March 12, 2004

FN7183.1

Single 400MHz Fixed Gain Amplifier with **Enable**

élantec.

The EL5196 and the EL5196A are fixed gain amplifiers with a bandwidth of 400MHz, making these amplifiers

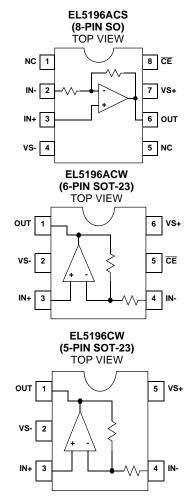
a Sheet

ideal for today's high speed video and monitor applications. These amplifiers feature internal gain setting resistors and can be configured in a gain of +1, -1 or +2. The same bandwidth is seen in both gain-of-1 and gain-of-2 applications.

The EL5196A also incorporates an enable and disable function to reduce the supply current to 100µA typical per amplifier. Allowing the CE pin to float or applying a low logic level will enable the amplifier.

The EL5196 is offered in the 5-pin SOT-23 package and the EL5196A is available in the 6-pin SOT-23 as well as the industry-standard 8-pin SO packages. Both operate over the industrial temperature range of -40°C to +85°C.

Pinouts



Features

- Gain selectable (+1, -1, +2)
- 400MHz -3dB BW (A_V = 1, 2)
- 9mA supply current
- Fast enable/disable (EL5196A only)
- Single and dual supply operation, from 5V to 10V or ±2.5V to ±5V
- Available in SOT-23 packages
- Triple (EL5396) available
- 200MHz, 4mA products available (EL5197 & EL5397)

Applications

- · Video amplifiers
- · Cable drivers
- **RGB** amplifiers
- Test equipment
- Instrumentation
- Current to voltage converters

Ordering Information

PART NUMBER	PACKAGE	TAPE & REEL	PKG. DWG. #	
EL5196CW-T7	5-Pin SOT-23	7" (3K pcs)	MDP0038	
EL5196CW-T7A	5-Pin SOT-23	7" (250 pcs)	MDP0038	
EL5196ACW-T7	6-Pin SOT-23	7"(3K pcs)	MDP0038	
EL5196ACW-T7A	6-Pin SOT-23	7"(250 pcs)	MDP0038	
EL5196ACS	8-Pin SO	-	MDP0027	
EL5196ACS-T7	7 8-Pin SO 7"		MDP0027	
EL5196ACS-T13	8-Pin SO	13"	MDP0027	

EL5196, EL5196A

Absolute Maximum Ratings (T_A = 25° C)

Supply Voltage between V _S + and V _S 11V	Pin Voltages
Maximum Continuous Output Current 50mA	Storage Temperature
Operating Junction Temperature	Operating Ambient Temperature
Power Dissipation See Curves	

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

$\textbf{Electrical Specifications} \qquad \text{V}_{S}\text{+} = \text{+5V}, \text{ V}_{S}\text{-} = \text{-5V}, \text{ R}_{L} = 150\Omega, \text{ T}_{A} = 25^{\circ}\text{ C unless otherwise specified}.$

PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
AC PERFORM	ANCE		<u>'</u>			"
BW	-3dB Bandwidth	A _V = +1		400		MHz
		A _V = -1		400		MHz
		A _V = +2		400		MHz
BW1	0.1dB Bandwidth			35		MHz
SR	Slew Rate	$V_O = -2.5V$ to +2.5V, $A_V = +2$	2400	2900		V/µs
t _S	0.1% Settling Time	V _{OUT} = -2.5V to +2.5V, AV = -1		9		ns
e _N	Input Voltage Noise			3.8		nV/√Hz
i _N -	IN- Input Current Noise			25		pA/√Hz
i _N +	IN+ Input Current Noise			55		pA/√Hz
dG	Differential Gain Error (Note 1)	A _V = +2		0.035		%
dP	Differential Phase Error (Note 2)	A _V = +2		0.04		0
DC PERFORM	ANCE		<u>'</u>		1	
Vos	Offset Voltage		-15	1	15	mV
T _C V _{OS}	Input Offset Voltage Temperature Coefficient	Measured from T _{MIN} to T _{MAX}		5		μV/°C
A _E	Gain Error	$V_O = -3V$ to $+3V$	-2	1.3	2	%
R _F , R _G	Internal R _F and R _G		320	400	480	Ω
INPUT CHARA	CTERISTICS					
CMIR	Common Mode Input Range		±3V	±3.3V		V
+I _{IN}	+ Input Current		-120	40	120	μA
-I _{IN}	- Input Current		-40	4	40	μA
R _{IN}	Input Resistance	at I _N +		27		kΩ
C _{IN}	Input Capacitance			0.5		pF
OUTPUT CHAP	RACTERISTICS					
Vo	Output Voltage Swing	$R_L = 150\Omega$ to GND	±3.4V	±3.7V		V
		$R_L = 1k\Omega$ to GND	±3.8V	±4.0V		V
I _{OUT}	Output Current	$R_L = 10\Omega$ to GND	95	120		mA
SUPPLY			·			
I _{SON}	Supply Current - Enabled	No load, V _{IN} = 0V	8	9	11	mA
I _{SOFF}	Supply Current - Disabled	No load, V _{IN} = 0V		100	150	μA
PSRR	Power Supply Rejection Ratio	DC, $V_S = \pm 4.75V$ to $\pm 5.25V$	55	75		dB

