



General Description

The MAX5387 dual, 256-tap, volatile, low-voltage linear taper digital potentiometer offers three end-to-end resistance values of $10k\Omega$, $50k\Omega$, and $100k\Omega$. Operating from a single +2.6V to +5.5V power supply, the device provides a low 35ppm/°C end-to-end temperature coefficient. The device features an I²C interface.

The small package size, low supply operating voltage, low supply current, and automotive temperature range of the MAX5387 make the device uniquely suitable for the portable consumer market, battery backup industrial applications, and the automotive market.

The MAX5387 is specified over the automotive -40°C to +125°C temperature range and is available in a 14-pin TSSOP package.

Features

- ◆ Dual, 256-Tap Linear Taper Positions
- ♦ Single +2.6V to +5.5V Supply Operation
- ♦ Low < 1µA Quiescent Supply Current
- ♦ 10kΩ, 50kΩ, 100kΩ End-to-End Resistance Values
- ♦ I²C-Compatible Interface
- ♦ Power-On Sets Wiper to Midscale
- **◆** -40°C to + 125°C Operating Temperature Range

Applications

Low-Voltage Battery Applications

Portable Electronics

Mechanical Potentiometer Replacement

Offset and Gain Control

Adjustable Voltage References/Linear Regulators

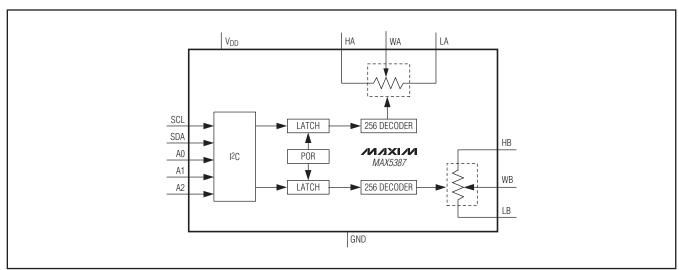
Automotive Electronics

Ordering Information

PART	PIN-PACKAGE	END-TO-END RESISTANCE ($k\Omega$)
MAX5387LAUD+	14 TSSOP	10
MAX5387MAUD+	14 TSSOP	50
MAX5387NAUD+	14 TSSOP	100

Note: All devices are specified over the -40°C to +125°C operating temperature range.

Functional Diagram



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⁺Denotes a lead(Pb)-free/RoHS-compliant package.

ABSOLUTE MAXIMUM RATINGS

V _{DD} to GND	0.3V to +6V	MAX5387N	±1mA
H_, W_, L_ to GND	0.3V to the lower of	Continuous Power Dissipation (TA = +70°C)	
	$(V_{DD} + 0.3V)$ and +6V	14-Pin TSSOP (derate 10mW/°C above +70	J°C)796.8mW
All Other Pins to GND	0.3V to +6V	Operating Temperature Range	40°C to +125°C
Continuous Current into H_, W_, and	d L_	Junction Temperature	+150°C
MAX5387L	±5mA	Storage Temperature Range	65°C to +150°C
MAX5387M	±2mA	Lead Temperature (soldering, 10s)	+300°C
0, 1, 1, 1, 1, 1, 4	1 1 M ' D '' "		1 16 11 1

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{DD} = +2.6V \text{ to } +5.5V, V_{H} = V_{DD}, V_{L} = GND, T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C, \text{ unless otherwise noted.}$ Typical values are at $V_{DD} = +5V, T_{A} = +25^{\circ}C.)$ (Note 1)

