## LMX2525

## PLLatinum ${ }^{\text {TM }}$ Dual Frequency Synthesizer System with Integrated VCOs

## General Description

LMX2525 is a highly integrated, high performance, low power frequency synthesizer system optimized for dualband Japan PDC mobile handsets. Using a proprietary digital phase locked loop technique, LMX2525 generates very stable, low noise local oscillator signals for up and down conversion in wireless communications devices.
LMX2525 includes dual voltage controlled oscillators (VCOs) for the upper and lower Japan PDC frequency bands, a loop filter, and a fractional-N RF PLL based on a delta sigma modulator. In concert, these blocks form a closed loop RF synthesizer system. The RF synthesizer system supports two frequency bands: PDC1500 and PDC800
Serial data is transferred to the device via a three-wire MICROWIRE interface (DATA, LE, CLK).
Operating supply voltage ranges from 2.5 V to 3.3 V . LMX2525 features low current consumption: 10 mA at 2.8 V when operating in the PDC800 mode.
LMX2525 is available in a 24 -pin leadless leadframe package (LLP).

## Features

- Small Size 5.0 mm X 4.0 mm X 0.75 mm 24-Pin LLP Package
- RF Synthesizer System Two Integrated VCOs Integrated Loop Filter Low Spurious, Low Phase Noise Fractional-N RF PLL Based on 10-Bit Delta Sigma Modulator Frequency Resolution Down to 20 kHz
- Supports Various Reference Frequencies 12.6 MHz, 14.4 MHz, 25.2MHz, and 26.0 MHz
- Fast Lock Time: 300 ss
- Low Current Consumption 10 mA at 2.8 V in PDC800 Mode
- 2.5 V to 3.3 V Operation
- Digital Filtered Lock Detect Output
- Hardware and Software Power Down Contro


## Applications

■ Japan PDC Systems at 800 MHz and 1500 MHz Frequency Bands.

Functional Block Diagram


20068907

## Connection Diagram



Note: Analog ground connected through exposed die attached pad.
20068902

## Pin Descriptions

| Pin Number | Name | I/O | Description |
| :---: | :---: | :---: | :--- |
| 1 | L1 | - | RF2 VCO tank pin. An external inductor is required between pins <br> L1 and L2 to set the resonant frequency of RF2 VCO (PDC800). |
| 2 | L2 | - | RF2 VCO tank pin. An external inductor is required between pins <br> L1 and L2 to set the resonant frequency of RF2 VCO (PDC800). |
| 3 | NC | - | Do not connect to any node on the printed circuit board. |
| 4 | NC | - | Do not connect to any node on the printed circuit board. |
| 5 | V $_{\text {DD }}$ | - | Supply voltage for RF analog circuitry |
| 6 | RF1out | O | RF output of RF1 VCO for PDC1500 |
| 7 | V $_{\text {DD }}$ | - | Supply voltage for RF analog circuitry |
| 8 | RF2out | O | RF output of RF2 VCO for PDC800 |
| 9 | V $_{\text {DD }}$ | - | Supply voltage for analog circuitry |
| 10 | VCC | - | Supply voltage for digital circuitry |
| 11 | GND | - | Ground for digital circuitry |
| 12 | VCC | - | Supply voltage for digital circuitry |
| 13 | LE | I | MICROWIRE Latch Enable |
| 14 | DATA | I | MICROWIRE Data |
| 15 | CLK | I | MICROWIRE Clock |
| 16 | CE | I | Chip enable control pin |
| 17 | BS | I | Band select control pin |
| 18 | LD | O | Lock detect pin |
| 19 | V | - | Supply voltage for digital circuitry |
| 20 | OSCin | I | Reference frequency input |
| 21 | GND | - | Ground for digital circuitry |
| 22 | VDD | - | Supply voltage for analog circuitry |
| 23 | GND | - | Ground for analog circuitry |
| 24 | V |  | - |
| SD | Supply voltage for RF analog circuitry |  |  |
|  |  |  |  |

## Ordering Information

| Part Number | RF1 Min. <br> $(\mathbf{M H z})$ | RF1 Max. <br> $(\mathbf{M H z})$ | RF1 <br> Center <br> $(M H z)$ | RF2 Min. <br> $(\mathbf{M H z})$ | RF2 Max. <br> $(\mathbf{M H z})$ | Package <br> Marking | Supplied As |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| LMX2525LQX1321 | 1270.22 | 1394.95 | $\sim 1321$ | 633.15 | 768.00 | 25251321 | 4500 units on <br> tape and reel |
| LMX2525LQ1321 | 1270.22 | 1394.95 | $\sim 1321$ | 633.15 | 768.00 | 25251321 | 1000 units on <br> tape and reel |

## Part Number Description



Typical Application Circuit (Note 1)


Note 1: Refer to RF2 VCO Tuning Range vs. External Inductance plot to aid in selecting the appropriate external inductance, PCB trace and L1, for the desired frequency range.

