

February 1984 Revised August 2000

# MM74HC4538 **Dual Retriggerable Monostable Multivibrator**

### **General Description**

The MM74HC4538 high speed monostable multivibrator (one shots) is implemented in advanced silicon-gate CMOS technology. They feature speeds comparable to low power Schottky TTL circuitry while retaining the low power and high noise immunity characteristic of CMOS circuits.

Each multivibrator features both a negative, A, and a positive, B, transition triggered input, either of which can be used as an inhibit input. Also included is a clear input that when taken low resets the one shot. The MM74HC4538 is retriggerable. That is, it may be triggered repeatedly while their outputs are generating a pulse and the pulse will be extended.

Pulse width stability over a wide range of temperature and supply is achieved using linear CMOS techniques. The output pulse equation is simply: PW = 0.7(R)(C) where PW is in seconds, R is in ohms, and C is in farads. This device is pin compatible with the CD4528, and the CD4538 one shots. All inputs are protected from damage due to static discharge by diodes to  $V_{\mbox{\footnotesize CC}}$  and ground.

#### **Features**

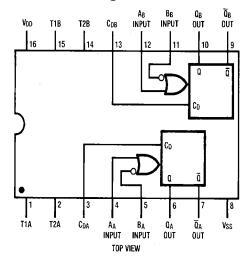
- Schmitt trigger on A and B inputs
- Wide power supply range: 2-6V
- Typical trigger propagation delay: 32 ns
- Fanout of 10 LS-TTL loads
- Low input current: 1 µA max

## **Ordering Code:**

Order Number	Package Number	Package Description
MM74HC4538M	M16A	16-Lead Small Outline Integrated Circuit (SOIC), JEDEC MS-012, 0.150 Narrow
MM74HC4538SJ	M16D	16-Lead Small Outline Package (SOP), EIAJ TYPE II, 5.3mm Wide
MM74HC4538N	N16E	16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300 Wide

Devices also available in Tape and Reel. Specify by appending the suffix letter "X" to the ordering code.

# **Connection Diagram**



# **Truth Table**

	Inputs	Outputs			
Clear	Α	В	Q	Q	
L	X	X	L	Н	
X	Н	X	L	Н	
Х	X	L	L	Н	
Н	L	$\downarrow$	丕	ъ	
Н	$\uparrow$	Н		~	

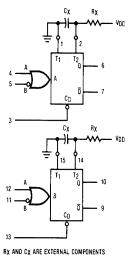
- H = HIGH Level
- L = LOW Level

  ↑ = Transition from LOW-to-HIGH
- $\downarrow = \text{Transition from HIGH-to-LOW}$
- \_\_ = One HIGH Level Pulse \_\_ = One LOW Level Pulse
- X = Irrelevant

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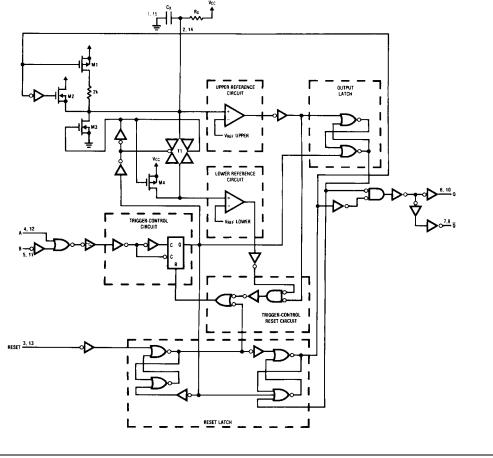
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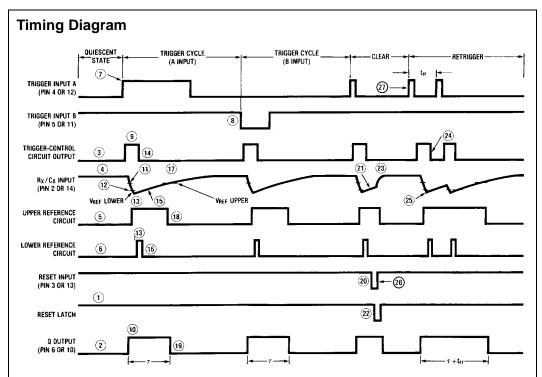
# **Block Diagrams**



Note: Pin 1 and Pin 15 must be hard-wired to GND.

