Freescale Semiconductor
Data Sheet: Product Preview

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MPC5553 Microcontroller Data Sheet

by: Microcontroller Division

This document provides electrical specifications, pin assignments, and package diagrams for the MPC5553 microcontroller device. For functional characteristics, refer to the MPC5553/MPC5554 *Microcontroller Reference Manual*.

Contents

1 Overview

The MPC5553 microcontroller (MCU) is a member of the MPC5500 family of microcontrollers built on the Power Architecture™ embedded technology. This family of parts contains many new features coupled with high performance CMOS technology to provide substantial reduction of cost per feature and significant performance improvement over the MPC500 family.

The host processor core of this device complies with the Power Architecture embedded category that is 100% user-mode compatible (with floating point library) with the original Power PC™ user instruction set architecture (UISA). The embedded architecture has enhancements that improve the performance in embedded applications. This core also has additional instructions, including digital signal processing (DSP) instructions, beyond the original Power PC instruction set. This family of parts

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contains many new features coupled with high performance CMOS technology to provide significant performance improvement over the MPC565.

The MPC5553 of the MPC5500 family has two levels of memory hierarchy. The fastest accesses are to the 8-kilobyte unified cache. The next level in the hierarchy contains the 64-kilobyte on-chip internal SRAM and 1.5 Mbyte internal Flash memory. The internal SRAM and flash memory can hold instructions and data. The external bus interface has been designed to support most of the standard memories used with the MPC5*xx* family.

The complex input/output timer functions of the MPC5500 family are performed by an enhanced time processor unit engine (eTPU). The eTPU engine controls 32 hardware channels. The eTPU has been enhanced over the TPU by providing 24-bit timers, double-action hardware channels, variable number of parameters per channel, angle clock hardware, and additional control and arithmetic instructions. The eTPU can be programmed using a high-level programming language.

The less complex timer functions of the MPC5500 family are performed by the enhanced modular input/output system (eMIOS). The eMIOS' 24 hardware channels are capable of single-action, double-action, pulse-width modulation (PWM), and modulus-counter operations. Motor control capabilities include edge-aligned and center-aligned PWM.

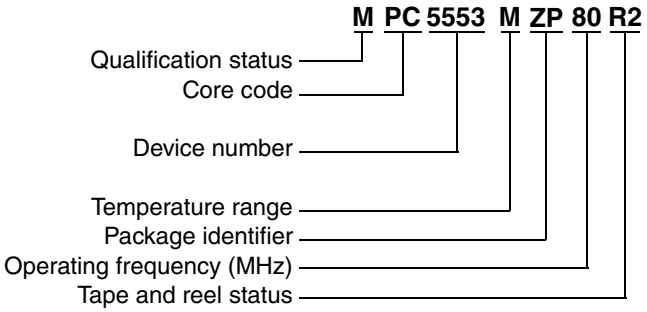
Off-chip communication is performed by a suite of serial protocols including controller area networks (FlexCANs), enhanced deserial/serial peripheral interfaces (DSPI), and enhanced serial communications interfaces (eSCIs). The DSPIs support pin reduction through hardware serialization and deserialization of timer channels and general-purpose input/output (GPIO) signals.

The MCU of the MPC5553 has an on-chip 40-channel enhanced queued dual analog-to-digital converter (eQADC).

The system integration unit (SIU) performs several chip-wide configuration functions. Pad configuration and general-purpose input and output (GPIO) are controlled from the SIU. External interrupts and reset control are also determined by the SIU. The internal multiplexer submodule (SIU_DISR) provides multiplexing of eQADC trigger sources, daisy chaining the DSPIs, and external interrupt signal multiplexing.

The Fast Ethernet (FEC) module is a RISC-based controller that supports both 10 and 100 Mbps Ethernet/IEEE® 802.3 networks and is compatible with three different standard MAC (media access controller) PHY (physical) interfaces to connect to an external Ethernet bus. The FEC supports the 10 or 100 Mbps MII (media independent interface), and the 10 Mbps-only with a seven-wire interface, which uses a subset of the MII signals. The upper 16-bits of the 32-bit external bus interface (EBI) are used to connect to an external Ethernet device. The FEC contains built-in transmit and receive message FIFOs and DMA support.

2 Ordering Information



Temperature Range
M = –40° C to 125° C

Package Identifier
ZP = 416PBGA SnPb
VR = 416PBGA Pb-free
VF = 208MAPBGA SnPb
VM = 208MAPBGA Pb-free
ZQ = 324PBGA SnPb
VZ = 324PBGA Pb-free

Operating Frequency
80 = 80 MHz
112 = 112 MHz
132 = 132 MHz

Tape and Reel Status
R2 = Tape and seel
(blank) = Trays

Qualification Status
P = Pre qualification
M = Full spec qualified

Note: Not all options are available on all devices. Refer to Table 1.

Figure 1. MPC5500 Family Part Number Example

Table 1. Orderable Part Numbers

Freescale Part Number[1]	Package Description	Speed (MHz)		Operating Temperature[2]	
		Nominal	Max[3] (f_{MAX})	Min (T_L)	Max (T_H)
MPC5553MVR132	MPC5553 Lead-free 416 package	132	132	–40° C	125° C
MPC5553MVR112		112	114		
MPC5553MVR80		80	82		
MPC5553MVZ132	MPC5553 Lead-free 324 package	132	132	–40° C	125° C
MPC5553MVZ112		112	114		
MPC5553MVZ80		80	82		
MPC5553MVM132	MPC5553 Lead-free 208 package	132	132	–40° C	125° C
MPC5553MVM112		112	114		
MPC5553MVM80		80	82		
MPC5553MZP132	MPC5553 Lead 416 package	132	132	–40° C	125° C
MPC5553MZP112		112	114		
MPC5553MZP80		80	82		
MPC5553MZQ132	MPC5553 Lead 324 package	132	132	–40° C	125° C
MPC5553MZQ112		112	114		
MPC5553MZQ80		80	82		

Table 1. Orderable Part Numbers (continued)

Freescale Part Number[1]	Package Description	Speed (MHz)		Operating Temperature[2]	
		Nominal	Max[3] (f_{MAX})	Min (T_L)	Max (T_H)
MPC5553MVF132	MPC5553 Lead 208 package	132	132	–40° C	125° C
MPC5553MVF112		112	114		
MPC5553MVF80		80	82		

1 All devices are PPC5553, rather than MPC5553, until the product qualifications. Not all configurations are available in the PPC parts.

2 The lowest operating temperature is referenced by T_L; the highest operating temperature is referenced by T_H.

3 Speed is the nominal maximum frequency. Max speed is the maximum speed allowed including any frequency modulation. 80 MHz parts allow for 80 MHz + 2% modulation. However, 132 MHz devices allow 128 MHz plus two percent frequency modulation only.

3 Electrical Characteristics

This section contains detailed information on power considerations, DC/AC electrical characteristics, and AC timing specifications for the MCU.

3.1 Maximum Ratings

Table 2. Absolute Maximum Ratings[1]

Spec	Characteristic	Symbol	Min	Max[2]	Unit
1	1.5 V core supply voltage [3]	V_{DD}	–0.3	1.7	V
2	Flash program/erase voltage	V_{PP}	–0.3	6.5	V
3	Flash core voltage	V_{DDF}	–0.3	1.7	V
4	Flash read voltage	V_{FLASH}	–0.3	4.6	V
5	SRAM standby voltage	V_{STBY}	–0.3	1.7	V
6	Clock synthesizer voltage	V_{DDSYN}	–0.3	4.6	V
7	3.3 V I/O buffer voltage	V_{DD33}	–0.3	4.6	V
8	Voltage regulator control input voltage	V_{RC33}	–0.3	4.6	V
9	Analog supply voltage (reference to V_{SSA})	V_{DDA}	–0.3	5.5	V
10	I/O supply voltage (fast I/O pads) [4]	V_{DDE}	–0.3	4.6	V
11	I/O supply voltage (slow and medium I/O pads) [4]	V_{DDEH}	–0.3	6.5	V
12	DC input voltage[5] V_{DDEH} powered I/O pads V_{DDE} powered I/O pads	V_{IN}	 –1.0[6] –1.0[6]	 6.5[7] 4.6[8]	V
13	Analog reference high voltage (reference to V_{RL})	V_{RH}	–0.3	5.5	V
14	V_{SS} differential voltage	$V_{SS} - V_{SSA}$	–0.1	0.1	V
15	V_{DD} differential voltage	$V_{DD} - V_{DDA}$	$-V_{DDA}$	V_{DD}	V
16	V_{REF} differential voltage	$V_{RH} - V_{RL}$	–0.3	5.5	V

Table 2. Absolute Maximum Ratings[1] (continued)

Spec	Characteristic	Symbol	Min	Max[2]	Unit
17	V_{RH} to V_{DDA} differential voltage	$V_{RH} - V_{DDA}$	–5.5	5.5	V
18	V_{RL} to V_{SSA} differential voltage	$V_{RL} - V_{SSA}$	–0.3	0.3	V
19	V_{DDEH} to V_{DDA} differential voltage	$V_{DDEH} - V_{DDA}$	$-V_{DDA}$	V_{DDEH}	V
20	V_{DDF} to V_{DD} differential voltage	$V_{DDF} - V_{DD}$	–0.3	0.3	V
21	This spec has been moved to Table 9, spec 43a.				
22	V_{SSSYN} to V_{SS} differential voltage	$V_{SSSYN} - V_{SS}$	–0.1	0.1	V
23	V_{RCVSS} to V_{SS} differential voltage	$V_{RCVSS} - V_{SS}$	–0.1	0.1	V
24	Maximum DC digital input current [9] (per pin, applies to all digital pins)[5]	I_{MAXD}	–2	2	mA
25	Maximum DC analog input current [10] (per pin, applies to all analog pins)	I_{MAXA}	–3	3	mA
26	Maximum operating temperature range [11] Die junction temperature	T_J	T_L	150.0	°C
27	Storage temperature range	T_{STG}	–55.0	150.0	°C
28	Maximum solder temperature [12]	T_{SDR}	—	260.0	°C
29	Moisture sensitivity level [13]	MSL	—	3	

1 Functional operating conditions are given in the DC electrical specifications. Absolute maximum ratings are stress ratings only, and functional operation at the maxima is not guaranteed. Stress beyond the listed maxima can affect device reliability or cause permanent damage to the device.

2 Absolute maximum voltages are currently maximum burn-in voltages. Absolute maximum specifications for device stress have not yet been determined.

3 1.5 V +/– 10% for proper operation. This parameter is specified at a maximum junction temperature of 150 °C.

4 All functional non-supply I/O pins are clamped to V_{SS} and V_{DDE}, or V_{DDEH}.

5 AC signal overshoot and undershoot of up to +/– 2.0 V of the input voltages is permitted for an accumulative duration of 60 hours over the complete lifetime of the device (injection current not limited for this duration).

6 Internal structures hold the voltage greater than –1.0 V if the injection current limit of 2 mA is met. Keep the negative DC current greater than –0.6 V on eTPUB[15] and SINB during the internal power-on reset (POR) state.

7 Internal structures hold the input voltage less than the maximum voltage on all pads powered by V_{DDEH} supplies, if the maximum injection current specification is met (2 mA for all pins) and V_{DDEH} is within the operating voltage specifications.

8 Internal structures hold the input voltage less than the maximum voltage on all pads powered by V_{DDE} supplies, if the maximum injection current specification is met (2 mA for all pins) and V_{DDE} is within the operating voltage specifications.

9 Total injection current for all pins (including both digital and analog) must not exceed 25 mA.

10 Total injection current for all analog input pins must not exceed 15 mA.

11 Lifetime operation at these specification limits is not guaranteed.

12 Solder profile per CDF-AEC-Q100.

13 Moisture sensitivity per JEDEC test method A112.

3.2 Thermal Characteristics

The shaded rows in the following table indicate information specific to a four-layer device.

Table 3. Thermal Characteristics

Spec	MPC5553 Thermal Characteristic	Symbol	Package			Unit
			208 MAPBGA	324 PBGA	416 PBGA	
1	Junction to ambient [1, 2], natural convection (one-layer board)	$R_{\theta JA}$	41	30	29	°C/W
2	Junction to ambient [1, 3], natural convection (four-layer board 2s2p)	$R_{\theta JA}$	25	21	21	°C/W
3	Junction to ambient (@200 ft./min., one-layer board)	$R_{\theta JMA}$	33	24	23	°C/W
4	Junction to ambient (@200 ft./min., four-layer board 2s2p)	$R_{\theta JMA}$	22	17	18	°C/W
5	Junction to board [4] (four-layer board 2s2p)	$R_{\theta JB}$	15	12	13	°C/W
6	Junction to case [5]	$R_{\theta JC}$	7	8	9	°C/W
7	Junction to package top [6], natural convection	Ψ_{JT}	2	2	2	°C/W

1 Junction temperature is a function of on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.

2 Per SEMI G38-87 and JEDEC JESD51-2 with the single-layer board horizontal.

3 Per JEDEC JESD51-6 with the board horizontal.

4 Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.

5 Indicates the average thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1) with the cold plate temperature used for the case temperature.

6 Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

3.2.1 General Notes for Specifications at Maximum Junction Temperature

An estimation of the device junction temperature, T_J, can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature for the package (^oC)

$R_{\theta JA}$ = junction to ambient thermal resistance ($^oC/W$)

P_D = power dissipation in the package (W)

The thermal resistance values used are based on the JEDEC JESD51 series of standards to provide consistent values for estimations and comparisons. The difference between the values determined for the single-layer (1s) board compared to a four-layer board that has two signal layers, a power and a ground plane (2s2p), demonstrate that the effective thermal resistance is not a constant. The thermal resistance depends on the:

- Construction of the application board (number of planes)
- Effective size of the board which cools the component
- Quality of the thermal and electrical connections to the planes
- Power dissipated by adjacent components

Connect all the ground and power balls to the respective planes with one via per ball. Using fewer vias to connect the package to the planes reduces the thermal performance. Thinner planes also reduce the thermal performance. When the clearance between the vias leave the planes virtually disconnected, the thermal performance is also greatly reduced.

As a general rule, the value obtained on a single-layer board is within the normal range for the tightly packed printed circuit board. The value obtained on a board with the internal planes is usually within the normal range if the application board has:

- One oz (35 micron nominal thickness) internal planes
- Components are well separated
- Overall power dissipation on the board is less than 0.02 W/cm^2

The thermal performance of any component depends on the power dissipation of the surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

T_J = junction temperature (^oC)

T_B = board temperature at the package perimeter ($^oC/W$)

$R_{\theta JB}$ = junction-to-board thermal resistance ($^oC/W$) per JESD51-8

P_D = power dissipation in the package (W)

When the heat loss from the package case to the air does not factor into the calculation, an acceptable value for the junction temperature is predictable. Ensure the application board is similar to the thermal test condition, with the component soldered to a board with internal planes.

The thermal resistance is expressed as the sum of a junction-to-case thermal resistance plus a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance ($^oC/W$)

$R_{\theta JC}$ = junction-to-case thermal resistance ($^oC/W$)

$R_{\theta CA}$ = case-to-ambient thermal resistance ($^oC/W$)

$R_{\theta JC}$ is device related and is not affected by other factors. The thermal environment can be controlled to change the case-to-ambient thermal resistance, $R_{\theta CA}$. For example, change the air flow around the device, add a heat sink, change the mounting arrangement on printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device. This description is most useful for packages with heat sinks where 90% of the heat flow is through the case to the heat sink to ambient. For most packages, a better model is required.

A more accurate two-resistor thermal model can be constructed from the junction-to-board thermal resistance and the junction-to-case thermal resistance. The junction-to-case thermal resistance describes when using a heat sink or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed circuit board. This model can be used to generate simple estimations and for computational fluid dynamics (CFD) thermal models.

To determine the junction temperature of the device in the application on a prototype board, use the thermal characterization parameter (Ψ_{JT}) to determine the junction temperature by measuring the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_T = thermocouple temperature on top of the package (^oC)

Ψ_{JT} = thermal characterization parameter ($^oC/W$)

P_D = power dissipation in the package (W)

The thermal characterization parameter is measured in compliance with the JESD51-2 specification using a 40-gauge type T thermocouple epoxied to the top center of the package case. Position the thermocouple so that the thermocouple junction rests on the package. A small amount of epoxy is placed on the thermocouple junction and approximately 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by the cooling effects of the thermocouple wire.

References:

Semiconductor Equipment and Materials International
805 East Middlefield Rd.
Mountain View, CA., 94043
(415) 964-5111

MIL-SPEC and EIA/JESD (JEDEC) specifications are available from Global Engineering Documents at 800-854-7179 or 303-397-7956.

JEDEC specifications are available on the web at http://www.jedec.org.

- 1. C.E. Triplett and B. Joiner, “An Experimental Characterization of a 272 PBGA Within an Automotive Engine Controller Module,” Proceedings of SemiTherm, San Diego, 1998, pp. 47–54.
- 2. G. Kromann, S. Shidore, and S. Addison, “Thermal Modeling of a PBGA for Air-Cooled Applications,” Electronic Packaging and Production, pp. 53–58, March 1998.
- 3. B. Joiner and V. Adams, “Measurement and Simulation of Junction to Board Thermal Resistance and Its Application in Thermal Modeling,” Proceedings of SemiTherm, San Diego, 1999, pp. 212–220.

3.3 Package

The MPC5553 is available in packaged form. Package options are listed in Section 2, "Ordering Information."

Refer to Section 4, "Mechanicals," for pinouts and package drawings.

3.4 EMI (Electromagnetic Interference) Characteristics

Table 4. EMI Testing Specifications[1]

Spec	Characteristic	Minimum	Typical	Maximum	Unit
1	Scan range	0.15	—	1000	MHz
2	Operating frequency	—	—	132	MHz
3	V_{DD} operating voltages	—	1.5	—	V
4	V_{DDSYN}, V_{RC33}, V_{DD33}, V_{FLASH}, V_{DDE} operating voltages	—	3.3	—	V
5	V_{PP}, V_{DDEH}, V_{DDA} operating voltages	—	5.0	—	V
6	Maximum amplitude	—	—	14[2] 32[3]	dBuV
7	Operating temperature	—	—	25	°C

1 EMI testing and I/O port waveforms per SAE J1752/3 issued 1995-03. Qualification testing was performed on the MPC5554 and applied to the MPC5500 family as generic EMI performance data.

2 Measured with single-chip EMI program.

3 Measured with expanded EMI program.

3.5 ESD Characteristics

Table 5. ESD Ratings[1, 2]

Characteristic	Symbol	Value	Unit
ESD for Human Body Model (HBM)		2000	V
HBM circuit description	R1	1500	Ω
	C	100	pF
ESD for Field Induced Charge Model (FDCM)		500 (all pins)	V
		750 (corner pins)	
Number of pulses per pin: Positive pulses (HBM) Negative pulses (HBM)	 — —	 1 1	 — —
Interval of pulses	—	1	second

1 All ESD testing conforms to CDF-AEC-Q100 Stress Test Qualification for Automotive Grade Integrated Circuits.

2 Device failure is defined as: if after exposure to ESD pulses, the device no longer meets the device specification requirements. Complete DC parametric and functional testing will be performed per applicable device specification at room temperature followed by hot temperature, unless specified otherwise in the device specification.

3.6 Voltage Regulator Controller (V_{RC}) and Power-On Reset (POR) Electrical Specifications

Table 6. VRC/POR Electrical Specifications

Spec	Characteristic	Symbol	Min	Max	Units
1	1.5 V (V_{DD}) POR negated (ramp up) 1.5 V (V_{DD}) POR asserted (ramp down)	V_POR15	1.1 1.1	1.35 1.35	V
2	3.3 V (V_{DDSYN}) POR negated (ramp up) 3.3 V (V_{DDSYN}) POR asserted (ramp down)	V_POR33	2.0 2.0	2.85 2.85	V
3	RESET pin supply (V_{DDEH6}) POR negated (ramp up)[1] RESET pin supply (V_{DDEH6}) POR asserted (ramp down)[1]	V_POR5	2.0 2.0	2.85 2.85	V
4	V_{RC33} voltage before the regulator controller allows the pass transistor to start turning on	V_TRANS_START	1.0	2.0	V
5	V_{RC33} voltage when the regulator controller allows the pass transistor to completely turn on[2, 3]	V_TRANS_ON	2.0	2.85	V
6	V_{RC33} voltage greater than the voltage at which the V_{RC} keeps the 1.5 V supply in regulation[4, 5]	$V\_V_{RC33REG}$	3.0	—	V
7	Current can be sourced by V_{RCCTL}	I_V_{RCCTL}[6]			mA
	– 40° C		11.0	—	mA
	25° C		9.0	—	mA
	150° C (Tj)		7.5	—	mA
8	Voltage differential during power up such that: V_{DD33} can lag V_{DDSYN} or V_{DDEH6}, before V_{DDSYN} and V_{DDEH6} reach the V_POR33 and V_POR5 minimums respectively.	V_{DD33}_LAG	—	1.0	V
9	Absolute value of slew rate on power supply pins		—	50	V/ms
10	Required gain: I_{DD} / I_V_{RCCTL} (@V_{DD} = 1.35 V, $f_{sys} = f_{MAX}$)[5, 7]	BETA[8]			
	– 40° C		55.0[9]	—	—
	25° C		58.0[9]	—	—
	150° C (Tj)		70.0[9]	500	—

1 V_{IL_S} (Table 9, Spec15) is guaranteed to scale with V_{DDEH6} down to V_POR5.

2 Supply full operating current for the 1.5 V supply when the 3.3 V supply reaches this range.

3 It is possible to reach the current limit during ramp up—do not treat this event as short circuit current.

4 At peak current for device.

5 Requires compliance with Freescale's recommended board requirements and transistor recommendations. Board signal traces/routing from the V_{RCCTL} package signal to the base of the external pass transistor and between the emitter of the pass transistor to the V_{DD} package signals must have a maximum of 100 nH inductance and minimal resistance (less than 1 Ω). V_{RCCTL} must have a nominal 1 μF phase compensation capacitor to ground. V_{DD} must have a 20 μF (nominal) bulk capacitor (greater than 4 μF over all conditions, including lifetime). Place high-frequency bypass capacitors consisting of eight 0.01 μF, two 0.1 μF, and one 1 μF capacitors around the package on the V_{DD} supply signals.

6 I_VRCCTL is measured at the following conditions: V_{DD} = 1.35 V, V_{RC33} = 3.1 V, V_VRCCTL = 2.2 V.

7 Values are based on IDD from high-use applications as explained in the IDD Electrical Specification.

8 BETA is measured on a per-part basis and is calculated as (IDD ÷ I_VRCCTL), and represents the worst-case external transistor BETA.

[9] Preliminary value. Final specification pending characterization.

3.7 Power-Up/Down Sequencing

Power sequencing between the 1.5 V power supply and V_{DDSYN} or the $\overline{RESET}$ power supplies is required if using an external 1.5 V power supply with V_{RC33} tied to ground (GND). To avoid power-sequencing, V_{RC33} must be powered up within the specified operating range, even if the on-chip voltage regulator controller is nit used. Refer to Section 3.7.2, "Power-Up Sequence (VRC33 Grounded)," and Section 3.7.3, "Power-Down Sequence (VRC33 Grounded)."

Power sequencing requires that V_{DD33} must reach a certain voltage where the values are read as ones before the POR signal negates. Refer to Section 3.7.1, "Input Value of Pins During POR Dependent on VDD33."

Although power sequencing is not required between V_{RC33} and V_{DDSYN} during power up, V_{RC33} must not lead V_{DDSYN} by more than 600 mV or lag by more than 100 mV for the V_{RC} stage turn-on to operate within specification. Higher spikes in the emitter current of the pass transistor occur if V_{RC33} leads or lags V_{DDSYN} by more than these amounts. The value of that higher spike in current depends on the board power supply circuitry and the amount of board level capacitance.