## Absolute Maximum Ratings (Notes 2, 3, 4) <br> If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

| Parameter | Symbol | Ratings | Units |
| :--- | :--- | :--- | :---: |
| Supply Voltage | $\mathrm{V}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{DD}}$ | -0.5 to 3.6 | V |
| Voltage on any pin <br> to GND | $\mathrm{V}_{\mathrm{I}}$ | -0.3 to $\mathrm{V}_{\mathrm{CC}}+0.3$ | V |
|  |  | -0.3 to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
| Storage Temperature Range | $\mathrm{T}_{\mathrm{STG}}$ | -65 to 150 | ${ }^{\circ} \mathrm{C}$ |

## Recommended Operating Conditions

| Parameter | Symbol | Min | Typ | Max | Unit |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Ambient Temperature | $\mathrm{T}_{\mathrm{A}}$ | -30 | 25 | 85 | ${ }^{\circ} \mathrm{C}$ |
| Supply Voltage (to GND) | $\mathrm{V}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{DD}}$ | 2.5 |  | 3.3 | V |

Note 2: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Recommended Operating Conditions indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, refer to the Electrical Characteristics section. The guaranteed specifications apply only for the conditions listed.
Note 3: This device is a high performance RF integrated circuit with an ESD rating $<2 \mathrm{kV}$ and is ESD sensitive. Handling and assembly of this device should be done at ESD protected workstations.
Note 4: GND $=0$ V.

Electrical Characteristics $\left(\mathrm{V}_{\mathrm{IN}}=2.8 \mathrm{~V}\right.$, refer to Typical Application Circuit; Limits in standard typeface are for $T_{A}=25^{\circ} \mathrm{C}$; Limits in boldface type apply over the operating temperature range from $-20^{\circ} \mathrm{C} \leq T_{A} \leq 75^{\circ} \mathrm{C}$ unless otherwise noted.)

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Icc PARAMETERS |  |  |  |  |  |  |
| $\mathrm{I}_{\mathrm{CC}}+\mathrm{I}_{\mathrm{DD}}$ | Supply Current (Note 5) | OB_CRL [1:0] = 10 |  | 10.6 | $\begin{aligned} & 12.0 \\ & 12.3 \end{aligned}$ | mA |
|  |  | OB_CRL [1:0] = 00 |  | 10.0 | $\begin{aligned} & 11.5 \\ & 11.8 \end{aligned}$ | mA |
| $\overline{I_{C C}+I_{D D}}$ | Supply Current (Note 6) | OB_CRL [1:0] = 10 |  | 15.0 | $\begin{aligned} & 16.5 \\ & 16.8 \end{aligned}$ | mA |
|  |  | OB_CRL [1:0] = 00 |  | 14.2 | $\begin{aligned} & \hline 15.6 \\ & 15.9 \end{aligned}$ | mA |
| $\mathrm{I}_{\text {PD }}$ | Power Down Current | $\begin{aligned} & \mathrm{CE}=\mathrm{LOW} \text { or } \\ & \text { RF_PD }=1 \end{aligned}$ |  |  | 20 | $\mu \mathrm{A}$ |
| REFERENCE OSCILLATOR PARAMETERS |  |  |  |  |  |  |
| $\mathrm{f}_{\text {OSCin }}$ | Reference Oscillator Input Frequency (Note 7) | Supports 12.6, 14.4, 25.2 and 26.0 MHz . | 12.6 | 14.4 | 26.0 | MHz |
| $\mathrm{V}_{\text {OSCin }}$ | Reference Oscillator Input Sensitivity |  |  | 0.5 | $\mathrm{V}_{\mathrm{Cc}}$ | Vp-p |
| RF1 VCO FOR PDC1500 |  |  |  |  |  |  |
| $\mathrm{f}_{\text {RF1 }{ }^{\text {out }}}$ | Frequency Range (Note 8) | RF1 VCO for PDC1500 | 1270.22 |  | 1394.95 | MHz |
| $\mathrm{P}_{\text {RF1 }{ }^{\text {out }}}$ | Output Power | OB_CRL [1:0] = 11 | -5 | -2 | 1 | dBm |
|  |  | OB_CRL [1:0] = 10 | -7 | -4 | -1 | dBm |
|  |  | OB_CRL [1:0] = 01 | -10 | -7 | -4 | dBm |
|  |  | OB_CRL [1:0] = 00 | -13 | -10 | -7 | dBm |
|  | Lock Time | Full frequency span within each band in High Speed Mode (HS = 1). |  |  | $\begin{gathered} 300 \\ (\text { Note 9) } \end{gathered}$ | $\mu \mathrm{s}$ |
|  |  | Between bands High Speed Mode (HS = 1). |  |  | $\begin{gathered} 300 \\ (\text { Note 9) } \end{gathered}$ | $\mu \mathrm{s}$ |
|  |  | Full frequency span within each band in Normal Mode (HS = 0). <br> (Note 11) |  |  | $\begin{gathered} 500 \\ (\text { Note 9) } \end{gathered}$ | $\mu \mathrm{s}$ |
|  |  |  |  |  | 375 <br> (Note 10) | $\mu \mathrm{s}$ |
|  |  | Between bands in Normal Mode (HS = 0). |  |  | 500 <br> (Note 9) | $\mu \mathrm{s}$ |
|  |  |  |  |  | $\begin{gathered} 400 \\ (\text { Note 10) } \end{gathered}$ | $\mu \mathrm{s}$ |
|  | RMS Phase Error |  |  | 1.3 |  | degrees |