# **Logic Diagram**





## **Circuit Operation**

The MM74HC4538 operates as follows (refer to logic diagram). In the quiescent state, the external timing capacitor,  $C_X$ , is charged to  $V_{CC}$ . When a trigger occurs, the Q output goes HIGH and  $C_X$  discharges quickly to the lower reference voltage ( $V_{REF}$  Lower =  $^1/_3$   $V_{CC}$ ).  $C_X$  then charges, through  $R_X$ , back up to the upper reference voltage ( $V_{REF}$  Upper =  $^2/_3$   $V_{CC}$ ), at which point the one-shot has timed out and the Q output goes LOW.

The following, more detailed description of the circuit operation refers to both the logic diagram and the timing diagram.

## QUIESCENT STATE

In the quiescent state, before an input trigger appears, the output latch is HIGH and the reset latch is HIGH (#1 in logic diagram).

Thus the Q output (pin 6 or 10) of the monostable multivibrator is LOW (#2, timing diagram).

The output of the trigger-control circuit is LOW (#3), and transistors M1, M2, and M3 are turned off. The external timing capacitor,  $C_X$ , is charged to  $V_{CC}$  (#4), and the upper reference circuit has a LOW output (#5). Transistor M4 is turned ON and transmission gate T1 is turned OFF. Thus the lower reference circuit has  $V_{CC}$  at the noninverting input and a resulting LOW output (#6).

In addition, the output of the trigger-control reset circuit is LOW.

#### TRIGGER OPERATION

The MM74HC4538 is triggered by either a rising-edge signal at input A (#7) or a falling-edge signal at input B (#8), with the unused trigger input and the Reset input held at the voltage levels shown in the Truth Table. Either trigger signal will cause the output of the trigger-control circuit to go HIGH (#9).

The trigger-control circuit going HIGH simultaneously initiates three events. First, the output latch goes LOW, thus taking the Q output of the HC4538 to a HIGH State (#10). Second, transistor M3 is turned on, which allows the external timing capacitor,  $C_{\rm X}$ , to rapidly discharge toward ground (#11). (Note that the voltage across  $C_{\rm X}$ appears at the input of the upper reference circuit comparator.) Third, transistor M4 is turned off and transmission gate T1 is turned ON, thus allowing the voltage across  $C_{\rm X}$  to also appear at the input of the lower reference circuit comparator

When  $C_X$  discharges to the reference voltage of the lower reference circuit (#12), the outputs of both reference circuits will be HIGH (#13). The trigger-control reset circuit goes HIGH, resetting the trigger-control circuit flip-flop to a LOW State (#14). This turns transistor M3 OFF again, allowing  $C_X$  to begin to charge back up toward  $V_{CC}$ , with a time constant  $t = R_X C_X$  (#15). In addition, transistor M4 is turned ON and transmission gate T1 is turned OFF. Thus a high voltage level is applied to the input of the lower reference circuit comparator, causing its output to go LOW (#16). The monostable multivibrator may be retriggered at any time after the trigger-control circuit goes LOW.

When  $C_X$  charges up to the reference voltage of the upper reference circuit (#17), the output of the upper reference circuit goes LOW (#18). This causes the output latch to

#### Circuit Operation (Continued)

toggle, taking the Q output of the HC4538 to a LOW State (#19), and completing the time-out cycle.

#### RESET OPERATION

A low voltage applied to the Reset pin always forces the Q output of the HC4538 to a LOW State.

The timing diagram illustrates the case in which reset occurs (#20) while  $C_{\rm X}$  is charging up toward the reference voltage of the upper reference circuit (#21). When a reset occurs, the output of the reset latch goes LOW (#22), turning ON transistor M1. Thus  $C_{\rm X}$  is allowed to quickly charge up to  $V_{\rm CC}$  (#23) to await the next trigger signal.

Recovery time is the required delay after reset goes inactive to a new trigger rising edge. On the diagram it is shown as (#26) to (#27).

