

# $I^2C^{\otimes}$ -Compatible, Wide Bandwidth, Triple 2 × 2 Crosspoint Switch

# ADG799A/ADG799G

## **FEATURES**

Bandwidth: 230 MHz

Low insertion loss and on resistance: 2.6  $\Omega$  typical

On resistance flatness: 0.3  $\Omega$  typical Single 3 V/5 V supply operation

3.3 V analog signal range (5 V supply, 75  $\Omega$  load) Low quiescent supply current: 1 nA typical Fast switching times:  $t_{ON} = 184$  ns,  $t_{OFF} = 180$  ns

I<sup>2</sup>C-compatible interface Compact 24-lead LFCSP

Two I<sup>2</sup>C-controllable logic outputs (ADG799G only)

**ESD** protection

4 kV human body model 200 V machine model

1 kV field-induced charged device model

## **APPLICATIONS**

RGB/YPbPr video switches HDTV Projection TV DVD-R/RW AV receivers

## **GENERAL DESCRIPTION**

The ADG799A/ADG799G are monolithic CMOS devices comprising three  $2\times 2$  crosspoint switches controllable via a standard I²C serial interface. The CMOS process provides ultralow power dissipation, yet offers high switching speed and low on resistance.

The on resistance profile is very flat over the full analog input range and wide bandwidth ensures excellent linearity and low distortion. These features, combined with a wide input signal range, make the ADG799A/ADG799G the ideal switching solution for a wide range of TV applications including RGB and YPbPr video switches for picture-in picture applications.

The switches conduct equally well in both directions when on. In the off condition, signal levels up to the supplies are blocked. The ADG799A/ADG799G switches exhibit break-before-make switching action. The ADG799G has two general-purpose logic output pins controlled by the I<sup>2</sup>C interface that can also be used to control other non-I<sup>2</sup>C-compatible devices such as video filters.

The integrated I<sup>2</sup>C interface provides a large degree of flexibility in the system design. It has three user-adjustable I<sup>2</sup>C address

#### Rev. 0

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## **FUNCTIONAL BLOCK DIAGRAMS**

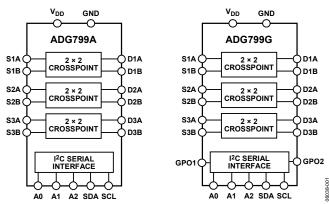


Figure 1. ADG799A and ADG799G

pins that allow up to eight devices on the same bus. This allows the user to expand the capability of the device by increasing the size of the switching array.

The ADG799A/ADG799G operate from single 3 V or 5 V supply voltages and are available in a compact, 4 mm  $\times$  4 mm body, 24-lead, Pb-free LFCSP.

# **PRODUCT HIGHLIGHTS**

- Wide bandwidth: 230 MHz.
- 2. Ultralow power dissipation.
- 3. Extended input signal range.
- 4. Integrated I<sup>2</sup>C serial interface.
- 5. Compact 4 mm × 4 mm, 24-lead, Pb-free LFCSP.
- 6. ESD protection tested as per ESD Association standards:
  - 4 kV HBM (ANSI/ESD STM5.1-2001)
  - 200 V MM (ANSI/ESD STM5.2-1999)
  - 1 kV FICDM (ANSI/ESD STM5.3.1-1999)

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# **REVISION HISTORY**

7/06—Revision 0: Initial Version

# **SPECIFICATIONS**

 $V_{\text{DD}}$  = 5 V  $\pm$  10%, GND = 0 V,  $T_{\text{A}}$  =  $-40^{\circ}\text{C}$  to +85°C, unless otherwise noted.

Table 1.

Parameter	Conditions	Min	Typ <sup>1</sup>	Max	Unit
ANALOG SWITCH					
Analog Signal Range <sup>2</sup>	$V_S = V_{DD}$ , $R_L = 1 M\Omega$	0		4	V
	$V_S = V_{DD}$ , $R_L = 75 \Omega$	0		3.3	V
On Resistance, Ron	$V_D = 0 \text{ V}, I_{DS} = -10 \text{ mA}, \text{ see Figure 22}$		2.6	5	Ω
	$V_D = 0 \text{ V to } 1 \text{ V, } I_{DS} = -10 \text{ mA, see Figure } 22$			5.5	Ω
On Resistance Matching Between Channels, ΔR <sub>ON</sub>	$V_D = 0 \text{ V, } I_{DS} = -10 \text{ mA}$		0.15	1.85	Ω
5	$V_D = 1 \text{ V, } I_{DS} = -10 \text{ mA}$			1.85	Ω
On Resistance Flatness, R <sub>FLAT(ON)</sub>	$V_D = 0 \text{ V to } 1 \text{ V, } I_{DS} = -10 \text{ mA}$		0.3	0.55	Ω
LEAKAGE CURRENTS					
Source Off Leakage (I <sub>S(OFF)</sub> )	$V_D = 4 \text{ V}/1 \text{ V}, V_S = 1 \text{ V}/4 \text{ V}, \text{ see Figure 23}$		±0.25		nA
Drain Off Leakage (I <sub>D(OFF)</sub> )	$V_D = 4 \text{ V/1 V}, V_S = 1 \text{ V/4 V}, \text{ see Figure 23}$		±0.25		nA
Channel On Leakage (I <sub>D(ON)</sub> , I <sub>S(ON)</sub> )	$V_D = V_S = 4 \text{ V/1 V, see Figure 24}$		±0.25		nA
DYNAMIC CHARACTERISTICS <sup>3</sup>	15 15 1.1,1.1,5001.1ga.10 2.1				
ton, tenable	$C_L = 35 \text{ pF, } R_L = 50 \Omega, V_S = 2 \text{ V, see Figure 28}$		184	240	ns
tor, tenable	$C_L = 35 \text{ pF}, R_L = 50 \Omega, V_S = 2 \text{ V}, \text{ see Figure 28}$ $C_L = 35 \text{ pF}, R_L = 50 \Omega, V_S = 2 \text{ V}, \text{ see Figure 28}$		180	235	ns
Break-Before-Make Time Delay, t <sub>D</sub>	$C_L = 35 \text{ pF}, R_L = 50 \Omega, V_{S1} = V_{S2} = 2 \text{ V},$ $C_L = 35 \text{ pF}, R_L = 50 \Omega, V_{S1} = V_{S2} = 2 \text{ V},$	1	3	233	ns
break before make fille beldy, to	$CL = 33 \text{ pr, } NL = 30 \Omega, VS1 = VS2 = 2 V,$ see Figure 29	'	,		113
I <sup>2</sup> C to GPO Propagation Delay, t <sub>H</sub> , t <sub>L</sub> (ADG799G only)				130	ns
Off Isolation	$f = 10 \text{ MHz}, R_L = 50 \Omega$ , see Figure 26		-60		dB
Channel-to-Channel Crosstalk	$f = 10 \text{ MHz}, R_L = 50 \Omega$ , see Figure 27		00		ab
Same Crosspoint Switch	7 = 10 Will2, N <sub>L</sub> = 30 12, 300 Figure 27		-50		dB
Different Crosspoint Switch			-80		dB
-3 dB Bandwidth	$R_L = 50 \Omega$ , see Figure 15		230		MHz
THD + N	$R_L = 300\Omega$ , see Figure 13		0.14		%
	$C_L = 1 \text{ nF, V}_S = 0 \text{ V, see Figure 30}$		4		
Charge Injection	CL = 1 fir, vs = 0 v, see Figure 30		13		pC pF
C <sub>S</sub> (OFF)					
C <sub>D</sub> (OFF)			17 25		pF
C <sub>D(ON)</sub> , C <sub>S(ON)</sub>	6 20111		35		pF
Power Supply Rejection Ratio, PSRR	f = 20 kHz		70		dB
Differential Gain Error	CCIR330 test signal		0.56		%
Differential Phase Error	CCIR330 test signal		0.79		Degree
LOGIC INPUTS <sup>3</sup>					
A0, A1, A2 Pins					
Input High Voltage, V <sub>INH</sub>		2.0			V
Input Low Voltage, V <sub>INL</sub>				0.8	٧
Input Current, I <sub>INL</sub> or I <sub>INH</sub>	$V_{IN} = 0 V \text{ to } V_{DD}$		0.005	±1	μΑ
Input Capacitance, C <sub>IN</sub>			3		рF
SCL, SDA Pins					
Input High Voltage, V <sub>INH</sub>		$0.7 \times V_{DD}$		$V_{DD} + 0.3$	V
Input Low Voltage, V <sub>INL</sub>		-0.3		$+0.3 \times V_{DD}$	V
Input Leakage Current, I <sub>IN</sub>	$V_{IN} = 0 V \text{ to } V_{DD}$		+0.005	±1	μΑ
Input Hysteresis			$0.05 \times V_{DD}$		V
Input Capacitance, C <sub>IN</sub>			3		pF
LOGIC OUTPUTS					
SDA Pin <sup>3</sup>					
Output Low Voltage, Vol	I <sub>SINK</sub> = 3 mA			0.4	٧
	I <sub>SINK</sub> = 6 mA			0.6	٧
Floating State Leakage Current				±1	μΑ
Floating State Output Capacitance				10	pF