Electrical Characteristics $\left(\mathrm{V}_{\mathrm{IN}}=2.8 \mathrm{~V}\right.$, refer to Typical Application Circuit; Limits in standard typeface are for $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$; Limits in boldface type apply over the operating temperature range from $-20^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 75{ }^{\circ} \mathrm{C}$ unless otherwise noted.) (Continued)


Electrical Characteristics $\left(\mathrm{V}_{\mathrm{IN}}=2.8 \mathrm{~V}\right.$, refer to Typical Application Circuit; Limits in standard typeface are for $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$; Limits in boldface type apply over the operating temperature range from $-20^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 75{ }^{\circ} \mathrm{C}$ unless otherwise noted.) (Continued)

| Symbol | Parameter | Condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RF2 VCO FOR PDC800 |  |  |  |  |  |  |
|  | Spurious Tones | @ $\leq 25 \mathrm{kHz}$ offset |  |  | -45 | dBc |
|  |  | @ 25 kHz < offset $\leq 50 \mathrm{kHz}$ |  |  | -60 | dBc |
|  |  | @ 50 kHz < offset $\leq 100 \mathrm{kHz}$ |  |  | -69 | dBc |
|  |  | @ offset > 100 kHz |  |  | -75 | dBc |

DIGITAL INTERFACE (DATA, CLK, LE, LD, CE, BS)

| $\mathrm{V}_{\mathrm{IH}}$ | High-Level Input Voltage |  | 0.8 V CC |  | $\mathrm{V}_{\mathrm{Cc}}$ | V |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | 0.8 V DD |  | $V_{\text {DD }}$ | V |
| $\mathrm{V}_{\text {IL }}$ | Low-Level Input Voltage |  | -0.3 |  | $0.2 \mathrm{~V}_{\mathrm{CC}}$ | V |
|  |  |  | -0.3 |  | $0.2 \mathrm{~V}_{\mathrm{DD}}$ | V |
| $\mathrm{I}_{\mathrm{IH}}$ | High-Level Input Current |  | -10 |  | 10 | $\mu \mathrm{A}$ |
| ILL | Low-Level Input Current |  | -10 |  | 10 | $\mu \mathrm{A}$ |
|  | Input Capacitance |  |  | 3 |  | pF |
|  | Rise/Fall Time |  |  | 30 |  | ns |
| $\mathrm{V}_{\mathrm{OH}}$ | High-Level Output Voltage |  | $\mathrm{V}_{\mathrm{CC}}-0.4$ |  |  | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}-0.4$ |  |  | V |
| $\overline{\mathrm{V}_{\mathrm{OL}}}$ | Low-Level Output Voltage |  |  |  | 0.4 | V |
|  | Output Capacitance |  |  |  | 5 | pF |

## MICROWIRE INTERFACE TIMING

| $\mathrm{t}_{\mathrm{CS}}$ | Data to Clock Set Up Time |  | 50 |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{t}_{\mathrm{CH}}$ | Data to Clock Hold Time |  | 10 |  |  |
| $\mathrm{t}_{\mathrm{CWH}}$ | Clock Pulse Width HIGH |  | 50 |  |  |
| $\mathrm{t}_{\mathrm{CWL}}$ | Clock Pulse Width LOW |  | 50 |  | ns |
| $\mathrm{t}_{\mathrm{DS}}$ | Latch Enable LOW to Data Set Up <br> Time |  | 50 |  | ns |
| $\mathrm{t}_{\mathrm{ES}}$ | Clock to Latch Enable Set Up Time |  | 50 |  | ns |
| $\mathrm{t}_{\mathrm{EW}}$ | Latch Enable Pulse Width |  | 50 |  | ns |