Furthermore, when all of the PORs negate, the system clock starts to toggle, adding another large increase of the current consumed by V_{RC33}. If V_{RC33} lags V_{DDSYN} by more than 100 mV, the increase in current consumed can drop V_{DD} low enough to assert the 1.5 V POR again. Oscillations are possible when the 1.5 V POR asserts and stops the system clock, causing the voltage on V_{DD} to rise until the 1.5 V POR negates again. All oscillations stop when V_{RC33} is powered sufficiently.

When powering down, V_{RC33} and V_{DDSYN} have no delta requirement to each other, because the bypass capacitors internal and external to the device are already charged. When not powering up or down, no delta between V_{RC33} and V_{DDSYN} is required for the V_{RC} to operate within specification.

There are no power up/down sequencing requirements to prevent issues such as latch-up, excessive current spikes, and so on. Therefore, the state of the I/O pins during power up/down varies depending on which supplies are powered.

Table 7 gives the pin state for the sequence cases for all pins with pad type pad_fc (fast type).

Table 7. Power Sequence Pin Status for Fast Pads

V_{DDE}	V_{DD33}	V_{DD}	POR	Pin Status for Fast Pad Output Driver
				pad_fc (fast)
Low	—	—	Asserted	Low
V_{DDE}	Low	Low	Asserted	High
V_{DDE}	Low	V_{DD}	Asserted	High
V_{DDE}	V_{DD33}	Low	Asserted	High impedance (Hi-Z)
V_{DDE}	V_{DD33}	V_{DD}	Asserted	Hi-Z
V_{DDE}	V_{DD33}	V_{DD}	Negated	Functional

Table 8 gives the pin state for the sequence cases for all pins with pad type pad_mh (medium type) and pad_sh (slow type).

Table 8. Power Sequence Pin Status for Medium / Slow Pads

V_{DDEH}	V_{DD}	POR	Pin Status for Medium and Slow Pad Output Driver pad_mh (medium) pad_sh (slow)
Low	—	Asserted	Low
V_{DDEH}	Low	Asserted	High impedance (Hi-Z)
V_{DDEH}	V_{DD}	Asserted	Hi-Z
V_{DDEH}	V_{DD}	Negated	Functional

3.7.1 Input Value of Pins During POR Dependent on V_{DD33}

When powering up the device, V_{DD33} must not lag the latest V_{DDSYN} *or* $\overline{\text{RESET}}$ power pin (V_{DDEH6}) by more than the V_{DD33} lag specification listed in Table 6, spec 8. This avoids accidentally selecting the bypass clock mode because the internal versions of PLLCFG[0:1] and $\overline{\text{RSTCFG}}$ are not powered and therefore cannot read the default state when POR negates. V_{DD33} can lag V_{DDSYN} *or* the $\overline{\text{RESET}}$ power pin (V_{DDEH6}), but cannot lag both by more than the V_{DD33} lag specification. This V_{DD33} lag specification applies during power up only. V_{DD33} has no lead or lag requirements when powering down.

3.7.2 Power-Up Sequence (V_{RC33} Grounded)

The 1.5 V V_{DD} power supply must rise to 1.35 V before the 3.3 V V_{DDSYN} power supply and the $\overline{\text{RESET}}$ power supply rises above 2.0 V. This ensures that digital logic in the PLL for the 1.5 V power supply does not begin to operate below the specified operation range lower limit of 1.35 V. Because the internal 1.5 V POR is disabled, the internal 3.3 V POR or the $\overline{\text{RESET}}$ power POR must hold the device in reset. Since they can negate as low as 2.0 V, V_{DD} must be within specification before the 3.3 V POR and the $\overline{\text{RESET}}$ POR negate.

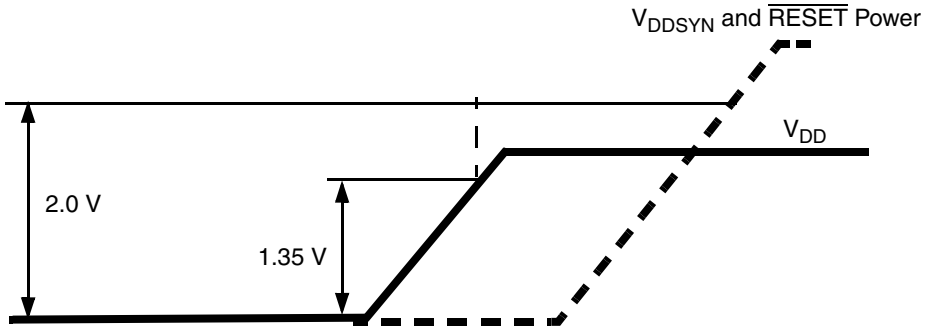


Figure 2. Power-Up Sequence (V_{RC33} Grounded)

3.7.3 Power-Down Sequence (V_{RC33} Grounded)

The only requirement for the power-down sequence when V_{RC33} is grounded is that if V_{DD} decreases to less than its operating range, V_{DDSYN} or the $\overline{\text{RESET}}$ power must decrease to less than 2.0 V before the V_{DD} power is allowed to increase to its operating range. This ensures that the digital 1.5 V logic, which is reset by the ORed POR only and can cause the 1.5 V supply to decrease below its specification, is reset properly.

3.8 DC Electrical Specifications

Table 9. DC Electrical Specifications

Spec	Characteristic	Symbol	Min	Max	Unit
1	Core supply voltage (average DC RMS voltage)	V_{DD}	1.35	1.65	V
2	I/O supply voltage (fast I/O)	V_{DDE}	1.62	3.6	V
3	I/O supply voltage (slow / medium I/O)	V_{DDEH}	3.0	5.25	V
4	3.3 V I/O buffer voltage	V_{DD33}	3.0	3.6	V
5	Voltage regulator control input voltage	V_{RC33}	3.0	3.6	V
6	Analog supply voltage[1]	V_{DDA}	4.5	5.25	V
8	Flash programming voltage[2]	V_{PP}	4.5	5.25	V
9	Flash read voltage	V_{FLASH}	3.0	3.6	V
10	SRAM standby voltage[3]	V_{STBY}	0.8	1.2	V
11	Clock synthesizer operating voltage	V_{DDSYN}	3.0	3.6	V
12	Fast I/O input high voltage	V_{IH_F}	$0.65 \times V_{DDE}$	$V_{DDE} + 0.3$	V
13	Fast I/O input low voltage	V_{IL_F}	$V_{SS} - 0.3$	$0.35 \times V_{DDE}$	V
14	Medium / slow I/O input high voltage	V_{IH_S}	$0.65 \times V_{DDEH}$	$V_{DDEH} + 0.3$	V
15	Medium / slow I/O input low voltage	V_{IL_S}	$V_{SS} - 0.3$	$0.35 \times V_{DDEH}$	V
16	Fast I/O input hysteresis	V_{HYS_F}	$0.1 \times V_{DDE}$		V
17	Medium / slow I/O input hysteresis	V_{HYS_S}	$0.1 \times V_{DDEH}$		V
18	Analog input voltage	V_{INDC}	$V_{SSA} - 0.3$	$V_{DDA} + 0.3$	V
19	Fast I/O output high voltage ($I_{OH\_F} = -2.0$ mA)	V_{OH_F}	$0.8 \times V_{DDE}$	—	V
20	Slow / medium I/O output high voltage ($I_{OH\_S} = -2.0$ mA)	V_{OH_S}	$0.8 \times V_{DDEH}$	—	V
21	Fast I/O output low voltage ($I_{OL\_F} = 2.0$ mA)	V_{OL_F}	—	$0.2 \times V_{DDE}$	V
22	Slow / medium I/O output low voltage ($I_{OL\_S} = 2.0$ mA)	V_{OL_S}	—	$0.2 \times V_{DDEH}$	V
23	Load capacitance (fast I/O)[4] DSC (SIU_PCR[8:9]) = 0b00 DSC (SIU_PCR[8:9]) = 0b01 DSC (SIU_PCR[8:9]) = 0b10 DSC (SIU_PCR[8:9]) = 0b11	C_L	 — — — —	 10 20 30 50	 pF pF pF pF

Table 9. DC Electrical Specifications (continued)

Spec	Characteristic	Symbol	Min	Max	Unit
24	Input capacitance (digital pins)	C_{IN}	—	7	pF
25	Input capacitance (analog pins)	C_{IN_A}	—	10	pF
26	Input capacitance (Shared digital and analog pins AN[12]_MA[0]_$\overline{SDS}$, AN[12]_MA[1]_SDO, AN[14]_MA[2]_SDI, and AN[15]_FCK)	C_{IN_M}	—	12	pF
27a	Operating current[5] 1.5 V supplies @ 132 MHz:				
	V_{DD} (including V_{DDF} max current)[6, 7] @1.65 V typical use	IDD	—	550[9]	mA
	V_{DD} (including V_{DDF} max current)[6, 7] @1.35 V typical use	IDD	—	450[9]	mA
	V_{DD} (including V_{DDF} max current) [7, 8] @1.65 V high use	IDD	—	600[9]	mA
	V_{DD} (including V_{DDF} max current)[7, 8]@1.35 V high use	IDD	—	490[9]	mA
27b	Operating current [5]1.5 V supplies @ 114 MHz:				
	V_{DD} (including V_{DDF} max current)[6, 7]@1.65 V typical use	IDD	—	460[9]	mA
	V_{DD} (including V_{DDF} max current)[6, 7]@1.35 V typical use	IDD	—	380[9]	mA
	V_{DD} (including V_{DDF} max current)[7, 8] @1.65 V high use	IDD	—	520[9]	mA
	V_{DD} (including V_{DDF} max current)[7, 8] @1.35 V high use	IDD	—	420[9]	mA
27c	Operating current[5] 1.5 V supplies @ 82 MHz:				
	V_{DD} (including V_{DDF} max current)[6, 7] @1.65 V typical use	IDD	—	350[9]	mA
	V_{DD} (including V_{DDF} max current)[6, 7] @1.35 V typical use	IDD	—	290[9]	mA
	V_{DD} (including V_{DDF} max current)[7, 8] @1.65 V high use	IDD	—	400[9]	mA
	V_{DD} (including V_{DDF} max current)[7, 8] @1.35 V high use	IDD	—	330[9]	mA
27d	Refer to Figure 3 for an interpolation of this data. [10]				
	IDD_{STBY} @ 25° C				
	V_{STBY} @ 0.8 V	IDD_{STBY}	—	20	µA
	V_{STBY} @ 1.0 V	IDD_{STBY}	—	30	µA
	V_{STBY} @ 1.2 V	IDD_{STBY}	—	50	µA
	IDD_{STBY} @ 60° C				
	V_{STBY} @ 0.8 V	IDD_{STBY}	—	70	µA
	V_{STBY} @ 1.0 V	IDD_{STBY}	—	100	µA
	V_{STBY} @ 1.2 V	IDD_{STBY}	—	200	µA
	IDD_{STBY} @ 150° C (Tj)				
	V_{STBY} @ 0.8 V	IDD_{STBY}	—	1200	µA
	V_{STBY} @ 1.0 V	IDD_{STBY}	—	1500	µA
	V_{STBY} @ 1.2 V	IDD_{STBY}	—	2000	µA
28	Operating current 3.3 V supplies @ 132 MHz				
	V_{DD33}[11]	IDD_{33}	—	2 + (values derived from procedure of Footnote [11])	mA
	V_{FLASH}	I_{VFLASH}	—	10	mA
	V_{DDSYN}	I_{DDSYN}	—	15	mA

Table 9. DC Electrical Specifications (continued)

Spec	Characteristic	Symbol	Min	Max	Unit
29	Operating current 5.0 V supplies (12 MHz ADCLK): V_{DDA} (V_{DDA0} + V_{DDA1}) Analog reference supply current (V_{RH}, V_{RL}) V_{PP}	 IDD_A I_{REF} I_{PP}	 — — —	 20.0 1.0 25.0	 mA mA mA
30	Operating current V_{DDE}[12] supplies: V_{DDEH1} V_{DDE2} V_{DDE3} V_{DDEH4} V_{DDE5} V_{DDEH6} V_{DDE7} V_{DDEH8} V_{DDEH9}	 IDD1 IDD2 IDD3 IDD4 IDD5 IDD6 IDD7 IDD8 IDD9	 — — — — — — — — —	 Refer to Footnote [12]	 mA mA mA mA mA mA mA mA mA
31	Fast I/O weak pullup current[13] 1.62–1.98 V 2.25–2.75 V 3.00–3.60 V	I_{ACT_F}	 10 20 20	 110 130 170	 µA µA µA
	Fast I/O weak pulldown current[13] 1.62–1.98 V 2.25–2.75 V 3.00–3.60 V		 10 20 20	 100 130 170	 µA µA µA
32	Slow / medium I/O weak pullup/down current[14] 3.0–3.6 V 4.5–5.5 V	I_{ACT_S}	 10 20	 150 170	 µA µA
33	I/O input leakage current[15]	I_{INACT_D}	– 2.5	2.5	µA
34	DC injection current (per pin)	I_{IC}	– 2.0	2.0	mA
35	Analog input current, channel off [16]	I_{INACT_A}	–150	150	nA
35a	Analog input current, shared analog / digital pins (AN[12], AN[13], AN[14], AN[15])	I_{INACT_AD}	– 2.5	2.5	µA
36	V_{SS} differential voltage[17]	$V_{SS} – V_{SSA}$	– 100	100	mV
37	Analog reference low voltage	V_{RL}	V_{SSA} – 0.1	V_{SSA} + 0.1	V
38	V_{RL} differential voltage	$V_{RL} – V_{SSA}$	–100	100	mV
39	Analog reference high voltage	V_{RH}	V_{DDA} – 0.1	V_{DDA} + 0.1	V
40	V_{REF} differential voltage	$V_{RH} – V_{RL}$	4.5	5.25	V
41	V_{SSSYN} to V_{SS} differential voltage	$V_{SSSYN} – V_{SS}$	–50	50	mV
42	V_{RCVSS} to V_{SS} differential voltage	$V_{RCVSS} – V_{SS}$	–50	50	mV
43	V_{DDF} to V_{DD} differential voltage[2]	$V_{DDF} – V_{DD}$	–100	100	mV
43a	V_{RC33} to V_{DDSYN} differential voltage	$V_{RC33} – V_{DDSYN}$	–0.1	0.1[18]	V

Table 9. DC Electrical Specifications (continued)

Spec	Characteristic	Symbol	Min	Max	Unit
44	Analog input differential signal range (with common mode 2.5 V)	V_{IDIFF}	– 2.5	2.5	V
45	Operating temperature range, ambient (packaged)	$T_A = (T_L$ to $T_H)$	T_L	T_H	°C
46	Slew rate on power-supply pins	—	—	50	V/ms

1 $| V_{DDA0} - V_{DDA1} |$ must be < 0.1 V.

2 V_{PP} can drop to 3.0 V during read operations.

3 During standby operation, if standby operation is not required, connect V_{STBY} to ground.

4 Applies to CLKOUT, external bus pins, and Nexus pins.

5 Maximum average RMS DC current.

6 Average current measured on Automotive benchmark.

7 Peak currents can be higher on specialized code.

8 High use current measured while running optimized SPE assembly code with all code and data 100% locked in cache (0% miss rate) with all channels of the eMIOS and eTPU running autonomously, plus the eDMA transferring data continuously from SRAM to SRAM. Higher currents are possible if an idle loop that crosses cache lines is run from cache. Design and write code to avoid this condition.

9 Preliminary. Final specification pending characterization.

10 Figure 3 shows an illustration of the IDD_{STBY} values interpolated for these temperature values.

11 Power requirements for the V_{DD33} supply depend on the frequency of operation and load of all I/O pins, and the voltages on the I/O segments. Refer to Table 11 for values to calculate power dissipation for specific operation.

12 Power requirements for each I/O segment are dependent on the frequency of operation and load of the I/O pins on a particular I/O segment, and the voltage of the I/O segment. Refer to Table 10 for values to calculate power dissipation for specific operation. The total power consumption of an I/O segment is the sum of the individual power consumptions for each pin on the segment.

13 Absolute value of current, measured at V_{IL} and V_{IH}.

14 Absolute value of current, measured at V_{IL} and V_{IH}.

15 Weak pullup/down inactive. Measured at $V_{DDE} = 3.6$ V and $V_{DDEH} = 5.25$ V. Applies to pad types: pad_fc, pad_sh, and pad_mh.

16 Maximum leakage occurs at maximum operating temperature. Leakage current decreases by approximately one-half for each 8 °C to 12 °C, in the ambient temperature range of 50 °C to 125 °C. Applies to pad types: pad_a and pad_ae.

17 V_{SSA} refers to both V_{SSA0} and V_{SSA1}. $| V_{SSA0} - V_{SSA1} |$ must be < 0.1 V.

18 Up to 0.6 V during power up and power down.

Figure 3 shows an approximate interpolation of the I_{STBY} worst-case specification to help estimate the values at different voltages and temperatures. The vertical lines inside the graph show the actual specifications listed in Table 9. Refer to the IDD_{STBY} specifications (27d) in Table 9 for more information.

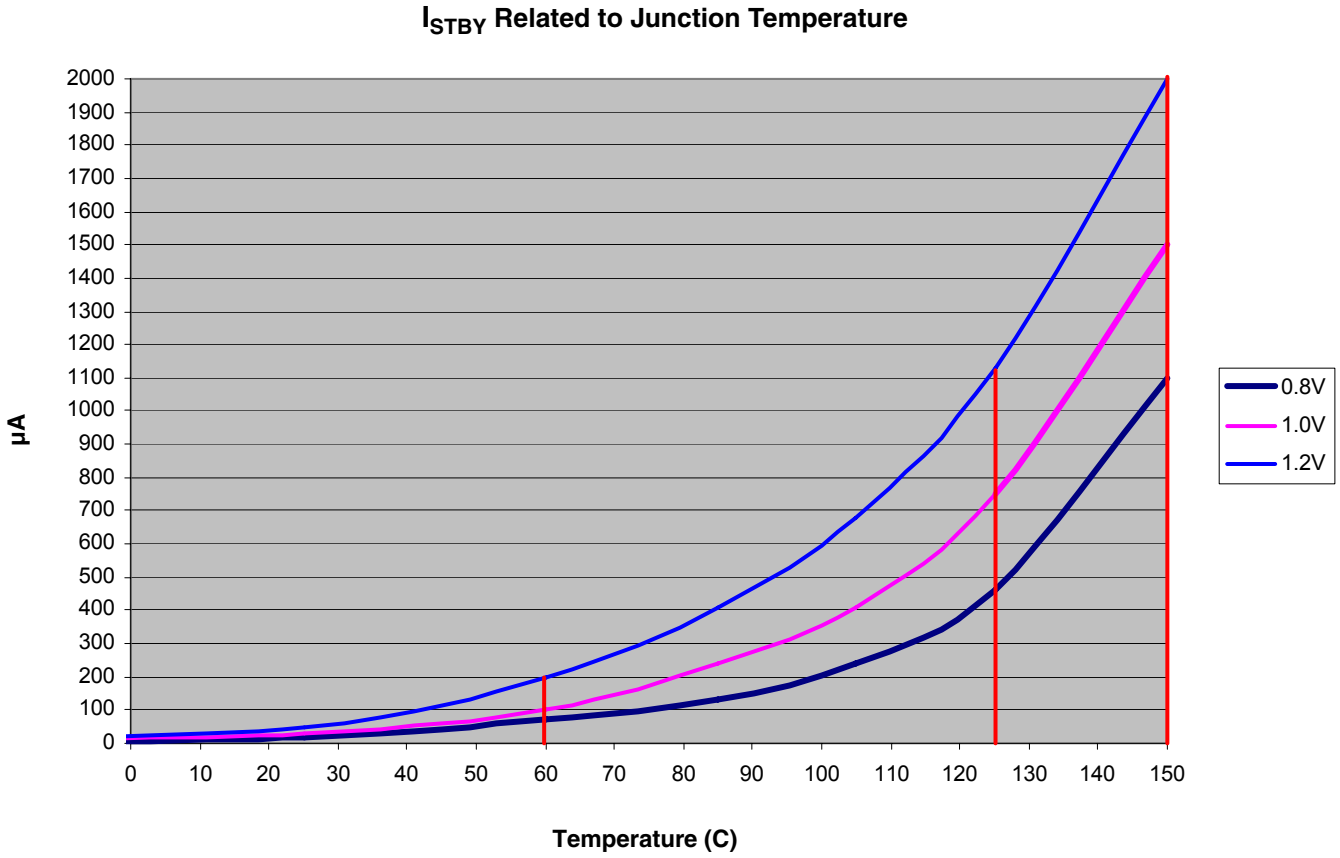


Figure 3. I_{STBY} Worst-case Specifications

3.8.1 I/O Pad Current Specifications

The power consumption of an I/O segment depends on the usage of the pins on a particular segment. The power consumption is the sum of all output pin currents for a particular segment. The output pin current can be calculated from Table 10 based on the voltage, frequency, and load on the pin. Use linear scaling to calculate pin currents for voltage, frequency, and load parameters that fall outside the values given in Table 10.

Table 10. I/O Pad Average DC Current[1]

Spec	Pad Type	Symbol	Frequency (MHz)	Load[2] (pF)	Voltage (V)	Drive Select / Slew Rate Control Setting	Current (mA)
1	Slow	IDRV_SH	25	50	5.25	11	8.0
2			10	50	5.25	01	3.2
3			2	50	5.25	00	0.7
4			2	200	5.25	00	2.4
5	Medium	IDRV_MH	50	50	5.25	11	17.3
6			20	50	5.25	01	6.5
7			3.33	50	5.25	00	1.1
8			3.33	200	5.25	00	3.9
9	Fast	IDRV_FC	66	10	3.6	00	2.8
10			66	20	3.6	01	5.2
11			66	30	3.6	10	8.5
12			66	50	3.6	11	11.0
13			66	10	1.98	00	1.6
14			66	20	1.98	01	2.9
15			66	30	1.98	10	4.2
16			66	50	1.98	11	6.7
17			56	10	3.6	00	2.4
18			56	20	3.6	01	4.4
19			56	30	3.6	10	7.2
20			56	50	3.6	11	9.3
21			56	10	1.98	00	1.3
22			56	20	1.98	01	2.5
23			56	30	1.98	10	3.5
24			56	50	1.98	11	5.7
25			40	10	3.6	00	1.7
26			40	20	3.6	01	3.1
27			40	30	3.6	10	5.1
28			40	50	3.6	11	6.6
29			40	10	1.98	00	1.0
30			40	20	1.98	01	1.8
31			40	30	1.98	10	2.5
32			40	50	1.98	11	4.0

1 These values are estimates from simulation and are not tested. Currents apply to output pins only.

2 All loads are lumped.

3.8.2 I/O Pad V_{DD33} Current Specifications

The power consumption of the V_{DD33} supply dependents on the usage of the pins on all I/O segments. The power consumption is the sum of all input and output pin V_{DD33} currents for all I/O segments. The output pin V_{DD33} current can be calculated from Table 11 based on the voltage, frequency, and load on all fast (pad_fc) pins. The input pin V_{DD33} current can be calculated from Table 11 based on the voltage, frequency, and load on all pad_sh and pad_sh pins. Use linear scaling to calculate pin currents for voltage, frequency, and load parameters that fall outside the values given in Table 11.