Parameter	Conditions	Min	Typ <sup>1</sup>	Max	Unit
GPO1 and GPO2 Pins					
Output Low Voltage, Vol	$I_{LOAD} = 2 \text{ mA}$			0.4	V
Output High Voltage, V <sub>OH</sub>	$I_{LOAD} = -2 \text{ mA}$	2.0			V
POWER REQUIREMENTS					
$I_{DD}$	Digital inputs = $0 \text{ V}$ or $V_{DD}$ , $I^2C$ interface inactive		0.001	1	μΑ
	$I^2C$ interface active, $f_{SCL} = 400 \text{ kHz}$			0.2	mA
	$I^2C$ interface active, $f_{SCL} = 3.4$ MHz			0.7	mA

 $<sup>^1</sup>$  All typical values are at  $T_A = +25\,^{\circ}\text{C}$ , unless otherwise stated.  $^2$  Guaranteed by initial characterization, not subject to production test.  $^3$  Guaranteed by design, not subject to production test.

 $V_{DD}$  = 3 V  $\pm$  10%, GND = 0 V,  $T_{\rm A}$  =  $-40^{\rm o}C$  to +85°C, unless otherwise noted.

Table 2.

Parameter	Conditions	Min	Typ <sup>1</sup>	Max	Unit
ANALOG SWITCH					
Analog Signal Range <sup>2</sup>	$V_S = V_{DD}$ , $R_L = 1 M\Omega$	0		2.2	٧
	$V_S = V_{DD}$ , $R_L = 75 \Omega$	0		1.7	V
On Resistance, R <sub>ON</sub>	$V_D = 0 \text{ V}$ , $I_{DS} = -10 \text{ mA}$ , see Figure 22		3	5.5	Ω
	$V_D = 0 \text{ V to } 1 \text{ V}, I_{DS} = -10 \text{ mA}, \text{ see Figure } 22$			8	Ω
On Resistance Matching Between Channels, $\Delta R_{ON}$	$V_D = 0 \text{ V}, I_{DS} = -10 \text{ mA}$		0.15	1.8	Ω
	$V_D = 1 \text{ V, } I_{DS} = -10 \text{ mA}$			2.1	Ω
On Resistance Flatness, R <sub>FLAT(ON)</sub>	$V_D = 0 \text{ V to } 1 \text{ V, } I_{DS} = -10 \text{ mA}$		0.3	2.8	Ω
LEAKAGE CURRENTS					
Source Off Leakage (I <sub>S(OFF)</sub> )	$V_D = 2 V/1 V$ , $V_S = 1 V/2 V$ , see Figure 23		±0.25		nA
Drain Off Leakage (ID(OFF))	$V_D = 2 V/1 V$ , $V_S = 1 V/2 V$ , see Figure 23		±0.25		nA
Channel On Leakage (I <sub>D(ON)</sub> , I <sub>S(ON)</sub> )	$V_D = V_S = 2 V/1 V$ , see Figure 24		±0.25		nA
DYNAMIC CHARACTERISTICS <sup>3</sup>					
t <sub>on</sub> , t <sub>enable</sub>	$C_L = 35 \text{ pF, } R_L = 50 \Omega, V_S = 2 \text{ V, see Figure 28}$		203	266	ns
toff, tdisable	$C_L = 35 \text{ pF, } R_L = 50 \Omega, V_S = 2 \text{ V, see Figure 28}$		200	260	ns
Break-Before-Make Time Delay, t <sub>D</sub>	$C_L = 35 \text{ pF, } R_L = 50 \Omega, V_{S1} = V_{S2} = 2 \text{ V,}$ see Figure 29	1	3		ns
I <sup>2</sup> C to GPO Propagation Delay, t <sub>H</sub> , t <sub>L</sub> (ADG799G only)				121	ns
Off Isolation	$f = 10 \text{ MHz}$ , $R_L = 50 \Omega$ , see Figure 26		-60		dB
Channel-to-Channel Crosstalk	$f = 10 \text{ MHz}$ , $R_L = 50 \Omega$ , see Figure 27				
Same Crosspoint Switch	_		-50		dB
Different Crosspoint Switch			-80		dB
–3 dB Bandwidth	$R_L = 50 \Omega$ , see Figure 15		210		MHz
THD + N	$R_L = 100 \Omega$		0.14		%
Charge Injection	$C_L = 1 \text{ nF, } V_S = 0 \text{ V, see Figure 30}$		2		рС
C <sub>S(OFF)</sub>			13		pF
C <sub>D(OFF)</sub>			17		pF
C <sub>D(ON)</sub> , C <sub>S(ON)</sub>			35		pF
Power Supply Rejection Ratio, PSRR	f = 20 kHz		70		dB
Differential Gain Error	CCIR330 test signal		0.66		%
Differential Phase Error	CCIR330 test signal		1		Degre
LOGIC INPUTS					
A0, A1, A2 Pins <sup>3</sup>					
Input High Voltage, V <sub>INH</sub>		2.0			٧
Input Low Voltage, V <sub>INL</sub>				0.8	٧
Input Current, linL or linh	$V_{IN} = 0 V to V_{DD}$		+0.005	±1	μΑ
Input Capacitance, C <sub>IN</sub>			3		pF
SCL, SDA Pins <sup>3</sup>					
Input High Voltage, V <sub>INH</sub>		$0.7 \times V_{DD}$		$V_{DD} + 0.3$	V
Input Low Voltage, VINL		-0.3		$+0.3 \times V_{DD}$	v
Input Leakage Current, I <sub>IN</sub>	$V_{IN} = 0 V \text{ to } V_{DD}$	3.3	0.005	±1	μA
Input Hysteresis			$0.05 \times V_{DD}$		V
Input Capacitance, C <sub>IN</sub>			3		pF
LOGIC OUTPUTS <sup>3</sup>					۲۰
SDA Pin					
	I <sub>SINK</sub> = 3 mA			0.4	V
( )LITOLIT LOW VOITAGE Voi		I		J. <del>T</del>	
Output Low Voltage, Vol	Icinic = 6 mA			0.6	W
Output Low Voltage, Vol	I <sub>SINK</sub> = 6 mA			0.6 ±1	V μA