Note 5: RF PLL and VCO in PDC800 mode.
Note 6: RF PLL and VCO in PDC1500 mode.
Note 7: The reference frequency must be programmed using the OSC_FREQ control bit. For other reference frequencies, please contact National Semiconductor.
Note 8: For other frequency ranges, please contact National Semiconductor.
Note 9: Lock time is defined as the time difference between the beginning of the frequency transition and the point at which the frequency remains within $+/-1 \mathrm{kHz}$ of the final frequency.
Note 10: Lock time is defined as the time difference between the beginning of the frequency transition and the point at which the frequency remains within $+/-3 \mathrm{kHz}$ of the final frequency.
Note 11: Lock time specification also applies to power up with MICROWIRE serial interface.
Note 12: All limits are guaranteed. All electrical characteristics having room temperature limits are tested during production with $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ or correlated using Statistical Quality Control (SQC) methods. All hot and cold limits are guaranteed by correlating the electrical characteristics to process and temperature variations and applying statistical process control.

MICROWIRE Interface Timing Diagram


Typical Performance Characteristics (Note 13)


Note 13: Typical performance characteristics do not guarantee specific performance limits. For guaranteed specifications, refer to the Electrical Characteristics section.

Note 14: The frequency range is defined as the difference between the highest frequency and the lowest frequency of a given unit. For a chosen external inductance, the typical frequency range equals the difference between the Typical Maximum Frequency and the Typical Minimum Frequency. Typical frequency range may be assumed on any unit with that chosen external inductance, even if the unit has worst case Maximum Frequency or worst case Minimum Frequency.

## Functional Description

## GENERAL

The LMX2525 is a highly integrated frequency synthesizer system for Japan PDC wireless communication systems. The LMX2525 supports dual band operation for 800 MHz and 1500 MHz .

The LMX2525 includes all functional blocks for the RF PLL including RF VCOs, frequency dividers, PFDs, and loop filters. Only external passive elements for the RF2 VCO tank and supply bypassing are required to complete the RF synthesizer.
The LMX2525 uses a patent pending Fractional-N synthesizer architecture based on a delta sigma modulator to support fine frequency resolution. Four of the most common reference frequencies for PDC applications, 12.6 MHz, 14.4 MHz , 25.2 MHz and 26.0 MHz , are supported. The unique feature of this architecture is its low spurious modulation effect.
The use of a fractional synthesizer based on a delta sigma modulator allows for faster lock and system set-up times, which reduces system power consumption. The loop filter is included on chip to minimize the external noise coupling and to reduce the form factor applicable to the board level application. Only one of the two RF VCOs is activated at a given time, and each output is provided through its own output pin.

## RF PLL SECTION

## Frequency Selection

The divide ratio can be calculated using the following equations:
$f_{\text {Vco }}=\left\{8 \times R F_{-} B+R F_{-} A+\left(R F_{-} F N / F D\right)\right\} \times\left(f_{\text {Osc }} / R\right)$ where (RF_A < RF_B) for PDC1500
$\mathrm{f}_{\mathrm{Vco}}=\left\{4 \times\right.$ RF_B + RF_A $\left.+\left(R F \_F N / F D\right)\right\} \times\left(f_{\mathrm{Osc}} / R\right)$ where (RF_A < RF_B) for PDC800
$\mathrm{f}_{\mathrm{Vco}}$ : Output frequency of voltage controlled oscillator (VCO) RF_B: Preset divide ratio of binary 4-bit programmable counter ( $2 \leq R F \_B \leq 15$ )
RF_A: Preset divide ratio of binary 3-bit swallow counter ( $0 \leq$ RF_A $\leq 7$ for PDC1500 and $0 \leq R F \_A \leq 3$ for PDC800)

RF_FN: Preset numerator of binary 10-bit modulus counter ( $0 \leq R F$ _FN < FD)
FD: Preset denominator for modulus counter (FD $=\mathrm{f}_{\mathrm{Osc}} /(R$ $X f_{\mathrm{CH}}$ ) where $\mathrm{f}_{\mathrm{CH}}$ is the channel spacing)
$f_{\text {osc }}$ : Reference oscillator frequency
R: Internal reference oscillator frequency divider (1 for 12.6
MHz and 14.4 MHz, 2 for 25.2 MHz and 26.0 MHz )
The denominator, FD , in the above equation is dependent on the channel spacing and reference oscillator frequency. The channel spacing will change based on the Rx/Tx and BS bits. Table 6 in the R0 Register section summarizes the values of FD.