PARAMETER	SYMBOL	С	ONDITIONS	MIN	TYP	MAX	UNITS	
Resolution	N			256			Тар	
DC PERFORMANCE (Voltage-D	river Mode)			,				
Integral Nonlinearity	INL	(Note 2)		-0.5	-	+0.5	LSB	
Differential Nonlinearity	DNL	(Note 2)		-0.5		+0.5	LSB	
Dual Code Matching		Register A = reg	ister B	-0.5		+0.5	LSB	
Ratiometric Resistor Tempco		(ΔV _W /V _W)/ΔT; no	load		+5		LSB	
			MAX5387L	-3	-2.5			
Full-Scale Error		Code = FFH	MAX5387M	-1	-0.5		LSB	
			MAX5387N	-0.5	-0.25			
			MAX5387L		+2.5	+3		
Zero-Scale Error		Code = 00H	MAX5387M		+0.5	+1.0	LSB	
			MAX5387N		+0.25	+0.5		
DC PERFORMANCE (Variable-I	Resistor Mod	e)						
			MAX5387L		±1.0	±2.5		
		V _{DD} > +2.6V	MAX5387M		±0.5	±1.0		
			MAX5387N		±0.25	±0.8]	
Integral Nonlinearity	R-INL		MAX5387L		±0.4	±1.5	LSB	
		V _{DD} > +4.75V	MAX5387M		±0.3	±0.75		
			MAX5387N		±0.25	±0.5]	
Differential Nonlinearity	R-DNL	V _{DD} > 2.6V (Note	e 3)	-0.5		+0.5	LSB	
DC PERFORMANCE (Resistor	Characteristic	s)						
Wiper Resistance (Note 4)	D.v.	V _{DD} > 2.6V		250	600	0		
Wiper Resistance (Note 4)	RwL	V _{DD} > 4.75V		150	200	Ω		
Terminal Capacitance	CH_, CL_	Measured to GN	D		10		рF	
Wiper Capacitance	Cw_	Measured to GN	D		50		рF	
End-to-End Resistor Tempco	TCR	No load			35		ppm/°C	
End-to-End Resistor Tolerance	ΔRHL	Wiper not conne	-25		+25	%		
AC PERFORMANCE								
Crosstalk		(Note 5)			-90		dB	
		Code = 08H,	MAX5387L		600			
-3dB Bandwidth	BW	10pF load,	MAX5387M		150		kHz	
		$V_{DD} = +2.6V$	MAX5387N		75			

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +2.6V \text{ to } +5.5V, V_{H} = V_{DD}, V_{L} = GND, T_{A} = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}, \text{ unless otherwise noted.}$ Typical values are at $V_{DD} = +5V, T_{A} = +25^{\circ}\text{C}$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Total Harmonic Distortion Plus Noise	THD+N	Measured at W; VH_ = 1VRMS at 1kHz		0.015		%
		MAX5387L		300		
Wiper Settling Time (Note 6)	ts	MAX5387M		1000		ns
		MAX5387N		2000		
POWER SUPPLIES		,				
Supply-Voltage Range	V _{DD}		2.6		5.5	V
Standby Current		Digital inputs = V _{DD} or GND		1		μΑ
DIGITAL INPUTS						
Minimum Input High Voltage	VIH		70			% x VDD
Maximum Input Low Voltage	VIL				30	% x V _{DD}
Input Leakage Current			-1		+1	μΑ
Input Capacitance				5		pF
TIMING CHARACTERISTICS (No	tes 7, 8)					
Maximum SCL Frequency	fscl				400	kHz
Setup Time for START Condition	tsu:sta		0.6			μs
Hold Time for START Condition	thd:Sta		0.6			μs
SCL High Time	tHIGH		0.6			μs
SCL Low Time	tLOW		1.3			μs
Data Setup Time	tsu:dat		100			ns
Data Hold Time	thd:dat		0			μs
SDA, SCL Rise Time	t _R				0.3	μs
SDA, SCL Fall	tF				0.3	μs
Setup Time for STOP Condition	tsu:sto		0.6			μs
Bus Free Time Between STOP and START Conditions	tBUF	Minimum power-up rate = 0.2V/μs	1.3			μs
Pulse-Suppressed Spike Width	tsp				50	ns
Capacitive Load for Each Bus	СВ				400	pF