Parameter	Conditions	Min	Typ <sup>1</sup>	Max	Unit
GPO1 and GPO2 Pins					
Output Low Voltage, Vol	$I_{LOAD} = 2 \text{ mA}$			0.4	V
Output High Voltage, V <sub>OH</sub>	$I_{LOAD} = -2 \text{ mA}$	2.0			V
POWER REQUIREMENTS					
$I_{DD}$	Digital inputs = $0 \text{ V or V}_{DD}$ , $I^2C$ interface inactive		0.001	1	μΑ
	$I^2C$ interface active, $f_{SCL} = 400 \text{ kHz}$			0.1	mA
	$I^2C$ interface active, $f_{SCL} = 3.4$ MHz			0.2	mA

 $<sup>^1</sup>$  All typical values are at  $T_A = +25\,^{\circ}\text{C}$ , unless otherwise stated.  $^2$  Guaranteed by initial characterization, not subject to production test.  $^3$  Guaranteed by design, not subject to production test.

# **I<sup>2</sup>C TIMING SPECIFICATIONS**

 $V_{DD} = 2.7 \text{ V}$  to 5.5 V; GND = 0 V;  $T_A = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ , unless otherwise noted (see Figure 2 for timing diagram).

Table 3.

Parameter <sup>1</sup>	Conditions	Min	Max	Unit	Description
$f_{SCL}$	Standard mode		100	kHz	Serial clock frequency
	Fast mode		400	kHz	
	High speed mode				
	$C_B = 100 \text{ pF max}$		3.4	MHz	
	$C_B = 400 \text{ pF max}$		1.7	MHz	
t <sub>1</sub>	Standard mode	4		μs	t <sub>нібн</sub> , SCL high time
	Fast mode	0.6		μs	
	High speed mode				
	$C_B = 100 \text{ pF max}$	60		ns	
	$C_B = 400 \text{ pF max}$	120		ns	
t <sub>2</sub>	Standard mode	4.7		μs	t <sub>LOW</sub> , SCL low time
	Fast mode	1.3		μs	
	High speed mode				
	$C_B = 100 \text{ pF max}$	160		ns	
	$C_B = 400 \text{ pF max}$	320		ns	
t <sub>3</sub>	Standard mode	250		ns	t <sub>SU;DAT</sub> , data setup time
	Fast mode	100		ns	
	High speed mode	10		ns	
t <sub>4</sub> <sup>2</sup>	Standard mode	0	3.45	μs	t <sub>HD;DAT</sub> , data hold time
	Fast mode	0	0.9	μs	
	High speed mode				
	$C_B = 100 \text{ pF max}$	0	703	ns	
	$C_B = 400 \text{ pF max}$	0	150	ns	
<b>t</b> <sub>5</sub>	Standard mode	4.7		μs	t <sub>SU;STA</sub> , setup time for a repeated start condition
	Fast mode	0.6		μs	
	High speed mode	160		ns	
t <sub>6</sub>	Standard mode	4		μs	t <sub>HD;STA</sub> , hold time (repeated) start condition
	Fast mode	0.6		μs	
	High speed mode	160		ns	
t <sub>7</sub>	Standard mode	4.7		μs	t <sub>BUF</sub> , bus free time between a stop and a start condition
	Fast mode	1.3		μs	
t <sub>8</sub>	Standard mode	4		μs	t <sub>SU;STO</sub> , setup time for stop condition
	Fast mode	0.6		μs	
	High speed mode	160		ns	
<b>t</b> <sub>9</sub>	Standard mode		1000	ns	t <sub>RDA</sub> , rise time of SDA signal
	Fast mode	20 + 0.1 C <sub>B</sub>	300	ns	
	High speed mode				
	$C_B = 100 \text{ pF max}$	10	80	ns	
	$C_B = 400 \text{ pF max}$	20	160	ns	
t <sub>10</sub>	Standard mode		300	ns	t <sub>FDA</sub> , fall time of SDA signal
	Fast mode	20 + 0.1 C <sub>B</sub>	300	ns	
	High speed mode				
	$C_B = 100 \text{ pF max}$	10	80	ns	
	$C_B = 400 \text{ pF max}$	20	160	ns	
t <sub>11</sub>	Standard mode		1000	ns	t <sub>RCL</sub> , rise time of SCL signal
	Fast mode	20 + 0.1 C <sub>B</sub>	300	ns	
	High speed mode				
	$C_B = 100 \text{ pF max}$	10	40	ns	
	$C_B = 400 \text{ pF max}$	20	80	ns	

Parameter <sup>1</sup>	Conditions	Min	Max	Unit	Description
t <sub>11A</sub>	Standard mode		1000	ns	t <sub>RCL1</sub> , rise time of SCL signal after a repeated start condition and after
	Fast mode	20 + 0.1 C <sub>B</sub>	300	ns	an acknowledge bit.
	High speed mode				
	$C_B = 100 pF max$	10	80	ns	
	$C_B = 400 \text{ pF max}$	20	160	ns	
t <sub>12</sub>	Standard mode		300	ns	t <sub>FCL</sub> , fall time of SCL signal
	Fast mode	20 + 0.1 C <sub>B</sub>	300	ns	
	High speed mode				
	$C_B = 100 \text{ pF max}$	10	40	ns	
	$C_B = 400 \text{ pF max}$	20	80	ns	
t <sub>SP</sub>	Fast mode	0	50	ns	Pulse width of suppressed spike
	High speed mode	0	10	ns	

 $<sup>^1</sup>$  Guaranteed by initial characterization.  $C_B$  refers to capacitive load on the bus line, tr and tf measured between 0.3  $V_{DD}$  and 0.7  $V_{DD}$ .  $^2$  A device must provide a data hold time for SDA in order to bridge the undefined region of the SCL falling edge.

