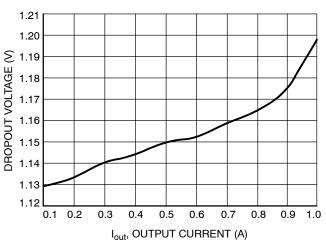


Figure 11. Dropout Voltage vs. Output Current

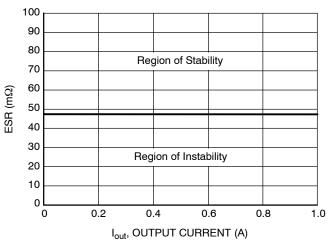


Figure 12. Equivalent Series Resistance vs.
Output Current – Electrolytic Capacitor

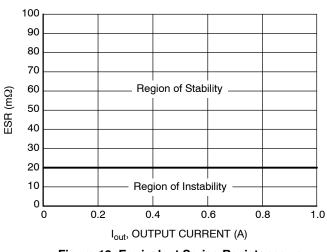


Figure 13. Equivalent Series Resistance vs.
Output Current – MLCC Capacitor

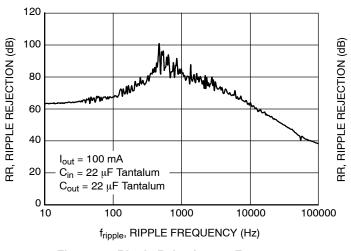


Figure 14. Ripple Rejection vs. Frequency – 1.5 V

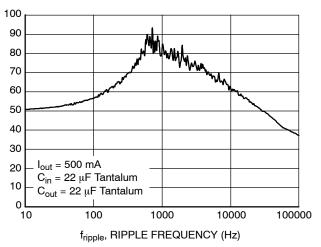


Figure 15. Ripple Rejection vs. Frequency – 1.5 V

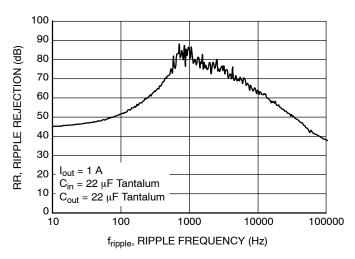


Figure 16. Ripple Rejection vs. Frequency – 1.5 V

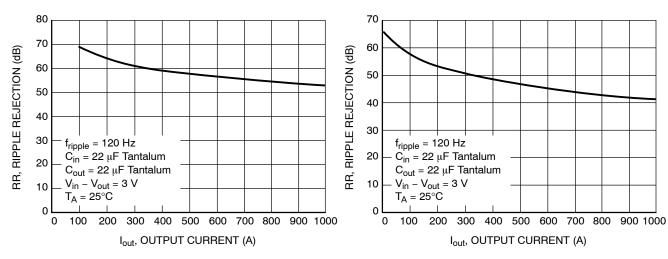


Figure 17. Ripple Rejection vs. Output Current – 1.5 V

Figure 18. Ripple Rejection vs. Output Current
- 5 V

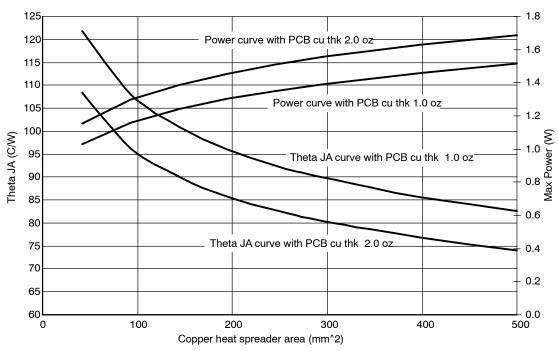


Figure 19. SOT-223 Thermal Resistance and Maximum Power Dissipation vs. P.C.B. Copper Length

APPLICATIONS INFORMATION

Introduction

The NCP1117LP is a low dropout positive fixed or adjustable mode regulator with 1 A output capability. This LDO is guaranteed to have a significant reduction in dropout voltage along with enhanced output voltage accuracy and temperature stability when compared to older industry standard three–terminal adjustable regulators.