#### **RETRIGGER OPERATION**

In the retriggerable mode, the HC4538 may be retriggered during timing out of the output pulse at any time after the trigger-control circuit flip-flop has been reset (#24). Because the trigger-control circuit flip-flop resets shortly after  $C_X$  has discharged to the reference voltage of the lower reference circuit (#25), the minimum retrigger time,  $t_{rr}$  is a function of internal propagation delays and the discharge time of  $C_X$ :

$$t_{rr}(ns) \approx 72 + \frac{V_{CC}(volts) \cdot C_X(pF)}{30.5}$$

at room temperature

#### POWER-DOWN CONSIDERATIONS

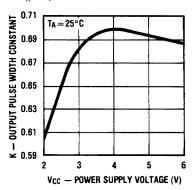
Large values of  $C_X$  may cause problems when powering down the MM74HC4538 because of the amount of energy stored in the capacitor. When a system containing this

device is powered down, the capacitor may discharge from  $V_{CC}$  through the input protection diodes at pin 2 or pin 14. Current through the protection diodes must be limited to 30 mA; therefore, the turn-off time of the  $V_{CC}$  power supply must not be faster than t =  $V_{CC}{}^{\bullet}C_X/(30 \text{ mA})$ . For example, if  $V_{CC}$  = 5V and  $C_X$  = 15  $\mu F$ , the  $V_{CC}$  supply must turn OFF no faster than t = (15V) $^{\bullet}(15~\mu F)/30~mA$  = 2.5 ms. This is usually not a problem because power supplies are heavily filtered and cannot discharge at this rate.

When a more rapid decrease of V<sub>CC</sub> to zero volts occurs, the MM74HC4538 may sustain damage. To avoid this possibility, use an external clamping diode, D<sub>X</sub>, connected from V<sub>CC</sub> to the C<sub>X</sub> pin.

#### **SET UP RECOMMENDATIONS**

 $\label{eq:minimum} \begin{aligned} &\text{Minimum} & &R_X=1 \text{ } k\Omega \\ &\text{Minimum} & &C_X=0 \text{ } pF. \end{aligned}$ 



# Absolute Maximum Ratings(Note 1)

(Note 2) Supply Voltage ( $V_{CC}$ ) -0.5 to +7.0V DC Input Voltage (V<sub>IN</sub>) -1.5 to  $V_{CC} + 1.5V$ DC Output Voltage ( $V_{OUT}$ ) -0.5 to V<sub>CC +</sub>0.5V Clamp Diode Current (I<sub>IK</sub>, I<sub>OK</sub>) ±20 mA DC Output Current, per pin (I<sub>OUT</sub>) ±25 mA DC  $V_{CC}$  or GND Current, per pin ( $I_{CC}$ ) ±50 mA Storage Temperature Range (T<sub>STG</sub>) -65°C to +150°C Power Dissipation (P<sub>D</sub>) 600 mW (Note 3) 500 mW S.O. Package only Lead Temperature (T<sub>L</sub>)

# Recommended Operating Conditions

	Min	Max	Units
Supply Voltage (V <sub>CC</sub> )	2	6	V
DC Input or Output Voltage	0	$V_{CC}$	V
$(V_{IN}, V_{OUT})$			
Operating Temperature Range (T <sub>A</sub> )	-40	+85	°C
Input Rise or Fall Times			
(Reset only)			
$(t_{\rm r}, t_{\rm f}) \ V_{\rm CC} = 2.0 V$		1000	ns
$V_{CC} = 4.5V$		500	ns
$V_{CC} = 6.0V$		400	ns

Note 1: Maximum Ratings are those values beyond which damage to the device may occur.

Note 2: Unless otherwise specified all voltages are referenced to ground.

Note 3: Power Dissipation Temperature Derating: Plastic "N" Package: –
12mW/°C from 65°C to 85°C.