# **Timing Diagram**

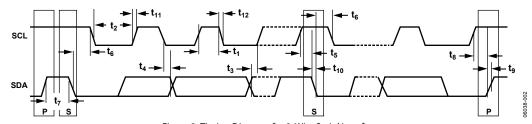


Figure 2. Timing Diagram for 2-Wire Serial Interface

# **ABSOLUTE MAXIMUM RATINGS**

 $T_A = 25$ °C, unless otherwise noted.

#### Table 4.

Table 4.	
Parameter	Rating
V <sub>DD</sub> to GND	−0.3 V to +6 V
Analog, Digital Inputs	$-0.3 \text{ V to V}_{DD} + 0.3 \text{ V or } 30 \text{ mA},$ whichever occurs first
Continuous Current, S or D Pins	100 mA
Peak Current, S or D Pins	300 mA (pulsed at 1 ms, 10% duty cycle max)
Operating Temperature Range	
Industrial (B Version)	−40°C to +85°C
Storage Temperature Range	−65°C to +150°C
Junction Temperature	150°C
$\theta_{JA}$ Thermal Impedance	
24-Lead LFCSP	30°C/W
Lead Temperature, Soldering (10 sec)	300°C
IR Reflow, Peak Temperature (<20 sec)	260°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Only one absolute maximum rating can be applied at any one time.

# **ESD CAUTION**

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



# PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

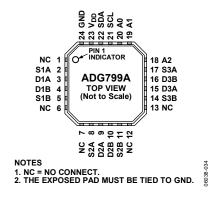


Figure 3. ADG799A Pin Configuration

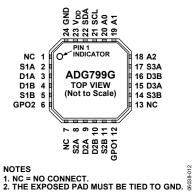


Figure 4. ADG799G Pin Configuration

**Table 5. Pin Function Descriptions** 

Pin No.	Mnemonic	Function
1	NC	Not internally connected.
2	S1A	A-Side Source Terminal for Crosspoint Switch 1. Can be an input or output.
3	D1A	A-Side Drain Terminal for Crosspoint Switch 1. Can be an input or output.
4	D1B	B-Side Drain Terminal for Crosspoint Switch 1. Can be an input or output.
5	S1B	B-Side Source Terminal for Crosspoint Switch 1. Can be an input or output.
6	NC/GPO2	Not internally connected (for the ADG799A) / General-Purpose Logic Output 2 (for the ADG799G).
7	NC	Not internally connected.
8	S2A	A-Side Source Terminal for Crosspoint Switch 2. Can be an input or output.
9	D2A	A-Side Drain Terminal for Crosspoint Switch 2. Can be an input or output.
10	D2B	B-Side Drain Terminal for Crosspoint Switch 2. Can be an input or output.
11	S2B	B-Side Source Terminal for Crosspoint Switch 2. Can be an input or output.
12	NC/GPO1	Not internally connected (for the ADG799A) / General-Purpose Logic Output 1 (for the ADG799G).
13	NC	Not internally connected.
14	S3B	B-Side Source Terminal for Crosspoint Switch 3. Can be an input or output.
15	D3A	A-Side Drain Terminal for Crosspoint Switch 3. Can be an input or output
16	D3B	B-Side Drain Terminal for Crosspoint Switch 3. Can be an input or output.
17	S3A	A-Side Source Terminal for Crosspoint Switch 3. Can be an input or output.
18	A2	Logic Input. Sets Bit A2 from the least significant bits of the 7-bit slave address.
19	A1	Logic Input. Sets Bit A1 from the least significant bits of the 7-bit slave address.
20	A0	Logic Input. Sets Bit A0 from the least significant bits of the 7-bit slave address.
21	SCL	Digital Input, Serial Clock Line. Open drain input used in conjunction with SDA to clock data into the device. External pull-up resistor required.
22	SDA	Digital Input/Output. Bidirectional open drain data line. External pull-up resistor required.
23	$V_{DD}$	Positive Power Supply Input.
24	GND	Ground (0 V) Reference.

# TYPICAL PERFORMANCE CHARACTERISTICS

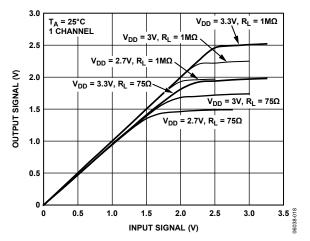


Figure 5. Analog Signal Range, 3 V Supply

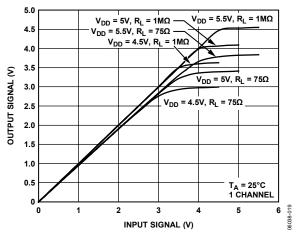


Figure 6. Analog Signal Range, 5 V Supply

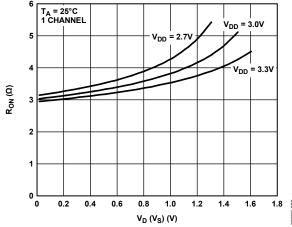


Figure 7. On Resistance vs.  $V_D$  ( $V_S$ ), 3 V Supply

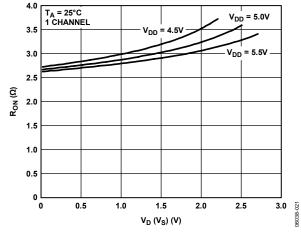


Figure 8. On Resistance vs.  $V_D$  ( $V_S$ ), 5 V Supply

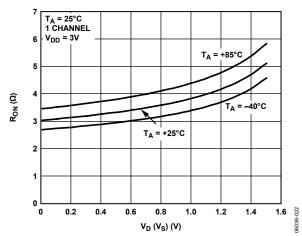


Figure 9. On Resistance vs.  $V_D$  ( $V_S$ ) for Various Temperatures, 3 V Supply

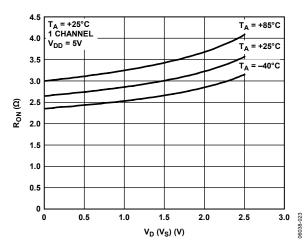


Figure 10. On Resistance vs.  $V_D$  ( $V_S$ ) for Various Temperatures, 5 V Supply

