1:6 PCI Clock Generator/ Fanout Buffer

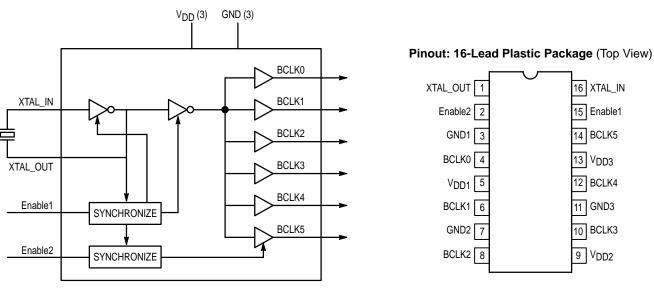
The MPC903, MPC904 and MPC905 are six output clock generation devices targeted to provide the clocks required in a 3.3V or 5.0V PCI environment. The device operates from a 3.3V supply and can interface to either a TTL input or an external crystal. The inputs to the device can be driven with 5.0V when the V_{CC} is at 3.3V. The outputs of the MPC903/904/905 meet all of the specifications of the PCI standard. The three devices are identical except in the function of the Output Enables.

- Six Low Skew Outputs
- Synchronous Output Enables for Power Management
- Low Voltage Operation
- XTAL Oscillator Interface
- 16-Lead SOIC Package
- 5.0V Tolerant Enable Inputs

The MPC903/904/905 device is targeted for PCI bus or processor bus environments with up to 12 clock loads. Each of the six outputs on the MPC903/904/905 can drive two series terminated 50 Ω transmission lines. This capability effectively makes the MPC903/904/905 a 1:12 fanout buffer.

The MPC903 offers two synchronous enable inputs to allow users flexibility in developing power management features for their designs. Both enable signals are active HIGH inputs. A logic '0' on the Enable1 input will pull all of the outputs into the logic '0' state and shut down the internal oscillator for a zero power sleep state. A logic '0' on the Enable2

input will disable only the BCLK5 output. The Enable2 input can be used to disable any high power device for system power savings during periods of inactivity. Both enable inputs are synchronized internal to the chip so that the output disabling will happen only when the outputs are already LOW. This feature guarantees no runt pulses will be generated during enabling and disabling. Note that when the MPC903 is re-enabled via the Enable1 pin, the user must allow for the oscillator to regain stability. Thus, the re-enabling of the chip cannot occur instantaneously. The MPC904 and MPC905 Enable functions are slightly different than the 903 and are outlined in the function tables on the following page.



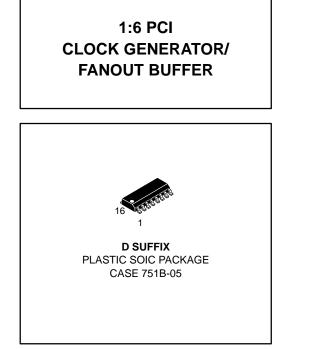
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MPC903

MPC904

MPC905

FUNCTION TABLE

		Outputs 0 to 4			Output 5			OSC (On/Off)		
ENABLE1	ENABLE2	MPC903	MPC904	MPC905	MPC903	MPC904	MPC905	MPC903	MPC904	MPC905
0	0	Low	Low	Low	Low	Low	Low	OFF	OFF	ON
0	1	Low	Low	Low	Low	Toggling	Toggling	OFF	ON	ON
1	0	Toggling	Toggling	Toggling	Low	Low	Low	ON	ON	ON
1	1	Toggling	Toggling	Toggling	Toggling	Toggling	Toggling	ON	ON	ON

ABSOLUTE MAXIMUM RATINGS*

Symbol	Parameter	Min	Max	Unit
V _{DD}	Supply Voltage	-0.5	4.6	V
VIN	Input Voltage	-0.5	V _{CC} + 0.5	V
T _{oper}	Operating Temperature Range	0	+70	°C
T _{stg}	Storage Temperature Range	-65	+150	°C
T _{sol}	Soldering Temperature Range (10 Sec)		+260	°C
Тј	Junction Temperature Range		+125	°C
P(E1=1)	Power Dissipation		TBD	mW
P(E1=0)	Power Dissipation		40	μW
ESD	Static Discharge Voltage	2000		V
ILatch	Latch Up Current	50		mA

* Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation should be restricted to the Recommended Operating Conditions.