- **Note 1:** All devices are 100% production tested at T_A = +25°C. Specifications overtemperature limits are guaranteed by design and characterization.
- Note 2: DNL and INL are measured with the potentiometer configured as a voltage-divider (Figure 1) with H_ = VDD and L_ = 0V. The wiper terminal is unloaded and measured with an ideal voltmeter.
- Note 3: R-DNL and R-INL are measured with the potentiometer configured as a variable resistor (Figure 1). DNL and INL are measured with the potentiometer configured as a variable resistor. H_ is unconnected and L_ = GND. For VDD = +5V, the wiper terminal is driven with a source current of 400μA for the $10k\Omega$ configuration, 80μ A for the $50k\Omega$ configuration, and 40μ A for the $100k\Omega$ configuration. For VDD = +2.6V, the wiper terminal is driven with a source current of 200μ A for the $10k\Omega$ configuration, 40μ A for the $50k\Omega$ configuration, and 20μ A for the $100k\Omega$ configuration.
- Note 4: The wiper resistance is the worst value measured by injecting the currents given in Note 3 into W_ with L_ = GND. Rw = (Vw VH)/Iw.
- Note 5: Drive HA with a 1kHz GND to VDD amplitude tone. LA = LB = GND. No load. WB is at midscale with a 10pF load. Measure WB.
- Note 6: The wiper settling time is the worst-case 0 to 50% rise time, measured between tap 0 and tap 127. H_ = V_{DD}, L_ = GND, and the wiper terminal is loaded with 10pF capacitance to ground.
- Note 7: Digital timing is guaranteed by design and characterization, not production tested.
- Note 8: The SCL clock period includes rise and fall times ($t_R = t_F$). All digital input signals are specified with $t_R = t_F = 2$ ns and timed from a voltage level of ($V_{IL} + V_{IH}$)/2.



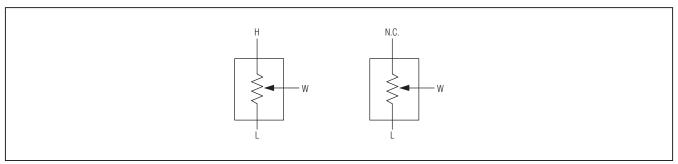
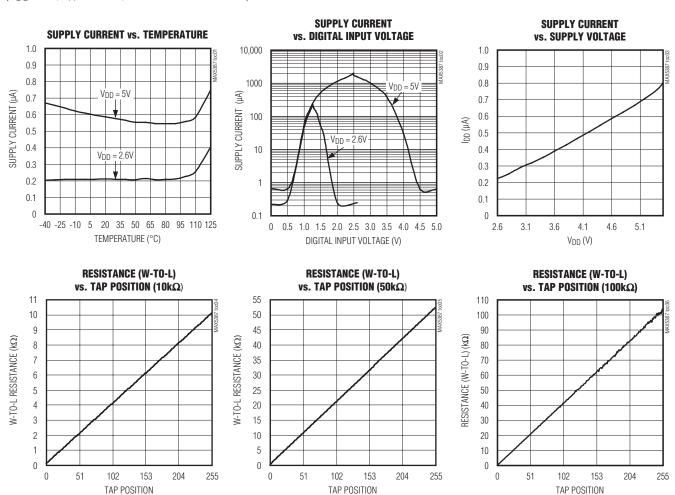