## VCO Frequency Tuning

The center frequency of the RF VCOs are determined by the resonant frequency of the tank circuit, illustrated in Figure 1. With an internal fixed bonding-wire inductor and an external inductance, the center frequency of the VCO is given as follows:

$$
\mathrm{f}_{\text {center }}=\frac{1}{2 \pi \sqrt{\left(L_{\text {fixed }}+L_{\text {external }}\right) \cdot C_{\text {total }}}}
$$

where $\mathrm{C}_{\text {total }}$ is the total capacitance of the VCO, including the parasitic capacitance and the nominal self-tuning capacitance. Note, the external inductance consists of the PCB traces and lumped element inductor. The output frequency tuning range can be optimized for the specific application by selecting the appropriate external inductance. Refer to RF2 VCO Tuning Range vs. External Inductance plot to aid in selecting the appropriate external inductance. Care should be taken to ensure proper frequency coverage when choosing the tolerance of the lumped element inductor. For the 1500 MHz band, the internal bonding-wires provide the necessary inductance to set the VCO center frequency.


## FIGURE 1. External Inductor Connection

In real implementation, the inductance of $\mathrm{L}_{\text {fixed }}$ and $\mathrm{L}_{\text {external }}$ can vary from its nominal value. The LMX2525 utilizes a built-in tracking algorithm to compensate for variations up to $\pm 15 \%$ and tunes the VCO to the required frequency. During the frequency acquisition period, the loop bandwidth is extended to achieve the frequency lock. After the frequency lock, the loop bandwidth of the PLL is set to the nominal value and the phase lock is achieved. The transition between the two operating modes is very smooth and extremely fast to meet the stringent PDC requirements for lock time and phase noise.

## POWER DOWN MODE

The LMX2525 includes the power down mode to reduce the power consumption. The LMX2525 enters the power down mode either by taking the CE pin LOW or by setting the RF_PD bit in the RO register. If the CE pin is set LOW, the circuit is powered down regardless of the register values. When the CE pin is HIGH, the RF_PD bit controls power to the RF circuitry. Data can be written to the registers even when the CE pin is set LOW. The following truth table summarizes the power down logic.

TABLE 1. Power Down Modes

| CE Pin | RF_PD Bit | Mode |
| :---: | :---: | :---: |
| HIGH | 0 | Active |
| HIGH | 1 | Not Active |
| LOW | 0 | Not Active |
| LOW | 1 | Not Active |

## Functional Description (Continued)

## BAND SELECT MODE

The BS pin and BS bit can be used to select one of the two RF VCO outputs. When using the BS pin, the BS bit must be set to 0 , and when using the BS bit, the BS pin must be tied to ground. When using the BS pin, the state of the input must exceed the minimum band select set up time prior to the LE signal transition. The truth table summarizing the band select logic is as follows:

TABLE 2. Band Select Modes

| BS Pin | BS Bit | Mode |
| :---: | :---: | :---: |
| HIGH | 0 | PDC1500 |
| LOW | 0 | PDC800 |
| LOW | 1 | PDC1500 |

## LOCK DETECT MODE

The LD output can be used to indicate the lock status of the PLL. Bit 6 in Register R1 determines the signal that appears
on the LD pin. When the PLL is not locked, the LD pin remains LOW. After obtaining phase lock, the LD pin will have a logical HIGH level. The LD output is always LOW when the LD register bit is 0 and in power down mode.

TABLE 3. Lock Detect Modes

| LD Bit | Mode |
| :---: | :---: |
| 0 | Disable (GND) |
| 1 | Enable |

TABLE 4. Lock Detect Logic

| RF PLL Section | LD Output |
| :---: | :---: |
| Locked | HIGH |
| Not Locked | LOW |



FIGURE 2. Lock Detect Timing Diagram Waveform (Notes 15, 16, 17, 18, 19)

Note 15: LD output becomes LOW when the phase error is larger than $t_{W 2}$ Note 16: LD output becomes HIGH when the phase error is less than $\mathrm{t}_{\mathrm{W} 1}$ for four or more consecutive cycles.
Note 17: Phase Error is determined on the leading edge. Only errors greater than $\mathrm{t}_{\mathrm{W} 1}$ and $\mathrm{t}_{\mathrm{W} 2}$ are labeled.