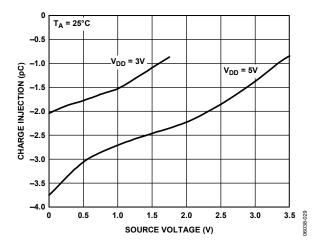


Figure 11. Charge Injection vs. Source Voltage

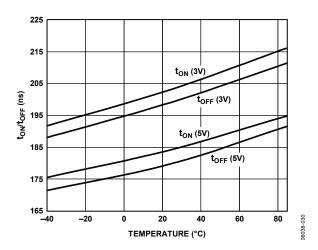


Figure 12. ton/toff vs. Temperature

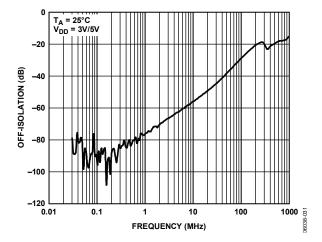


Figure 13. Off Isolation vs. Frequency

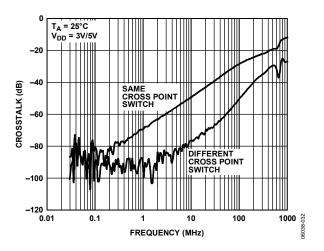


Figure 14. Crosstalk vs. Frequency

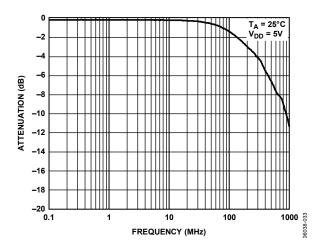


Figure 15. Bandwidth

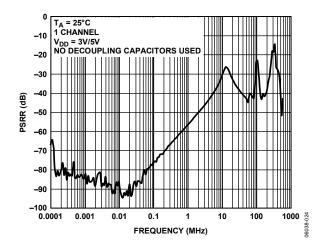


Figure 16. PSRR vs. Frequency

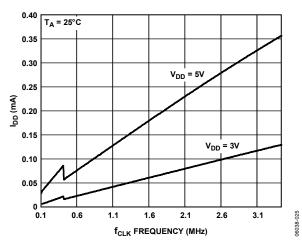


Figure 17. IDD vs. fclk Frequency

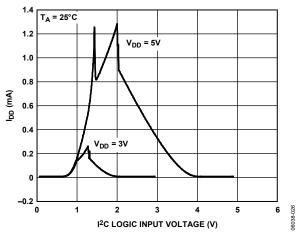


Figure 18. I<sub>DD</sub> vs. I<sup>2</sup>C Logic Input Voltage (SDA, SCL)

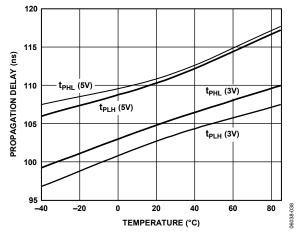


Figure 19. I<sup>2</sup>C to GPO Propagation Delay vs. Temperature (ADG799G Only)

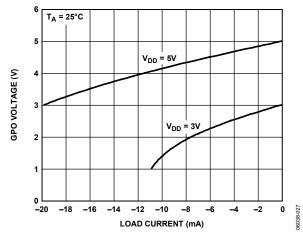


Figure 20. GPO V<sub>OH</sub> vs. Load Current

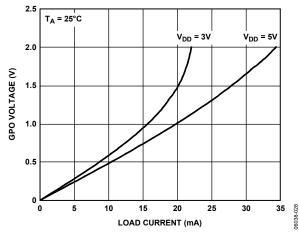


Figure 21. GPO Vol vs. Load Current

# **TEST CIRCUITS**

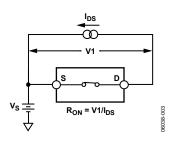


Figure 22. On Resistance

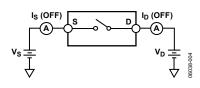


Figure 23. Off Leakage

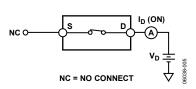


Figure 24. On Leakage

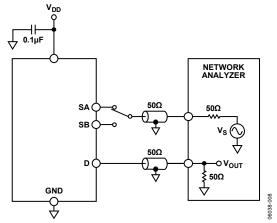


Figure 25. Bandwidth

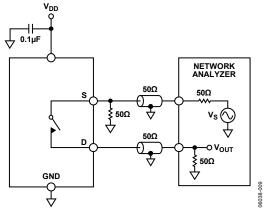


Figure 26. Off Isolation

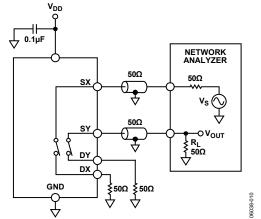


Figure 27. Channel-to-Channel Crosstalk

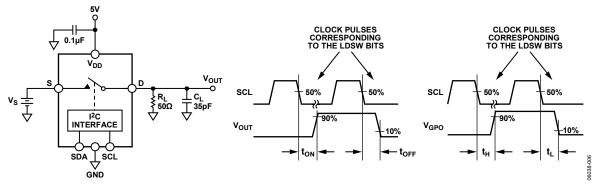


Figure 28. Switching Times

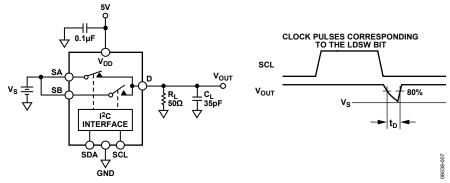


Figure 29. Break-Before-Make Time Delay

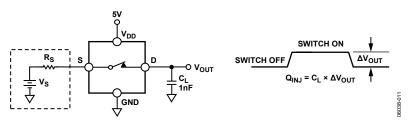


Figure 30. Charge Injection

# **TERMINOLOGY**

# On Resistance (Ron)

The series on-channel resistance measured between the S and D pins.

## On Resistance Match (ΔR<sub>ON</sub>)

The channel-to-channel matching of on resistance when channels are operated under identical conditions.

## On Resistance Flatness (R<sub>FLAT(ON)</sub>)

The variation of on resistance over the specified range produced by the specified analog input voltage change with a constant load current.

# Channel Off Leakage (IOFF)

The sum of leakage currents into or out of an off channel input.

# Channel On Leakage (IoN)

The current loss/gain through an on-channel resistance, creating a voltage offset across the device.

# Input Leakage Current ( $I_{\text{IN}}$ , $I_{\text{INL}}$ , $I_{\text{INH}}$ )

The current flowing into a digital input when a specified low level or high level voltage is applied to that input.

# Input/Output Off Capacitance (COFF)

The capacitance between an analog input and ground when the switch channel is off.

# Input/Output On Capacitance (CoN)

The capacitance between the inputs or outputs and ground when the switch channel is on.

#### Digital Input Capacitance (C<sub>IN</sub>)

The capacitance between a digital input and ground.

# Output On Switching Time (ton)

The time required for the switch channel to close. The time is measured from 50% of the falling edge of the LDSW bit to the time the output reaches 90% of the final value.

# Output Off Switching Time (toff)

The time required for the switch to open. The time is measured from 50% of the falling edge of the LDSW bit to the time the output reaches 10% of the final value.

# I<sup>2</sup>C to GPO Propagation Delay (t<sub>H</sub>, t<sub>L</sub>)

The time required for the logic value at the GPO pin to settle after loading a GPO command. The time is measured from 50% of the falling edge of the LDSW bit to the time the output reaches 90% of the final value for high and 10% for low.