These devices contain output current limiting, safe operating area compensation and thermal shutdown protection making them designer friendly for powering numerous consumer and industrial products. The NCP1117LP series is pin compatible with the older NCP1117.

Output Voltage

The typical application circuits for the fixed and adjustable output regulators are shown in Figures 20 and 21. The adjustable devices are floating voltage regulators. They develop and maintain the nominal 1.25 V reference voltage between the output and adjust pins. The reference voltage is programmed to a constant current source by resistor R1, and this current flows through R2 to ground to set the output voltage. The programmed current level is usually selected to be greater than the specified 5.0 mA minimum that is required for regulation. Since the adjust pin current, I_{adj} , is significantly lower and constant with respect to the programmed load current, it generates a small output voltage error that can usually be ignored. For the fixed output devices R1 and R2 are included within the device and the ground current I_{gnd} is 550 μ A (typ).

External Capacitors

Input bypass capacitor C_{in} may be required for regulator stability if the device is located more than a few inches from the power source. This capacitor will reduce the circuit's sensitivity when powered from a complex source impedance and significantly enhance the output transient response. The input bypass capacitor should be mounted with the shortest possible track length directly across the regulator's input and ground terminals. A $10~\mu F$ ceramic or tantalum capacitor should be adequate for most applications.

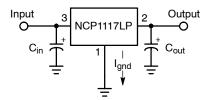


Figure 20. Fixed Output Regulator

Frequency compensation for the regulator is provided by capacitor C_{out} and its use is mandatory to ensure output stability. A minimum capacitance value of 4.7 μF with an equivalent series resistance (ESR) that is within the limits of 20 m Ω to 20 Ω is required. The capacitor type can be ceramic, tantalum, or aluminum electrolytic as long as it meets the minimum capacitance value and ESR limits over the circuit's entire operating temperature range. Higher values of output capacitance can be used to enhance loop stability and transient response with the additional benefit of reducing output noise.

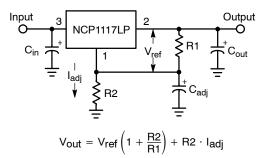


Figure 21. Adjustable Output Regulator

The output ripple will increase linearly for fixed and adjustable devices as the ratio of output voltage to the reference voltage increases. For example, with a 5 V regulator, the output ripple will increase by 5 V/1.25 V or 4 and the ripple rejection will decrease by 20 log of this ratio or 12 dB. The loss of ripple rejection can be restored to the values shown with the addition of bypass capacitor C_{adj} , shown in Figure 21. The reactance of C_{adj} at the ripple frequency must be less than the resistance of R1. The value of R1 can be selected to provide the minimum required load current to maintain regulation and is usually in the range of $100~\Omega$ to $200~\Omega$

$$C_{adj} > \frac{1}{2\pi \cdot f_{ripple} \cdot R1}$$

The minimum required capacitance can be calculated from the above formula. When using the device in an application that is powered from the AC line via a transformer and a full wave bridge, the value for C_{adj} is:

$$f_{ripple} \,=\, 120$$
 Hz, R1 = 120 $\Omega, \; then \; C_{adj} \,>\, 11.1 \; \mu F$

The value for C_{adj} is significantly reduced in applications where the input ripple frequency is high. If used as a post regulator in a switching converter under the following conditions:

$$f_{ripple}$$
 = 50 kHz, R1 = 120 Ω , then C_{adj} > 0.027 μF

Protection Diodes

The NCP1117LP family has two internal low impedance diode paths that normally do not require protection when used in the typical regulator applications. The first path connects between V_{out} and V_{in} , and it can withstand a peak surge current of about 15 A. Normal cycling of V_{in} cannot generate a current surge of this magnitude. Only when V_{in} is shorted or crowbarred to ground and C_{out} is greater than 50 μ F, it becomes possible for device damage to occur. Under these conditions, diode D1 is required to protect the device. The second path connects between C_{adj} and V_{out} , and it can withstand a peak surge current of about 150 mA. Protection diode D2 is required if the output is shorted or crowbarred to ground and C_{adj} is greater than 1.0 μ F.