Figure 1. Voltage-Divider and Variable Resistor Configurations

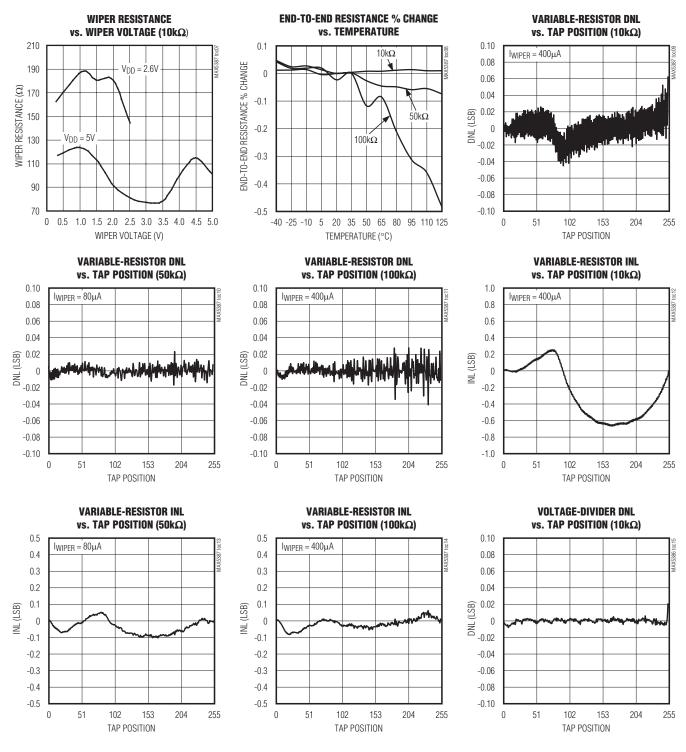
Typical Operating Characteristics

 $(V_{DD} = 5V, T_A = +25^{\circ}C, unless otherwise noted.)$



_Typical Operating Characteristics (continued)

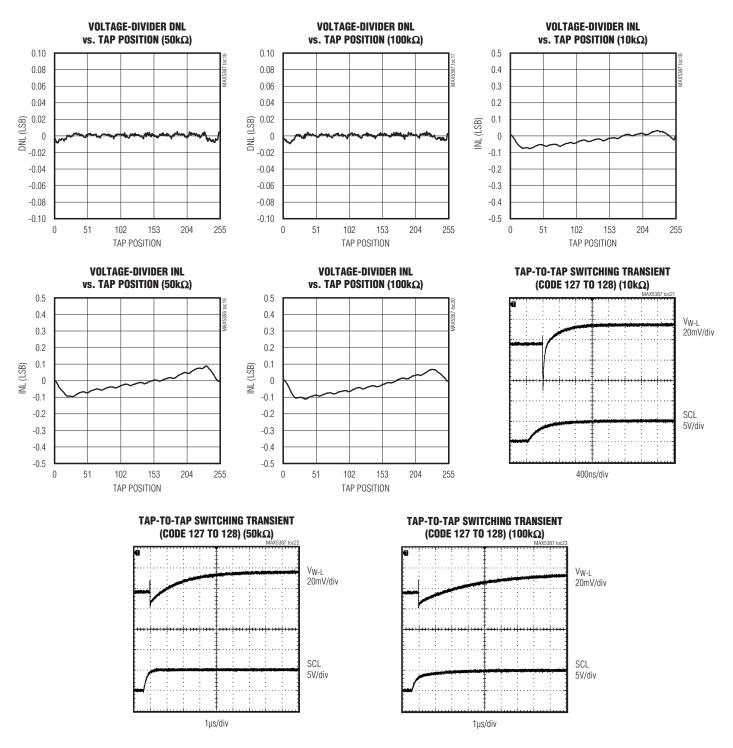
 $(V_{DD} = 5V, T_A = +25^{\circ}C, unless otherwise noted.)$



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Typical Operating Characteristics (continued)

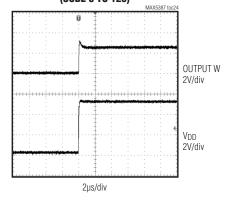
 $(V_{DD} = 5V, T_A = +25^{\circ}C, unless otherwise noted.)$



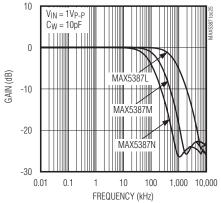
Typical Operating Characteristics (continued)