EL5196, EL5196A

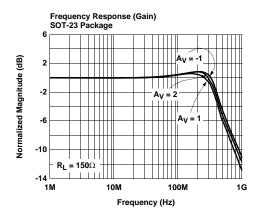
Electrical Specifications V_S + = +5V, V_S - = -5V, R_L = 150 Ω , T_A = 25° C unless otherwise specified. (Continued)

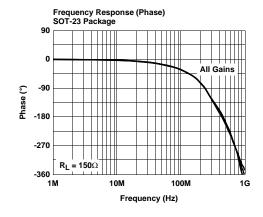
PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
-IPSR	- Input Current Power Supply Rejection	DC, $V_S = \pm 4.75V$ to $\pm 5.25V$	-2		2	μA/V
ENABLE (EL51	96A ONLY)					
t _{EN}	Enable Time			40		ns
t _{DIS}	Disable Time			600		ns
I _{IHCE}	CE Pin Input High Current	CE = V _S +		0.8	6	μΑ
I _{ILCE}	CE Pin Input Low Current	CE = V _S -		0	-0.1	μΑ
V _{IHCE}	CE Input High Voltage for Disable		V _S + - 1			V
V _{ILCE}	CE Input Low Voltage for Enable				V _S + - 3	V

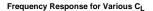
NOTES:

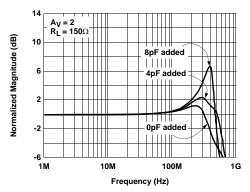
- 1. Standard NTSC test, AC signal amplitude = $286mV_{P-P}$, f = 3.58MHz.
- 2. Measured from the application of the $\overline{\text{CE}}$ logic signal until the output voltage is at the 50% point between initial and final values.

Typical Performance Curves

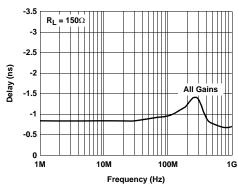




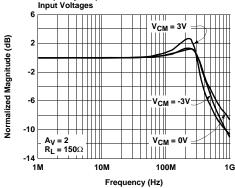




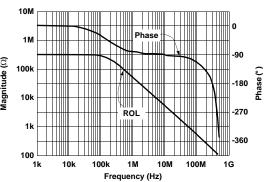


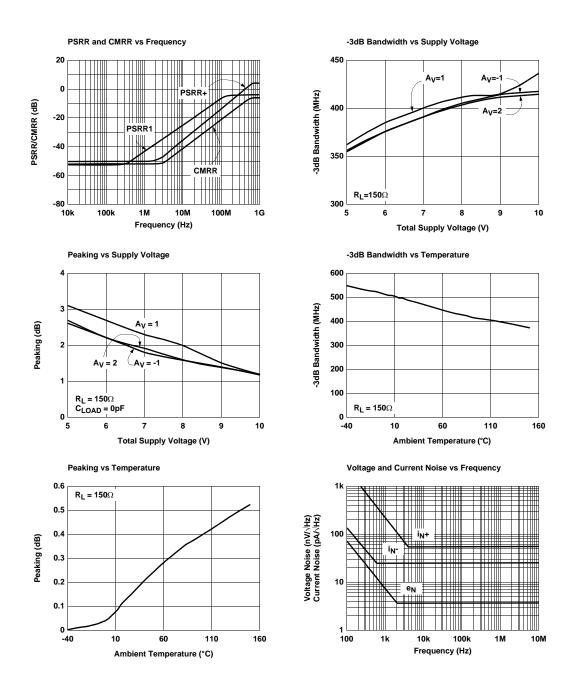


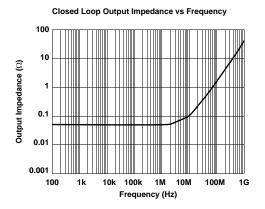
Frequency Response for Various Common-Mode Input Voltages

