103 dB

9.5 KHz

2.2V, 5V, 10V

0.95 µA/Amplifier

-0.3V to V<sup>+</sup> -0.9V

Power consumption



# LMC6442 **Dual Micropower Rail-to-Rail Output Single Supply Operational Amplifier General Description** Features

The LMC6442 is ideal for battery powered systems, where very low supply current (less than one microamp per amplifier) and Rail-to-Rail output swing is required. It is characterized for 2.2V to 10V operation, and at 2.2V supply, the LMC6442 is ideal for single (Li-Ion) or two cell (NiCad or alkaline) battery systems.

The LMC6442 is designed for battery powered systems that require long service life through low supply current, such as smoke and gas detectors, and pager or personal communications systems.

Operation from single supply is enhanced by the wide common mode input voltage range which includes the ground (or negative supply) for ground sensing applications. Very low (5 fA, typical) input bias current and near constant supply current over supply voltage enhance the LMC6442's performance near the end-of-life battery voltage.

Designed for closed loop gains of greater than plus two (or minus one), the amplifier has typically 9.5 KHz GBWP (Gain Bandwidth Product). Unity gain can be used with a simple compensation circuit, which also allows capacitive loads of up to 300 pF to be driven, as described in the Application Notes section.

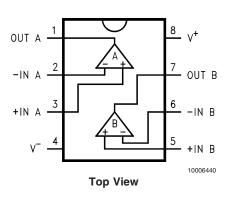
(Typical,  $V_S = 2.2V$ )

- Output Swing to within 30 mV of supply rail
- High voltage gain
- Gain Bandwidth Product
- Guaranteed for:
- Low Supply Current
- Input Voltage Range
- 2.1 µW/Amplifier
- Stable for  $A_V \ge +2$  or  $A_V \le -1$

## Applications

- Portable instruments
- Smoke/gas/CO/fire detectors
- Pagers/cell phones
- Instrumentation
- Thermostats
- Occupancy sensors
- Cameras
- Active badges

# **Connection Diagram**



# Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ESD Tolerance (Note 2)	2 kV
Differential Input Voltage	±Supply Voltages
Voltage at Input/Output Pin	$(V^+)$ + 0.3V, $(V^-)$ – 0.3V
Supply Voltage (V <sup>+</sup> – V <sup>-</sup> ):	16V
Current at Input Pin (Note 10)	±5 mA
Current at Output Pin(Notes 3, 7)	±30 mA
Lead Temp. (soldering 10 sec)	260°C
Storage Temp. Range:	–65°C to +150°C

# 2.2V Electrical Characteristics

Junction Temp. (Note 4)

# Operating Ratings(Note 1)

Supply Voltage	$1.8V \le V_S \le 11V$
Junction Temperature	$-40^{\circ}C < T_{J} < +85^{\circ}C$
Range: LMC6442AI, LMC6442I	
Thermal Resistance $(\theta_{JA})$	
M Package, 8-pin Surface	193°C/W
Mount	
N Package, 8-pin Molded DIP	115°C/W

150°C

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 2.2V$ ,  $V^- = 0V$ ,  $V_{CM} = V_O = V^{+/2}$ , and  $R_L = 1 \text{ M}\Omega$  to  $V^{+/2}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	<b>Typ</b> (Note 5)	LMC6442AI Limit (Note 6)	LMC6442I Limit (Note 6)	Units
DC Electr	ical Characteristics	I				1
V <sub>OS</sub>	Input Offset Voltage		-0.75	±3 <b>±4</b>	±7 <b>±8</b>	mV max
TCV <sub>OS</sub>	Temp. coefficient of input offset voltage		0.4			µV/°C
I <sub>B</sub>	Input Bias Current	(Note 14)	0.005	4	4	pA max
I <sub>os</sub>	Input Offset Current	(Note 14)	0.0025	2	2	pA max
CMRR	Common Mode Rejection Ratio	$-0.1V \le V_{CM} \le 0.5V$	92	67 <b>67</b>	67 <b>67</b>	dB min
C <sub>IN</sub>	Common Mode Input Capacitance		4.7			pF
PSRR	Power Supply Rejection Ratio	V <sub>S</sub> = 2.5 V to 10V	95	75 <b>75</b>	75 <b>75</b>	dB min
V <sub>CM</sub>	V <sub>CM</sub> Input Common-Mode Voltage Range		1.3	1.05 <b>0.95</b>	1.05 <b>0.95</b>	V min
		CMRR ≥ 50 dB	-0.3	-0.2 0	-0.2 0	V max
A <sub>V</sub>	Large Signal Voltage	Sourcing (Note 11)	100			
	Gain	Sinking (Note 11)	94			dB min
		V <sub>O</sub> = 0.22V to 2V	103	80	80	] ''''''
Vo	Output Swing	V <sub>ID</sub> = 100 mV (Note 13)	2.18	2.15 <b>2.15</b>	2.15 <b>2.15</b>	V min
	V <sub>ID</sub> = -100 mV (Note 13)	22	60 <b>60</b>	60 <b>60</b>	mV max	
I <sub>SC</sub>	Output Short Circuit Current	Sourcing, V <sub>ID</sub> = 100 mV (Notes 12, 13)	50	18 17	18 <b>17</b>	μA
		Sinking, V <sub>ID</sub> = -100 mV (Notes 12, 13)	50	20 <b>19</b>	20 <b>19</b>	min
I <sub>S</sub>	Supply Current (2 amplifiers)	R <sub>L</sub> = open	1.90	2.4 <b>3.0</b>	2.6 <b>3.2</b>	μA
		V <sup>+</sup> = 1.8V, R <sub>L</sub> = open	2.10			– max