Table 11. V_{DD33} Pad Average DC Current[1]

Spec	Pad Type	Symbol	Frequency (MHz)	Load[2] (pF)	V_{DD33} (V)	V_{DDE} (V)	Drive Select	Current (mA)
Inputs								
1	Slow	I33_SH	66	0.5	3.6	5.5	NA	0.003
2	Medium	I33_MH	66	0.5	3.6	5.5	NA	0.003
Outputs								
3	Fast	I33_FC	66	10	3.6	3.6	00	0.35
4			66	20	3.6	3.6	01	0.53
5			66	30	3.6	3.6	10	0.62
6			66	50	3.6	3.6	11	0.79
7			66	10	3.6	1.98	00	0.35
8			66	20	3.6	1.98	01	0.44
9			66	30	3.6	1.98	10	0.53
10			66	50	3.6	1.98	11	0.7
11			56	10	3.6	3.6	00	0.30
12			56	20	3.6	3.6	01	0.45
13			56	30	3.6	3.6	10	0.52
14			56	50	3.6	3.6	11	0.67
15			56	10	3.6	1.98	00	0.30
16			56	20	3.6	1.98	01	0.37
17			56	30	3.6	1.98	10	0.45
18			56	50	3.6	1.98	11	0.60
19			40	10	3.6	3.6	00	0.21
20			40	20	3.6	3.6	01	0.31
21			40	30	3.6	3.6	10	0.37
22			40	50	3.6	3.6	11	0.48
23			40	10	3.6	1.98	00	0.21
24			40	20	3.6	1.98	01	0.27
25			40	30	3.6	1.98	10	0.32
26			40	50	3.6	1.98	11	0.42

[1] These values are estimated from simulation and not tested. Currents apply to output pins for the fast pads only and to input pins for the slow and medium pads only.

[2] All loads are lumped.

3.9 Oscillator and FMPLL Electrical Characteristics

Table 12. FMPLL Electrical Specifications

(V_{DDSYN} = 3.0–3.6 V; V_{SS} = V_{SSSYN} = 0.0 V; T_A = T_L to T_H)

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
1	PLL reference frequency range: Crystal reference External reference Dual controller (1:1 mode)	 $f_{ref\_crystal}$ f_{ref_ext} $f_{ref\_1:1}$	 8 8 24	 20 20 $f_{sys} \div 2$	MHz
2	System frequency [1]	f_{sys}	$f_{ICO(MIN)} \div 2^{RFD}$	f_{MAX} [2]	MHz
3	System clock period	t_{CYC}	—	$1 \div f_{sys}$	ns
4	Loss of reference frequency [3]	f_{LOR}	100	1000	kHz
5	Self clocked mode (SCM) frequency [4]	f_{SCM}	7.4	17.5	MHz
6	EXTAL input high voltage crystal mode [5] All other modes (dual controller (1:1), bypass, external reference)	V_{IHEXT} V_{IHEXT}	V_{XTAL} + 0.4 V [($V_{DDE5} \div 2$) + 0.4 V]	— —	V V
7	EXTAL input low voltage crystal mode [6] All other modes (dual controller (1:1), bypass, external reference)	V_{ILEXT} V_{ILEXT}	— —	V_{XTAL} – 0.4 V [($V_{DDE5} \div 2$) – 0.4 V]	V V
8	XTAL current [7]	I_{XTAL}	0.8	3	mA
9	Total on-chip stray capacitance on XTAL	C_{S_XTAL}	—	1.5	pF
10	Total on-chip stray capacitance on EXTAL	C_{S_EXTAL}	—	1.5	pF
11	Crystal manufacturer's recommended capacitive load	C_L	Refer to crystal specification	Refer to crystal specification	pF
12	Discrete load capacitance to connect to EXTAL	C_{L_EXTAL}	—	$(2 \times C_L) - C_{S\_EXTAL} - C_{PCB\_EXTAL}$ [8]	pF
13	Discrete load capacitance to connect to XTAL	C_{L_XTAL}	—	$(2 \times C_L) - C_{S\_XTAL} - C_{PCB\_XTAL}$ [8]	pF
14	PLL lock time[9]	t_{lpll}	—	750	µs
15	Dual controller (1:1) clock skew (between CLKOUT and EXTAL) [10, 11]	t_{skew}	–2	2	ns
16	Duty cycle of reference	t_{DC}	40	60	%
17	Frequency un-LOCK range	f_{UL}	– 4.0	4.0	% f_{SYS}
18	Frequency LOCK range	f_{LCK}	– 2.0	2.0	% f_{SYS}

Table 12. FMPLL Electrical Specifications (continued)

(V_{DDSYN} = 3.0–3.6 V; V_{SS} = V_{SSSYN} = 0.0 V; T_A = T_L to T_H)

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
19	CLKOUT period jitter,[12, 13] measured at f_{SYS} maximum peak-to-peak jitter (clock edge to clock edge) Long term jitter (averaged over 2 ms interval)	C_{JITTER}	 — —	 5.0 0.01	% f_{CLKOUT}
20	Frequency modulation range limit [14] (do not exceed f_{sys} maximum)	C_{MOD}	0.8	2.4	$\%f_{SYS}$
21	ICO frequency $f_{ico} = [f_{ref} \times (MFD + 4)] \div (PREDIV + 1)$[15]	f_{ico}	48	f_{sys}	MHz
22	Predivider output frequency (to PLL)	f_{PREDIV}	4	f_{MAX}	MHz

1 All internal registers retain data at 0 Hz.

2 Up to the maximum frequency rating of the device (refer to Table 1).

3 Loss of reference frequency is defined as the reference frequency detected internally, which transitions the PLL into self-clocked mode.

4 The PLL operates at self-clocked mode (SCM) frequency when the reference frequency falls below f_{LOR}. SCM frequency is measured on the CLKOUT ball with the divider set to divide-by-two of the system clock.
NOTE: In SCM, the MFD and PREDIV have no effect and the RFD is bypassed.

5 Use the EXTAL input high voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators). ($V_{extal} - V_{xtal}$) must be ≥ 400 mV for the oscillator's comparator to produce the output clock.

6 Use the EXTAL input low voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators). ($V_{xtal} - V_{extal}$) must be ≥ 400 mV for the oscillator's comparator to produce the output clock.

7 I_{xtal} is the oscillator bias current out of the XTAL pin with both EXTAL and XTAL pins grounded.

8 C_{PCB_EXTAL} and C_{PCB_XTAL} are the measured PCB stray capacitances on EXTAL and XTAL, respectively.

9 This specification applies to the period required for the PLL to relock after changing the MFD frequency control bits in the synthesizer control register (SYNCR). From power up with crystal oscillator reference, the lock time also includes the crystal startup time.

10 PLL is operating in 1:1 PLL mode.

11 V_{DDE} = 3.0–3.6 V

12 Jitter is the average deviation from the programmed frequency measured over the specified interval at maximum f_{sys}. Measurements are made with the device powered by filtered supplies and clocked by a stable external clock signal. Noise injected into the PLL circuitry via V_{DDSYN} and V_{SSSYN} and variation in crystal oscillator frequency increase the jitter percentage for a given interval. CLKOUT divider is set to divide-by-two.

13 Values are with frequency modulation disabled. If frequency modulation is enabled, jitter is the sum of (jitter + Cmod).

14 Modulation depth selected must not result in f_{sys} value greater than the f_{sys} maximum specified value.

15 $f_{sys} = f_{ico} \div (2^{RFD})$.

3.10 eQADC Electrical Characteristics

Table 13. eQADC Conversion Specifications (Operating)

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
1	ADC clock (ADCLK) frequency[1]	F_{ADCLK}	1	12	MHz
2	Conversion cycles Differential Single ended	CC	 13 + 2 (15) 14 + 2 (16)	 13 + 128 (141) 14 + 128 (142)	ADCLK cycles
3	Stop mode recovery time[2]	T_{SR}	10	—	µs
4	Resolution[3]	—	1.25	—	mV
5	INL: 6 MHz ADC clock	INL6	–4	4	Counts[3]
6	INL: 12 MHz ADC clock	INL12	–8	8	Counts
7	DNL: 6 MHz ADC clock	DNL6	–3 [4]	3 [4]	Counts
8	DNL: 12 MHz ADC clock	DNL12	–6 [4]	6[4]	Counts
9	Offset error with calibration	OFFWC	–4 [5]	4 [5]	Counts
10	Full-scale gain error with calibration	GAINWC	–8 [6]	8 [6]	Counts
11	Disruptive input injection current [7, 8, 9, 10]	I_{INJ}	–1	1	mA
12	Incremental error due to injection current. All channels have same 10 kΩ < Rs <100 kΩ Channel under test has Rs = 10 kΩ, $I_{INJ} = I_{INJMAX}$, I_{INJMIN}	E_{INJ}	–4	4	Counts
13	Total Unadjusted Error for single ended conversions with calibration[11, 12, 13, 14, 15]	TUE	–4	4	Counts

1 Conversion characteristics vary with F_{ADCLK} rate. Reduced conversion accuracy occurs at maximum F_{ADCLK} rate. The maximum value is based on 800 KS/s and the minimum value is based on 20 MHz oscillator clock frequency divided by a maximum 16 factor.

2 Stop mode recovery time begins when the ADC control register enable bits are set until the ADC is ready to perform conversions.

3 At $V_{RH} - V_{RL}$ = 5.12 V, one least significant bit (LSB) = 1.25, mV = one count.

4 Guaranteed 10-bit monotonicity.

5 The absolute value of the offset error without calibration ≤ 100 counts.

6 The absolute value of the full scale gain error without calibration ≤ 120 counts.

7 Below disruptive current conditions, the channel being stressed has conversion values of: 0x3FF for analog inputs greater than V_{RH}, and 0x000 for values less than V_{RL}. This assumes that $V_{RH} \leq V_{DDA}$ and $V_{RL} \geq V_{SSA}$ due to the presence of the sample amplifier. Other channels are not affected by non-disruptive conditions.

8 Exceeding the limit can cause a conversion error on both stressed and unstressed channels. Transitions within the limit do not affect device reliability or cause permanent damage.

9 Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values using $V_{POSCLAMP} = V_{DDA}$ + 0.5 V and $V_{NEGCLAMP}$ = – 0.3 V, then use the larger of the calculated values.

10 Condition applies to two adjacent pads on the internal pad.

11 The TUE specification is always less than the sum of the INL, DNL, offset, and gain errors due to canceling errors.

12 TUE does not apply to differential conversions.

13 Measured at 6 MHz ADC clock. TUE with a 12 MHz ADC clock is: –16 counts < TUE < 16 counts.

14 TUE includes all internal device errors such as internal reference variation (75% Ref, 25% Ref).

15 Depending on the input impedance, the analog input leakage current (DC Electrical specification 35a) can affect the actual TUE measured on analog channels AN[12], AN[13], AN[14], AN[15].

3.11 H7Fa Flash Memory Electrical Characteristics

Table 14. Flash Program and Erase Specifications[1]

Spec	Flash Program Characteristic	Symbol	Min	Typical	Initial Max[2]	Max[3]	Unit
3	Doubleword (64 bits) program time[4]	$T_{dwprogram}$	—	10	—	500	µs
4	Page program time[4]	$T_{pprogram}$	—	22	44[5]	500	µs
7	16 Kbyte block pre-program and erase time	$T_{16kpperase}$	—	325	525	5000	ms
9	48 Kbyte block pre-program and erase time	$T_{48kpperase}$	—	435	525	5000	ms
10	64 Kbyte block pre-program and erase time	$T_{64kpperase}$	—	525	675	5000	ms
8	128 Kbyte block pre-program and erase time	$T_{128kpperase}$	—	675	1800	15,000	ms
11	Minimum operating frequency for program and erase operations[6]	—	25	—	—	—	MHz

1 Typical program and erase times assume nominal supply values and operation at 25 °C.

2 Initial factory condition: ≤ 100 program/erase cycles, 25 °C, typical supply voltage, 80 MHz minimum system frequency.

3 The maximum erase time occurs after the specified number of program/erase cycles. This maximum value is characterized but not guaranteed.

4 Actual hardware programming times. This does not include software overhead.

5 Page size is 256 bits (8 words).

6 Read frequency of the flash can be up to the maximum operating frequency of the device. There is no minimum read frequency condition.

Table 15. Flash EEPROM Module Life (Full Temperature Range)

Spec	Characteristic	Symbol	Min	Typical[1]	Unit
1a	Number of program/erase cycles per block for 16 Kbyte, 48 Kbyte, and 64 Kbyte blocks over the operating temperature range (T_J)	P/E	100,000	—	cycles
1b	Number of program/erase cycles per block for 128 Kbyte blocks over the operating temperature range (T_J)	P/E	10,000	100,000	cycles
2	Data retention Blocks with 0–1,000 P/E cycles Blocks with 1,001–100,000 P/E cycles	Retention	 20 5	—	years

1 Typical endurance is evaluated at 25° C. Product qualification is performed to the minimum specification. For additional information on the Freescale definition of typical endurance, refer to engineering bulletin EB619 *Typical Endurance for Nonvolatile Memory*.

Table 16 shows the FLASH_BIU settings versus frequency of operation. Refer to the device reference manual for definitions of these bit fields.

Table 16. FLASH_BIU Settings vs. Frequency of Operation

Maximum Frequency (MHz)	APC	RWSC	WWSC	DPFEN	IPFEN	PFLIM	BFEN
Up to and including 82 MHz[1]	0b001	0b001	0b01	0b00, 0b01, or 0b11[2]	0b00, 0b01, or 0b11[2]	0b000 to 0b110[3]	0b0, 0b1[4]
Up to and including 102 MHz[5]	0b001	0b010	0b01	0b00, 0b01, or 0b11[2]	0b00, 0b01, or 0b11[2]	0b000 to 0b110[3]	0b0, 0b1[4]
Up to and including132 MHz[6]	0b010	0b011	0b01	0b00, 0b01, or 0b11[2]	0b00, 0b01, or 0b11[2]	0b000 to 0b110[3]	0b0, 0b1[4]
Default setting after reset	0b111	0b111	0b11	0b00	0b00	0b000	0b0

1 Allows for 80 MHz system clock with 2% frequency modulation.
2 For maximum flash performance, set to 0b11.
3 For maximum flash performance, set to 0b110.
4 For maximum flash performance, set to 0b1.
5 Allows for 100 MHz system clock with 2% frequency modulation.
6 Allows for 128 MHz system clock with 2% frequency modulation.

3.12 AC Specifications

3.12.1 Pad AC Specifications

Table 17. Pad AC Specifications (V_{DDEH} = 5.0 V, V_{DDE} = 1.8 V)[1]

Spec	Pad	SRC/DSC (binary)	Out Delay[2, 3, 4] (ns)	Rise / Fall[4, 5] (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	26	15	50
			82	60	200
		01	75	40	50
			137	80	200
		00	377	200	50
			476	260	200
2	Medium high voltage (MH)	11	16	8	50
			43	30	200
		01	34	15	50
			61	35	200
		00	192	100	50
			239	125	200

Table 17. Pad AC Specifications (V_{DDEH} = 5.0 V, V_{DDE} = 1.8 V)[1] (continued)

Spec	Pad	SRC/DSC (binary)	Out Delay[2, 3, 4] (ns)	Rise / Fall[4, 5] (ns)	Load Drive (pF)
3	Fast	00	3.1	2.7	10
		01		2.5	20
		10		2.4	30
		11		2.3	50
4	Pullup/down (3.6 V max)	—	—	7500	50
5	Pullup/down (5.5 V max)	—	—	9000	50

1 These are worst case values that are estimated from simulation and not tested. The values in the table are simulated at F_{SYS} = 132 MHz, V_{DD} = 1.35–1.65 V, V_{DDE} = 1.62–1.98 V, V_{DDEH} = 4.5–5.5 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, $T_A = T_L$ to T_H.

2 This parameter is supplied for reference and is guaranteed by design and tested.

3 Out delay is shown in Figure 4. Add a maximum of one system clock to the output delay for delay with respect to system clock.

4 Delay and rise and fall are measured to 20% or 80% of the respective signal.

5 This parameter is guaranteed by characterization before qualification rather than 100% tested.

Table 18. De-rated Pad AC Specifications (V_{DDEH} = 3.3 V, V_{DDE} = 3.3 V)[1]

Spec	Pad	SRC/DSC (binary)	Out Delay[2, 3, 4] (ns)	Rise/Fall[3, 5] (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	39	23	50
			120	87	200
		01	101	52	50
			188	111	200
		00	507	248	50
			597	312	200
2	Medium high voltage (MH)	11	23	12	50
			64	44	200
		01	50	22	50
			90	50	200
		00	261	123	50
			305	156	200
3	Fast	00	3.2	2.4	10
		01		2.2	20
		10		2.1	30
		11		2.1	50
4	Pullup/down (3.6 V max)	—	—	7500	50
5	Pullup/down (5.5 V max)	—	—	9500	50

1 These are worst-case values that are estimated from simulation and not tested. The values in the table are simulated at: F_{SYS} = 132 MHz; V_{DD} = 1.35–1.65 V; V_{DDE} = 3.0–3.6 V; V_{DDEH} = 3.0–3.6 V; V_{DD33} and V_{DDSYN} = 3.0–3.6 V; and $T_A = T_L$ to T_H.

2 This parameter is supplied for reference and is guaranteed by design and tested.
3 The delay, and the rise and fall, are measured to 20% or 80% of the respective signal.
4 Out delay is shown in Figure 4. Add a maximum of one system clock to the output delay for delay with respect to system clock.
5 This parameter is guaranteed by characterization before qualification rather than 100% tested.

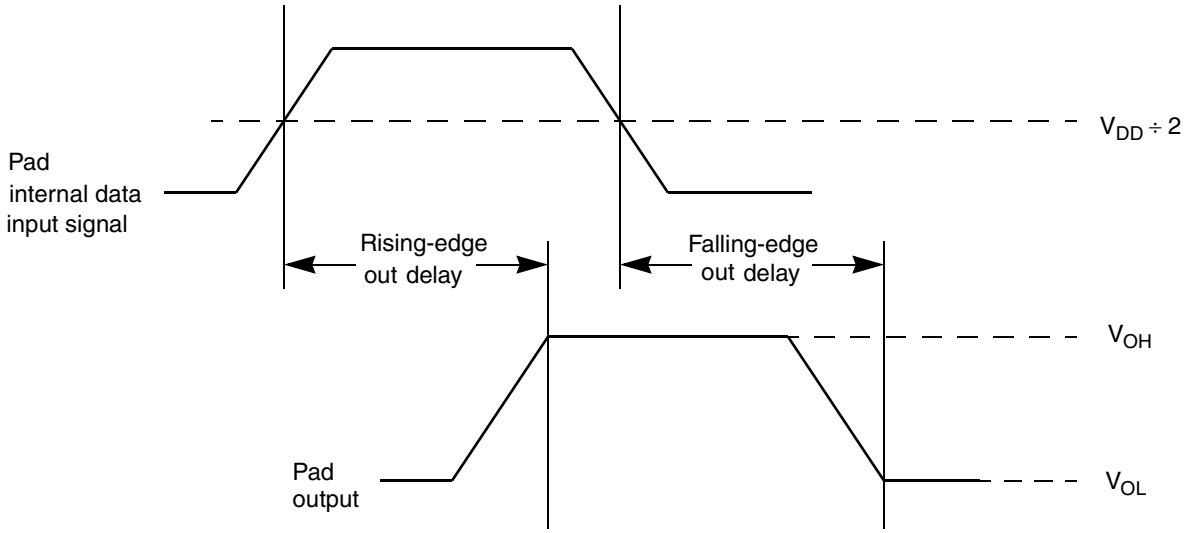


Figure 4. Pad Output Delay

3.13 AC Timing

3.13.1 Reset and Configuration Pin Timing

Table 19. Reset and Configuration Pin Timing[1]

Spec	Characteristic	Symbol	Min	Max	Unit
1	$\overline{\text{RESET}}$ pulse width	t_{RPW}	10	—	t_{CYC}
2	$\overline{\text{RESET}}$ glitch detect pulse width	t_{GPW}	2	—	t_{CYC}
3	PLLCFG, BOOTCFG, WKPCFG, $\overline{\text{RSTCFG}}$ setup time to $\overline{\text{RSTOUT}}$ valid	t_{RCSU}	10	—	t_{CYC}
4	PLLCFG, BOOTCFG, WKPCFG, $\overline{\text{RSTCFG}}$ hold time from $\overline{\text{RSTOUT}}$ valid	t_{RCH}	0	—	t_{CYC}

1 Reset timing specified at: F_{SYS} = 132 MHz; V_{DDEH} = 3.0–5.25 V; V_{DD} = 1.35–1.65 V; and $T_A = T_L$ to T_H.

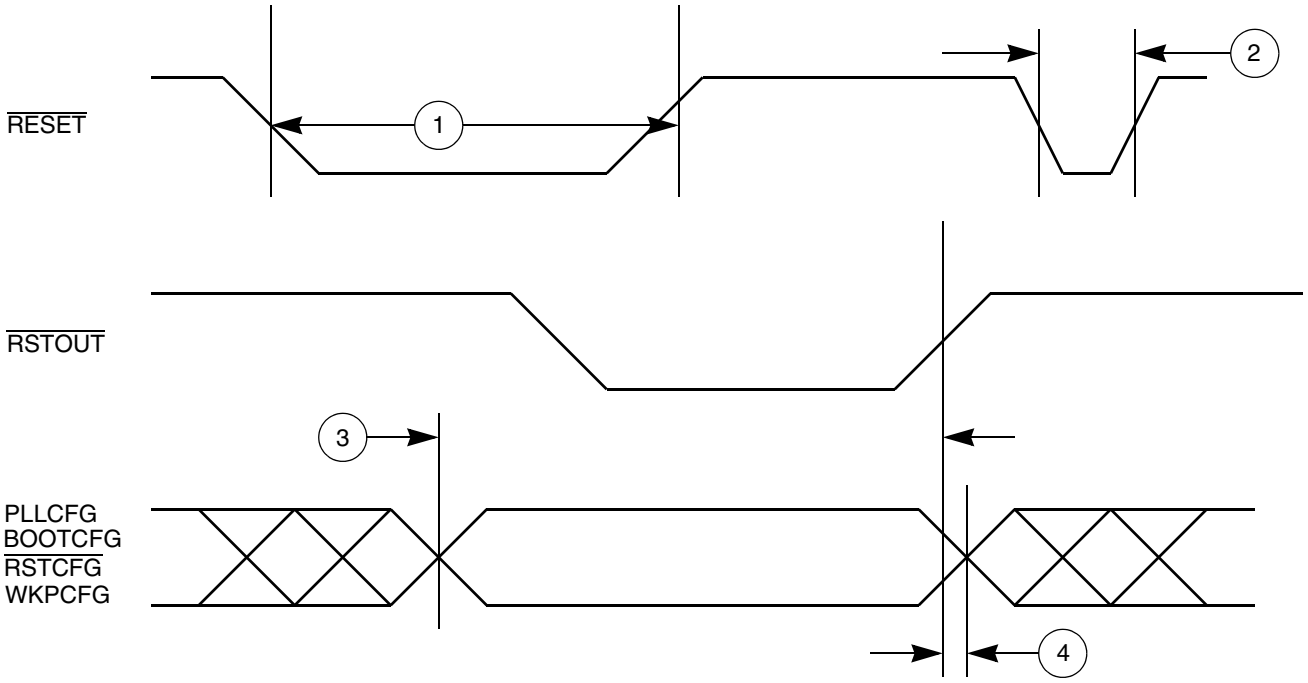


Figure 5. Reset and Configuration Pin Timing

3.13.2 IEEE 1149.1 Interface Timing

Table 20. JTAG Pin AC Electrical Characteristics[1]

Spec	Characteristic	Symbol	Min	Max	Unit
1	TCK cycle time	t_{JCYC}	100	—	ns
2	TCK clock pulse width (measured at $V_{DDE} \div 2$)	t_{JDC}	40	60	ns
3	TCK rise and fall times (40% to 70%)	$t_{TCKRISE}$	—	3	ns
4	TMS, TDI data setup time	t_{TMSS}, t_{TDIS}	5	—	ns
5	TMS, TDI data hold time	t_{TMSH}, t_{TDIH}	25	—	ns
6	TCK low to TDO data valid	t_{TDOV}	—	20	ns
7	TCK low to TDO data invalid	t_{TDOI}	0	—	ns
8	TCK low to TDO high impedance	t_{TDOHZ}	—	20	ns
9	JCOMP assertion time	t_{JCMPPW}	100	—	ns
10	JCOMP setup time to TCK low	t_{JCMPS}	40	—	ns
11	TCK falling-edge to output valid	t_{BSDV}	—	50	ns
12	TCK falling-edge to output valid out of high impedance	t_{BSDVZ}	—	50	ns
13	TCK falling-edge to output high impedance (Hi-Z)	t_{BSDHZ}	—	50	ns
14	Boundary scan input valid to TCK rising-edge	t_{BSDST}	50	—	ns
15	TCK rising-edge to boundary scan input invalid	t_{BSDHT}	50	—	ns

[1] These specifications apply to JTAG boundary scan only. JTAG timing specified at V_{DD} = 1.35–1.65 V, V_{DDE} = 3.0–3.6 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, $T_A = T_L$ to T_H, and CL = 30 pF with DSC = 0b10, SRC = 0b11. Refer to Table 21 for functional specifications.