## Total Harmonic Distortion + Noise (THD + N)

The ratio of the harmonic amplitudes plus noise of a signal to the fundamental.

#### -3 dB Bandwidth

The frequency at which the output is attenuated by 3 dB.

#### Off Isolation

The measure of unwanted signal coupling through an off switch.

#### Crosstalk

The measure of unwanted signal that is coupled through from one channel to another because of parasitic capacitance.

# **Charge Injection**

The measure of the glitch impulse transferred from the digital input to the analog output during on/off switching.

## **Differential Gain Error**

The measure of how much color saturation shift occurs when the luminance level changes. Both attenuation and amplification can occur; therefore, the largest amplitude change between any two levels is specified and expressed in %.

#### **Differential Phase Error**

The measure of how much hue shift occurs when the luminance level changes. It can be a negative or positive value and is expressed in degrees of subcarrier phase.

# Input High Voltage (V<sub>INH</sub>)

The minimum input voltage for Logic 1.

## Input Low Voltage (VINL)

The maximum input voltage for Logic 0.

# Output High Voltage (Voh)

The minimum output voltage for Logic 1.

# Output Low Voltage (Vol.)

The maximum output voltage for Logic 0.

 $I_{DD}$ 

Positive supply current.

# THEORY OF OPERATION

The ADG799A/ADG799G are monolithic CMOS device comprising three  $2 \times 2$  crosspoint switches controllable via a standard I<sup>2</sup>C serial interface. The CMOS process provides ultralow power dissipation, yet offers high switching speed and low on resistance.

The on resistance profile is very flat over the full analog input range, and wide bandwidth ensures excellent linearity and low distortion. These features, combined with a wide input signal range, make the ADG799A/ADG799G an ideal switching solution for a wide range of TV applications.

The switches conduct equally well in both directions when on. In the off condition, signal levels up to the supplies are blocked. The integrated serial I<sup>2</sup>C interface controls the operation of the crosspoint switches (ADG799A/ADG799G) and general-purpose logic pins (ADG799G only).

The ADG799A/ADG799G have many attractive features, such as the ability to individually control each switch, the option of reading back the status of any switch. The ADG799G has two general-purpose logic output pins controllable through the I<sup>2</sup>C interface. The following sections describe these features in more detail.

## I<sup>2</sup>C SERIAL INTERFACE

The ADG799A/ADG799G are controlled via an I<sup>2</sup>C-compatible serial bus interface (refer to the *I*<sup>2</sup>C-Bus Specification available from Philips Semiconductor) that allows the part to operate as a slave device (no clock is generated by the ADG799A/ADG799G). The communication protocol between the I<sup>2</sup>C master and the device operates as follows:

- The master initiates data transfer by establishing a start condition (defined as a high-to-low transition on the SDA line while SCL is high). This indicates that an address/data stream follows. All slave devices connected to the bus respond to the start condition and shift in the next eight bits, consisting of a seven bit address (MSB first) plus an R/W bit. This bit determines the direction of the data flow during the communication between the master and the addressed slave device.
- The slave device whose address corresponds to the transmitted address responds by pulling the SDA line low during the ninth clock pulse (this is known as the acknowledge bit).

At this stage, all other devices on the bus remain idle while the selected device waits for data to be written to, or read from, its serial register. If the  $R/\overline{W}$  bit is set high, the master reads from the slave device. However, if the  $R/\overline{W}$  bit is set low, the master writes to the slave device.

- 3. Data transmits over the serial bus in sequences of nine clock pulses (eight data bits followed by an acknowledge bit). The transitions on the SDA line must occur during the low period of the clock signal, SCL, and remain stable during the high period of SCL. Otherwise, a low-to-high transition when the clock signal is high can be interpreted as a stop event that ends the communication between the master and the addressed slave device.
- 4. After transferring all data bytes, the master establishes a stop condition, defined as a low-to-high transition on the SDA line while SCL is high. In write mode, the master pulls the SDA line high during the 10<sup>th</sup> clock pulse to establish a stop condition. In read mode, the master issues a no acknowledge for the ninth clock pulse (the SDA line remains high). The master then brings the SDA line low before the 10<sup>th</sup> clock pulse, and then high during the 10<sup>th</sup> clock pulse to establish a stop condition.

## I<sup>2</sup>C ADDRESS

The ADG799A/ADG799G each have a seven-bit  $I^2C$  address. The four most significant bits are internally hardwired while the last three bits (A0, A1, and A2) are user-adjustable. This allows the user to connect up to eight ADG799A/ADG799Gs to the same bus. The  $I^2C$  bit map shows the configuration of the seven-bit address.

# Seven-Bit I<sup>2</sup>C Address Bit Configuration

MSB						LSB
1	0	1	0	A2	A1	A0

# WRITE OPERATION

When writing to the ADG799A/ADG799G, the user must begin with an address byte and  $R/\overline{W}$  bit. Next, the switch acknowledges that it is prepared to receive data by pulling SDA low. Data is loaded into the device as a 16-bit word under the control of a serial clock input, SCL. Figure 31 illustrates the entire write sequence for the ADG799A/ADG799G. The first data byte (AX7 to AX0) controls the status of the crosspoint switches and the GPO pins, while the LDSW and RESETB bits from the second byte controls the operation mode of the device. Table 6 shows a list of all commands supported by the ADG799A/ADG799G with the corresponding byte that needs to be loaded during a write operation.

To achieve the desired configuration, one or more commands can be loaded into the device. Any combination of the commands listed in Table 6 can be used with the following restrictions:

- The commands referring to more than one switch overwrite any previous command.
- When a sequence of successive commands affect the same element (that is, the switch or GPO pin), only the last command is executed.

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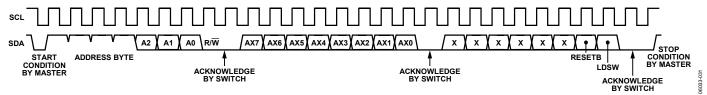