Note 18: $t_{W 1}$ is 5 ns for PDC1500 and 10 ns for PDC800. $\mathrm{t}_{\mathrm{W} 2}$ is 10 ns for both bands.
Note 19: The lock detect comparison occurs with every $64^{\text {th }}$ cycle of $f_{R}$ and $\mathrm{f}_{\mathrm{N}}$.


FIGURE 3. Lock Detect Flow Diagram

LOCKING MODES
Two frequency-locking modes are provided: a Normal mode and a High Speed mode for faster lock times. The HS bit in register R0 controls the locking mode.

TABLE 5. Locking Modes

| HS Bit | Mode |
| :---: | :---: |
| 0 | Normal mode |
| 1 | High Speed mode |

## MICROWIRE INTERFACE

The programmable register set is accessed via the MICROWIRE serial interface. The interface is comprised of three signal pins: CLK, DATA, and LE (Latch Enable). Serial data is clocked into the 24-bit shift register on the rising edge of the clock. The last bits decode the internal control register address. When the latch enable (LE) transitions from LOW to HIGH, data stored in the shift registers is loaded into the corresponding control register. The data is loaded MSB first.

## Programming Description

## GENERAL PROGRAMMING INFORMATION

The serial interface has a 24 -bit shift register to store the incoming data bits temporarily. The incoming data is first loaded into the shift register from MSB to LSB. The data is shifted at the rising edge of the clock signal. When the latch enable signal transitions from LOW to HIGH, the data stored in shift register is transferred to the proper register depending on the address bit setting. The selection of the particular register is determined by the control bits indicated in boldface text.
At initial start-up, the MICROWIRE loading requires three default words (registers R2, loaded first, to R0, loaded last). After the device has been initially programmed, the RF VCO frequency can be changed using a single register (RO).
The control register content map describes how the bits within each control register are allocated to the specific control functions.

## COMPLETE REGISTER MAP

|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $\begin{gathered} \hline \text { LSB } \\ \hline 0 \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { ত্ত } \\ & \text { © } \end{aligned}$ | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
| Ro (Default) | $\begin{aligned} & \mathrm{RX} / \\ & \mathrm{TX} \end{aligned}$ | $\begin{aligned} & \mathrm{RF}_{-} \\ & \mathrm{PD} \end{aligned}$ | HS | 0 | BS | $\begin{aligned} & \hline \text { RF_B } \\ & {[3: 0]} \end{aligned}$ |  |  |  | $\begin{aligned} & \text { RF_A } \\ & {[2: 0]} \end{aligned}$ |  |  | $\begin{gathered} \text { RF_FN } \\ {[9: 0]} \end{gathered}$ |  |  |  |  |  |  |  |  |  | 0 | 0 |
| R1 (Default) | $\begin{aligned} & \hline \text { SPI_ } \\ & \mathrm{DEF} \end{aligned}$ | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | LD |  |  | OS <br> FR <br> [1: |  | 0 | 1 |
| R2 (Default) | 1 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 |
| R3 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 1 | 1 |
| R4 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 0 | 1 | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 |
| R5 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |

Note: R0 control register will be used when hot start frequency change.
Note: Boldface text represent address bits.

## Programming Description (Continued)

## RO REGISTER

The R0 register address bits (R0 [1:0]) are "00".
The $R x / T x$ bit selects between receive and transmit modes and, in conjunction with the band select bit (BS), the channel spacing to be synthesized.
The RF_PD bit selects the power down mode of the RF PLL and selected VCO.
The HS bit selects between normal and high speed locking mode.
The BS bit determines which of the two internal VCOs (PDC800 or PDC1500) is active.
The RF N counter consists of the 4-bit programmable counter (RF_B counter), the 3-bit swallow counter (RF_A counter) and the 10-bit delta sigma modulator (RF_FN counter). The equations for calculating the counter values are presented below.

RO REGISTER

|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 |  | 1 | 0 |
|  | Data Field |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Address <br> Field |  |
| R0 <br> (Default) | $\begin{aligned} & \mathrm{RX} / \\ & \mathrm{TX} \end{aligned}$ | $\begin{aligned} & \mathrm{RF}_{-} \\ & \mathrm{PD} \end{aligned}$ | HS | 0 | BS |  | $\begin{aligned} & \hline \text { RF_B } \\ & {[3: 0]} \end{aligned}$ |  |  | $\begin{gathered} \hline \text { RF_A } \\ {[2: 0]} \end{gathered}$ |  |  |  |  |  | $\begin{gathered} \hline \text { RF_FN } \\ {[9: 0]} \end{gathered}$ |  |  |  |  |  |  | 0 | 0 |