 $(V_{DD} = 5V, T_A = +25^{\circ}C, unless otherwise noted.)$

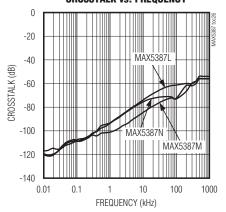
MAX5387 POWER-ON WIPER TRANSIENT (CODE 0 TO 128)



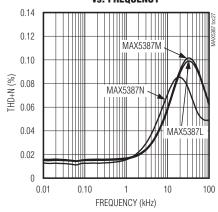
MIDSCALE FREQUENCY RESPONSE



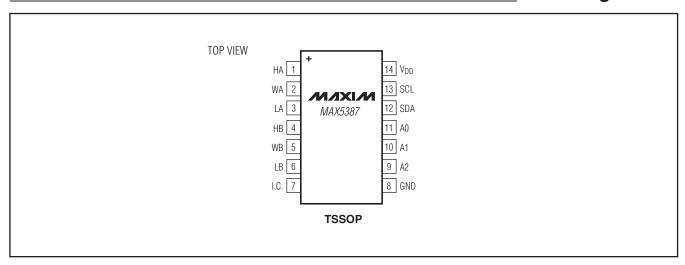
CROSSTALK vs. FREQUENCY



TOTAL HARMONIC DISTORTION PLUS NOISE vs. Frequency



Pin Configuration



Pin Description

PIN	NAME	FUNCTION
1	НА	Resistor A High Terminal. The voltage at HA can be higher or lower than the voltage at LA. Current can flow into or out of HA.
2	WA	Resistor A Wiper Terminal
3	LA	Resistor A Low Terminal. The voltage at LA can be higher or lower than the voltage at HA. Current can flow into or out of LA.
4	НВ	Resistor B High Terminal. The voltage at HB can be higher or lower than the voltage at LB. Current can flow into or out of HB.
5	WB	Resistor B Wiper Terminal
6	LB	Resistor B Low Terminal. The voltage at LB can be higher or lower than the voltage at HB. Current can flow into or out of LB.
7	I.C.	Internally Connected. Connect to GND.
8	GND	Ground
9	A2	Address Input 2. Connect to V _{DD} or GND.
10	A1	Address Input 1. Connect to V _{DD} or GND.
11	A0	Address Input 0. Connect to V _{DD} or GND.
12	SDA	I ² C-Compatible Serial-Data Input/Output. A pullup resistor is required.
13	SCL	I ² C-Compatible Serial-Clock Input. A pullup resistor is required.
14	V _{DD}	Power-Supply Input. Bypass VDD to GND with a 0.1µF capacitor close to the device.

Detailed Description

The MAX5387 dual, 256-tap, volatile, low-voltage linear taper digital potentiometer offers three end-to-end resistance values of $10k\Omega$, $50k\Omega$, and $100k\Omega$. The potentiometer consists of 255 fixed resistors in series between terminals H_ and L_. The potentiometer wiper, W_, is programmable to access any one of the 256 tap points on the resistor string.

The potentiometers are programmable independently of each other. The MAX5387 features an I²C interface.

I²C Digital Interface

The I²C interface contains a shift register that decodes the command and address bytes, routing the data to the appropriate control registers. Data written to a control register immediately updates the wiper position. Wipers A and B power up in midposition, D[7:0] = 80H.

Serial Addressing

The MAX5387 operates as a slave device that receives data through an I²C-/SMBusTM-compatible 2-wire serial interface. The interface uses a serial-data access (SDA) line and a serial-clock line (SCL) to achieve bidirectional communication between master(s) and slave(s). A

master, typically a microcontroller, initiates all data transfers to the MAX5387, and generates the SCL clock that synchronizes the data transfer (Figure 2).