				(	(	
AC Elect	rical Characteristics					
SR	Slew Rate (Note 8)		2.2			V/ms
GBWP	Gain-Bandwidth Product		9.5			KHz
φ <sub>m</sub>	Phase Margin	(Note 15)	63			deg
Unless o	ectrical Charact	its guaranteed for $T_J = 25^{\circ}C, V^+$	= 5V, V <sup>-</sup> = 0V, V	$V_{\rm CM} = V_{\rm O} = V^{+/2}$	and $R_L = 1 M\Omega$	to V+/2.
Symbol	Parameter	Conditions	<b>Typ</b> (Note 5)	LMC6442AI Limit (Note 6)	LMC6442I Limit (Note 6)	Units
DC Elect	rical Characteristics	J			, , ,	1
V <sub>OS</sub>	Input Offset Voltage		-0.75	±3 <b>±4</b>	±7 <b>±8</b>	mV max
TCV <sub>os</sub>	Temp. coefficient of input offset voltage		0.4			µV/°C
I <sub>B</sub>	Input Bias Current	(Note 14)	0.005	4	4	pA max
I <sub>os</sub>	Input Offset Current	(Note 14)	0.0025	2	2	pA max
CMRR	Common Mode Rejection Ratio	$-0.1V \le V_{CM} \le 3.5V$	102	70 <b>70</b>	70 <b>70</b>	dB min
C <sub>IN</sub>	Common Mode Input Capacitance		4.1			pF
PSRR	Power Supply Rejection Ratio	V <sub>S</sub> = 2.5 V to 10V	95	75 <b>75</b>	75 <b>75</b>	dB min
V <sub>CM</sub>	Input Common-Mode Voltage Range		4.1	3.85 <b>3.75</b>	3.85 <b>3.75</b>	V min
		CMRR ≥ 50 dB	-0.4	-0.2 0	-0.2 <b>0</b>	V max
A <sub>V</sub>	Large Signal Voltage	Sourcing (Note 11)	100			
	Gain	Sinking (Note 11)	94			- dB
		$V_{\rm O} = 0.5 V$ to 4.5V	103	80	80	- min
Vo	Output Swing	V <sub>ID</sub> = 100 mV (Note 13)	4.99	4.95 <b>4.95</b>	4.95 <b>4.95</b>	V min
		V <sub>ID</sub> = -100 mV (Note 13)	20	50 <b>50</b>	50 <b>50</b>	mV max
I <sub>SC</sub>	Output Short Circuit Current	Sourcing, $V_{ID} = 100 \text{ mV}$ (Notes 12, 13)	500	300 200	300 <b>200</b>	μA
		Sinking, V <sub>ID</sub> = -100 mV (Notes 12, 13)	350	200 <b>150</b>	200 <b>150</b>	min
ls	Supply Current (2 amplifiers)	R <sub>L</sub> = open	1.90	2.4 <b>3.0</b>	2.6 <b>3.2</b>	µA max

Parameter

Symbol

**2.2V Electrical Characteristics** (Continued) Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 2.2V$ ,  $V^- = 0V$ ,  $V_{CM} = V_O = V^+/2$ , and  $R_L = 1 \text{ M}\Omega$  to  $V^+/2$ . Boldface limits apply at the temperature extremes.