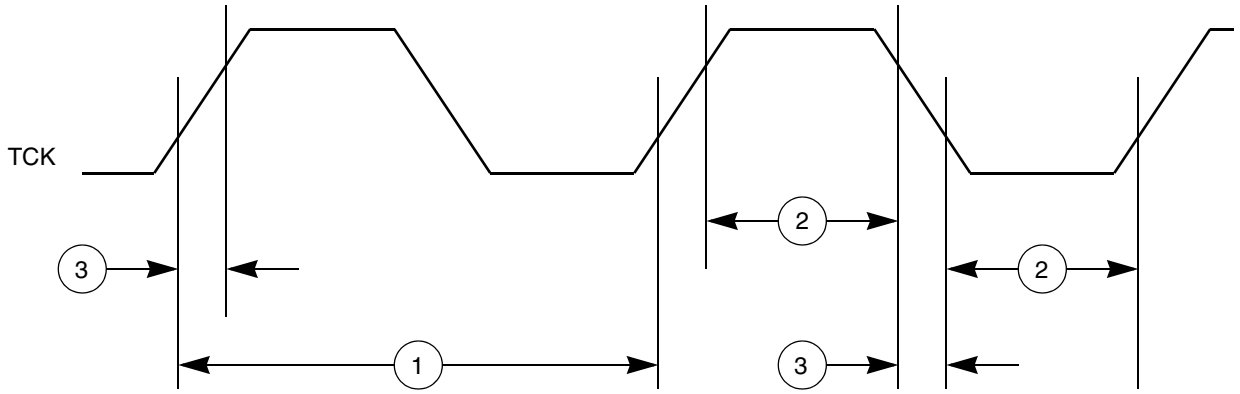


Figure 6. JTAG Test Clock Input Timing

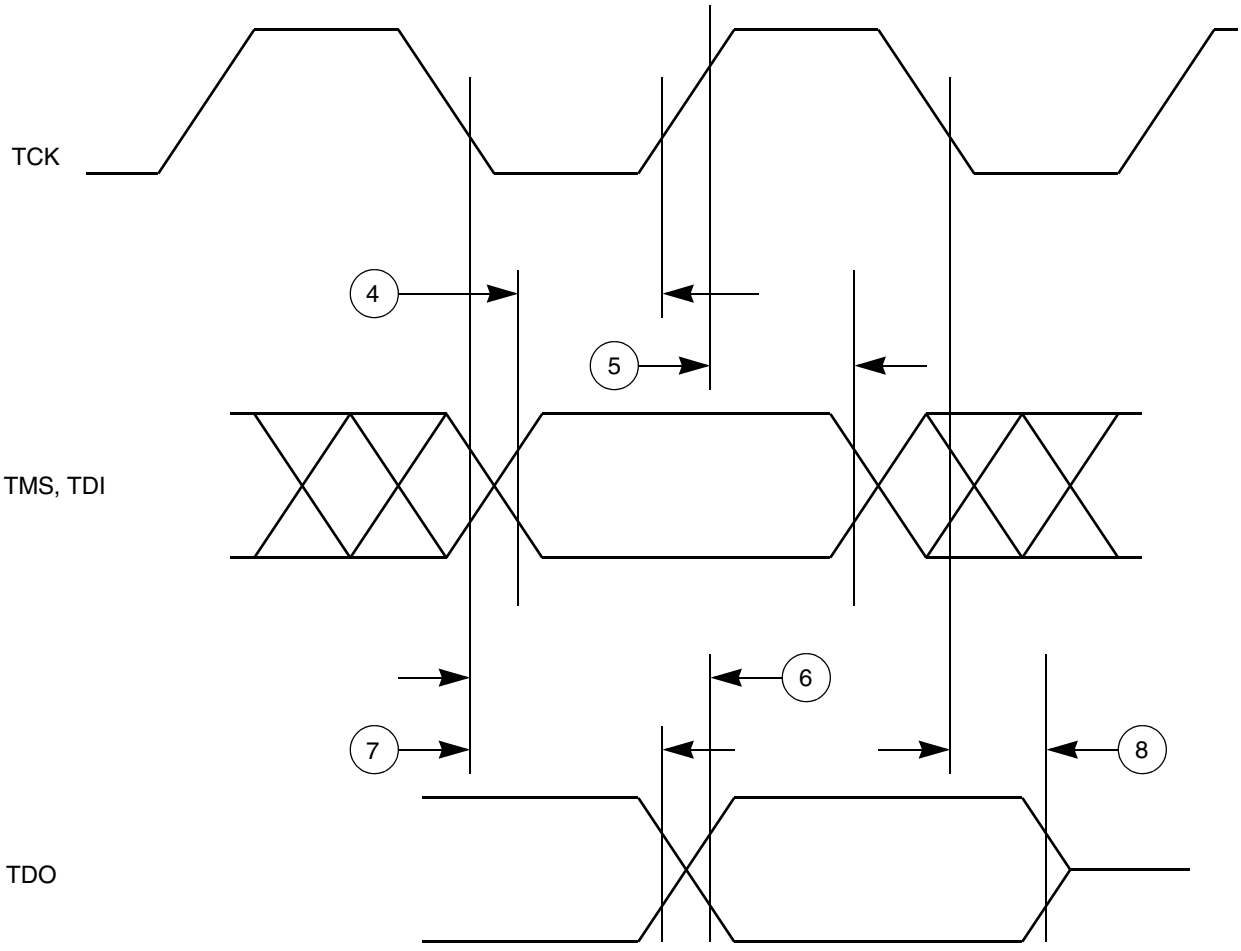


Figure 7. JTAG Test Access Port Timing

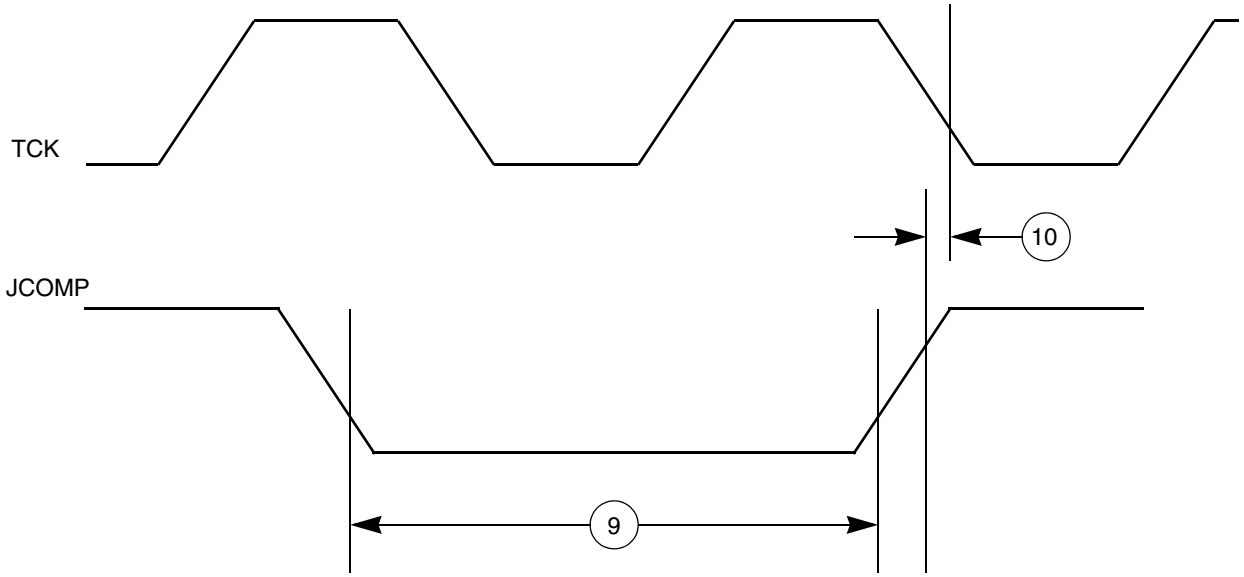


Figure 8. JTAG JCOMP Timing

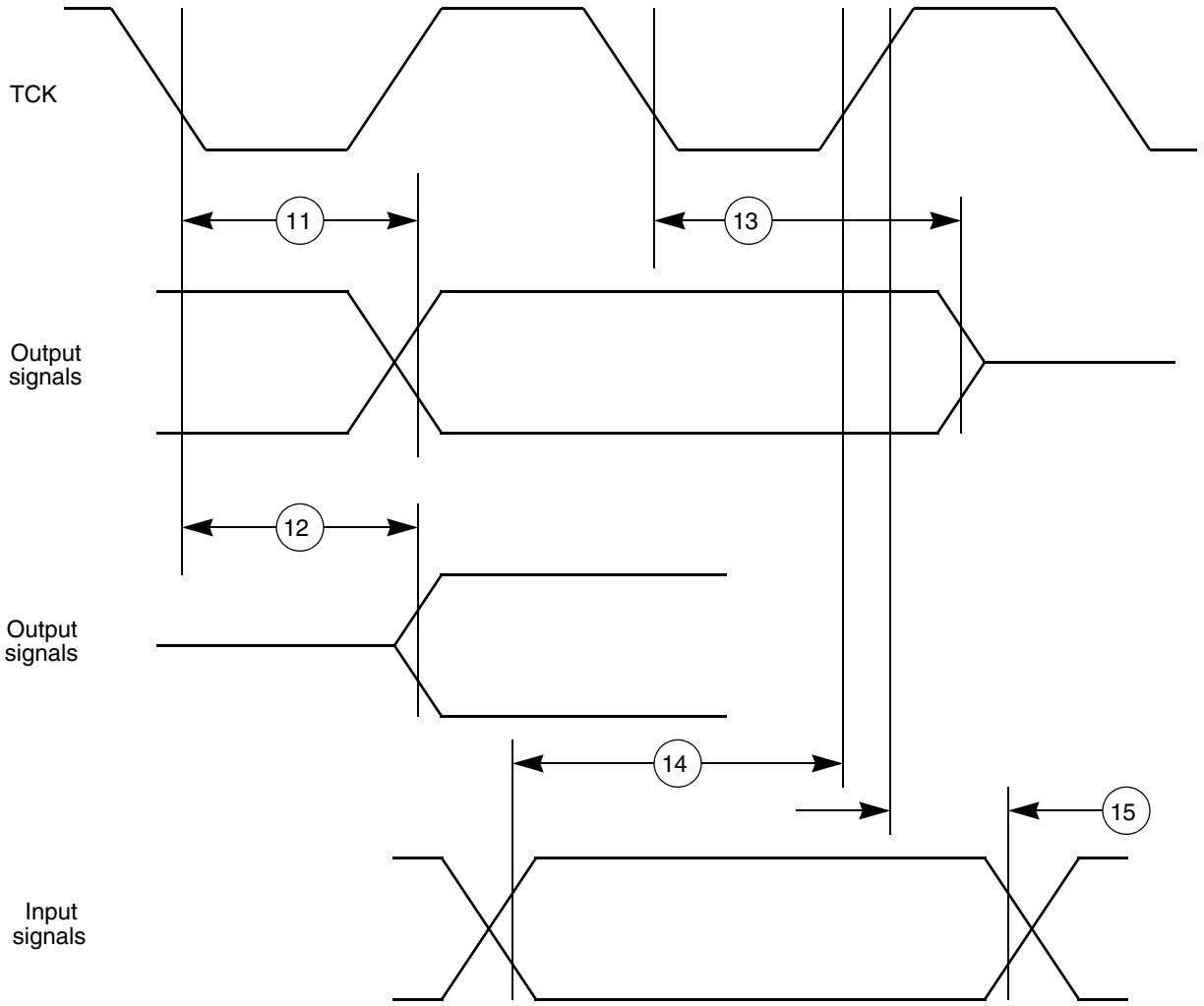


Figure 9. JTAG Boundary Scan Timing

3.13.3 Nexus Timing

Table 21. Nexus Debug Port Timing[1]

Spec	Characteristic	Symbol	Min	Max	Unit
1	MCKO cycle time	t_{MCYC}	1[2]	8	t_{CYC}
2	MCKO duty cycle	t_{MDC}	40	60	%
3	MCKO low to MDO data valid[3]	t_{MDOV}	–1.5	3.0	ns
4	MCKO low to $\overline{\text{MSEO}}$ data valid[3]	t_{MSEOV}	–1.5	3.0	ns
5	MCKO low to $\overline{\text{EVTO}}$ data valid[3]	t_{EVTOV}	–1.5	3.0	ns
6	$\overline{\text{EVTI}}$ pulse width	t_{EVTIPW}	4.0	—	t_{TCYC}
7	$\overline{\text{EVTO}}$ pulse width	t_{EVTOPW}	1	—	t_{MCYC}
8	TCK cycle time	t_{TCYC}	4[4]	—	t_{CYC}
9	TCK duty cycle	t_{TDC}	40	60	%
10	TDI, TMS data setup time	t_{NTDIS}, t_{NTMSS}	8	—	ns
11	TDI, TMS data hold time	t_{NTDIH}, t_{NTMSH}	5	—	ns
12	TCK low to TDO data valid V_{DDE} = 2.25–3.0 V V_{DDE} = 3.0–3.6 V	t_{JOV}	 0 0	 12 9	 ns ns
13	$\overline{\text{RDY}}$ valid to MCKO[5]	—	—	—	—

1 JTAG specifications apply when used for debug functionality. All Nexus timing relative to MCKO is measured from 50% of MCKO and 50% of the respective signal. Nexus timing specified at V_{DD} = 1.35–1.65 V, V_{DDE} = 2.25–3.6 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, T_A = T_L to T_H, and CL = 30 pF with DSC = 0b10.

2 The Nexus AUX port runs up to 82 MHz. Set NPC_PCR[MCKO_DIV] to divide-by-two if the system frequency is greater than 82 MHz.

3 MDO, $\overline{\text{MSEO}}$, and $\overline{\text{EVTO}}$ data is held valid until the next MCKO low cycle occurs.

4 Limit the maximum frequency to approximately 16 MHz (V_{DDE} = 2.25–3.0 V) or 22 MHz (V_{DDE} = 3.0–3.6 V) to meet the timing specification for t_{JOV} of [0.2 x t_{JCYC}] as outlined in the IEEE-ISTO 5001-2003 specification.

5 The $\overline{\text{RDY}}$ pin timing is asynchronous to MCKO. The timing is guaranteed by design to function correctly.

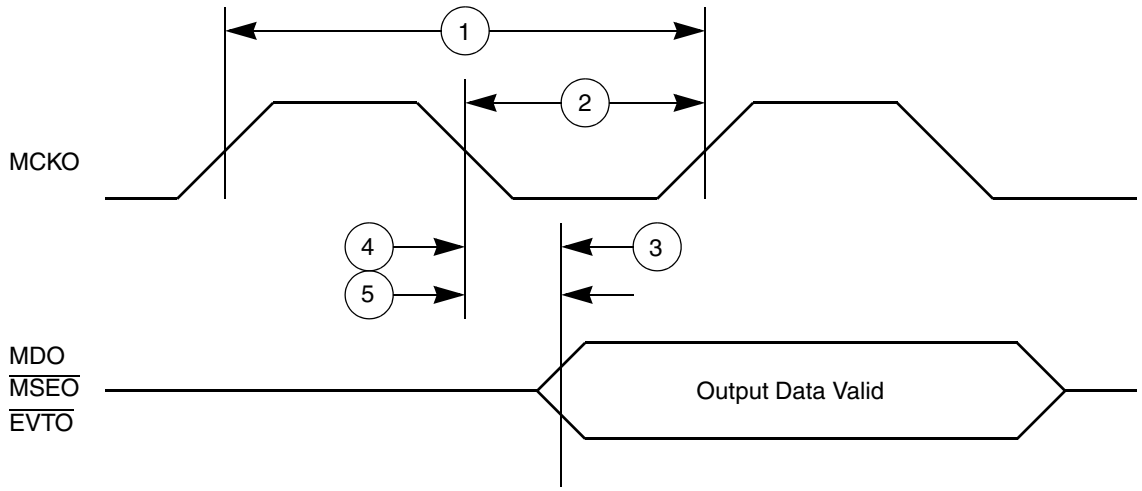


Figure 10. Nexus Output Timing

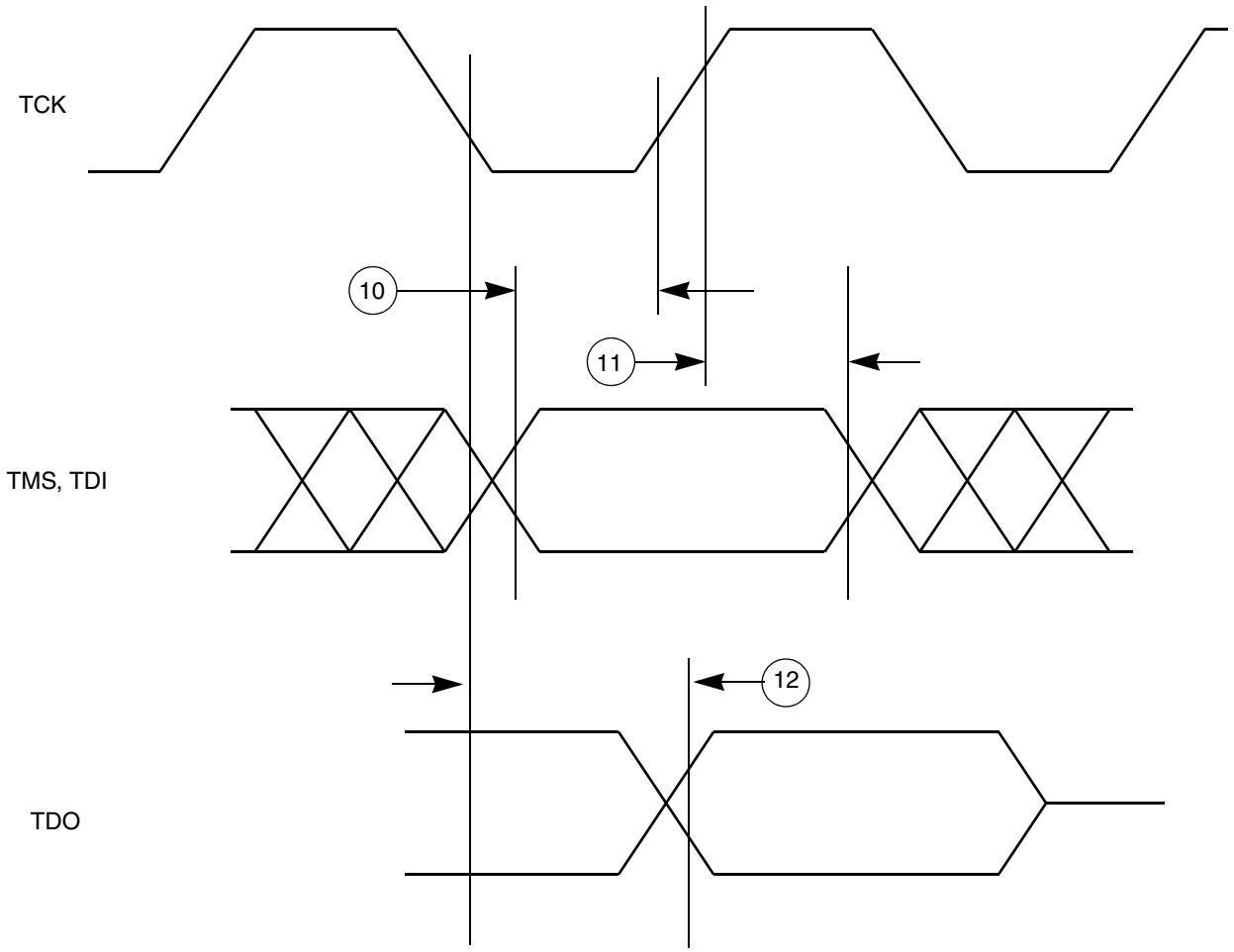


Figure 11. Nexus TDI, TMS, TDO Timing

3.13.4 External Bus Interface (EBI) Timing

Table 22. Bus Operation Timing[1]

Spec	Characteristic and Description	Symbol	40 MHz (ext. bus)[2]		56 MHz (ext. bus)[2]		66 MHz (ext. bus)[2]		Unit	Notes
			Min	Max	Min	Max	Min	Max		
1	CLKOUT period	T_C	25.0	—	17.9	—	15.2	—	ns	Signals are measured at 50% V_{DDE}.
2	CLKOUT duty cycle	t_{CDC}	45%	55%	45%	55%	45%	55%	T_C	
3	CLKOUT rise time	t_{CRT}	—	—[3]	—	—[3]	—	—[3]	ns	
4	CLKOUT fall time	t_{CFT}	—	—[3]	—	—[3]	—	—[3]	ns	

Table 22. Bus Operation Timing[1] (continued)

Spec	Characteristic and Description	Symbol	40 MHz (ext. bus)[2]		56 MHz (ext. bus)[2]		66 MHz (ext. bus)[2]		Unit	Notes
			Min	Max	Min	Max	Min	Max		
5	CLKOUT positive edge to output signal invalid or Hi-Z (hold time) External bus interface ADDR[8:31] $\overline{BDIP}$ $\overline{CS}$[0:3] DATA[0:31] $\overline{OE}$ RD_$\overline{WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ $\overline{WE}$/$\overline{BE}$[0:3]	t_{COH}	1.0[4] 1.5	—	1.0[4] 1.5	—	1.0[4] 1.5	—	ns	EBTS=0 EBTS=1 Hold time selectable via SIU_ECCR[EBTS] bit.
	Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_$\overline{CS}$[0, 2:3] CAL_DATA[0:15] CAL_$\overline{WE}$/$\overline{BE}$[0:1]	t_{CCOH}	1.0[5] 1.5	—	1.0[4] 1.5	—	1.0[4] 1.5	—	ns	EBTS=0 EBTS=1 Hold time selectable via SIU_ECCR[EBTS] bit.
6	CLKOUT positive edge to output signal valid (output delay) External bus interface ADDR[8:31] $\overline{BDIP}$ $\overline{CS}$[0:3] DATA[0:31] $\overline{OE}$ RD_$\overline{WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ $\overline{WE}$/$\overline{BE}$[0:3]	t_{COV}	—	10.0[4] 11.0	—	7.5[4] 8.5	—	6.0[4] 7.0	ns	EBTS=0 EBTS=1 Output valid time selectable via SIU_ECCR[EBTS] bit.
6a	CLKOUT positive edge to output signal valid (output delay) Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_$\overline{CS}$[0, 2:3] CAL_DATA[0:15] CAL_$\overline{WE}$/$\overline{BE}$[0:1]	t_{CCOV}	—	11.0[4] 12.0	—	8.5[4] 9.5	—	7.0[4] 8.0	ns	EBTS=0 EBTS=1 Output valid time selectable via SIU_ECCR[EBTS] bit.