# DC Electrical Characteristics (Note 4)

(Soldering 10 seconds)

Symbol	Parameter	Conditions	V <sub>CC</sub>	$T_A =$	25°C	$T_A = -40 \text{ to } 85^{\circ}\text{C}$	$T_A = -55 \text{ to } 125^{\circ}\text{C}$	Units	
Syllibol	rarameter	Conditions	• 66	Тур	Guara	anteed Limits		Units	
V <sub>IH</sub>	Minimum HIGH Level Input		2.0V		1.5	1.5	1.5	V	
	Voltage		4.5V		3.15	3.15	3.15	V	
			6.0V		4.2	4.2	4.2	V	
V <sub>IL</sub>	Maximum LOW Level Input		2.0V		0.5	0.5	0.5	V	
	Voltage		4.5V		1.35	1.35	1.35	V	
			6.0V		1.8	1.8	1.8	V	
V <sub>OH</sub>	Minimum HIGH Level Output	$V_{IN} = V_{IH}$ or $V_{IL}$							
	Voltage	$ I_{OUT}  \le 20 \ \mu A$	2.0V	2.0	1.9	1.9	1.9	V	
			4.5V	4.5	4.4	4.4	4.4	V	
			6.0V	6.0	5.9	5.9	5.9	V	
		$V_{IN} = V_{IH}$ or $V_{IL}$							
		$ I_{OUT}  \le 4.0 \text{ mA}$	4.5V		3.98	3.84	3.7	V	
		$ I_{OUT}  \le 5.2 \text{ mA}$	6.0V		5.48	5.34	5.2	V	
V <sub>OL</sub>	Maximum LOW Level Output	$V_{IN} = V_{IH}$ or $V_{IL}$	2.0V	0	0.1	0.1	0.1	V	
	Voltage	$ I_{OUT}  \le 20 \ \mu A$	4.5V	0	0.1	0.1	0.1	V	
			6.0V	0	0.1	0.1	0.1	V	
		$V_{IN} = V_{IH}$ or $V_{IL}$							
		$ I_{OUT}  \le 4.0 \text{ mA}$	4.5V		0.26	0.33	0.4	V	
		$ I_{OUT}  \le 5.2 \text{ mA}$	6.0V		0.26	0.33	0.4	V	
I <sub>IN</sub>	Maximum Input Current (Pins 2, 14) (Note 5)	$V_{IN} = V_{CC}$ or GND	6.0V		±0.1	±1.0	±1.0	μА	
I <sub>IN</sub>	Maximum Input Current (all other pins)	$V_{IN} = V_{CC}$ or GND	6.0V		±0.1	±1.0	±1.0	μΑ	
Icc	Maximum Active Supply	Pins 2, 14 = 0.5 V <sub>CC</sub>							
Active	Current	Q1, Q2 = HIGH	6.0V		150	250	400	μΑ	
		$V_{IN} = V_{CC}$ or GND							
I <sub>CC</sub>	Maximum Quiescent Supply	Pins 2, 14 = OPEN							
Quiescent	Current	Q1, Q2 = LOW	6.0V		130	220	350	μΑ	
		$V_{IN} = V_{CC}$ or GND							
					1	1			

260°C

Note 4: For a power supply of 5V  $\pm$ 10% the worst case output voltages ( $V_{OH}$ , and  $V_{OL}$ ) occur for HC at 4.5V. Thus the 4.5V values should be used when designing with this supply. Worst case  $V_{IH}$  and  $V_{IL}$  occur at  $V_{CC} = 5.5V$  and 4.5V respectively. (The  $V_{IH}$  value at 5.5V is 3.85V.) The worst case leakage current ( $I_{IN}$ ,  $I_{CC}$ , and  $I_{OZ}$ ) occur for CMOS at the higher voltage and so the 6.0V values should be used.

# DC Electrical Characteristics (Continued) Note 5: The device must be set up with 3 steps before measuring $I_{\text{IN}}$ :

	Clear	Α	В
1.	Н	L	Н
2.	Н	Н	Н
3.	Н	L	Н

## **AC Electrical Characteristics**

 $V_{CC} = 5V$ ,  $T_A = 25^{\circ}$  C,  $C_L = 15$  pF,  $t_r = t_f = 6$  ns

Symbol Parameter		Conditions	Тур	Limit	Units
t <sub>PLH</sub>	Maximum Propagation Delay A, or B to Q		23	45	ns
t <sub>PHL</sub>	Maximum Propagation Delay A, or B to $\overline{\mathbb{Q}}$		26	50	ns
t <sub>PHL</sub>	Maximum Propagation Delay Clear to Q		23	45	ns
t <sub>PLH</sub>	Maximum Propagation Delay Clear to Q		26	50	ns
t <sub>W</sub>	Minimum Pulse Width A, B or Clear		10	16	ns