Figure 31. ADG799A/ADG799G Write Operation

## Table 6. ADG799A/ADG799G Command List

AX7	AX6	AX5	AX4	AX3	AX2	AX1	AX0	Addressed Switch /GPO Pin
0	1	1	0	0	0	0	0	\$1A/D1A, \$1B/D1B, \$2A/D2A, \$2B/D2B, \$3A/D3A, \$3B/D3B off
1	1	1	0	0	0	0	0	S1A/D1A, S1B/D1B, S2A/D2A, S2B/D2B, S3A/D3A, S3B/D3B on
0	1	1	0	0	0	0	1	S1A/D1B, S1B/D1A, S2A/D2B, S2B/D2A, S3A/D3B, S3B/D3A off
1	1	1	0	0	0	0	1	S1A/D1B, S1B/D1A, S2A/D2B, S2B/D2A, S3A/D3B, S3B/D3A on
0	1	1	0	0	0	1	0	S1A/D1A and S1A/D1B, S2A/D2A and S2A/D2B, S3A/D3A and S3A/D3B off
1	1	1	0	0	0	1	0	S1A/D1A and S1A/D1B, S2A/D2A and S2A/D2B, S3A/D3A and S3A/D3B on
0	1	1	0	0	0	1	1	S1B/D1A and S1B/D1B, S2B/D2A and S2B/D2B, S3B/D3A and S3B/D3B off
1	1	1	0	0	0	1	1	S1B/D1A and S1B/D1B, S2B/D2A and S2B/D2B, S3B/D3A and S3B/D3B on
0	1	1	0	0	1	0	0	S1A/D1A and S1B/D1A, S2A/D2A and S2B/D2A, S3A/D3A and S3B/D3A off
1	1	1	0	0	1	0	0	S1A/D1A and S1B/D1A, S2A/D2A and S2B/D2A, S3A/D3A and S3B/D3A on
)	1	1	0	0	1	0	1	S1A/D1B and S1B/D1B, S2A/D2B and S2B/D2B, S3A/D3B and S3B/D3B off
1	1	1	0	0	1	0	1	S1A/D1B and S1B/D1B, S2A/D2B and S2B/D2B, S3A/D3B and S3B/D3B on
)	1	1	0	0	1	1	0	S1A/D18 and 318/D18, 32A/D28 and 328/D28, 33A/D38 and 338/D38 on
1	1		0	0		1	0	S1A/D1A on
ا م	1	1	-	0	1	· ·		
) 1	1		0	0	1	1	1	S1A/D1B off S1A/D1B on
)	1	1	0		0	0	1	S1B/D1A off
		1		1	1		0	
1 )	1	1	0	1	0	0	0	S1B/D1A on
,	1	1	0	1	1	0	1	S1B/D1B off
	1	1	0	1	0	0	1	\$1B/D1B on
)	1	1	0	1	0	1	0	S2A/D2A off
l -	1	1	0	1	0	1	0	S2A/D2A on
)	1	1	0	1	0	1	1	S2A/D2B off
1	1	1	0	1	0	1	1	S2A/D2B on
)	1	1	0	1	1	0	0	S2B/D2A off
1	1	1	0	1	1	0	0	S2B/D2A on
0	1	1	0	1	1	0	1	S2B/D2B off
1	1	1	0	1	1	0	1	S2B/D2B on
)	1	1	0	1	1	1	0	S3A/D3A off
1	1	1	0	1	1	1	0	S3A/D3A on
)	1	1	0	1	1	1	1	S3A/D3B off
l	1	1	0	1	1	1	1	S3A/D3B on
)	1	1	1	0	0	0	0	S3B/D3A off
	1	1	1	0	0	0	0	S3B/D3A on
)	1	1	1	0	0	0	1	S3B/D3B off
١	1	1	1	0	0	0	1	S3B/D3B on
K <sup>1</sup>	1	1	1	0	0	1	0	Crosspoint Switch 1 disabled (All switches connected to D1A and D1B are off)
<b>K</b> <sup>1</sup>	1	1	1	0	0	1	1	Crosspoint Switch 2 disabled (All switches connected to D2A and D2B are off)
<b>〈</b> ¹	1	1	1	0	1	0	0	Crosspoint Switch 3 disabled (All switches connected to D3A and D3B are off)
)	1	1	1	0	1	0	1	GPO1 low for ADG799G/Reserved for ADG799A
l	1	1	1	0	1	0	1	GPO1 high for ADG799G/Reserved for ADG799A
)	1	1	1	0	1	1	0	GPO2 low for ADG799G/Reserved for ADG799A
l	1	1	1	0	1	1	0	GPO2 high for ADG799G/Reserved for ADG799A
)	1	1	1	0	1	1	1	GPO1 and GPO2 low for ADG799G/Reserved for ADG799A
l	1	1	1	0	1	1	1	GPO1 and GPO2 high for ADG799G/Reserved for ADG799A
0	1	1	1	1	1	1	1	All muxes disabled (all switches are off)
1	1	1	1	1	1	1	1	Reserved

 $<sup>^{1}</sup>$  X = Logic state does not matter.

# **LDSW BIT**

The LDSW bit allows the user to control the way the device executes the commands loaded during the write operations. The ADG799A/ADG799G execute all the commands loaded between two successive write operations that have set the LDSW bit high.

Setting the LDSW high for every write cycle ensures that the device executes the command immediately after the LDSW bit is loaded into the device. This setting can be used when the desired configuration can be achieved by sending a single command or when the switches and/or GPO pins are not required to be updated at the same time. When the desired configuration requires multiple commands with simultaneous updates, the LDSW bit should be set low while loading the commands except for the last one when the LDSW bit should be set high. Once the last command with LDSW = high is loaded, the device simultaneously executes all commands received since the last update.

## **POWER ON/SOFTWARE RESET**

The ADG799A/ADG799G has a software reset function implemented by the RESETB bit from the second data byte written to the device. For normal operation of the crosspoint switch and GPO pins, this bit should be set high. When RESETB = low or after power-up, the switches from all crosspoint switch pins are turned off (open) and the GPO pins are set low.

## **READ OPERATION**

When reading data back from the ADG799A/ADG799G, the user must begin with an address byte and R/W bit. The switch then acknowledges that it is prepared to transmit data by pulling SDA low. Following this acknowledgement, the ADG799A/ADG799G transmit two bytes on the next clock edges. These bytes contain the status of the switches, and each byte is followed by an acknowledge bit. A logic high bit represents a switch in the on (close) state while a low represents a switch in the off (open) state. For the GPO pins (ADG799G only), the bit represents the logic value of the pin. Figure 32 illustrates the entire read sequence.

The bit maps accompanying Figure 32 show the relationship between the elements of the ADG799A and ADG799G (that is, the switches and GPO pins) and the bits that represent their status after a completed read operation.

# ADG799A Bit Map

RB15	RB14	RB13	RB12	RB11	RB10	RB9	RB8	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0
S1A/D1A	S1B/D1A	S1A/D1B	S1B/D1B	S2A/D2A	S2B/D2A	S2A/D2B	S2B/D2B	S3A/D3A	S3B/D3A	S3A/D3B	S3B/D3B	-	-	-	-

## ADG799G Bit Map

RB15	RB14	RB13	RB12	RB11	RB10	RB9	RB8	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0
S1A/D1A	S1B/D1A	S1A/D1B	S1B/D1B	S2A/D2A	S2B/D2A	S2A/D2B	S2B/D2B	S3A/D3A	S3B/D3A	S3A/D3B	S3B/D3B	GPO1	GPO2	-	-

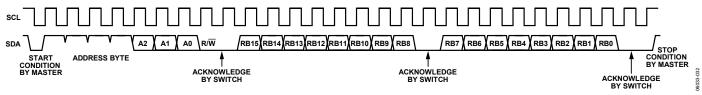


Figure 32. ADG799A/ADG799G Read Operation

# **EVALUATION BOARD**

The EVAL-ADG799GEB allows designers to evaluate the high performance of the device with minimum effort.