Conditions

Тур

(Note 5)

LMC6442AI

Limit

(Note 6)

LMC6442I

Limit

(Note 6)

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LMC6442

Units

## 5V Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}$ C, V<sup>+</sup> = 5V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sub>O</sub> = V <sup>+</sup>/2, and R<sub>L</sub> = 1 M $\Omega$  to V<sup>+</sup>/2. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LMC6442AI Limit (Note 6)	LMC6442I Limit (Note 6)	Units
AC Electr	ical Characteristics					
SR	Slew Rate (Note 8)		4.1	2.5	2.5	V/ms
GBWP	Gain-Bandwidth		10			KHz
	Product					
φ <sub>m</sub>	Phase Margin	(Note 15)	64			deg
THD	Total Harmonic	$A_V = +2$ , f = 100 Hz,	0.08			%
	Distortion	$R_L = 10 M\Omega$ , $V_{OUT} = 1 V_{PP}$				

# **10V Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 10V$ ,  $V^- = 0V$ ,  $V_{CM} = V_O = V^+/2$ , and  $R_L = 1 \text{ M}\Omega$  to  $V^+/2$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	<b>Typ</b> (Note 5)	LMC6442AI Limit (Note 6)	LMC6442I Limit (Note 6)	Units
DC Electr	ical Characteristics					
V <sub>os</sub>	Input Offset Voltage		-1.5	±3 ±4	±7 <b>±8</b>	mV max
TCV <sub>OS</sub>	Temp. coefficient of input offset voltage		0.4			µV/°C
I <sub>B</sub>	Input Bias Current	(Note 14)	0.005	4	4	pA max
I <sub>os</sub>	Input Offset Current	(Note 14)	0.0025	2	2	pA max
CMRR	Common Mode Rejection Ratio	$-0.1V \le V_{CM} \le 8.5V$	105	70 <b>70</b>	70 <b>70</b>	dB min
C <sub>IN</sub>	Common Mode Input Capacitance		3.5			pF
PSRR	Power Supply Rejection Ratio	V <sub>S</sub> = 2.5 V to 10V	95	75 <b>75</b>	75 <b>75</b>	dB min
V <sub>CM</sub>	Input Common-Mode Voltage Range	CMRR ≥ 50 dB	9.1	8.85 <b>8.75</b>	8.85 <b>8.75</b>	V min
			-0.4	-0.2 0	-0.2 0	V max
A <sub>V</sub>	Large Signal Voltage	Sourcing (Note 11)	120			dB
	Gain	Sinking (Note 11)	100			min
		$V_{\rm O} = 0.5V$ to 9.5V	104	80	80	
Vo	Output Swing	V <sub>ID</sub> = 100 mV (Note 13)	9.99	9.97 <b>9.97</b>	9.97 <b>9.97</b>	V min
		V <sub>ID</sub> = -100 mV (Note 13)	22	50 <b>50</b>	50 <b>50</b>	mV max
I <sub>sc</sub>	Output Short Circuit Current	Sourcing, V <sub>ID</sub> = 100 mV (Notes 12, 13)	2100	1200 <b>1000</b>	1200 <b>1000</b>	μA
		Sinking, V <sub>ID</sub> = -100 mV (Notes 12, 13)	900	600 <b>500</b>	600 <b>500</b>	min
ls	Supply Current (2 amplifiers)	R <sub>L</sub> = open	1.90	2.4 <b>3.0</b>	2.6 <b>3.2</b>	μA max

### 10V Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 10V$ ,  $V^- = 0V$ ,  $V_{CM} = V_O = V^+/2$ , and  $R_L = 1 M\Omega$  to  $V^+/2$ . Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	<b>Typ</b> (Note 5)	LMC6442AI Limit (Note 6)	LMC6442I Limit (Note 6)	Units
AC Electr	ical Characteristics	·	·			
SR	Slew Rate(Note 8)		4.1	2.5	2.5	V/ms
GBWP	Gain-Bandwidth Product		10.5			KHz
φ <sub>m</sub>	Phase Margin	(Note 15)	68			deg
e <sub>n</sub>	Input-Referred Voltage Noise	R <sub>L</sub> = open f = 10 Hz	170			nV/√Hz
İn	Input-Referred Current Noise	$R_{L} = open$ f = 10 Hz	0.0002			pA/√Hz
	Crosstalk Rejection	(Note 9)	85			dB

# **Electrical Characteristics (continued)**

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics. Note 2: Human body model, 1.5 kΩ in series with 100 pF.