#### **AC Electrical Characteristics**

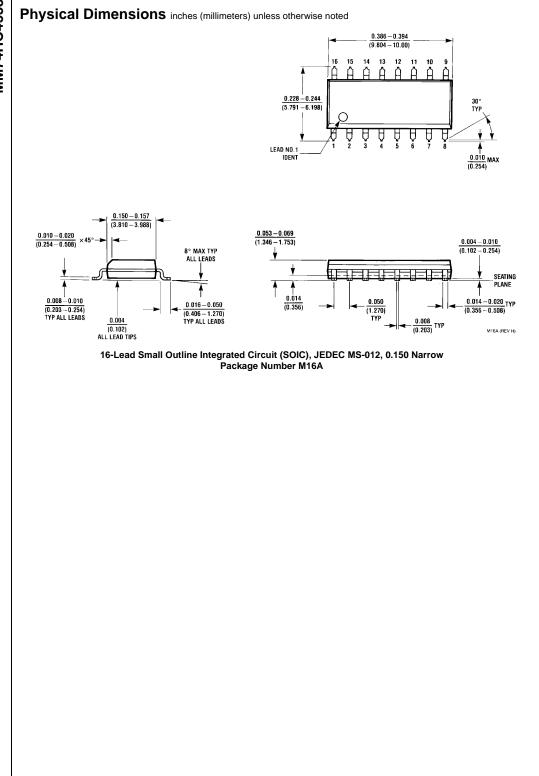
 $C_L = 50 \text{ pF}, t_r = t_f = 6 \text{ ns} \text{ (unless otherwise specified)}$ 

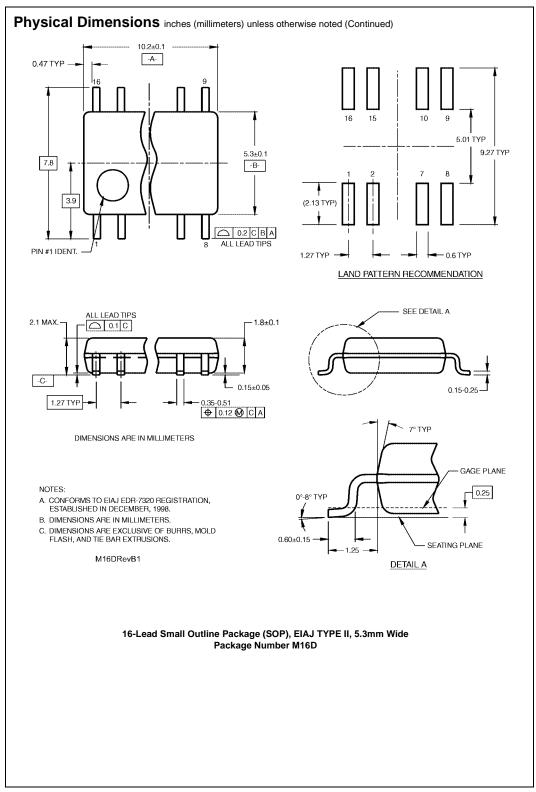
Symbol	Parameter	Condit	iono	V <sub>CC</sub>	T <sub>A</sub> =25°C		T <sub>A</sub> =-40 to 85°C	T <sub>A</sub> = -55 to 125°C	Units
	Farameter	Condit	Conditions		Тур	Guar	anteed Limits		
t <sub>PLH</sub>	Maximum Propagation			2.0V	100	250	315	373	ns
	Delay A, or B to Q			4.5V	25	50	63	75	ns
				6.0V	21	43	54	63	ns
t <sub>PHL</sub>	Maximum Propagation			2.0V	110	275	347	410	ns
	Delay A, or B to Q			4.5V	28	55	69	82	ns
				6.0V	23	47	59	70	ns
t <sub>PHL</sub>	Maximum Propagation			2.0V	100	250	315	373	ns
	Delay Clear to Q			4.5V	25	50	63	75	ns
				6.0V	21	43	54	63	ns
t <sub>PLH</sub>	Maximum Propagation			2.0V	110	275	347	410	ns
	Delay Clear to Q			4.5V	28	55	69	82	ns
				6.0V	23	47	59	70	ns
t <sub>TLH</sub> , t <sub>THL</sub>	Maximum Output			2.0V	30	75	95	110	ns
	Rise and Fall			4.5V	10	15	19	22	ns
	Time			6.0V	8	13	16	19	ns
t <sub>r</sub> , t <sub>f</sub>	Maximum Input			2.0V		1000	1000	1000	ns
	Rise and Fall			4.5V		500	500	500	ns
	Time (Reset only)			6.0V		400	400	400	ns
t <sub>W</sub>	Minimum Pulse Width			2.0V		80	101	119	ns
	A, B, Clear			4.5V		16	20	24	ns
				6.0V		14	17	20	ns
t <sub>REC</sub>	Minimum Recovery			2.0V	-5	0	0	0	ns
	Time, Clear			4.5V		0	0	0	ns
	Inactive to A or B			6.0V		0	0	0	ns
t <sub>WQ</sub>	Output Pulse Width	C <sub>X</sub> = 12 pF	Min	3.0V	283	190			ns
		$R_X = 1 k\Omega$		5.0V	147	120			ns
			Max	3.0V	283	400			ns
				5.0V	147	185			ns
t <sub>WQ</sub>	Output Pulse Width	C <sub>X</sub> = 100 pF	Min	3.0V	1.2				μs
		$R_X = 10 \text{ k}\Omega$		5.0V	1.0				μs
			Max	3.0V	1.2				μs
				5.0V	1.0				μs