The MAX5387 SDA line operates as both an input and an open-drain output. The SDA line requires a pullup resistor, typically 4.7k Ω . The MAX5387 SCL line operates only as an input. The SCL line requires a pullup resistor (typically 4.7k Ω) if there are multiple masters on the 2-wire interface, or if the master in a single-master system provides an open-drain SCL output.

Each transmission consists of a START (S) condition (Figure 3) sent by a master, followed by the MAX5387 7-bit slave address plus the NOP/W bit (Figure 6), 1 command byte and 1 data byte, and finally a STOP (P) condition (Figure 3).

START and STOP Conditions

SCL and SDA remain high when the interface is inactive. A master controller signals the beginning of a transmission with a START condition by transitioning SDA from high to low while SCL is high. The master controller issues a STOP condition by transitioning the SDA from low to high while SCL is high, after finishing communicating with the slave. The bus is then free for another transmission.

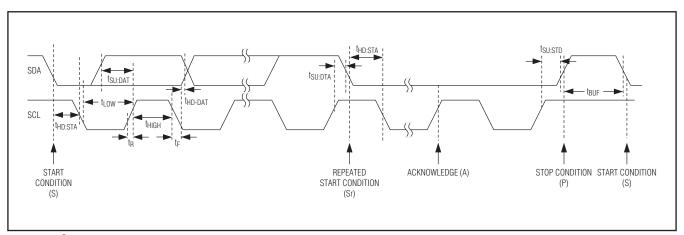


Figure 2. I²C Serial Interface Timing Diagram

SMBus is a trademark of Intel Corp.



Bit Transfer

One data bit is transferred during each clock pulse. The data on the SDA line must remain stable while SCL is high. See Figure 4.

Acknowledge

The acknowledge bit is a clocked 9th bit that the recipient uses to handshake receipt of each byte of data. See Figure 5. Each byte transferred requires a total of nine bits. The master controller generates the 9th clock pulse, and the recipient pulls down SDA during the acknowledge clock pulse, so the SDA line remains stable low during the high period of the clock pulse.

Slave Address

The MAX5387 includes a 7-bit slave address (Figure 6). The 8th bit following the 7th bit of the slave address is the NOP/\overline{W} bit. Set the NOP/\overline{W} bit low for a write command and high for a no-operation command. The device does not support readback.

The device provides three address inputs (A0, A1, and A2), allowing up to eight devices to share a common bus (Table 1). The first 4 bits (MSBs) of the factory-set slave addresses are always 0101. A2, A1, and A0 set the next 3 bits of the slave address. Connect each address input to VDD or GND. Each device must have a unique address to share a common bus.