Note 3: Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C. Output currents in excess of ±30 mA over long term may adversely affect reliability.

Note 4: The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(MAX)} - T_A)/|\theta_{JA}|$ . All numbers apply for packages soldered directly into a PC board.

Note 5: Typical Values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis unless otherwise specified.

Note 7: Do not short circuit output to V<sup>+</sup>, when V<sup>+</sup> is greater than 13V or reliability will be adversely affected.

Note 8: Slew rate is the slower of the rising and falling slew rates.

Note 9: Input referred, V<sup>+</sup> = 10V and R<sub>L</sub> = 10 M $\Omega$  connected to 5V. Each amp excited in turn with 1 KHz to produce about 10 V<sub>PP</sub> output.

Note 10: Limiting input pin current is only necessary for input voltages that exceed absolute maximum input voltage ratings.

**Note 11:**  $R_L$  connected to V<sup>+</sup>/2. For Sourcing Test,  $V_O > V^+/2$ . For Sinking tests,  $V_O < V^+/2$ .

Note 12: Output shorted to ground for sourcing, and shorted to V+ for sinking short circuit current test.

Note 13:  $V_{ID}$  is differential input voltage referenced to inverting input.

Note 14: Limits guaranteed by design.

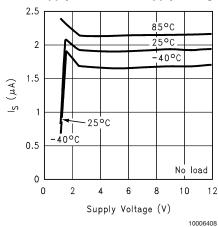
Note 15: See the Typical Performance Characteristics and Application Notes sections for more details.

LMC6442

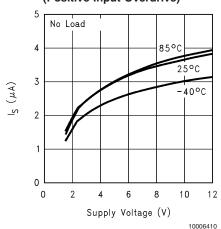
# LMC6442

# **Typical Performance Characteristics** $V_s = 5V$ , Single Supply, $T_A = 25^{\circ}C$ unless otherwise specified

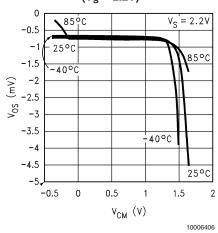
#### Total Supply Current vs. Supply Voltage



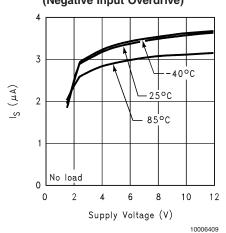
Total Supply Current vs. Supply Voltage (Positive Input Overdrive)



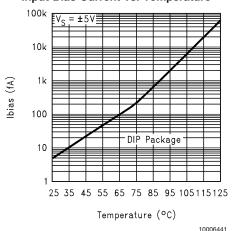
Offset Voltage vs. Common Mode Voltage ( $V_S = 2.2V$ )



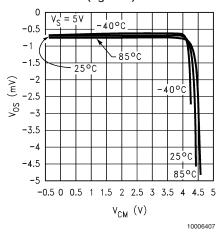
Total Supply Current vs. Supply Voltage (Negative Input Overdrive)



Input Bias Current vs. Temperature

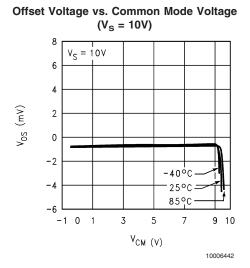


Offset Voltage vs. Common Mode Voltage  $(V_s = 5V)$ 

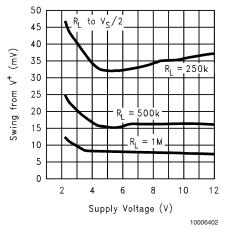


# **Typical Performance Characteristics** $V_s = 5V$ , Single Supply, $T_A = 25^{\circ}C$ unless otherwise

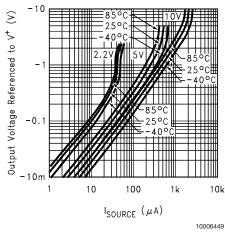
specified (Continued)

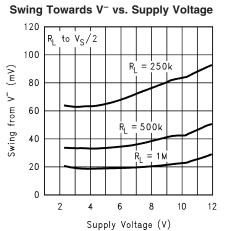


Swing Towards V<sup>+</sup> vs. Supply Voltage



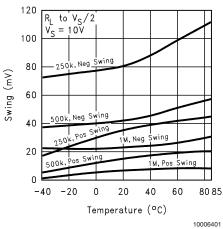
Output Source Current vs. Output Voltage