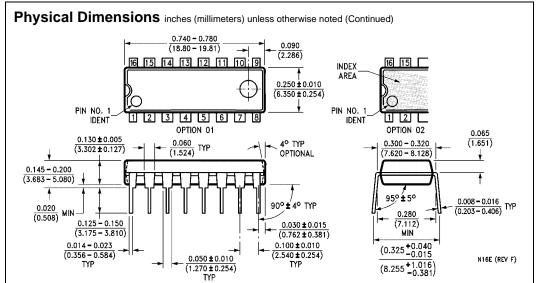
# AC Electrical Characteristics (Continued)

Symbol	Parameter	Conditions		V <sub>CC</sub>	T <sub>A</sub> =25°C		T <sub>A</sub> =-40 to 85°C	$T_A = -55 \text{ to } 125^{\circ}\text{C}$	Units
Cymbol				- 00	Тур	Guar	anteed Limits		Oille
t <sub>WQ</sub>	Output Pulse Width	C <sub>X</sub> = 1000 pF	Min	3.0V	10.5	9.4			μs
		$R_X = 10 \text{ k}\Omega$		5.0V	10.0	9.3			μs
			Max	3.0V	10.5	11.6			μs
				5.0V	10.0	10.7			μs
t <sub>WQ</sub>	Output Pulse Width	$C_X = 0.1 \mu F$	Min	5.0V		0.63	0.602	0.595	ms
		$R_X = 10k$	Max	5.0V		0.77	0.798	0.805	ms
C <sub>IN</sub>	Maximum Input				25				pF
	Capacitance (Pins 2 & 14)								
C <sub>IN</sub>	Maximum Input				5	10	10	10	pF
	Capacitance (other inputs)								
C <sub>PD</sub>	Power Dissipation	(per one shot)			150				pF
	Capacitance (Note 6)								
$\Delta t_{WQ}$	Pulse Width Match								
	Between Circuits in				±1				%
	Same Package								

Note 6:  $C_{PD}$  determines the no load dynamic consumption,  $P_D = C_{PD} \ V_{CC}^2 f + I_{CC} \ V_{CC}$ , and the no load dynamic current consumption,  $I_S = V_{CC} \ f + I_{CC} \ V_{CC}$ 







16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300 Wide Package Number N16E

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