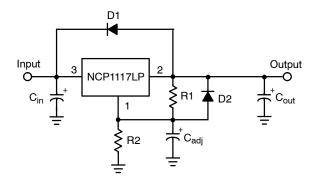


Figure 22. Protection Diode Placement

A combination of protection diodes D1 and D2 may be required in the event that V_{in} is shorted to ground and C_{adj} is greater than 50 μF . The peak current capability stated for the internal diodes are for a time of 100 μs with a junction temperature of 25°C. These values may vary and are to be used as a general guide.

Load Regulation

The NCP1117LP series is capable of providing excellent load regulation; but since these are three terminal devices, only partial remote load sensing is possible. There are two conditions that must be met to achieve the maximum available load regulation performance. The first is that the top side of programming resistor R1 should be connected as close to the regulator case as practicable. This will minimize the voltage drop caused by wiring resistance RW + from appearing in series with reference voltage that is across R1. The second condition is that the ground end of R2 should be connected directly to the load. This allows true Kelvin sensing where the regulator compensates for the voltage drop caused by wiring resistance RW –.

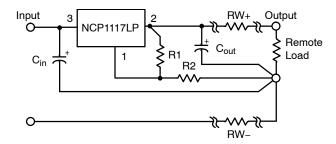


Figure 23. Load Sensing

Thermal Considerations

This series contains an internal thermal limiting circuit that is designed to protect the regulator in the event that the maximum junction temperature is exceeded. When activated, typically at 165°C, the regulator output switches off and then back on as the die cools. As a result, if the device is continuously operated in an overheated condition, the output will appear to be oscillating. This feature provides protection from a catastrophic device failure due to accidental overheating. It is not intended to be used as a substitute for proper heatsinking. The maximum device power dissipation can be calculated by:

$$P_D = \frac{T_J(max) - T_A}{R_{\theta JA}}$$

The devices are available in surface mount SOT–223 package. This package has an exposed metal tab that is specifically designed to reduce the junction to air thermal resistance, $R_{\theta J A}$, by utilizing the printed circuit board copper as a heat dissipater. Figure 19 shows typical $R_{\theta J A}$ values that can be obtained from a square pattern using economical single sided 1.0 oz and 2.0 oz copper board material. The final product thermal limits should be tested and quantified in order to insure acceptable performance and reliability. The actual $R_{\theta J A}$ can vary considerably from the graphs shown. This will be due to any changes made in the copper aspect ratio of the final layout, adjacent heat sources, and air flow.

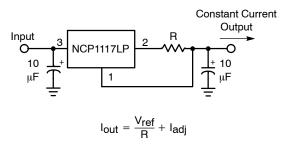


Figure 24. Constant Current Regulator

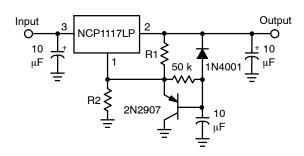
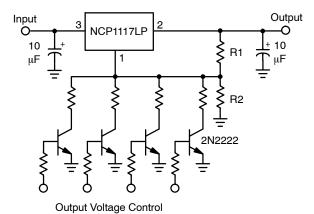


Figure 25. Slow Turn-On Regulator



Resistor R2 sets the maximum output voltage. Each transistor reduces the output voltage when turned on.