| Name | Functions |
| :--- | :--- |
| RX/TX | RX/TX Mode |
|  | $0=$ Rx |
| $1=$ Tx |  |


| Counter Name | Symbol | Functions |
| :---: | :---: | :---: |
| Modulus Counter | RF_FN | RF N Divider$\begin{aligned} & N=8 \times R F \_B+R F \_A+R F \_F N / F D(P D C 1500) \\ & N=4 \times R F \_B+R F \_A+R F \_F N / F D(P D C 800) \end{aligned}$ |
| Programmable Counter | RF_B |  |
| Swallow Counter | RF_A |  |

## Programming Description (Continued)

## PULSE SWALLOW FUNCTION

$f_{V C O}=\left\{8 \times R F_{-} B+R F \_A+\left(R F \_F N / F D\right)\right\} \times f_{\text {Osc }} / R$ where (RF_A < RF_B) for PDC1500
$f_{\text {vco }}=\left\{4 \times\right.$ RF_B + RF_A $\left.+\left(R F \_F N / F D\right)\right\} \times f_{\text {Osc }} / R$ where (RF_A < RF_B) for PDC800
$\mathrm{f}_{\mathrm{Vco}}$ : Output frequency of voltage controlled oscillator (VCO)
RF_B: Preset divide ratio of binary 4-bit programmable counter ( $2 \leq R F \_B \leq 15$ )
RF_A: Preset divide ratio of binary 3-bit swallow counter ( $0 \leq R F \_A \leq 7$ for PDC1500 and $0 \leq R F \_A \leq 3$ for PDC800)
RF_FN: Preset numerator of binary 10 -bit modulus counter ( $0 \leq R F \_F N<F D$ )
FD: Preset denominator for modulus counter ( $F D=f_{\mathrm{OsC}} /\left(\mathrm{Rxf} \mathrm{f}_{\mathrm{CH}}\right)$ where $\mathrm{f}_{\mathrm{CH}}$ is the channel spacing)
$\mathrm{f}_{\mathrm{Osc}}$ : Reference oscillator frequency
R: Internal reference oscillator frequency divider

| OSC_FREQ [1:0] | Reference Oscillator Frequency (MHz) | R Divider |
| :---: | :---: | :---: |
| 00 | 12.6 | 1 |
| 01 | 14.4 | 1 |
| 10 | 25.2 | 2 |
| 11 | 26.0 | 2 |

The value of the denominator (FD) is depended on the channel spacing and reference oscillator frequency. Table 6 summarizes the denominator values based on the settings of the $R x / T x, B S$, and OSC_FREQ [1:0] bits.

TABLE 6. Demonimator Values

| $\mathbf{R x} / \mathbf{T x}$ | $\mathbf{B S}$ | OSC_FREQ [1:0] | Reference Oscillator Frequency (MHz) | $\mathbf{R}$ | $\mathbf{f}_{\mathbf{C H}}(\mathbf{k H z})$ | Denominator(FD) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 00 | 12.6 | 1 | 25.0 | 504 |
| 0 | 0 | 01 | 14.4 | 1 | 25.0 | 576 |
| 0 | 0 | 10 | 25.2 | 2 | 25.0 | 504 |
| 0 | 0 | 11 | 26.0 | 2 | 25.0 | 520 |
| 0 | 1 | 00 | 12.6 | 1 | 25.0 | 504 |
| 0 | 1 | 01 | 14.4 | 1 | 25.0 | 576 |
| 0 | 1 | 10 | 25.2 | 2 | 25.0 | 504 |
| 0 | 1 | 11 | 26.0 | 2 | 25.0 | 520 |
| 1 | 0 | 00 | 12.6 | 1 | 20.0 | 630 |
| 1 | 0 | 01 | 25.2 | 1 | 20.0 | 720 |
| 1 | 0 | 10 | 26.0 | 2 | 20.0 | 630 |
| 1 | 0 | 11 | 12.6 | 2 | 20.0 | 650 |
| 1 | 1 | 00 | 14.4 | 1 | 22.22 | 567 |
| 1 | 1 | 01 | 25.2 | 1 | 22.22 | 648 |
| 1 | 1 | 10 | 26.0 | 2 | 22.22 | 567 |
| 1 | 1 | 11 |  | 2 | 22.22 | 585 |

## Programming Description (Continued)

## R1 REGISTER

The R1 register address bits (R1 [1:0]) are "01".
The SPI_DEF bit allows for the programming of words R3 to R5. Under most circumstances, the SPI_DEF bit should be set to 1.