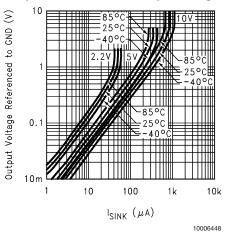


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Swing From Rail(s) vs. Temperature

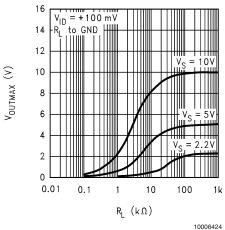


Output Sink Current vs. Output Voltage

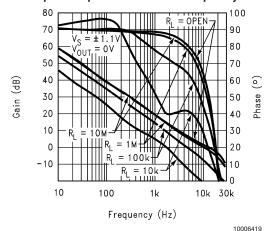


**Typical Performance Characteristics**  $V_s = 5V$ , Single Supply,  $T_A = 25^{\circ}C$  unless otherwise specified (Continued)

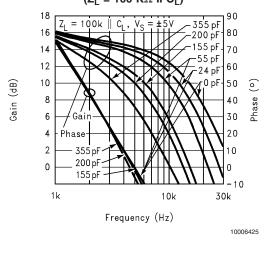
### Maximum Output Voltage vs. Load Resistance



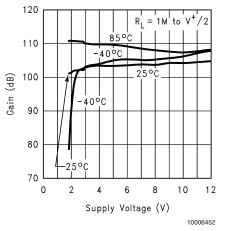
Open Loop Gain/Phase vs. Frequency



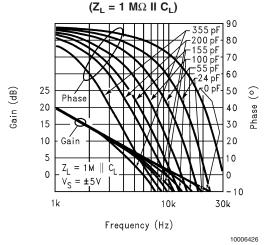
Open Loop Gain/Phase vs. Frequency For Various C<sub>L</sub> (Z<sub>L</sub> = 100 K $\Omega$  II C<sub>L</sub>)



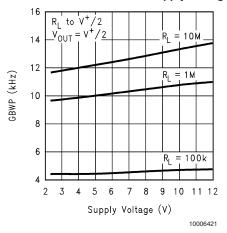
Large Signal Voltage Gain vs. Supply Voltage



Open Loop Gain/Phase vs. Frequency For Various  $C_L$ 

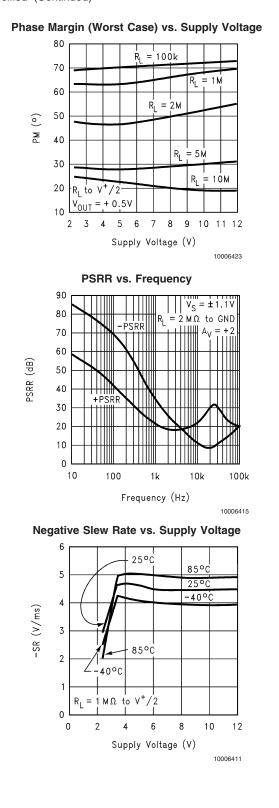


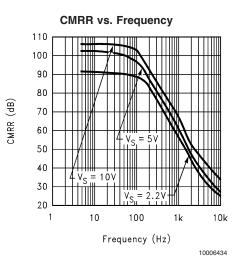
#### Gain Bandwidth Product vs. Supply Voltage



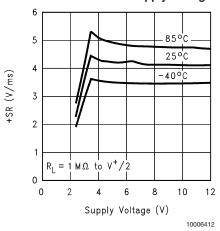
# LMC6442

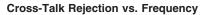
# **Typical Performance Characteristics** $V_s = 5V$ , Single Supply, $T_A = 25^{\circ}C$ unless otherwise specified (Continued)

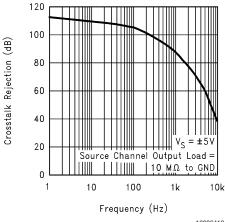




Positive Slew Rate vs. Supply Voltage

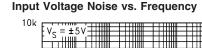


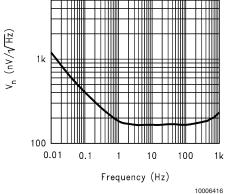


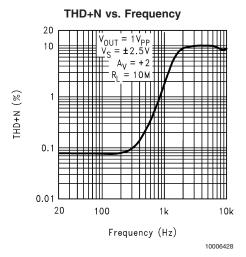




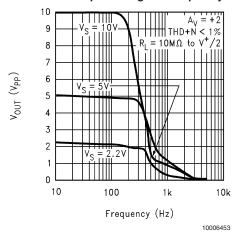
**Typical Performance Characteristics**  $V_s = 5V$ , Single Supply,  $T_A = 25^{\circ}C$  unless otherwise specified (Continued)