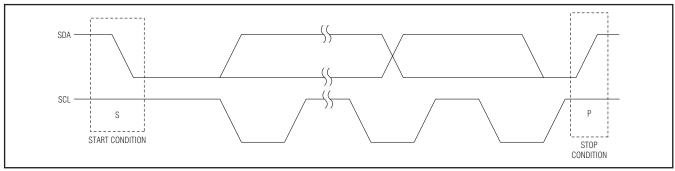


Figure 3. START and STOP Conditions

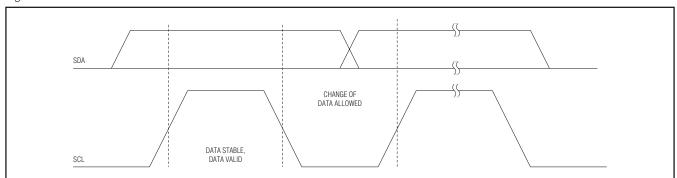


Figure 4. Bit Transfer

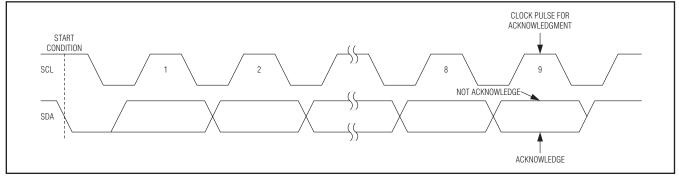


Figure 5. Acknowledge

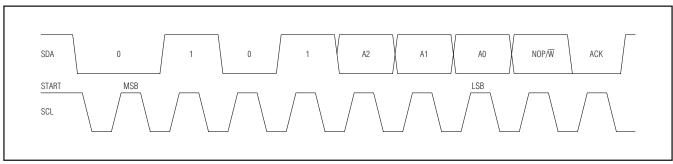


Figure 6. Slave Address

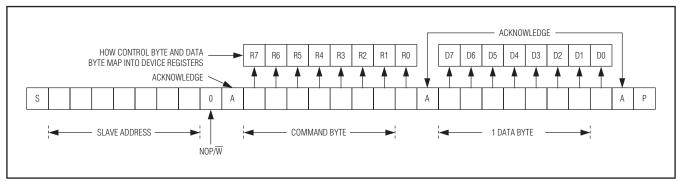


Figure 7. Command and Single Data Byte Received

Message Format for Writing

Write to the devices by transmitting the device's slave address with NOP/\overline{W} (eighth bit) set to zero, followed by at least 2 bytes of information. The first byte of information is the command byte. The second byte is the data byte. The data byte goes into the internal register of the device as selected by the command byte (Figure 7 and Table 2).

Command Byte

Use the command byte to select the destination of the wiper data. See Table 2.

Command Descriptions

REG A: The data byte writes to register A and the wiper of potentiometer A moves to the appropriate position. D[7:0] indicates the position of the wiper. D[7:0] = 00h

moves the wiper to the position closest to LA. D[7:0] = FFh moves the wiper to the position closest to HA. D[7:0] is 80h following power-on.

Table 1. Slave Addresses

AD	DRESS INP	SLAVE ADDRESS							
A2	A 1	A0	SLAVE ADDRESS						
GND	GND	GND	0101000						
GND	GND	VDD	0101001						
GND	V _{DD}	GND	0101010						
GND	VDD	VDD	0101011						
V _{DD}	GND	GND	0101100						
VDD	GND	VDD	0101101						
V _{DD}	V _{DD} V _{DD}		0101110						
VDD	VDD	VDD	0101111						

Table 2. I²C Command Byte Summary

			ADDRESS BYTE							COMMAND BYTE								DATA BYTE											
		1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	27	
SCL CYCLE NO.	START (S)	A6	A5	A4	А3	A2	A1	A0	W	ACK (A)	R7	R6	R5	R4	R3	R2	R1	R0	ACK (A)	D7	D6	D5	D4	D3	D2	D1	D0	ACK (A)	STOP (P)
REG A		0	1	0	1	A2	A1	A0	0		0	0	0	1	0	0	0	1		D7	D6	D5	D4	D3	D2	D1	D0		
REG B		0	1	0	1	A2	A1	A0	0		0	0	0	1	0	0	1	0		D7	D6	D5	D4	D3	D2	D1	D0		
REGS A AND B		0	1	0	1	A2	A1	A0	0		0	0	0	1	0	0	1	1		D7	D6	D5	D4	D3	D2	D1	D0		

REG B: The data byte writes to register B and the wiper of potentiometer B moves to the appropriate position. D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wiper to the position closest to LB. D[7:0] = FFh moves the wiper to the position closest to HB. D[7:0] is 80h following power-on.

REGS A and B: The data byte writes to registers A and B and the wipers of potentiometers A and B move to the appropriate position. D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wipers to the position closest to L_. D[7:0] = FFh moves the wipers to the position closest to H_. D[7:0] is 80h following power-on.

Applications Information

Variable Gain Amplifier

Figure 8 shows a potentiometer adjusting the gain of a noninverting amplifier. Figure 9 shows a potentiometer adjusting the gain of an inverting amplifier.