Figure 26. Digitally Controlled Regulator

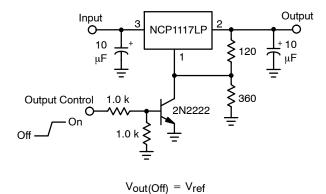


Figure 27. Regulator with Shutdown

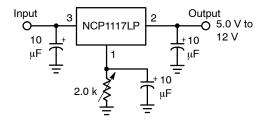
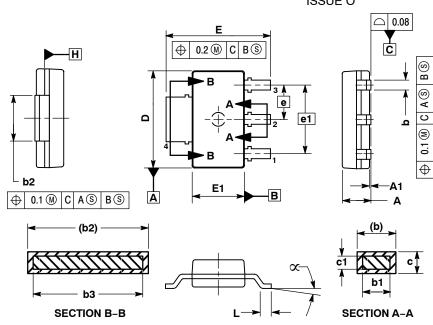


Figure 28. Adjusting Output of Fixed Voltage Regulators

PACKAGE DIMENSIONS

SOT-223 ST SUFFIX CASE 318H-01 ISSUE O



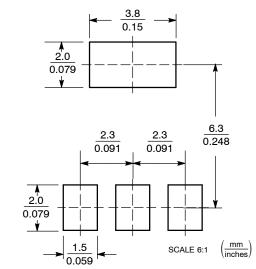
NOTES:

- NOTES:

 1. DIMENSIONS ARE IN MILLIMETERS.
 2. INTERPRET DIMENSIONS AND TOLERANCES PER ASME Y14.5M, 1994.
 3. DIMENSION E1 DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.23 PER
- SIDE.
 4. DIMENSIONS b AND b2 DO NOT INCLUDE 4. DIMENSIONS b AND b2 DO NOT INCLUDE
 DAMBAR PROTRUSION. ALLOWABLE DAMBAR
 PROTRUSION SHALL BE 0.08 TOTAL IN EXCESS
 OF THE b AND b2 DIMENSIONS AT MAXIMUM
 MATERIAL CONDITION.
 5. TERMINAL NUMBERS ARE SHOWN FOR
 REFERENCE ONLY.
 6. DIMENSIONS D AND E1 ARE TO BE DETERMINED
 AT DATIM B1 ANF H
- AT DATUM PLANE H.

	MILLIMETERS		
DIM	MIN MAX		
Α	1.80		
A1	0.02	0.11	
b	0.60 0.88		
b1	0.60 0.80		
b2	2.90 3.10		
b3	2.90 3.05		
C	0.24 0.35		
c1	0.24 0.30		
D	6.30 6.70		
Е	6.70 7.30		
E1	3.30 3.70		
е	2.30		
e1	4.60		
L	0.25		
\propto	0° 10°		

SOLDERING FOOTPRINT*



*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

ON Semiconductor and was are registered trademarks of Semiconductor Components Industries, LLC (SCILLC). SCILLC reserves the right to make changes without further notice to any products herein. SCILLC makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does SCILLC assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. "Typical" parameters which may be provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. SCILLC does not convey any license under its patent rights nor the rights of others. SCILLC products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other applications intended to support or sustain life, or for any other application in which the failure of the SCILLC product could create a situation where personal injury or death may occur. Should Buyer purchase or use SCILLC products for any such unintended or unauthorized application, Buyer shall indemnify and hold SCILLC and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that SCILLC was negligent regarding the design or manufacture of the part. SCILLC is an Equal Opportunity/Affirmative Action Employer. This literature is subject to all applicable copyright laws and is not for resale in any manner.

PUBLICATION ORDERING INFORMATION

LITERATURE FULFILLMENT:

Literature Distribution Center for ON Semiconductor P.O. Box 5163, Denver, Colorado 80217 USA Phone: 303-675-2175 or 800-344-3860 Toll Free USA/Canada Fax: 303-675-2176 or 800-344-3867 Toll Free USA/Canada Email: orderlit@onsemi.com N. American Technical Support: 800–282–9855 Toll Free USA/Canada

Europe, Middle East and Africa Technical Support: Phone: 421 33 790 2910 Japan Customer Focus Center Phone: 81-3-5773-3850 ON Semiconductor Website: www.onsemi.com

Order Literature: http://www.onsemi.com/orderlit

For additional information, please contact your local Sales Representative

NCP1117LP/D