The evaluation kit includes a printed circuit board populated with the ADG799G. The evaluation board can be used to evaluate the performance of both the ADG799A and ADG799G. It interfaces to the USB port of a PC, or it can be used as a standalone evaluation board. Software is available with the evaluation board that allows the user to easily program the ADG799G through the USB port. Schematics of the evaluation board are shown in Figure 33 and Figure 34. The software runs on any PC with Microsoft\* Windows\* 2000 or Windows XP and a minimum screen resolution of 1200 × 768.

## **USING THE ADG799G EVALUATION BOARD**

The ADG799G evaluation kit is a test system designed to simplify the evaluation of the device. Each input/output of the part comes with a socket specifically chosen for easy audio/video evaluation. A data sheet is also available with the evaluation board offering full information on how to operate the evaluation board.

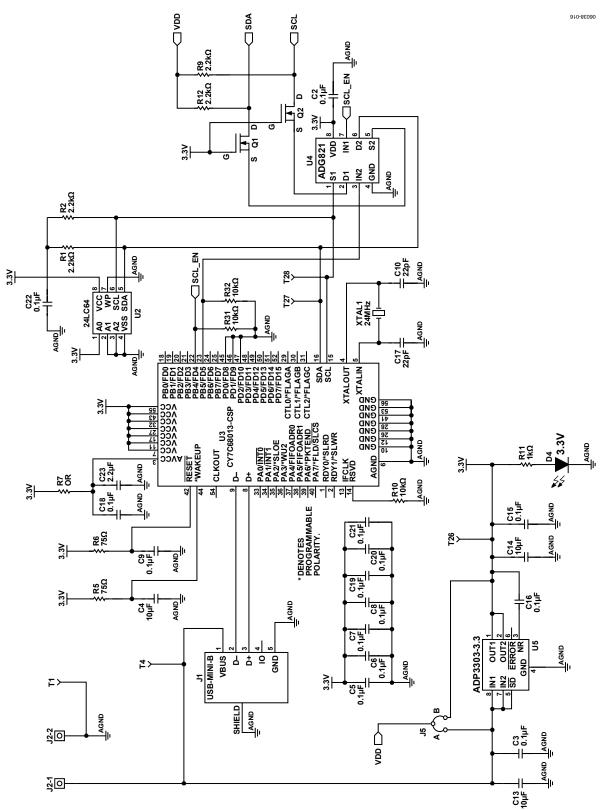
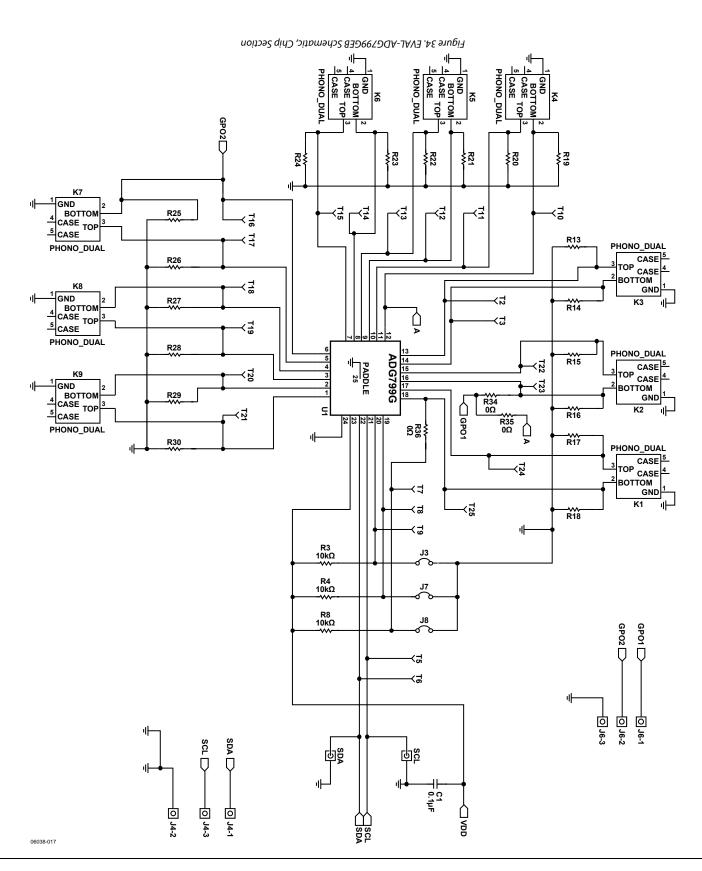
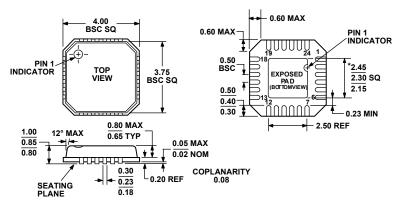


Figure 33. EVAL-ADG799GEB Schematic, USB Controller Section

# **366730A/A66730A**



# **OUTLINE DIMENSIONS**



## \*COMPLIANT TO JEDEC STANDARDS MO-220-VGGD-2 EXCEPT FOR EXPOSED PAD DIMENSION

Figure 35. 24-Lead Lead Frame Chip Scale Package [LFCSP\_VQ] 4 mm × 4 mm Body, Very Thin Quad (CP-24-2) Dimensions shown in millimeters

# **ORDERING GUIDE**

OIID EIIII I G GOID E						
Model	Temperature Range	I <sup>2</sup> C Speed	Package Description	Package Option		
ADG799ABCPZ-REEL <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz	24-Lead LFCSP_VQ	CP-24-2		
ADG799ABCPZ-500RL7 <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz	24-Lead LFCSP_VQ	CP-24-2		
ADG799ACCPZ-REEL <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz, 3.4 MHz	24-Lead LFCSP_VQ	CP-24-2		
ADG799ACCPZ-500RL7 <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz, 3.4 MHz	24-Lead LFCSP_VQ	CP-24-2		
ADG799GBCPZ-REEL <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz	24-Lead LFCSP_VQ	CP-24-2		
ADG799GBCPZ-500RL7 <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz	24-Lead LFCSP_VQ	CP-24-2		
ADG799GCCPZ-REEL <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz, 3.4 MHz	24-Lead LFCSP_VQ	CP-24-2		
ADG799GCCPZ-500RL7 <sup>1</sup>	-40°C to +85°C	100 kHz, 400 kHz, 3.4 MHz	24-Lead LFCSP_VQ	CP-24-2		
EVAL-ADG799GEB <sup>2</sup>			Evaluation Board			

 $<sup>^{1}</sup>$  Z = Pb-free part.

<sup>&</sup>lt;sup>2</sup> Evaluation board is RoHS compliant.

ADG799A/ADG799G		
NOTES		

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