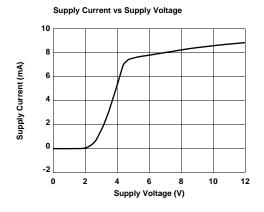


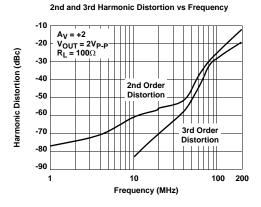
Transimpedance (ROL) vs Frequency

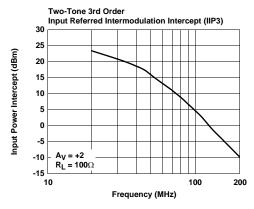


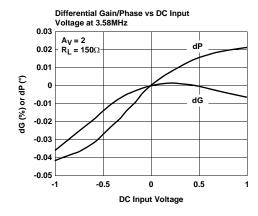


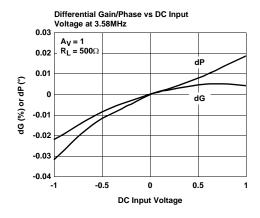


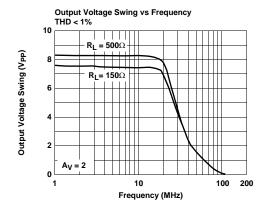


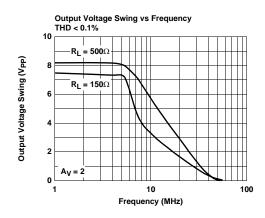


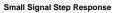


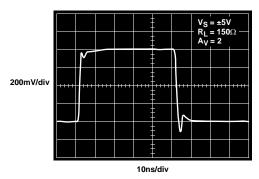




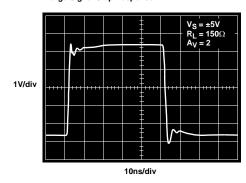




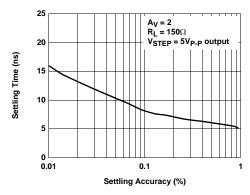




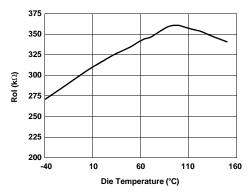


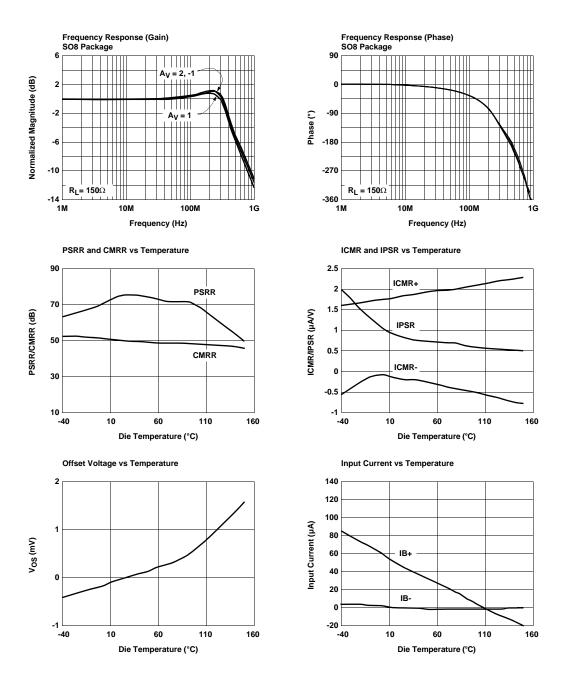


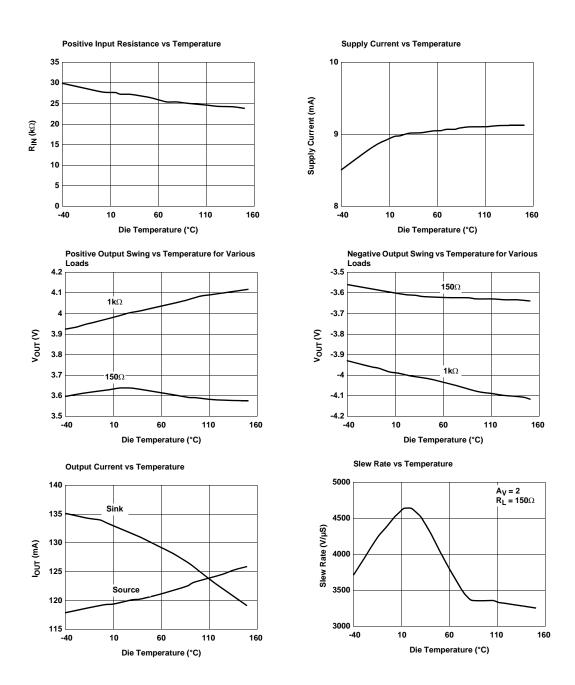
Settling Time vs Settling Accuracy

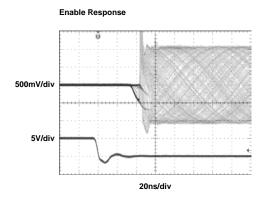


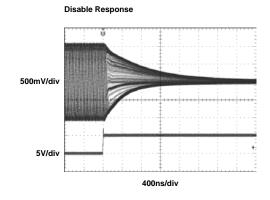
Transimpedance (Rol) vs Temperature

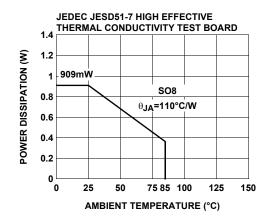


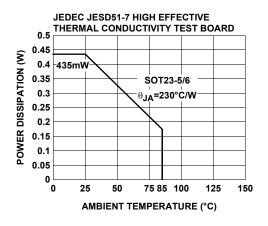


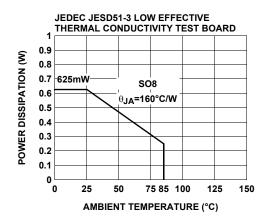


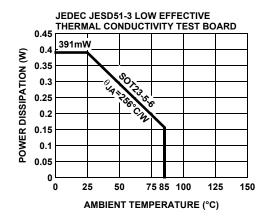












Pin Descriptions

8-PIN SO	5-PIN SOT-23	6-PIN SOT-23	PIN NAME	FUNCTION	EQUIVALENT CIRCUIT
1, 5			NC	Not connected	
2	4	4	IN-	Inverting input	IN+ R _F IN-
3	3	3	IN+	Non-inverting input	(See circuit 1)
4	2	2	V _S -	Negative supply	
6	1	1	OUT	Output	Circuit 2
7	5	6	V _S +	Positive supply	
8		5	ĈĒ	Chip enable	Circuit 3

Applications Information

Product Description

The EL5196 is a fixed gain amplifier that offers a wide -3dB bandwidth of 400MHz and a low supply current of 9mA per amplifier. The EL5196 works with supply voltages ranging from a single 5V to 10V and they are also capable of swinging to within 1V of either supply on the output. This combination of high bandwidth and low power, together with aggressive pricing make the EL5196 the ideal choice for many low-power/high-bandwidth applications such as portable, handheld, or battery-powered equipment.