Table 22. Bus Operation Timing[1] (continued)

Spec	Characteristic and Description	Symbol	40 MHz (ext. bus)[2]		56 MHz (ext. bus)[2]		66 MHz (ext. bus)[2]		Unit	Notes
			Min	Max	Min	Max	Min	Max		
7	Input signal valid to CLKOUT positive edge (setup time) External bus interface ADDR[8:31] $\overline{BDIP}$ DATA[0:31] $\overline{OE}$ RD_$\overline{WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ $\overline{WE/BE}$[0:3]	t_{CIS}	10.0	—	7.0	—	5.0	—	ns	
7a	Input signal valid to CLKOUT positive edge (setup time) Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_$\overline{CS}$[0, 2:3] CAL_DATA[0:15] CAL_$\overline{WE}$/$\overline{BE}$[0:1]	t_{CCIS}	11.0	—	8.0	—	6.0	—	ns	
8	CLKOUT positive edge to input signal invalid (hold time) External bus interface ADDR[8:31] $\overline{BDIP}$ DATA[0:31] $\overline{OE}$ RD_$\overline{WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ $\overline{WE}$/$\overline{BE}$[0:3]	t_{CIH}	1.0	—	1.0	—	1.0	—	ns	
	Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_$\overline{CS}$[0, 2:3] CAL_DATA[0:15] CAL_$\overline{WE}$/$\overline{BE}$[0:1]	t_{CCIH}	1.0	—	1.0	—	1.0	—	ns	

1 EBI timing specified at V_{DD} = 1.35–1.65 V, V_{DDE} = 1.6–3.6 V (unless stated otherwise), V_{DD33} and V_{DDSYN} = 3.0–3.6 V, T_A = T_L to T_H, and CL = 30 pF with DSC = 0b10.

2 The external bus is limited to half the speed of the internal bus.

3 Refer to fast pad timing in Table 17 and Table 18 (different values for 1.8 V and 3.3 V).

4 The EBTS = 0 timings are tested and valid at V_{DDE} = 2.25–3.6 V only, whereas EBTS = 1 timings are tested and valid at V_{DDE} = 1.6–3.6 V.

5 The EBTS = 0 timings are tested and valid at V_{DDE} = 2.25–3.6 V only, whereas EBTS = 1 timings are tested and valid at V_{DDE} = 1.6–3.6 V.

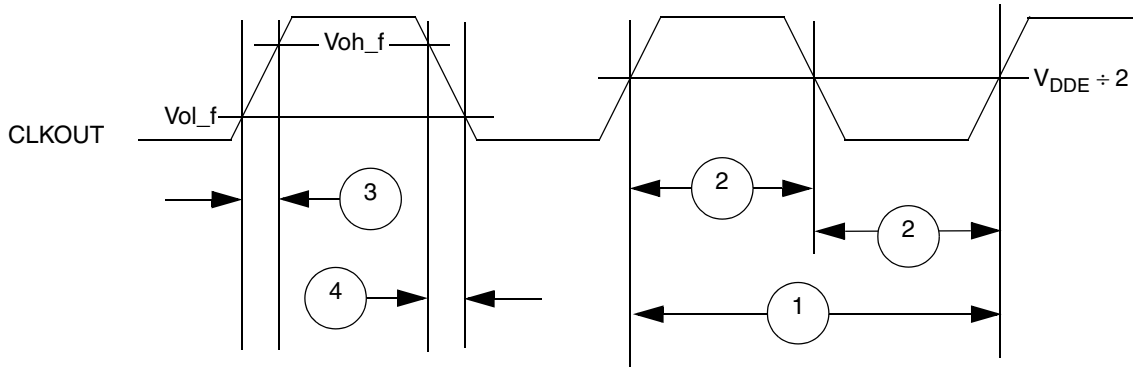


Figure 12. CLKOUT Timing

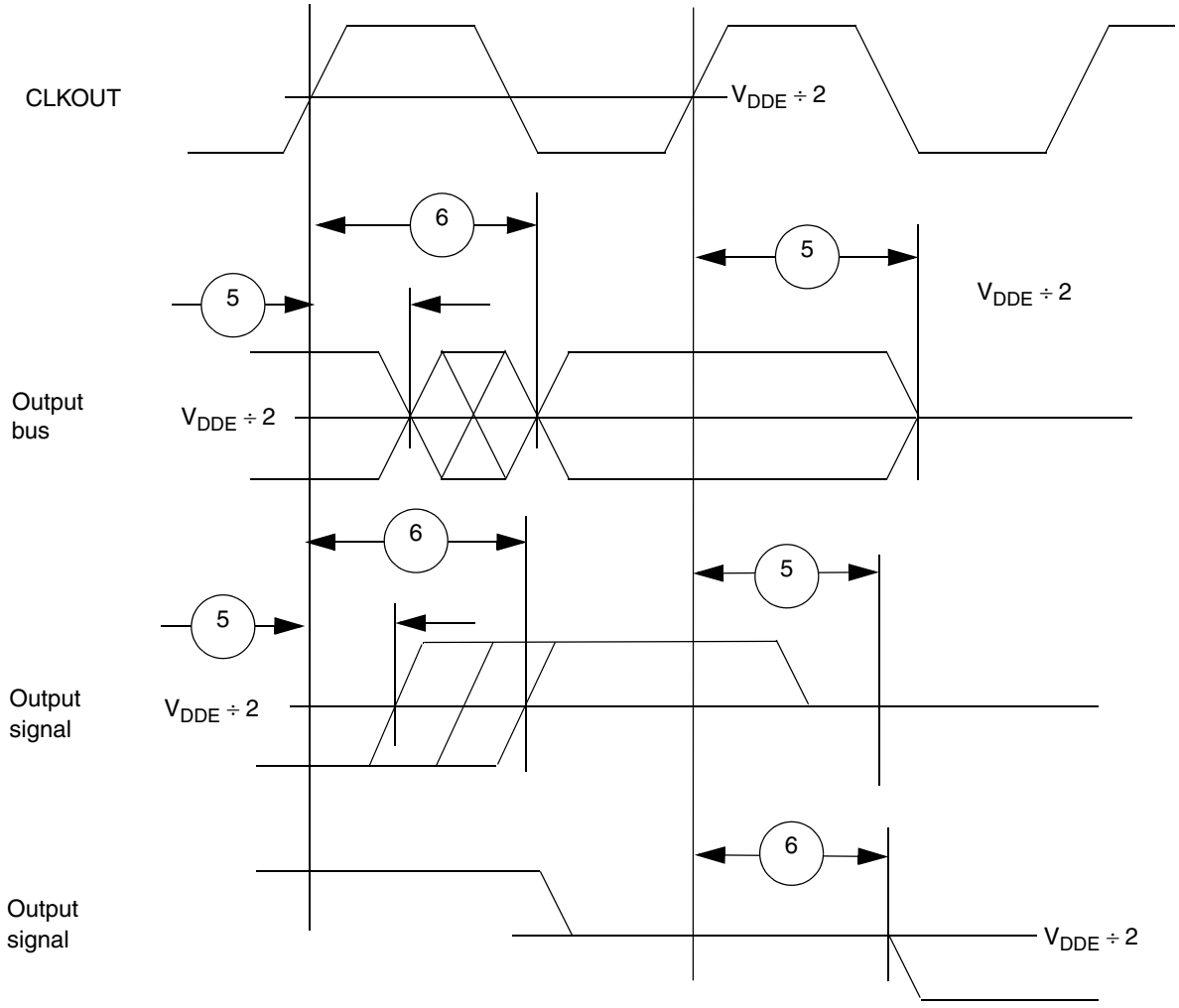


Figure 13. Synchronous Output Timing

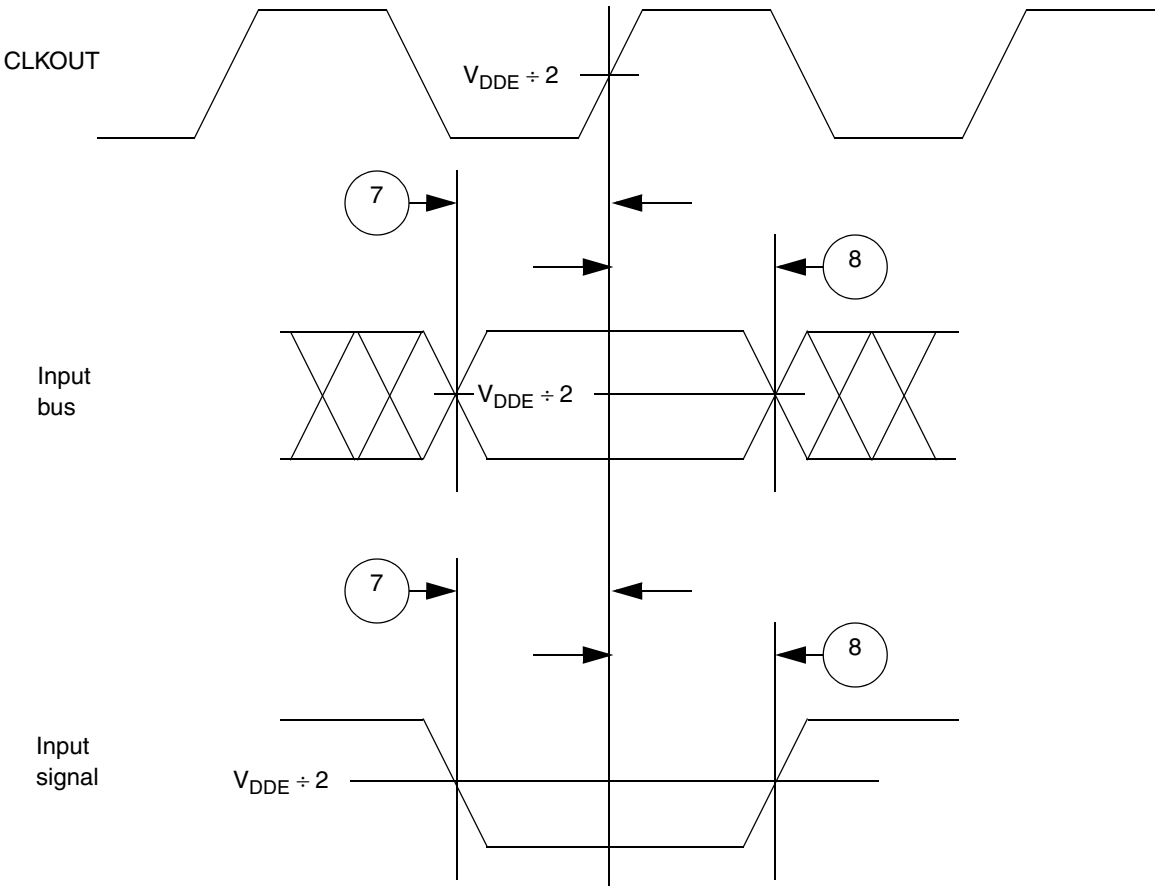


Figure 14. Synchronous Input Timing

3.13.5 External Interrupt Timing (IRQ Signals)

Table 23. External Interrupt Timing[1]

Spec	Characteristic	Symbol	Min	Max	Unit
1	IRQ pulse-width low	t_{IPWL}	3	—	t_{CYC}
2	IRQ pulse-width high	T_{IPWH}	3	—	t_{CYC}
3	IRQ edge-to-edge time[2]	t_{ICYC}	6	—	t_{CYC}

1 IRQ timing specified at F_{SYS} = 132 MHz, V_{DD} = 1.35–1.65 V, V_{DDEH} = 3.0–5.5 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, $T_A = T_L$ to T_H, and CL = 200pF with SRC = 0b11.

2 Applies when IRQ signals are configured for rising-edge or falling-edge events, but not both.

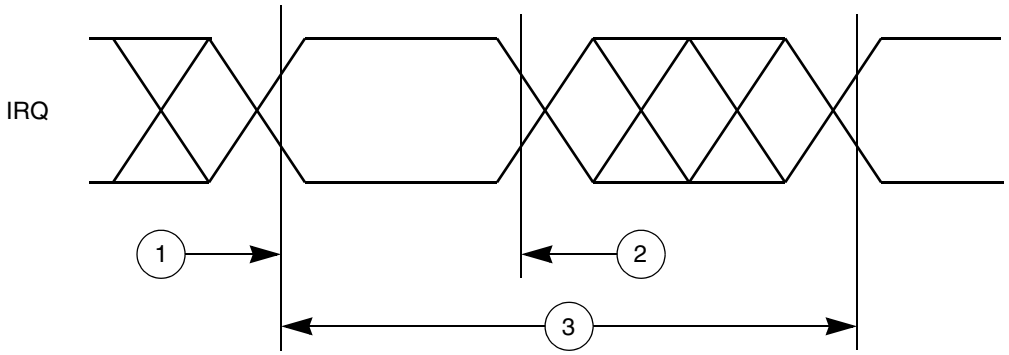


Figure 15. External Interrupt Timing

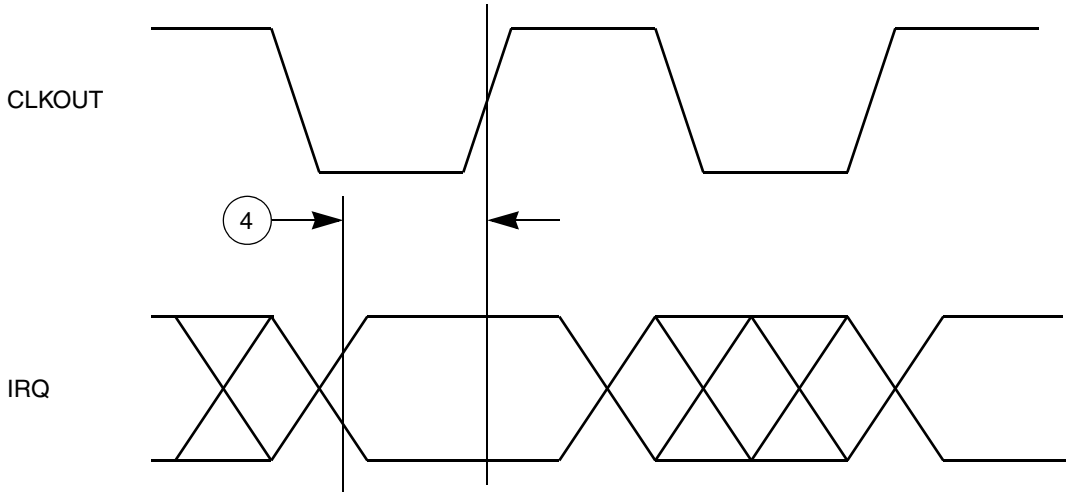


Figure 16. External Interrupt Setup Timing

3.13.6 eTPU Timing

Table 24. eTPU Timing[1]

Spec	Characteristic	Symbol	Min	Max	Unit
1	eTPU input channel pulse width	t_{ICPW}	4	—	t_{CYC}
2	eTPU output channel pulse width	t_{OCPW}	2	—	t_{CYC}

1 eTPU timing specified at F_{SYS} = 132 MHz, V_{DD} = 1.35–1.65 V, V_{DDEH} = 3.0–5.5 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, $T_A = T_L$ to T_H, and CL = 200 pF with SRC = 0b11.

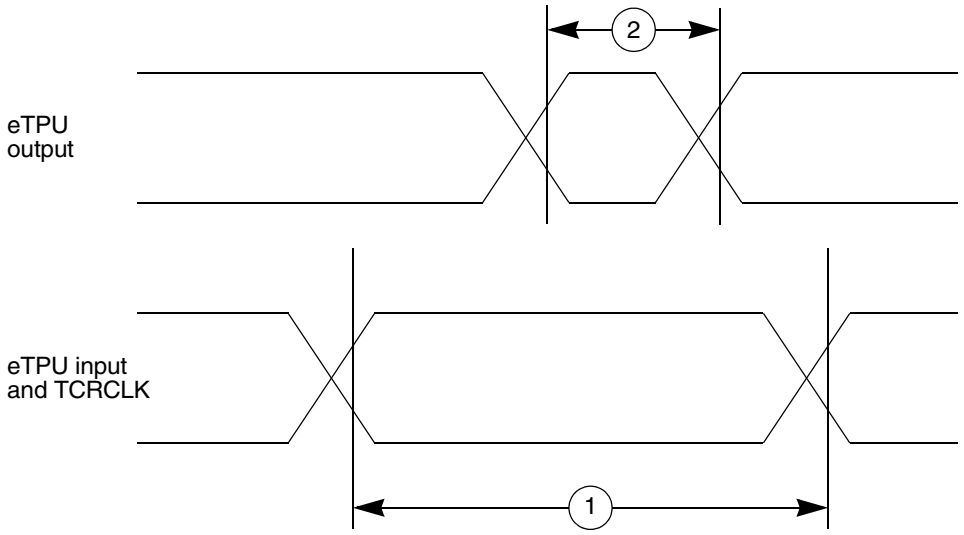


Figure 17. eTPU Timing

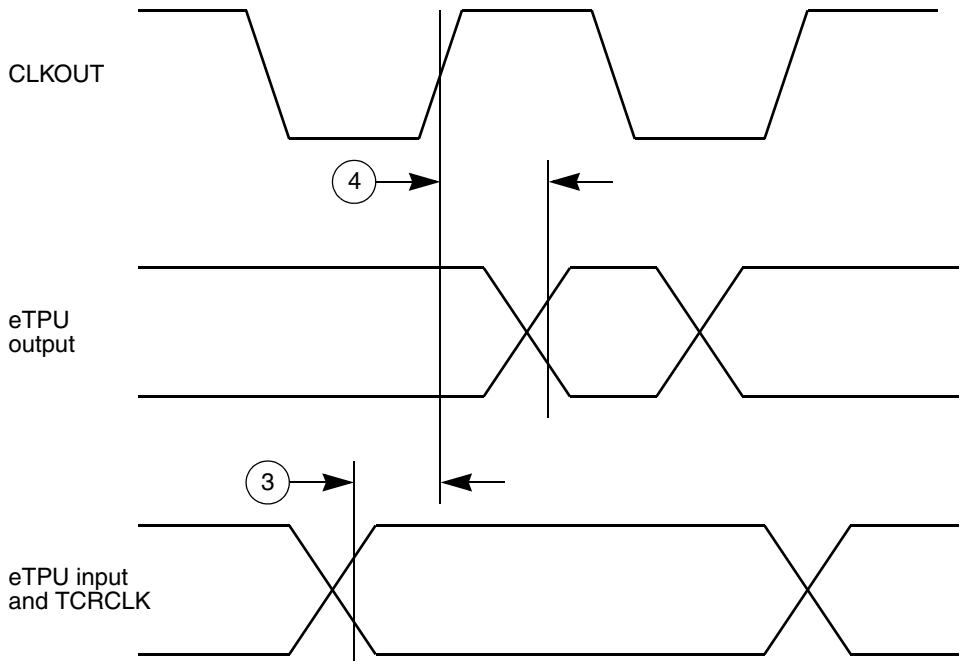


Figure 18. eTPU Input/Output Timing

3.13.7 eMIOS (MTS) Timing

Table 25. MTS Timing[1]

Spec	Characteristic	Symbol	Min	Max	Unit
1	eMIOS (MTS) input pulse width	t_{MIPW}	4	—	t_{CYC}
2	eMIOS (MTS) output pulse width	t_{MOPW}	1	—	t_{CYC}

[1] MTS timing specified at F_{SYS} = 132 MHz, V_{DD} = 1.35–1.65 V, V_{DDEH} = 3.0–5.5 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, $T_A = T_L$ to T_H, and CL = 50 pF with SRC = 0b11.

3.13.8 DSPI Timing

Table 26. DSPI Timing[1]

Spec	Characteristic	Symbol	80 MHz		112 MHz		132 MHz		Unit
			Min	Max	Min	Max	Min	Max	
1	SCK cycle time[2,3]	t_{SCK}	25 ns	2.9 ms	17.9 ns	2.0 ms	15.2 ns	1.7 ms	—
2	PCS to SCK delay[4]	t_{CSC}	23	—	15	—	13	—	ns
3	After SCK delay[5]	t_{ASC}	22	—	14	—	12	—	ns
4	SCK duty cycle	t_{SDC}	($t_{SCK} \div 2$) – 2 ns	($t_{SCK} \div 2$) + 2 ns	—	—	—	—	ns
5	Slave access time ($\overline{SS}$ active to SOUT driven)	t_A	—	25	—	25	—	25	ns
6	Slave SOUT disable time ($\overline{SS}$ inactive to SOUT Hi-Z, or invalid)	t_{DIS}	—	25	—	25	—	25	ns
7	PCSx to $\overline{PCSS}$ time	t_{PCSC}	4	—	4	—	4	—	ns
8	$\overline{PCSS}$ to PCSx time	t_{PASC}	5	—	5	—	5	—	ns
9	Data setup time for inputs	t_{SUI}							
	Master (MTFE = 0)		20	—	20	—	20	—	ns
	Slave		2	—	2	—	2	—	ns
	Master (MTFE = 1, CPHA = 0)[6]		–4	—	3	—	6	—	ns
	Master (MTFE = 1, CPHA = 1)		20	—	20	—	20	—	ns
10	Data hold time for inputs	t_{HI}							
	Master (MTFE = 0)		–4	—	–4	—	–4	—	ns
	Slave		7	—	7	—	7	—	ns
	Master (MTFE = 1, CPHA = 0)[6]		21	—	14	—	12	—	ns
	Master (MTFE = 1, CPHA = 1)		–4	—	–4	—	–4	—	ns
11	Data valid (after SCK edge)	t_{SUO}							
	Master (MTFE = 0)		—	5	—	5	—	5	ns
	Slave		—	25	—	25	—	25	ns
	Master (MTFE = 1, CPHA = 0)		—	18	—	14	—	13	ns
	Master (MTFE = 1, CPHA = 1)		—	5	—	5	—	5	ns
12	Data hold time for outputs	t_{HO}							
	Master (MTFE = 0)		–5	—	–5	—	–5	—	ns
	Slave		5.5	—	5.5	—	5.5	—	ns
	Master (MTFE = 1, CPHA = 0)		8	—	4	—	3	—	ns
	Master (MTFE = 1, CPHA = 1)		–5	—	–5	—	–5	—	ns

1 All DSPI timing specifications use the fastest slew rate (SRC = 0b11) on pad type M or MH. DSPI signals using pad types of S or SH have an additional delay based on the slew rate. DSPI timing is specified at V_{DD} = 1.35–1.65 V, V_{DDEH} = 3.0–5.5 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, $T_A = T_L$ to T_H, and CL = 50 pF with SRC = 0b11.

2 The minimum SCK cycle time restricts the baud rate selection for the given system clock rate. These numbers are calculated based on two MPC55xx devices communicating over a DSPI link.