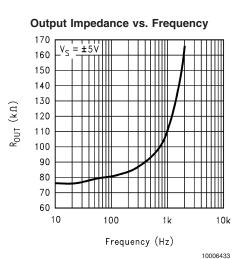




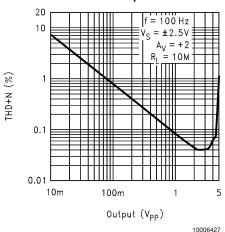


Maximum Output Swing vs. Frequency

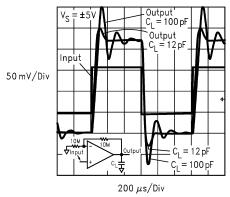




THD+N vs. Amplitude



Small Signal Step Response ( $A_V = +2$ ) ( $C_L = 12 \text{ pF}$ , 100 pF)

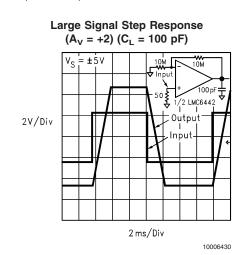


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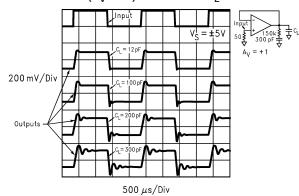


# LMC6442

# **Typical Performance Characteristics** $V_s = 5V$ , Single Supply, $T_A = 25^{\circ}C$ unless otherwise specified (Continued)

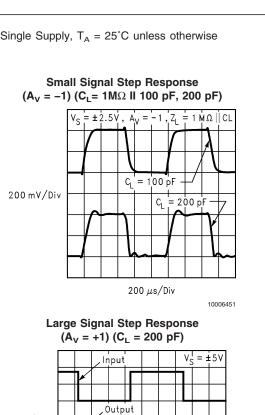






10006431

5V/Div



<u>Inp</u>ut

⊳ ₹ MC64

2 ms/Div

150k ¥ 300 pF ¥ Outpu

200 pF

10006432

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## Applications Information

#### USING LMC6442 IN UNITY GAIN APPLICATIONS

LMC6442 is optimized for maximum bandwidth and minimal external components when operating at a minimum closed loop gain of +2 (or -1). However, it is also possible to operate the device in a unity gain configuration by adding external compensation as shown in *Figure 1*:

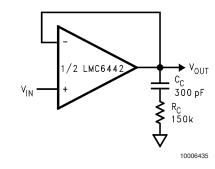


FIGURE 1.  $A_v$  = +1 Operation by adding  $C_c$  and  $R_c$ 

Using this compensation technique it is possible to drive capacitive loads of up to 300 pF without causing oscillations (see the Typical Performance Characteristics for step response plots). This compensation can also be used with other gain settings in order to improve stability, especially when driving capacitive loads (for optimum performance,  $R_C$  and  $C_C$  may need to be adjusted).

#### **USING "T" NETWORK**

Compromises need to be made whenever high gain inverting stages need to achieve a high input impedance as well. This is especially important in low current applications which tend to deal with high resistance values. Using a traditional inverting amplifier, gain is inversely proportional to the resistor value tied between the inverting terminal and input while the input impedance is equal to this value. For example, in order to build an inverting amplifier with an input impedance of 10M $\Omega$  and a gain of 100, one needs to come up with a feedback resistor of 1000 M $\Omega$  -an expensive task.

An alternate solution is to use a "T" Network in the feedback path, as shown in *Figure 2*.

Closed loop gain, A<sub>V</sub> is given by:

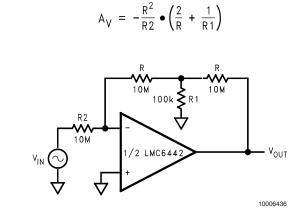


FIGURE 2. "T" Network Used to Replace High Value Resistor

It must be noted, however, that using this scheme, the realizable bandwidth would be less than the theoretical maximum. With feedback factor,  $\beta$ , defined as:

$$\beta \cong \frac{R2}{R2 + R} \bullet \frac{R1}{R1 + R}$$
 for R2 >> R1

#### $\textit{BW}(-\textit{3 dB}) \approx \textit{GBWP} \bullet \beta$

In this case, assuming a GBWP of about 10 KHz, the expected BW would be around 50 Hz (vs. 100 Hz with the conventional inverting amplifier).