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Min	Max	Unit
T _A	Ambient Temperature Range	0	70	°C
V _{DD}	Positive Supply Voltage (Functional Range)	3.0	3.6	V
tDCin	T _{high} (at XTAL_IN Input) T _{Iow} (at XTAL_IN Input)	0.44T 1 0.44T 1	0.56T 1 0.56T 1	T = Period

1. When using External Source for reference, requirement to meet PCI clock duty cycle requirement on the output.

DC CHARACTERISTICS (T_A = 0–70°C; V_{DD} = $3.3V \pm 0.3V$)

Symbol	Characteristic		Min	Тур	Max	Unit	Condition
VIH	High Level Input Voltage		2.0		5.5 2	V	
VIL	Low Level Input Voltage				0.8	V	
VOH	High Level Output Voltage		2.4			V	I _{OH} = -36mA ¹
VOL	Low Level Output Voltage				0.4	V	I _{OL} = 36mA ¹
IIH	Input High Current				2.5 2	μA	
۱ _{IL}	Input Low Current				2.5	μA	
ICC	Power Supply Current	DC 33MHz 66MHz		20 37 78	45 95	μA mA mA	
C _{IN}	Input Capacitance	XTAL_IN Others			9.0 4.5	pF	

 The MPC903/904/905 outputs can drive series terminated or parallel terminated 50Ω (or 50Ω to V_{CC}/2) transmission lines on the incident edge (see Applications Info).

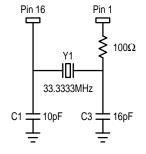
XTAL_IN input will sink up to 10mA when driven to 5.5V. There are no reliability concerns associated with the condition. Note that the Enable1 input must be a logic HIGH. Do not take the Enable1 input to a logic LOW with >V_{CC} volts on the XTAL_IN input.

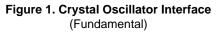
Symbol	Characteristic	Min	Тур	Max	Unit	Condition
F _{max}	Maximum Operating Using External Cry Frequency Using External Clock Sou			50 100	MHz	
tpw	Output Pulse Width HIGH (Above 2. LOW (Below 0. HIGH (Above 2. LOW (Below 0. LOW (Below 0.	8V) 0.40T 1 0V) 0.45T 2		0.60T 1 0.60T 1 0.55T 2 0.55T 2		T = Periods
tper	Output Period	T – 400ps				T = Desired Period
t _{os}	Output-to-Output Skew Rising Ed Falling Ed	-		400 500	ps	
t _r , t _f	Rise/Fall Times (Slew Rate)	1		4	V/ns	Series Terminated Transmission Lines
^t EN	Enable Time Enal Enal			5 4	ms Cycles	
^t DIS	Disable Time Enal Enal			4 4	Cycles	
A _{osc}	XTAL_IN to XTAL_OUT Oscillator Gain	6			db	
Phase	Loop Phase Shift Modulo 360° +	30			Degrees	

AC CHARACTERISTICS (T_A = 0–70°C; V_{DD} = $3.3V \pm 0.3V$)

1. Assuming input duty cycle specs from Recommended Operationg Conditions table are met.

2. Assuming external crystal or 50% duty cycle external reference is used.





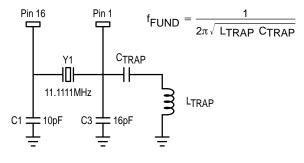
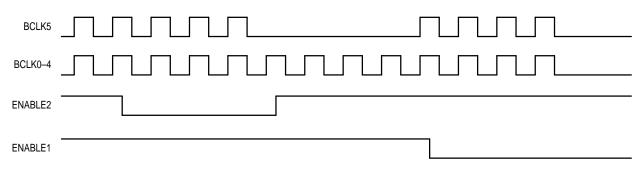