3 The actual minimum SCK cycle time is limited by pad performance.

4 The maximum value is programmable in DSPI_CTARx[PSSCK] and DSPI_CTARx[CSSCK].

5 The maximum value is programmable in DSPI_CTARx[PASC] and DSPI_CTARx[ASC].

6 This number is calculated using the SMPL_PT bit field in DSPI_MCR set to 0b10.

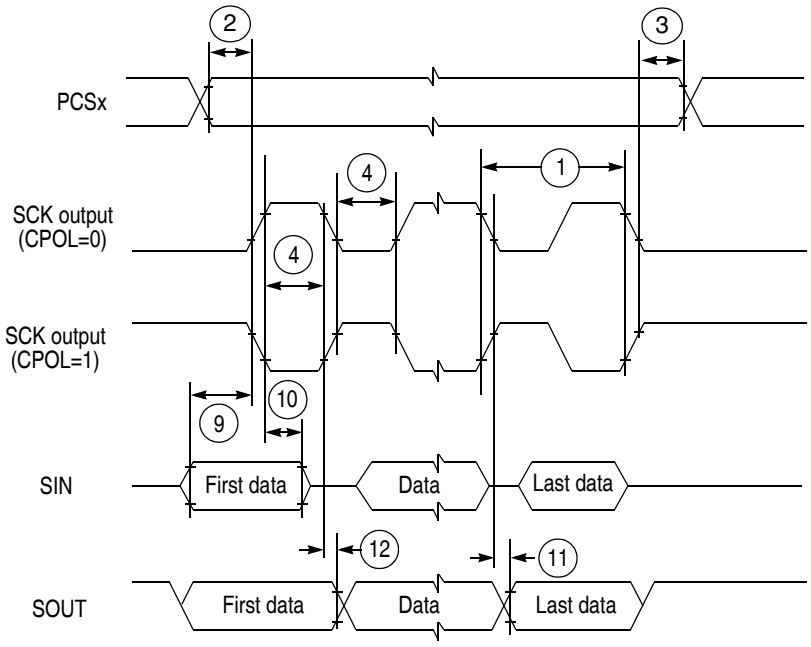


Figure 19. DSPI Classic SPI Timing—Master, CPHA = 0

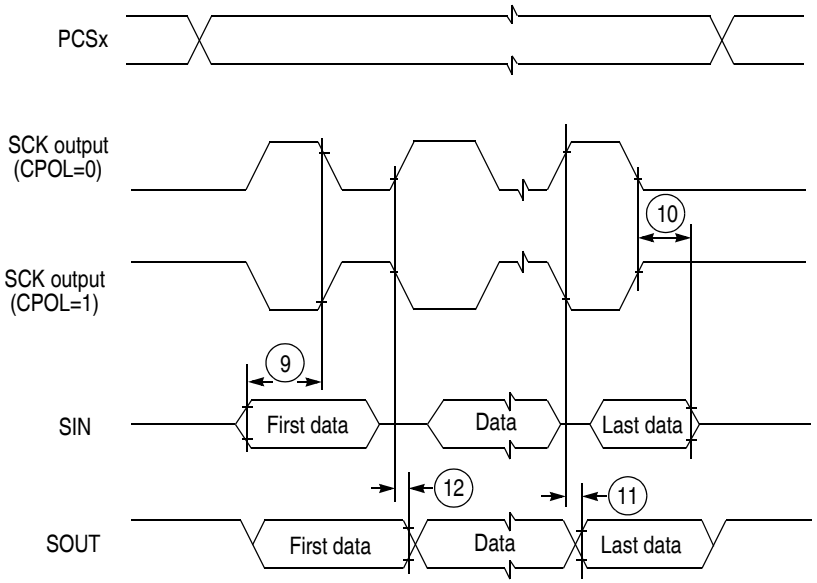


Figure 20. DSPI Classic SPI Timing—Master, CPHA = 1

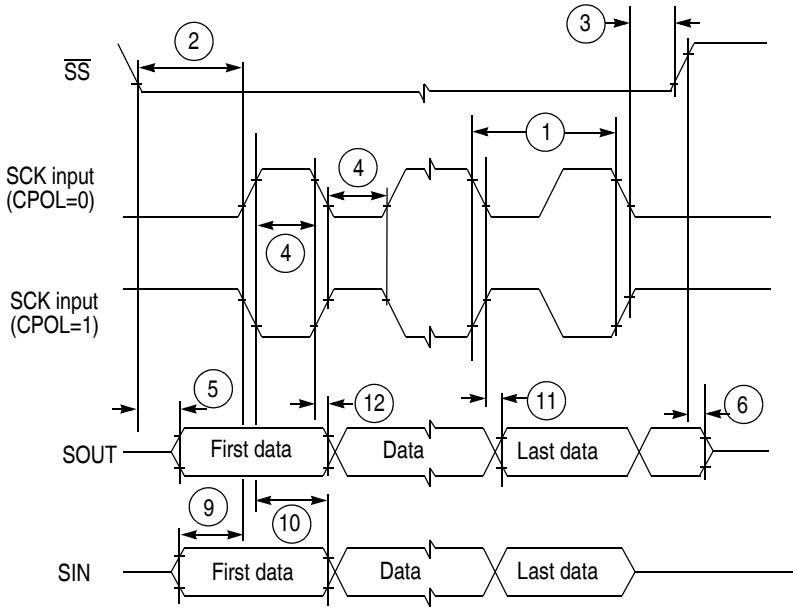


Figure 21. DSPI Classic SPI Timing—Slave, CPHA = 0

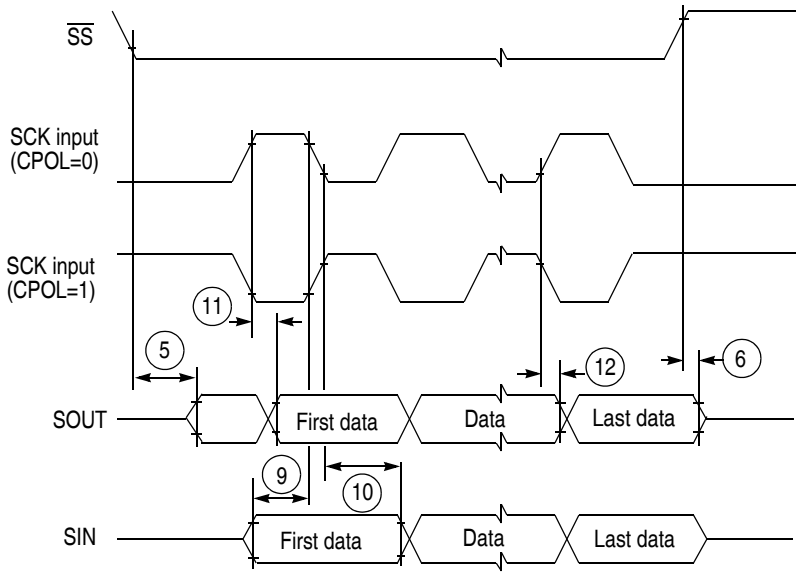


Figure 22. DSPI Classic SPI Timing—Slave, CPHA = 1

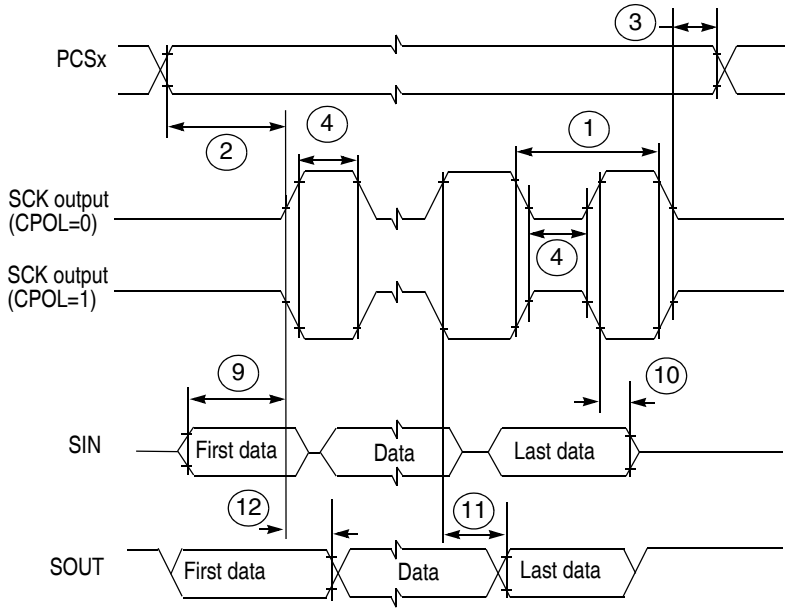


Figure 23. DSPI Modified Transfer Format Timing—Master, CPHA = 0

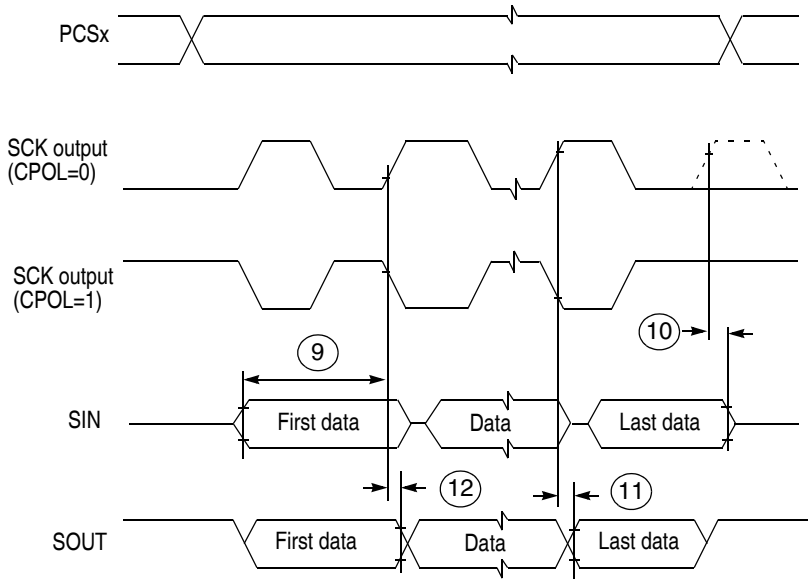


Figure 24. DSPI Modified Transfer Format Timing—Master, CPHA = 1

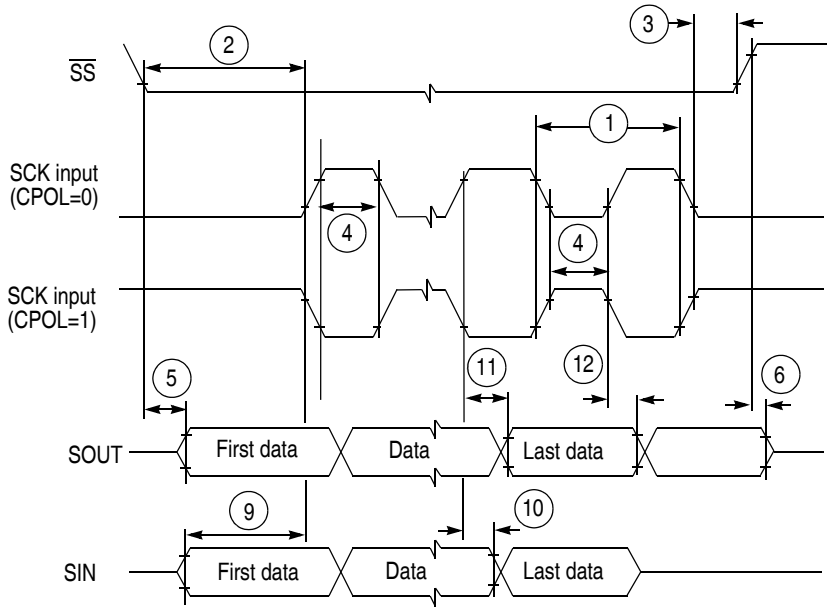


Figure 25. DSPI Modified Transfer Format Timing—Slave, CPHA =0

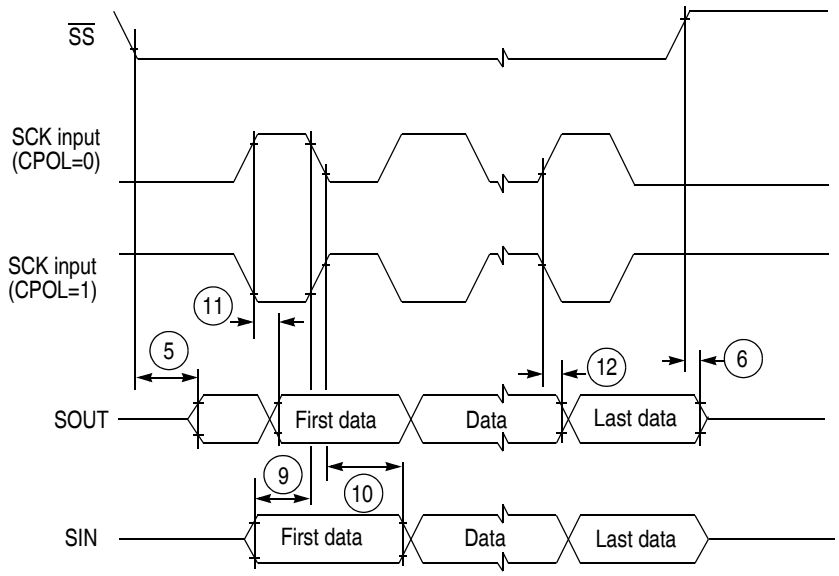


Figure 26. DSPI Modified Transfer Format Timing—Slave, CPHA =1

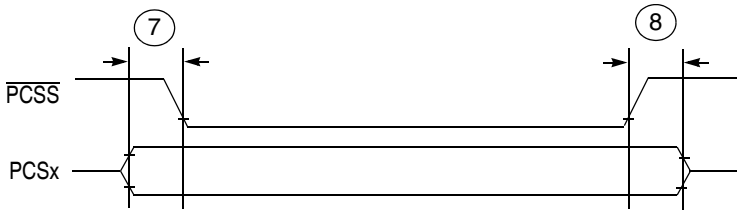


Figure 27. DSPI PCS Strobe ($\overline{\text{PCSS}}$) Timing

3.13.9 eQADC SSI Timing

Table 27. EQADC SSI Timing Characteristics (Pads at 3.3 V or 5.0 V)

CLOAD = 25 pF on all outputs. Pad drive strength set to maximum.						
Spec	**Rating**	**Symbol**	**Minimum**	**Typical**	**Maximum**	**Unit**
2	FCK period ($t_{FCK} = 1 \div f_{FCK}$) [1, 2]	t_{FCK}	2	—	17	t_{SYS_CLK}
3	Clock (FCK) high time	t_{FCKHT}	$t_{SYS\_CLK} - 6.5$	—	$9 \times (t_{SYS\_CLK} + 6.5)$	ns
4	Clock (FCK) low time	t_{FCKLT}	$t_{SYS\_CLK} - 6.5$	—	$8 \times (t_{SYS\_CLK} + 6.5)$	ns
5	SDS lead / lag time	t_{SDS_LL}	–7.5	—	+7.5	ns
6	SDO lead / lag time	t_{SDO_LL}	–7.5	—	+7.5	ns
7	EQADC data setup time (inputs)	t_{EQ_SU}	22	—	—	ns
8	EQADC data hold time (inputs)	t_{EQ_HO}	1	—	—	ns

1 $\overline{SS}$ timing specified at F_{SYS} = 132 MHz, V_{DD} = 1.35–1.65 V, V_{DDEH} = 3.0–5.5 V, V_{DD33} and V_{DDSYN} = 3.0–3.6 V, $T_A = T_L$ to T_H, and CL = 50 pF with SRC = 0b11. Maximum operating frequency varies depending on track delays, master pad delays, and slave pad delays.

2 FCK duty is not 50% when it is generated through the division of the system clock by an odd number.

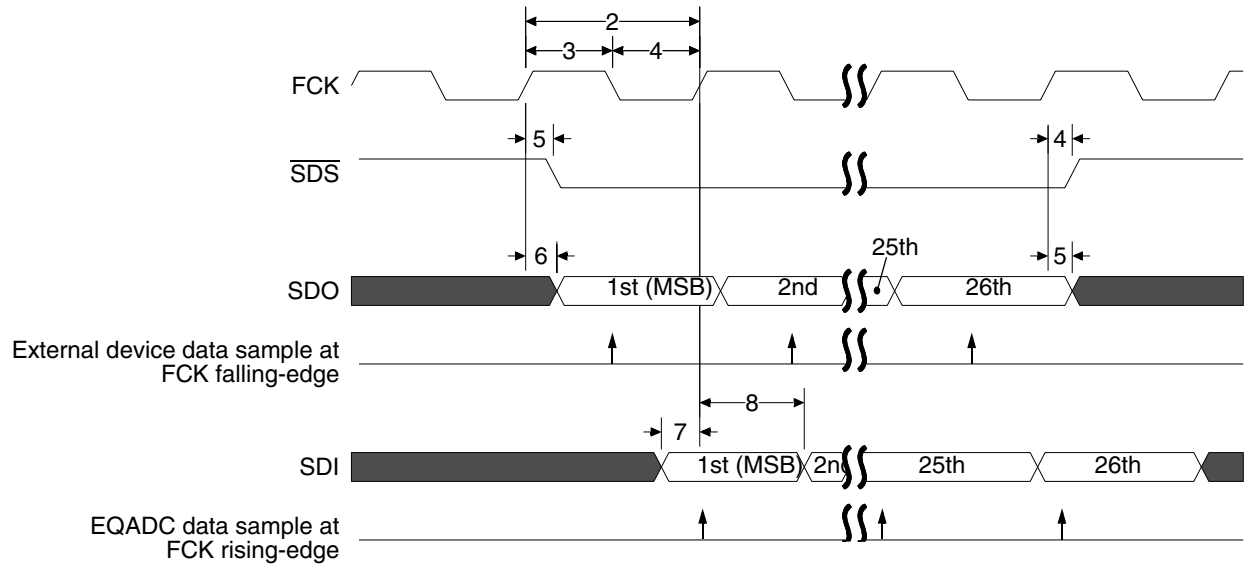


Figure 28. EQADC SSI Timing

3.14 Fast Ethernet AC Timing Specifications

Media Independent Interface (MII) Fast Ethernet Controller (FEC) signals use transistor-to-transistor logic (TTL) signal levels compatible with devices operating at 3.3 V. The timing specifications for the MII FEC signals are independent of the system clock frequency (part speed designation).

3.14.1 MII FEC Receive Signal Timing FEC_RXD[3:0], FEC_RX_DV, FEC_RX_ER, and FEC_RX_CLK

The receive functions correctly up to an FEC_RX_CLK maximum frequency of 25 MHz plus one percent. There is no minimum frequency requirement. The processor clock frequency must exceed four times the FEC_RX_CLK frequency.

Table 28 lists MII FEC receive channel timings.

Table 28. MII FEC Receive Signal Timing

Spec	Characteristic	Min	Max	Unit
1	FEC_RXD[3:0], FEC_RX_DV, FEC_RX_ER to FEC_RX_CLK setup	5	—	ns
2	FEC_RX_CLK to FEC_RXD[3:0], FEC_RX_DV, FEC_RX_ER hold	5	—	ns
3	FEC_RX_CLK pulse-width high	35%	65%	FEC_RX_CLK period
4	FEC_RX_CLK pulse-width low	35%	65%	FEC_RX_CLK period

Figure 29 shows MII FEC receive signal timings listed in Table 28.

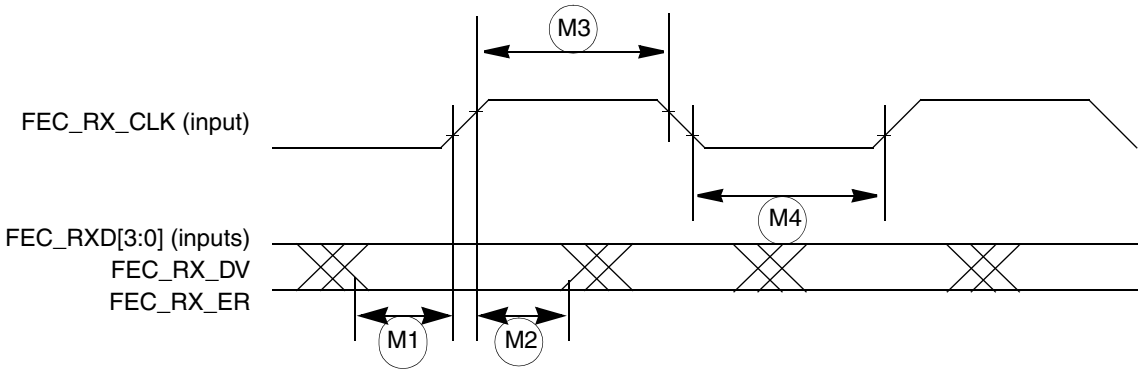


Figure 29. MII FEC Receive Signal Timing Diagram

3.14.2 MII FEC Transmit Signal Timing FEC_TXD[3:0], FEC_TX_EN, FEC_TX_ER, FEC_TX_CLK

The transmitter functions correctly up to the FEC_TX_CLK maximum frequency of 25 MHz plus one percent. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed twice the FEC_TX_CLK frequency.

The transmit outputs (FEC_TXD[3:0], FEC_TX_EN, FEC_TX_ER) can be programmed to transition from either the rising- or falling-edge of TX_CLK, and the timing is the same in either case. These options allow the use of non-compliant MII PHYs.

Refer to the Fast Ethernet Controller (FEC) chapter of the device reference manual for details of this option and how to enable it.

Table 29 lists MII FEC transmit channel timings.

Table 29. MII FEC Transmit Signal Timing

Spec	Characteristic	Min	Max	Unit
5	FEC_TX_CLK to FEC_TXD[3:0], FEC_TX_EN, FEC_TX_ER invalid	5	—	ns
6	FEC_TX_CLK to FEC_TXD[3:0], FEC_TX_EN, FEC_TX_ER valid	—	25	ns
7	FEC_TX_CLK pulse-width high	35%	65%	FEC_TX_CLK period
8	FEC_TX_CLK pulse-width low	35%	65%	FEC_TX_CLK period

Figure 30 shows MII FEC transmit signal timings listed in Table 29.

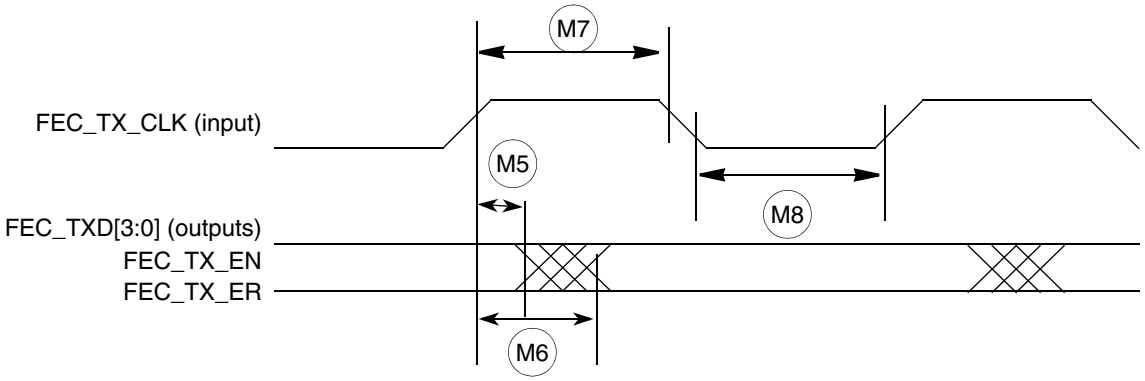


Figure 30. MII FEC Transmit Signal Timing Diagram

3.14.3 MII FEC Asynchronous Inputs Signal Timing FEC_CRS and FEC_COL

Table 30 lists MII FEC asynchronous input signal timing.