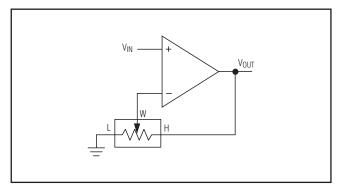


Figure 8. Variable Gain Noninverting Amplifier

Adjustable Dual Regulator

Figure 10 shows an adjustable dual linear regulator using a dual potentiometer as two variable resistors.

Adjustable Voltage Reference

Figure 11 shows an adjustable voltage reference circuit using a potentiometer as a voltage-divider.

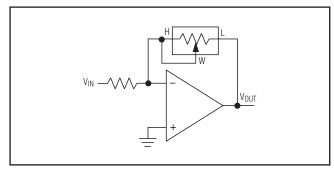


Figure 9. Variable Gain Inverting Amplifier

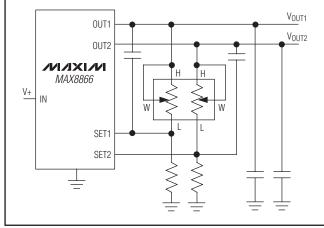


Figure 10. Adjustable Dual Linear Regulator

Variable Gain Current to Voltage Converter

Figure 12 shows a variable gain current to voltage converter using a potentiometer as a variable resistor.

LCD Bias Control

Figure 13 shows a positive LCD bias control circuit using a potentiometer as a voltage-divider.

Figure 14 shows a positive LCD bias control circuit using a potentiometer as a variable resistor.

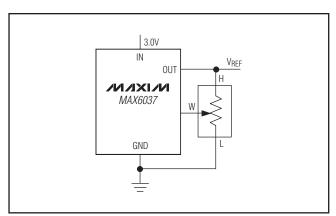


Figure 11. Adjustable Voltage Reference

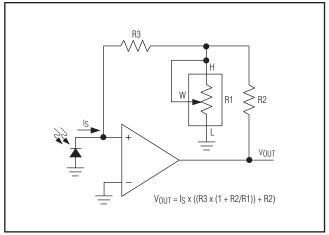


Figure 12. Variable Gain I-to-V Converter

Programmable Filter

Figure 15 shows a programmable filter using a dual potentiometer.

Offset-Voltage Adjustment Circuit

Figure 16 shows an offset-voltage adjustment circuit using a dual potentiometer.

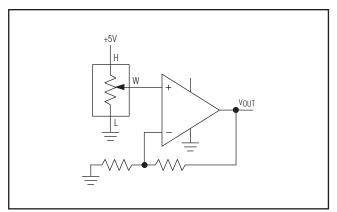


Figure 13. Positive LCD Bias Control Using a Voltage-Divider

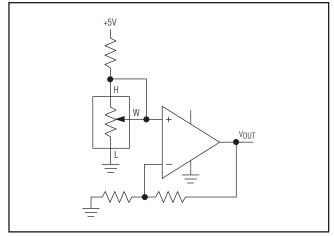


Figure 14. Positive LCD Bias Control Using a Variable Resistor

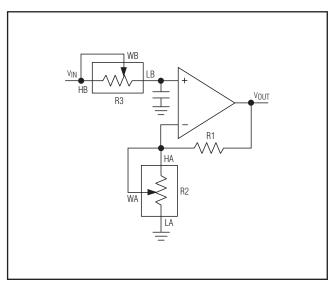


Figure 15. Programmable Filter

PROCESS: BICMOS

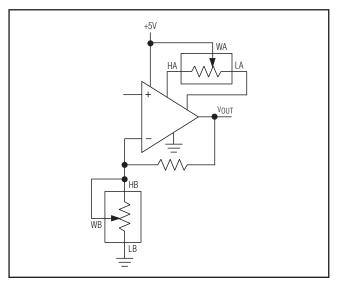


Figure 16. Offset-Voltage Adjustment Circuit

Process Information

_Package Information

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.				
14 TSSOP	U14+1	<u>21-0066</u>				

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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