For varying bandwidth and higher gains, consider the EL5191 with 1GHz on a 9mA supply current or the EL5193 with 300MHz on a 4mA supply current. Versions include single, dual, and triple amp packages with 5-pin SOT-23, 16-pin QSOP, and 8-pin or 16-pin SO outlines.

Power Supply Bypassing and Printed Circuit Board Layout

As with any high frequency device, good printed circuit board layout is necessary for optimum performance. Low impedance ground plane construction is essential. Surface mount components are recommended, but if leaded components are used, lead lengths should be as short as possible. The power supply pins must be well bypassed to reduce the risk of oscillation. The combination of a $4.7\mu F$ tantalum capacitor in parallel with a $0.01\mu F$ capacitor has been shown to work well when placed at each supply pin.

Disable/Power-Down

The EL5196A amplifier can be disabled placing its output in a high impedance state. When disabled, the amplifier supply current is reduced to < 150µA. The EL5196A is disabled when its $\overline{\text{CE}}$ pin is pulled up to within 1V of the positive supply. Similarly, the amplifier is enabled by floating or pulling its $\overline{\text{CE}}$ pin to at least 3V below the positive supply. For $\pm 5\text{V}$ supply, this means that an EL5196A amplifier will be enabled when $\overline{\text{CE}}$ is 2V or less, and disabled when $\overline{\text{CE}}$ is above 4V. Although the logic levels are not standard TTL, this choice of logic voltages allows the EL5196A to be enabled by tying $\overline{\text{CE}}$ to ground, even in 5V single supply applications. The $\overline{\text{CE}}$ pin can be driven from CMOS outputs.