Looking at the problem from a different view, with  $R_F$  defined by  $A_V$ •Rin, one could select a value for R in the "T" Network and then determine R1 based on this selection:

$$R1 = \frac{R^2}{R_F - 2R}$$

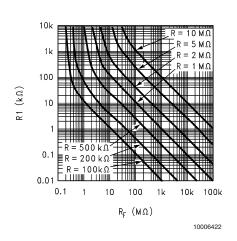


FIGURE 3. "T" Network Values for Various Values of R

For convenience, Figure 3 shows R1 vs.  $\rm R_{F}$  for different values of R.

#### DESIGN CONSIDERATIONS FOR CAPACITIVE LOADS

As with many other opamps, the LMC6442 is more stable at higher closed loop gains when driving a capacitive load. Figure 4 shows minimum closed loop gain versus load capacitance, to achieve less than 10% overshoot in the output small signal response. In addition, the LMC6442 is more stable when it provides more output current to the load and when its output voltage does not swing close to V<sup>-</sup>.

The LMC6442 is more tolerant to capacitive loads when the equivalent output load resistance is lowered or when output voltage is 1V or greater from the V<sup>-</sup> supply. The capacitive load drive capability is also improved by adding an isolating resistor in series with the load and the output of the device. *Figure 5* shows the value of this resistor for various capacitive loads ( $A_V = -1$ ), while limiting the output to less than 10 % overshoot.

Referring to the Typical Performance Characteristics plot of Phase Margin (Worst Case) vs. Supply Voltage, note that Phase Margin increases as the equivalent output load resistance is lowered. This plot shows the expected Phase Margin when the device output is very close to V<sup>-</sup>, which is the least stable condition of operation. Comparing this Phase Margin value to the one read off the Open Loop Gain/Phase vs. Frequency plot, one can predict the improvement in

# Applications Information (Continued)

Phase Margin if the output does not swing close to V<sup>-</sup>. This dependence of Phase Margin on output voltage is minimized as long as the output load,  $R_L$ , is about 1M $\Omega$  or less.

**Output Phase Reversal:** The LMC6442 is immune against this behavior even when the input voltages exceed the common mode voltage range.

**Output Time Delay:** Due to the ultra low power consumption of the device, there could be as long as 2.5 ms of time delay from when power is applied to when the device output reaches its final value.