Table 30. MII FEC Asynchronous Inputs Signal Timing

Spec	Characteristic	Min	Max	Unit
9	FEC_CRS, FEC_COL minimum pulse width	1.5	—	FEC_TX_CLK period

Figure 31 shows MII FEC asynchronous input timing listed in Table 30.

Figure 31. MII FEC Asynchronous Inputs Timing Diagram

3.14.4 MII FEC Serial Management Channel Timing FEC_MDIO and FEC_MDC

Table 31 lists MII FEC serial management channel timings. The FEC functions correctly with a maximum FEC_MDC frequency of 2.5 MHz.

Table 31. MII FEC Serial Management Channel Timing

Spec	Characteristic	Min	Max	Unit
10	FEC_MDC falling-edge to FEC_MDIO output invalid (minimum propagation delay)	0	—	ns
11	FEC_MDC falling-edge to FEC_MDIO output valid (maximum propagation delay)	—	25	ns
12	FEC_MDIO (input) to FEC_MDC rising-edge setup	10	—	ns
13	FEC_MDIO (input) to FEC_MDC rising-edge hold	0	—	ns
14	FEC_MDC pulse-width high	40%	60%	FEC_MDC period
15	FEC_MDC pulse-width low	40%	60%	FEC_MDC period

Figure 32 shows MII FEC serial management channel timings listed in Table 31.

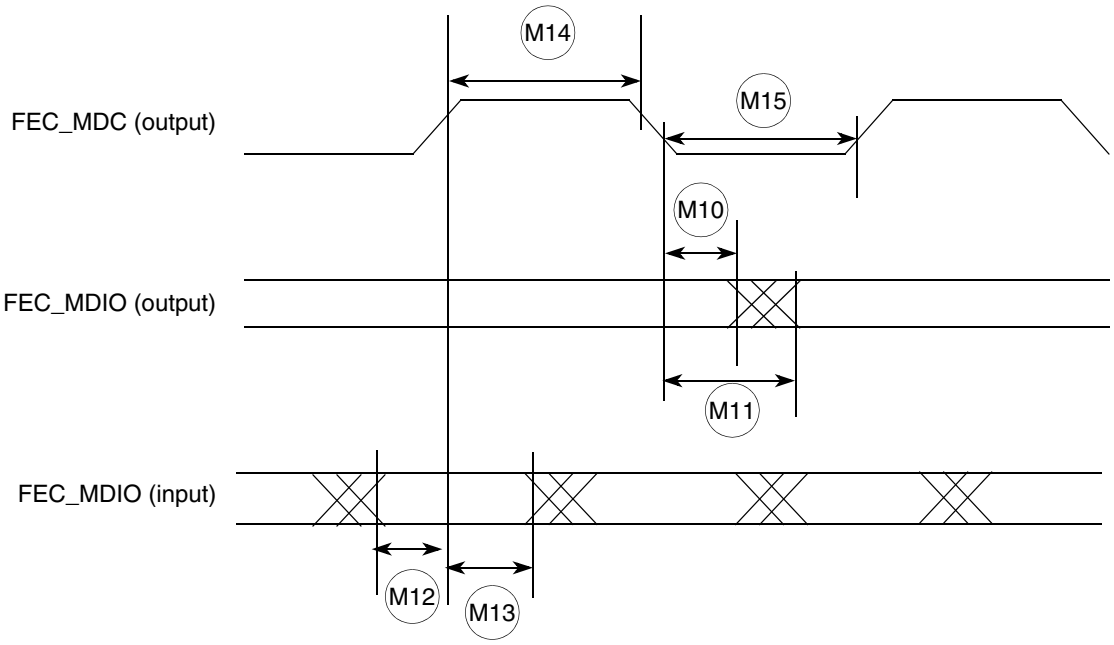


Figure 32. MII FEC Serial Management Channel Timing Diagram

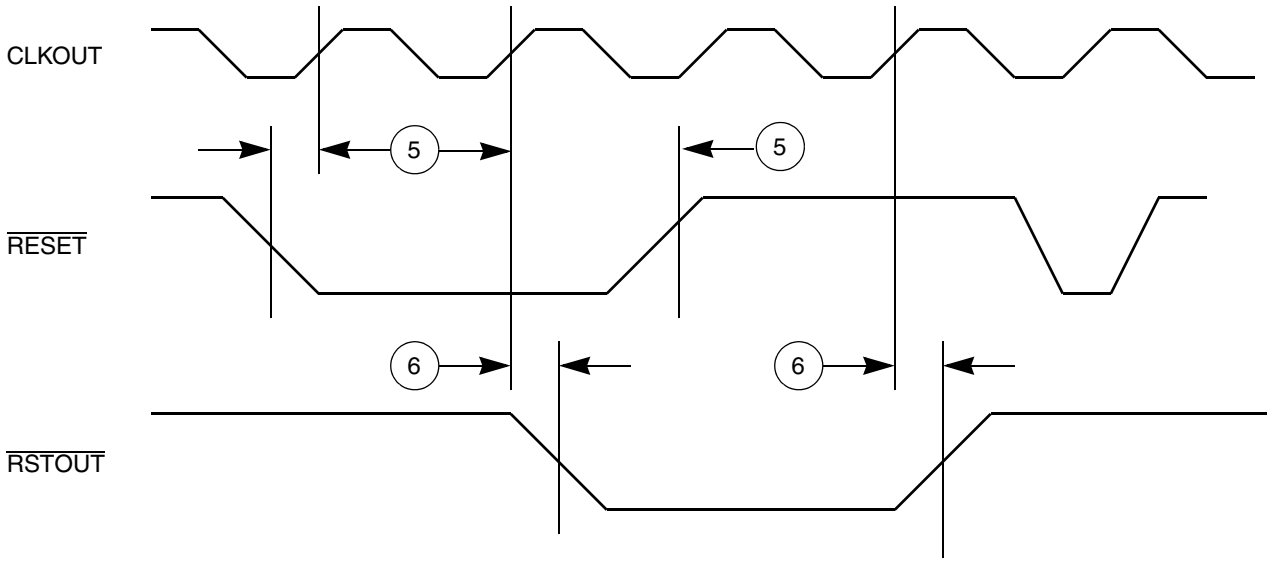


Figure 33. Reset and Configuration Pin Timing

4 Mechanicals

4.1 Pinouts

4.1.1 MPC5553 416 PBGA Pinout

Figure 34, Figure 35, and Figure 36 show the pinout for the MPC5553 416 PBGA package. While the MPC5553 and the MPC5554/MPC5565/MPC5566 are pin-compatible, the MPC5553 BGA is shown to highlight the balls that are not connected to any signal on the MPC5553 (the eTPUB[0:31] and TSIZ[0:1]). The alternate Fast Ethernet Controller (FEC) signals that are multiplexed with the data bus are not shown for the MPC5553.

NOTE

Some pins have names that include functions that are not available on all MPC55*xx* devices. For example, ball R25 of the 416 BGA package is named 'SINA,' but the MPC5553 does not have a DSPI A module. In this case, the SINA pin can only be used for its alternate functions of GPIO[94] or PCSC[2]. Refer to the specific device Reference Manual for functions available on each device.

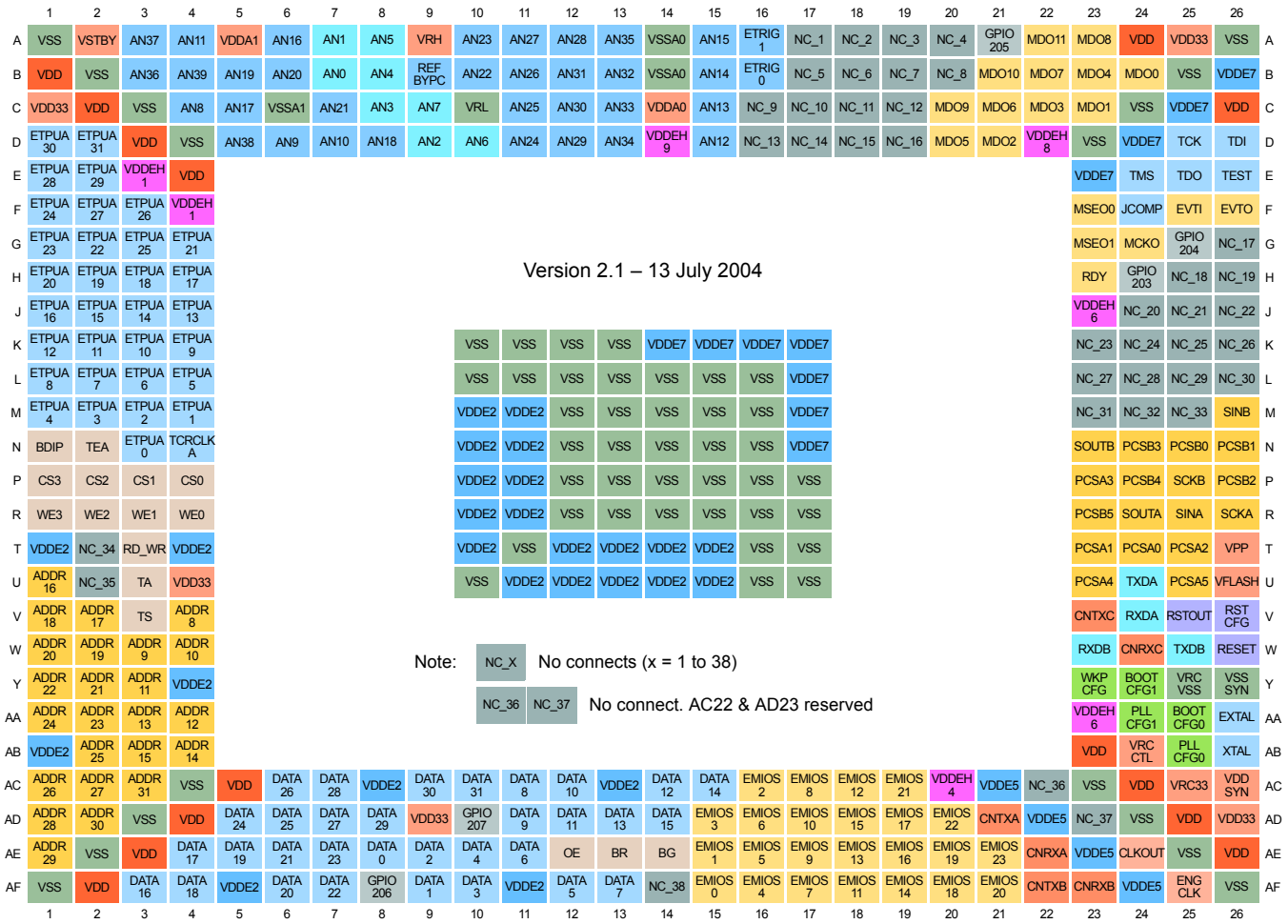


Figure 34. MPC5553 416 Package

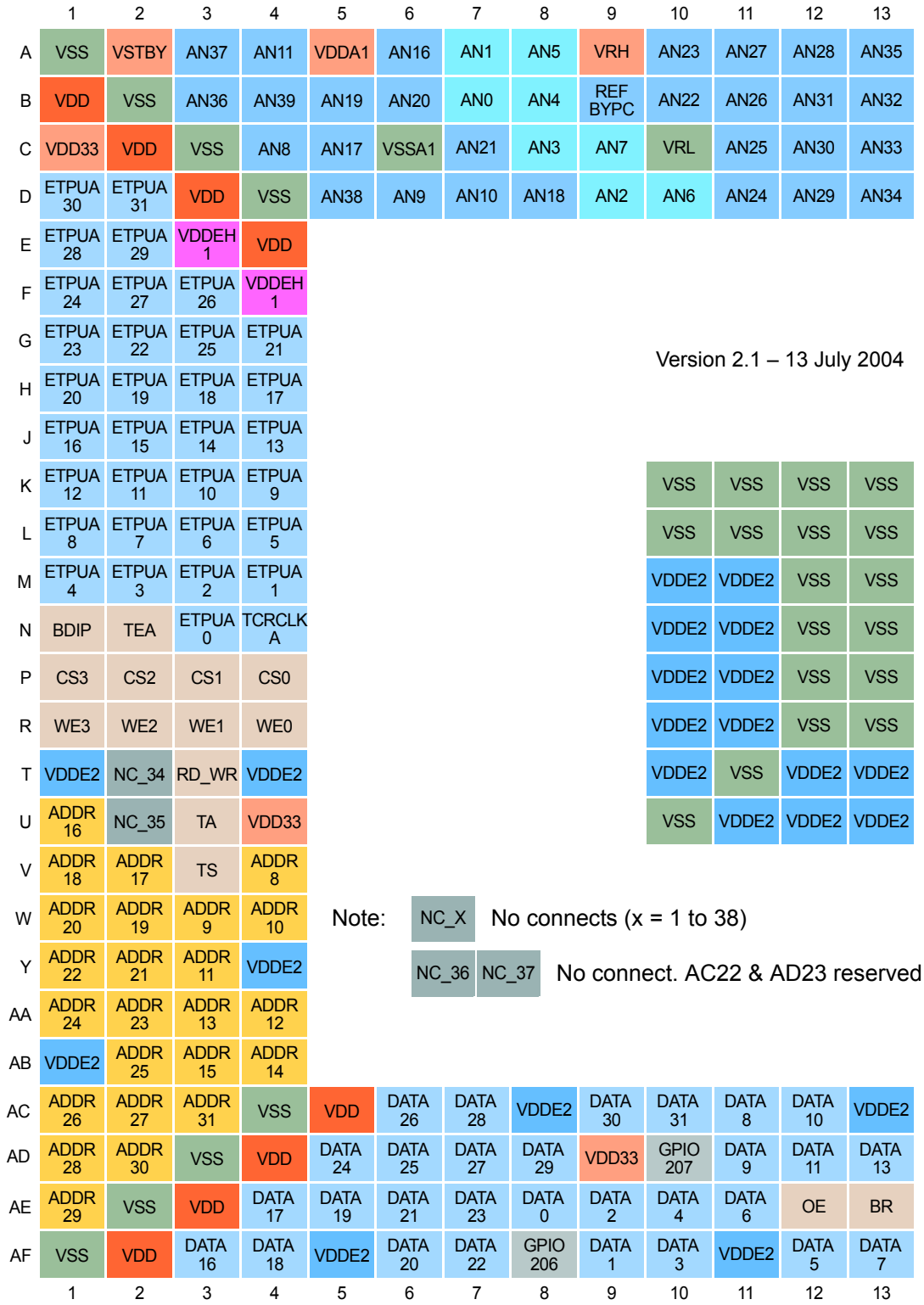


Figure 35. MPC5553 416 Package, Left Side

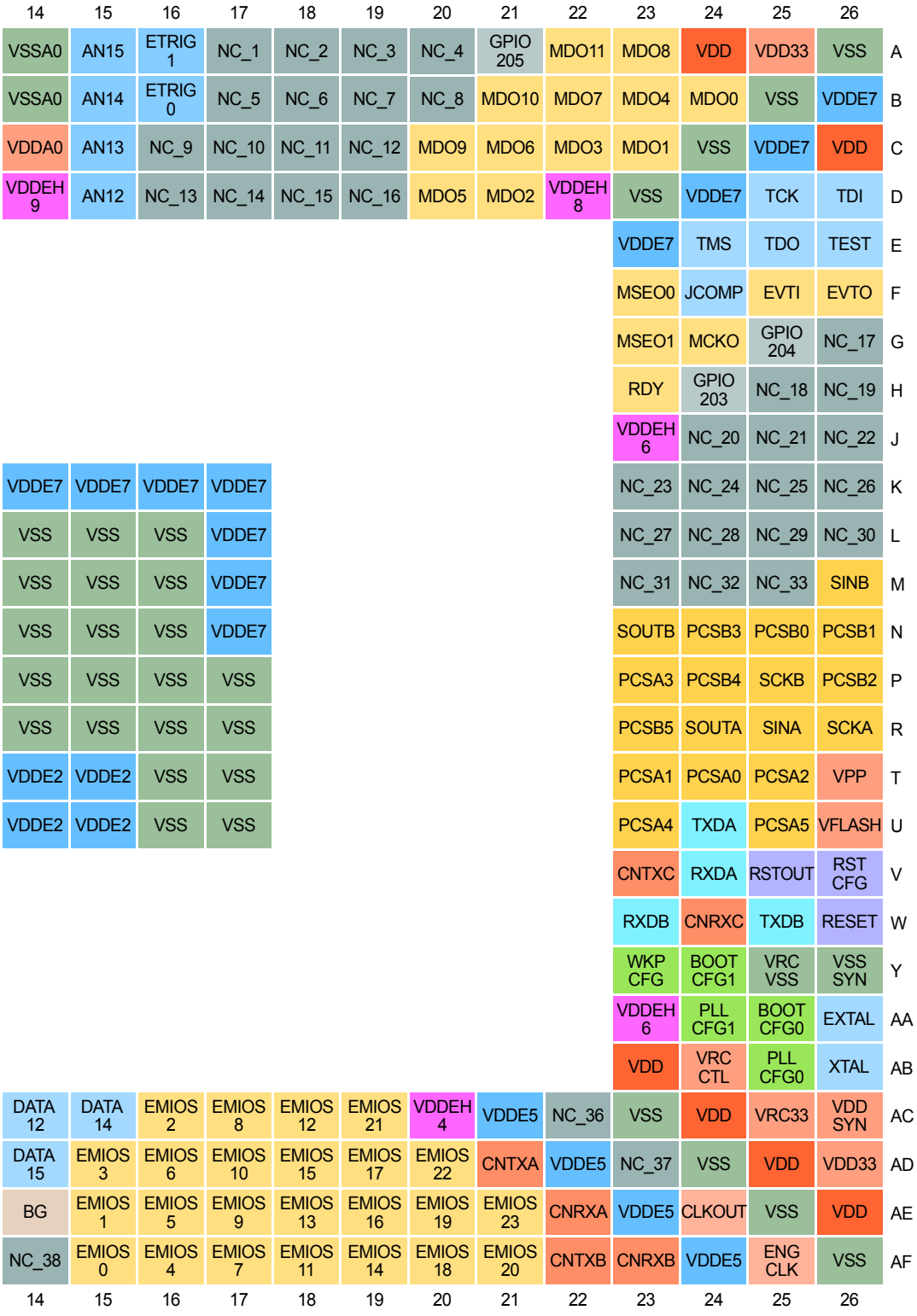


Figure 36. MPC5553 416 Package, Right Side

4.1.2 MPC5553 324 PBGA Pinout

Figure 37 is a pinout for the MPC5553 324 PBGA package.

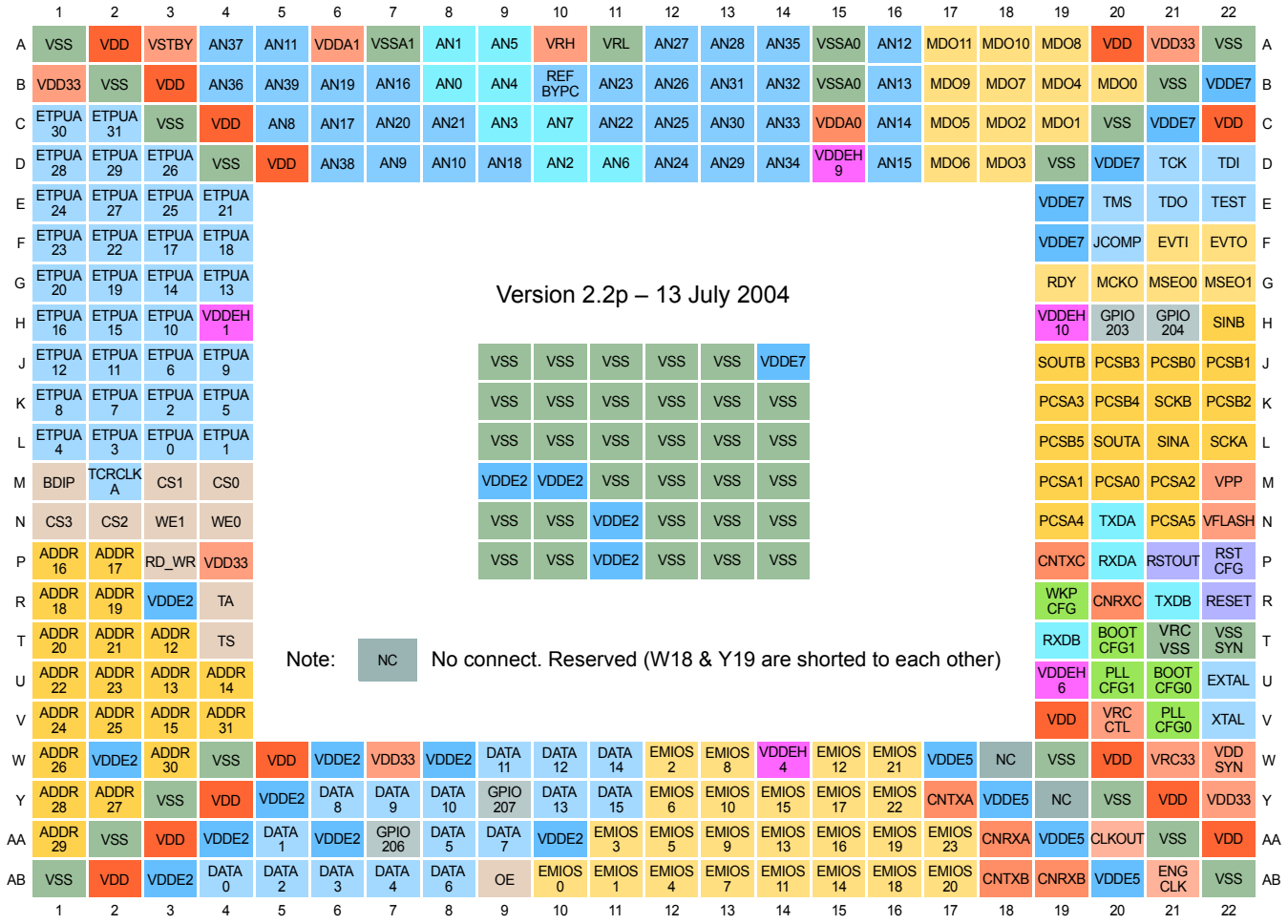


Figure 37. MPC5553 324 Package

4.1.3 MPC5553 208 MAP BGA Pinout

Figure 38 is a pinout for the MPC55MPC5553 208 MAP BGA package.

NOTES

V_{DDEH10} and V_{DDEH6} are connected internally on the 208-ball package and are listed as V_{DDEH6}.