The LD bit sets the function of the lock detect pin. Enabling the lock detect function provides a digital lock detect output of the active RF synthesizer at the LD pin.
The OB_CRL [1:0] bits determine the power level of the RF output buffer. The power level can be adjusted to best meet the system requirement. Refer to the Electrical Characteristics section for power output specifications.
The reference frequency selection bits, OSC_FREQ [1:0], are used to set the reference clock and R divider for use with one of the following reference frequencies: $12.6 \mathrm{MHz}, 14.4 \mathrm{MHz}, 25.2 \mathrm{MHz}$ or 26.0 MHz . The LMX2525 uses the OSC_FREQ bits along with the $B S$ and $R X / T X$ bits to determine the correct divide ratios needed to meet the required channel spacing for the mode of operation selected. Refer to Table 6 for a summary of denominator values.

R1 REGISTER

|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | Data Field |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Address <br> Field |  |
| R1 <br> (Default) | $\begin{aligned} & \hline \text { SPI_ } \\ & \mathrm{DEF} \end{aligned}$ | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | LD | OB <br> CRL <br> [1:0] |  |  |  | 0 | 1 |


| Name | Functions |
| :--- | :--- |
| SPI_DEF | Default Register Selection |
|  | $0=$ OFF (Use values set in R0 to R5) |
|  | $1=$ ON (Use default values set in R0 to R2) |
| LD | Lock Detect |
|  | $0=$ Disable (GND) |
|  | $1=$ Enable |
| OB_CRL [1:0] | Output Buffer Control |
|  | $00=$ Minimum Output Power |
|  | $01=$ |
|  | $10=$ |
|  | $11=$ Maximum Output Power |
| OSC_FREQ [1:0] | Reference Frequency Selection |
|  | $00=12.6 \mathrm{MHz}$ |
|  | $01=14.4 \mathrm{MHz}$ |
|  | $10=25.2 \mathrm{MHz}$ |
|  | $11=26.0 \mathrm{MHz}$ |

## Programming Description (Continued)

## R2 REGISTER

The R2 register address bits (R2 [1:0]) are " 10 ".
R2 REGISTER

|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 1 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | Data Field |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Address <br> Field |  |
| R2 <br> (Default) | 1 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | 0 |  | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 |

## R3 REGISTER

The R3 register address bits (R3 [2:0]) are " 011 ". This register is only written to if the SPI_DEF bit is set to 0 .
R3 REGISTER

|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | Data Field |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Address Field |  |  |
| R3 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 1 | 1 |

## R4 REGISTER

The R4 register address bits (R4 [3:0]) are " 0111 ". This register is only written to if the SPI_DEF bit is set to 0 .
R4 REGISTER

|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | Data Field |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Address <br> Field |  |  |  |
| R4 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 0 | 1 | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 |

## R5 REGISTER

The R5 register address bits (R5 [4:0]) are "01111". This register is only written to if the SPI_DEF bit is set to 0 .
R5 REGISTER

|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | Data Field |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Address Field |  |  |  |  |
| R5 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |

Physical Dimensions inches (millimeters) unless otherwise noted


24-Pin Leadless Leadframe Package (LLP)
Order Number LMX2525
NSC Package Number LQA24A

## LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT AND GENERAL COUNSEL OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

## BANNED SUBSTANCE COMPLIANCE

National Semiconductor certifies that the products and packing materials meet the provisions of the Customer Products Stewardship Specification (CSP-9-111C2) and the Banned Substances and Materials of Interest Specification (CSP-9-111S2) and contain no "Banned Substances" as defined in CSP-9-111S2.

| 1 National Semiconductor | National Semiconductor | National Semiconductor | National Semiconductor |
| :---: | :---: | :---: | :---: |
| Americas Customer | Europe Customer Support Center | Asia Pacific Customer | Japan Customer Support Center |
| Support Center | Fax: +49 (0) 180-530 8586 | Support Center | Fax: 81-3-5639-7507 |
| Email: new.feedback@nsc.com | Email: europe.support@ nsc.com | Email: ap.support@nsc.com | Email: jpn.feedback@nsc.com |
| Tel: 1-800-272-9959 | Deutsch Tel: +49 (0) 6995086208 |  | Tel: 81-3-5639-7560 |
|  | English Tel: +44 (0) 8702402171 |  |  |
| www.national.com | Français Tel: +33 (0) 141918790 |  |  |

National does not assume any responsibility for use of any circuitry described, no circuit patent licenses are implied and National reserves the right at any time without notice to change said circuitry and specifications.