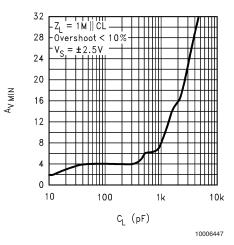


FIGURE 4. Minimum Operating Gain vs. Capacitive Load

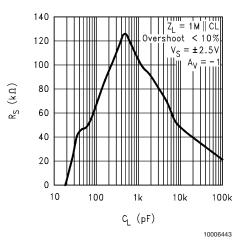
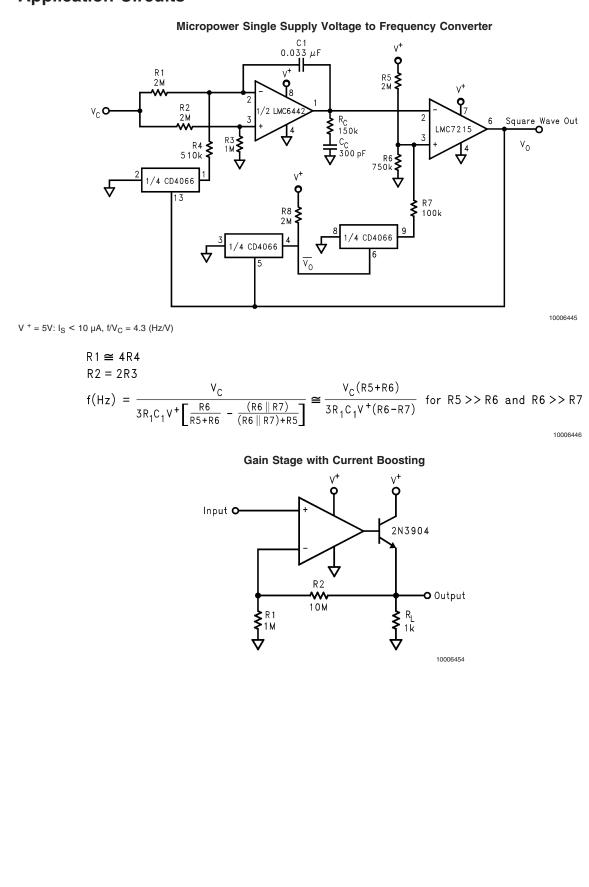
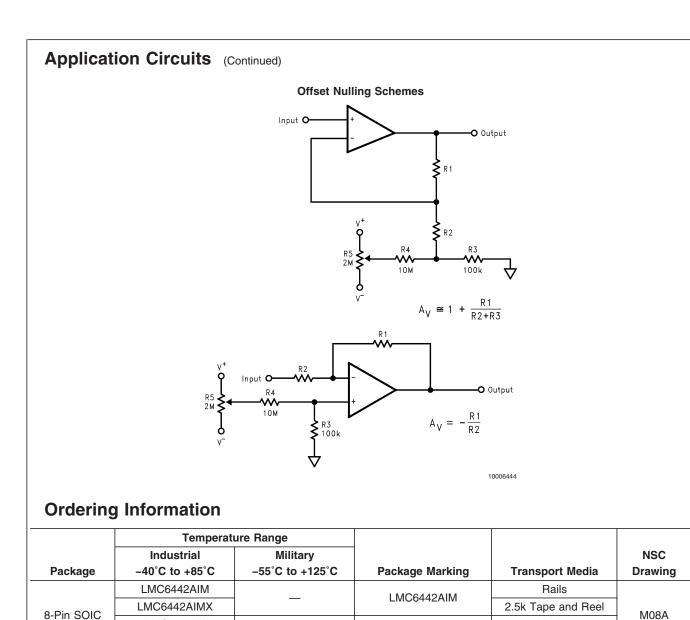


FIGURE 5. Isolating Resistor Value vs Capacitive Load

# **Application Circuits**

LMC6442





N08E

J08A

WG10A

Rails

2.5k Tape and Reel

Rails

Rails

Trays

LMC6442

LMC6442IM

LMC6442IN

LMC6442AMJ-QML

5962-9761301QPA LMC6442AMWG-Q

9761301QXA

LMC6442IM

LMC6442IMX

LMC6442IN

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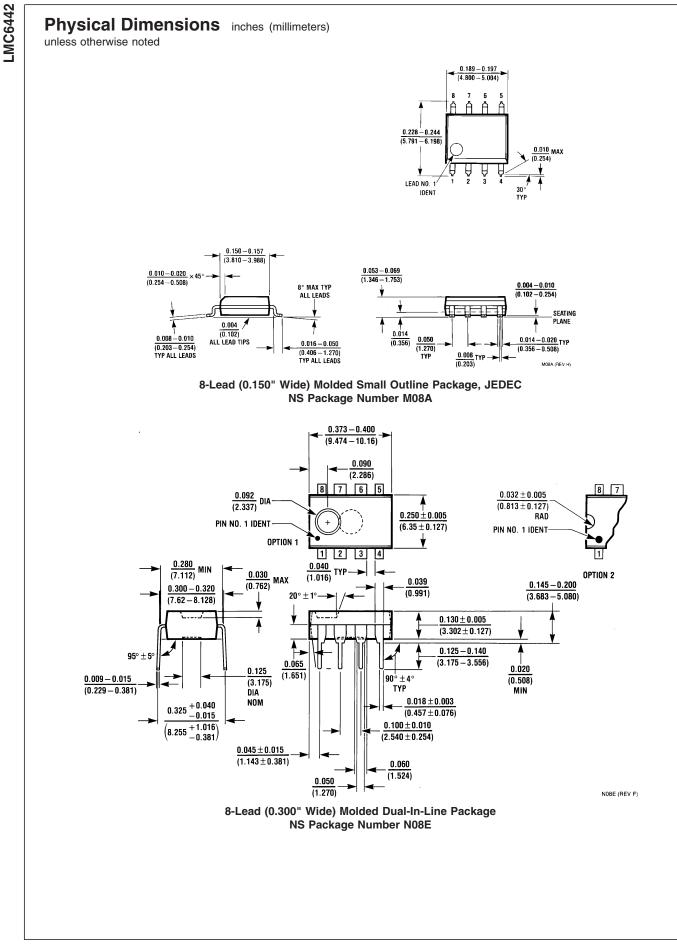
5962-9761301QPA

5962-9761301QXA

8-Pin DIP

8-Pin CDIP

10-Pin SOIC



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