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	
A	VSS	AN9	AN11	VDDA1	VSSA1	AN1	AN5	VRH	VRL	AN27	VSSA0	AN12	MDO2	MDO0	VDD33	VSS	A
B	VDD	VSS	AN38	AN21	AN0	AN4	REF BYPC	AN22	AN25	AN28	VDDA0	AN13	MDO3	MDO1	VSS	VDD	B
C	VSTBY	VDD	VSS	AN17	AN34	AN16	AN3	AN7	AN23	AN32	AN33	AN14	AN15	VSS	MSEO0	TCK	C
D	VDD33	AN39	VDD	VSS	AN18	AN2	AN6	AN24	AN30	AN31	AN35	VDDEH 9	VSS	TMS	EVTO	TEST	D
E	ETPUA 30	ETPUA 31	AN37	VDD									VDDE7	TDI	EVTI	MSEO1	E
F	ETPUA 28	ETPUA 29	ETPUA 26	AN36				8 June 2005p					VDDEH 6	TDO	MCKO	JCOMP	F
G	ETPUA 24	ETPUA 27	ETPUA 25	ETPUA 21			VSS	VSS	VSS	VSS			SOUTB	PCSB3	SINB	PCSB0	G
H	ETPUA 23	ETPUA 22	ETPUA 17	ETPUA 18			VSS	VSS	VSS	VSS			PCSA3	PCSB4	PCSB2	PCSB1	H
J	ETPUA 20	ETPUA 19	ETPUA 14	ETPUA 13			VSS	VSS	VSS	VSS			PCSB5	TXDA	PCSA2	SCKB	J
K	ETPUA 16	ETPUA 15	ETPUA 7	VDDEH 1			VSS	VSS	VSS	VSS			CNTXC	RXDA	RSTOUT	VPP	K
L	ETPUA 12	ETPUA 11	ETPUA 6	TCRCLK A									TXDB	CNRXC	WKP CFG	RESET	L
M	ETPUA 10	ETPUA 9	ETPUA 1	ETPUA 5									RXDB	PLL CFG0	BOOT CFG1	VSS SYN	M
N	ETPUA 8	ETPUA 4	ETPUA 0	VSS	VDD	VDD33	EMIOS 2	EMIOS 10	VDDEH 4	EMIOS 12	EMIOS 21	VDD33	VSS	VRC CTL	PLL CFG1	EXTAL	N
P	ETPUA 3	ETPUA 2	VSS	VDD	GPIO 207	VDDE2	EMIOS 6	EMIOS 8	EMIOS 16	EMIOS 17	EMIOS 22	CNTXA	VDD	VSS	VRC33	XTAL	P
R	CS0	VSS	VDD	GPIO 206	EMIOS 4	EMIOS 3	EMIOS 9	EMIOS 11	EMIOS 14	EMIOS 19	EMIOS 23	CNRXA	CNRXB	VDD	VSS	VDD SYN	R
T	VSS	VDD	OE	EMIOS 0	EMIOS 1	EMIOS 5	EMIOS 7	EMIOS 13	EMIOS 15	EMIOS 18	EMIOS 20	CNTXB	VDDE5	ENG CLK	VDD	VSS	T
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	

Figure 38. MPC5553 208 Package

4.2 Package Dimensions

4.2.1 MPC5553 416-Pin Package

The package drawings of the MPC5553 416 pin TEPBGA package are shown in Figure 39.

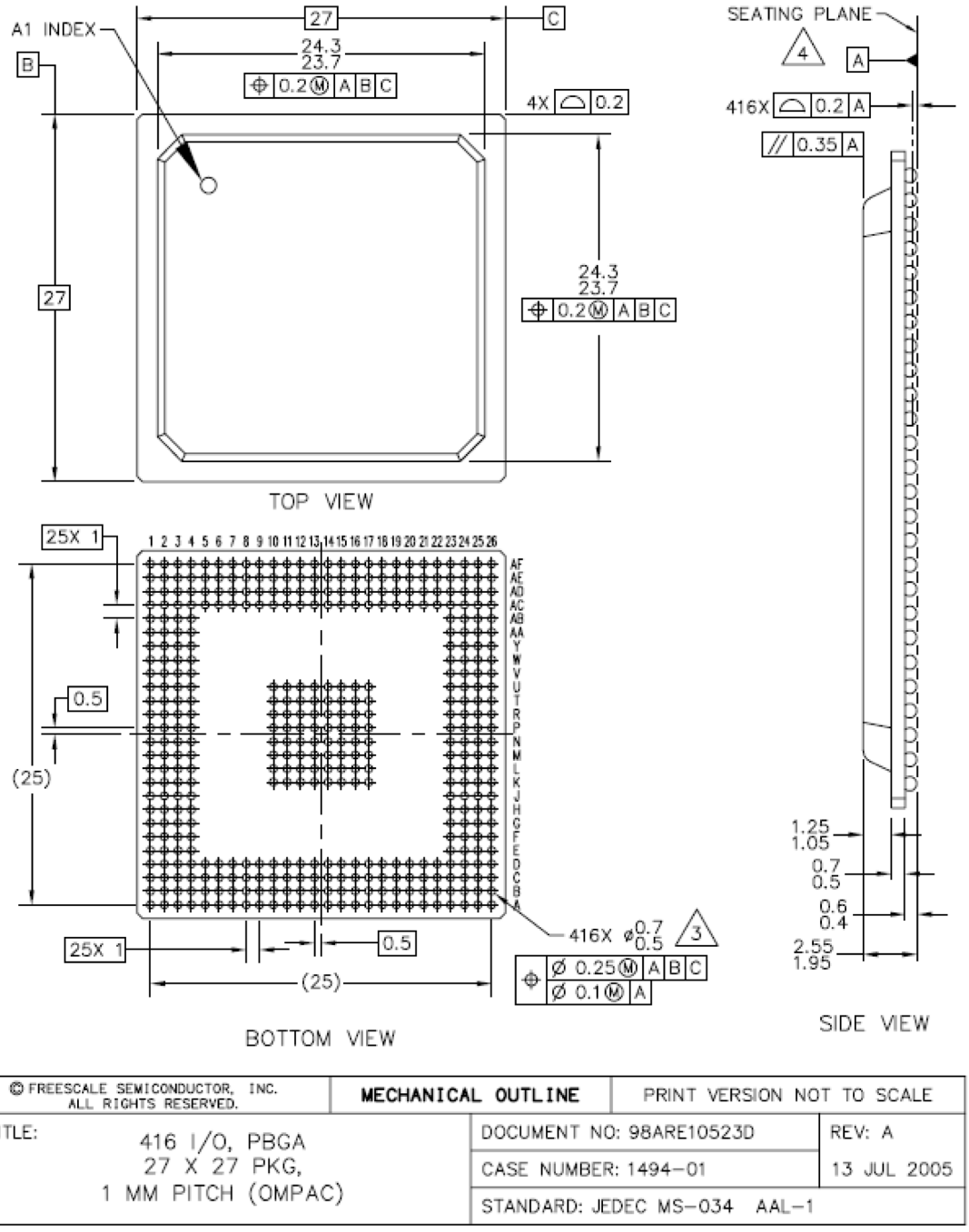


Figure 39. MPC5553 416 TEPBGA Package

NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M–1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.

	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: 416 I/O, PBGA 27 X 27 PKG, 1 MM PITCH (OMPAC)	DOCUMENT NO: 98ARE10523D	REV: A
	CASE NUMBER: 1494–01	13 JUL 2005
	STANDARD: JEDEC MS–034 AAL–1	

Figure 39. MPC5553 416 TEPBGA Package (continued)

4.2.2 MPC5553 324-Pin Package

The package drawings of the MPC5553 324-pin TEPBGA package are shown in Figure 40.

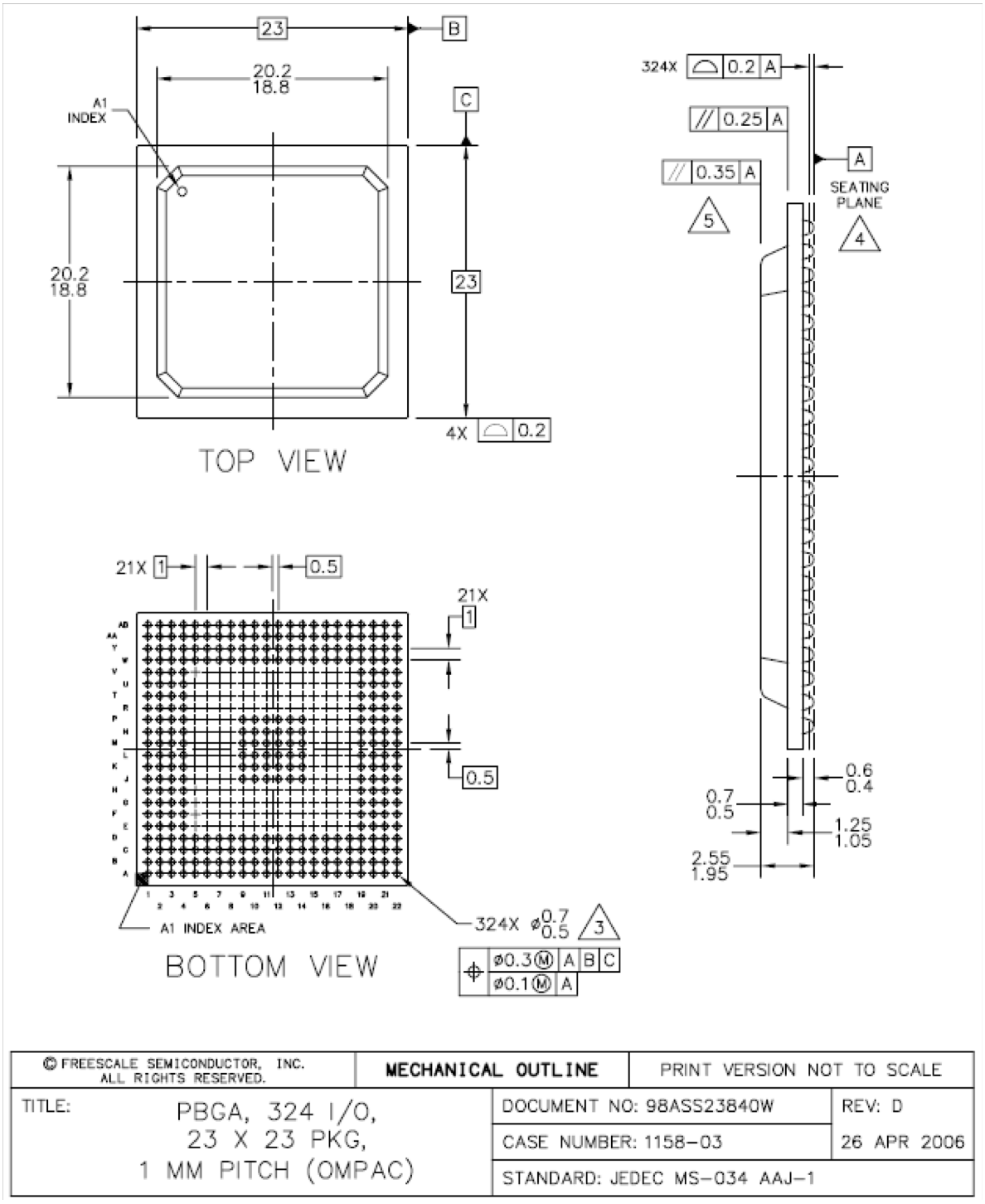


Figure 40. MPC5553 324 TEPBGA Package

NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M–1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.

MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE

TITLE:	DOCUMENT NO: 98ASS23840W	REV: D
PBGA, 324 I/O, 23 X 23 PKG, 1 MM PITCH (OMPAC)	CASE NUMBER: 1158–03	26 APR 2006
	STANDARD: JEDEC MS–034 AAJ–1	

Figure 40. MPC5553 324 TEPBGA Package (continued)

4.2.3 MPC5553 208-Pin Package

The package drawings of the MPC5553 208-pin MAP BGA package are shown in Figure 41.

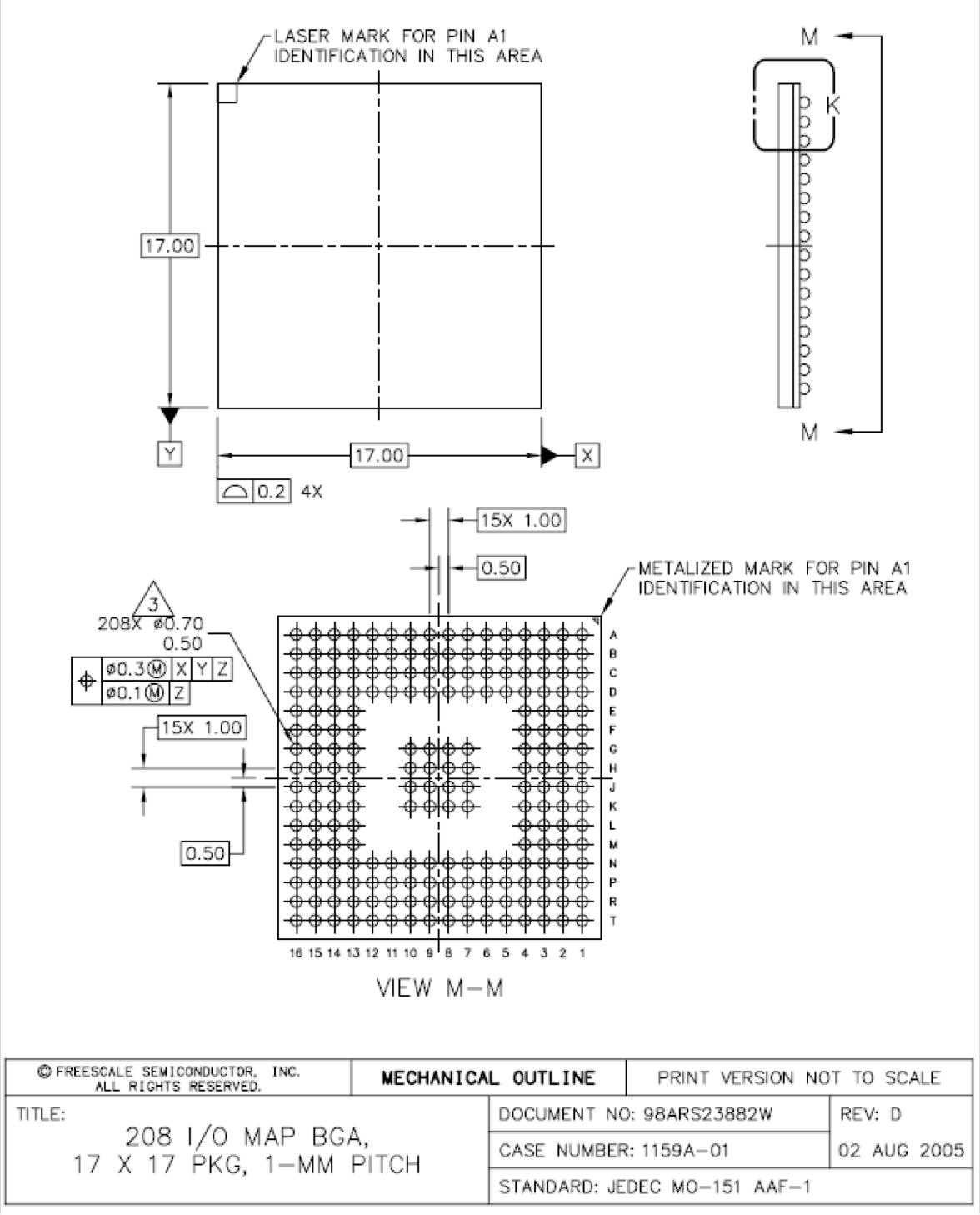
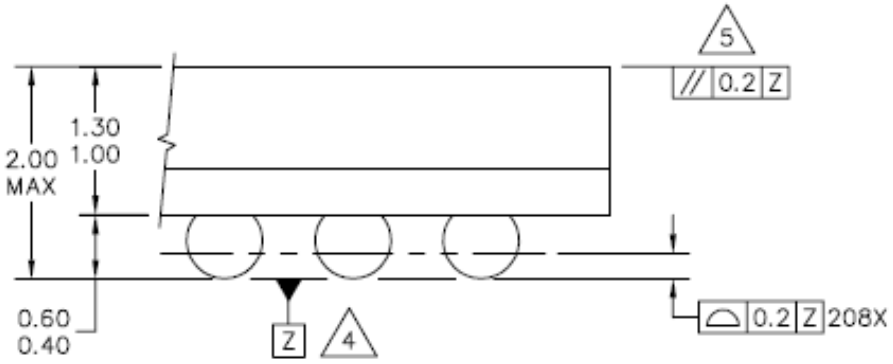


Figure 41. MPC5553 208 MAP BGA Package



DETAIL K
(ROTATED 90° CLOCKWISE)

NOTES:

1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. INTERPRET DIMENSIONS AND TOLERANCES PER ASME Y14.5M–1994.
3. DIMENSION b IS MEASURED AT THE MAXIMUM SOLDER BALL DIAMETER, PARALLEL TO DATUM PLANE Z.
4. DATUM Z (SEATING PLANE) IS DEFINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.

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TITLE: 208 I/O MAP BGA, 17 X 17 PKG, 1–MM PITCH	DOCUMENT NO: 98ARS23882W	REV: D
	CASE NUMBER: 1159A–01	02 AUG 2005
	STANDARD: JEDEC MO–151 AAF–1	

Figure 41. MPC5553 208 MAP BGA Package (continued)

5 MPC5553 Revision History

Table 32 provides a revision history of the MPC5553 Data Sheet.

Table 32. MPC5553 Revision History

Revision	Location(s)	Substantive Change(s)
Rev. 0		This is the first released version of this document.
Rev. 1	Table 1	Footnote added to Freescale Part Number column.
	Table 2	Footnotes 6, 8, and 9 changed from 1mA to 2mA.
	Figure 39, Figure 40, Figure 41	Second page of package drawings added.
	Figure 37	Removed note about pin R1 in the figure and added a Note above it instead.
Rev. 1.1	Throughout	Editorial changes: subscripting, simplifying language.

Table 33 is the new format for the Revision History and changes continue from Table 32.

Table 33. MPC5553 Revision History (*continued*)

Revision	Author	Date	Substantive Change(s)
Rev. 1.1	NH	02/02/07	Changes per RD initial review: • Changed the values in Table 14 for the H7Fa Flash pre-program and erase times. Typical and Initial Max values changed. • Typical Values — 16 Kbytes: from 265 to 325 48 Kbytes: from 340 to 435 64 Kbytes: from 400 to 525 128 Kbytes: from 500 to 675 • Initial Max Values — 16 Kbytes: from 400 to 525 48 Kbytes: from 400 to 525 64 Kbytes: from 500 to 675 128 Kbytes: from 1250 to 1800
Rev. 1.1	NH	02/06/07	Changes per RD second review: • Added Figure 3 to show interpolated IDD_{STBY} values listed in Table 9. • Table 9 DC Electrical Specifications: Changed wording of footnote 3. Spec 28: Corrected conditional text error showing wrong frequency. Spec 29: Deleted frequency information. • Table 6 FMPLL Electrical Characteristics: Grouped (2 x CI) in Specs 12 and 13. • Table 7 Power Sequence Pin Status for Fast Pads, updated paragraph. • Table 8 Power Sequence Pin Status, updated preceding paragraph. • Section 3.7.1, “Input Value of Pins During POR Dependent on VDD33 Updated paragraph to remove redundancy, • Table 16 Flash BIU Settlings: Changed wording of footnote from “Can be changed after Analysis and Characterization” to “These values may change after characterization.” • Table 17and Table 18: Deleted the words ‘not’ from footnote 2. Changed from ‘This parameter is supplied for reference and is not guaranteed by design and not tested’ to ‘This parameter is supplied for reference and is guaranteed by design and tested.’ • Table 22 Bus Operation Timing: Specs 5 and 6: corrected format to show the bus timing values for various frequencies with EBTS bit = 0 and EBTS bit = 1. Specs 6 and 7: Added the calibrations signals: CAL_ADDR, CAL_$\overline{WE}$/$\overline{BE}$, CAL_$\overline{CS}$, CAL_DATA. • Table 26 DSPI Timing: Added to beginning of footnote 1 ‘All DSPI timing specifications use the fastest slew rate (SRC=0b11) on pad type M or MH. DSPI signals using pad types of S or SH have an additional delay based on the slew rate.’ • Table 27 EQADC SS Timing Characteristics: combined footnotes 1 and 2. Moved footnotes 1 and 2 to Spec 2 and deleted Spec 1.

Table 33. MPC5553 Revision History (*continued*)

Revision	Author	Date	Substantive Change(s)
Rev 2.0	NH	02/07/07	Changes per RD sign-off review: • Changed paragraph preceding Table 7 Power Sequence Pin Status for the Fast Pad: From: Although there are no power up/down sequencing requirements to prevent issues like latch-up, excessive current spikes, etc., the state of the I/O pins during power up/down varies depending on power. Prior to exiting POR, the pads are in a high impedance state (Hi-Z). To: There are no power up/down sequencing requirements to prevent issues such as latch-up, excessive current spikes, and so on. Therefore, the state of the I/O pins during power up/down varies depending on which supplies are powered. • Section 3.7.1, "Input Value of Pins During POR Dependent on VDD33," changed From: To avoid accidentally selecting the bypass clock because PLLCFG[0:1] and RSTCFG are not treated as ones (1s) when POR negates, VDD33 must not lag VDDSYN and the RESET pin power (VDDEH6) when powering the device by more than the VDD33 lag specification in Table 6. VDD33 individually can lag either VDDSYN or the RESET power pin (VDDEH6) by more than the VDD33 lag specification. VDD33 can lag one of the VDDSYN or VDDEH6 supplies, but cannot lag both by more than the VDD33 lag specification. This VDD33 lag specification only applies during power up. VDD33 has no lead or lag requirements when powering down. To: When powering the device, VDD33 must not lag VDDSYN and the RESET power pin (VDDEH6) by more than the VDD33 lag specification listed in Table 6. This avoids accidentally selecting the bypass clock mode because the internal versions of PLLCFG[0:1] and RSTCFG are not powered and therefore cannot read the default state when POR negates. VDD33 can lag VDDSYN or the RESET power pin (VDDEH6), but cannot lag both by more than the VDD33 lag specification. This VDD33 lag specification only applies during power up. VDD33 has no lead or lag requirements when powering down. • Table 22 Bus Operation Timing: Added the correct pins to the calibration signals: CAL_ADDR[10:11, 27:30], CAL_$\overline{WE}$/$\overline{BE}$[0:1], CAL_$\overline{CS}$[0, 2:3], and CAL_DATA[0:15]. Added calibration signals to Specs 5 and 8. • Corrected the following EBI signals: Specs 7 and 8: Added the following signals to Specs 7 and 8 the EBI section: $\overline{OE}$, RD_$\overline{WR}$, and $\overline{BDIP}$. Broke out Spec 6 CLKOUT Posedge to output signal valid into Spec 6 for the EBI signals, and Spec 6a for the calibration signals, Broke out Spec 7 Input Signal Valid to CLKOUT Posedge into Spec 7 for the EBI signals, and Spec 7a for the calibration signals. • Section 3.7.3, "Power-Down Sequence (VRC33 Grounded)" Deleted the underscore in ORed_POR to become ORed POR.
Rev 2.0	NH	2/09/07	Table 22 Bus Operation Timing: Removed references to CAL_$\overline{OE}$, CAL_RD_$\overline{WR}$, and CAL_$\overline{TS}$ because they really use the EBI signals $\overline{OE}$, RD_$\overline{WR}$, and $\overline{TS}$ on the MPC5553.

Table 33. MPC5553 Revision History (*continued*)

Revision	Author	Date	Substantive Change(s)
Rev 2.0	NH	2/27/07	Per RD comments: Table 2 Absolute Maximum Ratings: changed footnote 6 from: Keep the negative DC current greater than 0.6 V on eTPUB[15] and SINB during the internal power-on reset (POR) state. To: Keep the negative DC current greater than –0.6 V on eTPUB[15] and SINB during the internal power-on reset (POR) state. Figure 38 MPC5553 208 Map BGA Pinout: Deleted two lines referring to the $\overline{CS}$[0] signal ball assignment for the 208.
Rev 2.0	NH	3/1/07	Corrected the signal names in Section 3.14, “Fast Ethernet AC Timing Specifications” to include the FEC_ prefix for the signal name. Waiting on response from Jim Eifert, Randy Dees, Jeffery Hopkins, and Bill Terry about the following Bugs filed against the Data Sheets: 1474, 1480, 1482, 1483, 1811, 1815, 1884, 2254, 2419, 2717, 2873 before preparing for final sign-off again.

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