Gain Setting

The EL5196A is built with internal feedback and gain resistors. The internal feedback resistors have equal value; as a result, the amplifier can be configured into gain of +1, -1, and +2 without any external resistors. Figure 1 shows the amplifier in gain of +2 configuration. The gain error is ±2% maximum. Figure 2 shows the amplifier in gain of -1 configuration. For gain of +1, IN+ and IN- should be connected together as shown in Figure 3. This configuration avoids the effects of any parasitic capacitance on the IN- pin. Since the internal feedback and gain resistors change with

temperature and process, external resistor should not be used to adjust the gain settings.

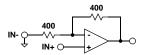


FIGURE 1. $A_V = +2$

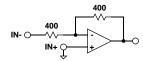


FIGURE 2. $A_V = -1$

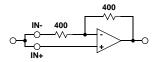


FIGURE 3. $A_V = +1$

Supply Voltage Range and Single-Supply Operation

The EL5196 has been designed to operate with supply voltages having a span of greater than or equal to 5V and less than 11V. In practical terms, this means that the EL5196 will operate on dual supplies ranging from ±2.5V to ±5V. With single-supply, the EL5196 will operate from 5V to 10V.

As supply voltages continue to decrease, it becomes necessary to provide input and output voltage ranges that can get as close as possible to the supply voltages. The EL5196 has an input range which extends to within 2V of either supply. So, for example, on ±5V supplies, the EL5196 has an input range which spans ±3V. The output range of the EL5196 is also quite large, extending to within 1V of the supply rail. On a ±5V supply, the output is therefore capable of swinging from -4V to +4V. Single-supply output range is larger because of the increased negative swing due to the

external pull-down resistor to ground. Figure 4 shows an AC-coupled, gain of +2, +5V single supply circuit configuration.

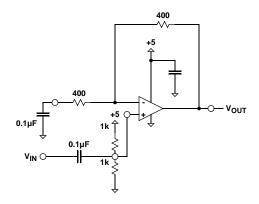


FIGURE 4.

Video Performance

For good video performance, an amplifier is required to maintain the same output impedance and the same frequency response as DC levels are changed at the output. This is especially difficult when driving a standard video load of 150Ω , because of the change in output current with DC level. Previously, good differential gain could only be achieved by running high idle currents through the output transistors (to reduce variations in output impedance.) These currents were typically comparable to the entire 9mA supply current of each EL5196 amplifier. Special circuitry has been incorporated in the EL5196 to reduce the variation of output impedance with current output. This results in dG and dP specifications of 0.0035% and 0.04°, while driving 150Ω at a gain of 2.

Video performance has also been measured with a 500Ω load at a gain of +1. Under these conditions, the EL5196 has dG and dP specifications of 0.03% and 0.05°, re spectively.

Output Drive Capability

In spite of its low 9mA of supply current, the EL5196 is capable of providing a minimum of ± 95 mA of output current with a minimum of ± 95 mA of output drive.

Driving Cables and Capacitive Loads

When used as a cable driver, double termination is always recommended for reflection-free performance. For those applications, the back-termination series resistor will decouple the EL5196 from the cable and allow extensive

capacitive drive. However, other applications may have high capacitive loads without a back-termination resistor. In these applications, a small series resistor (usually between 5Ω and 50Ω) can be placed in series with the output to eliminate most peaking.

Current Limiting

The EL5196 has no internal current-limiting circuitry. If the output is shorted, it is possible to exceed the Absolute Maximum Rating for output current or power dissipation, potentially resulting in the destruction of the device.

Power Dissipation

With the high output drive capability of the EL5196, it is possible to exceed the 125°C Absolu te Maximum junction temperature under certain very high load current conditions. Generally speaking when R_L falls below about $25\Omega,$ it is important to calculate the maximum junction temperature $(T_{\mbox{\scriptsize JMAX}})$ for the application to determine if power supply voltages, load conditions, or package type need to be modified for the EL5196 to remain in the safe operating area. These parameters are calculated as follows:

$$T_{JMAX} = T_{MAX} + (\theta_{JA} \times n \times PD_{MAX})$$

where:

T_{MAX} = Maximum ambient temperature

 θ_{JA} = Thermal resistance of the package

n = Number of amplifiers in the package

PD_{MAX} = Maximum power dissipation of each amplifier in the package

PD_{MAX} for each amplifier can be calculated as follows:

$$PD_{MAX} = (2 \times V_{S} \times I_{SMAX}) + \left[(V_{S} - V_{OUTMAX}) \times \frac{V_{OUTMAX}}{R_{L}} \right]$$

where:

 $V_S = Supply voltage$

I_{SMAX} = Maximum supply current of 1A

V_{OUTMAX} = Maximum output voltage (required)

R_L = Load resistance

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