Figure 2. Crystal Oscillator Interface (3rd Overtone)

Table 1. Crystal Specifications

Parameter	Value				
Crystal Cut	Fundamental AT Cut				
Resonance	Parallel Resonance*				
Frequency Tolerance	±75ppm at 25°C				
Frequency/Temperature Stability	±150pm 0 to 70°C				
Operating Range	0 to 70°C				
Shunt Capacitance	5–7pF				
Equivalent Series Resistance (ESR)	50 to 80Ω				
Correlation Drive Level	100µW				
Aging	5ppm/Yr (First 3 Years)				

MPC903 MPC904 MPC905





APPLICATIONS INFORMATION

Driving Transmission Lines

The MPC903/904/905 clock driver was designed to drive high speed signals in a terminated transmission line environment. To provide the optimum flexibility to the user the output drivers were designed to exhibit the lowest impedance possible. With an output impedance of less than 10Ω the drivers can drive either parallel or series terminated transmission lines. For more information on transmission lines the reader is referred to application note AN1091 in the Timing Solutions brochure (BR1333/D).

In most high performance clock networks point–to–point distribution of signals is the method of choice. In a point–to–point scheme either series terminated or parallel terminated transmission lines can be used. The parallel technique terminates the signal at the end of the line with a 50Ω resistance to VCC/2. This technique draws a fairly high level of DC current and thus only a single terminated line can be driven by each output of the MPC903/904/905 clock driver. For the series terminated case however there is no DC current draw, thus the outputs can drive multiple series terminated lines. Figure 4 illustrates an output driving a single series terminated line vs two series terminated lines in parallel. When taken to its extreme the fanout of the MPC903/904/905 clock driver is effectively doubled due to its capability to drive multiple lines.

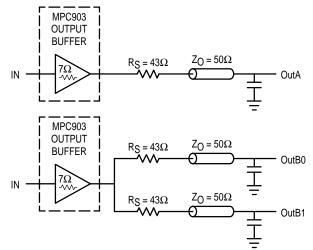


Figure 4. Single versus Dual Transmission Lines

The waveform plots of Figure 5 show the simulation results of an output driving a single line vs two lines. In both cases the drive capability of the MPC903/904/905 output buffers is more than sufficient to drive 50Ω transmission lines on the incident edge. Note from the delay measurements in the simulations a delta of only 43ps exists between the two differently loaded outputs. This suggests that the dual line

driving need not be used exclusively to maintain the tight output-to-output skew of the MPC903. The output waveform in Figure 5 shows a step in the waveform, this step is caused by the impedance mismatch seen looking into the driver. The parallel combination of the 43Ω series resistor plus the output impedance does not match the parallel combination of the line impedances. The voltage wave launched down the two lines will equal:

VL = VS (Zo / Rs + Ro +Zo) = 3.0 (25/53.5) = 1.40V

At the load end the voltage will double, due to the near unity reflection coefficient, to 2.8V. It will then increment towards the quiescent 3.0V in steps separated by one round trip delay (in this case 4.0ns).

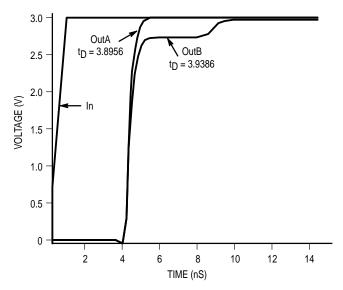


Figure 5. Single versus Dual Waveforms

Since this step is well above the threshold region it will not cause any false clock triggering, however designers may be uncomfortable with unwanted reflections on the line. To better match the impedances when driving multiple lines the situation in Figure 6 should be used. In this case the series terminating resistors are reduced such that when the parallel combination is added to the output buffer impedance the line impedance is perfectly matched.

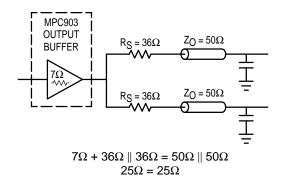
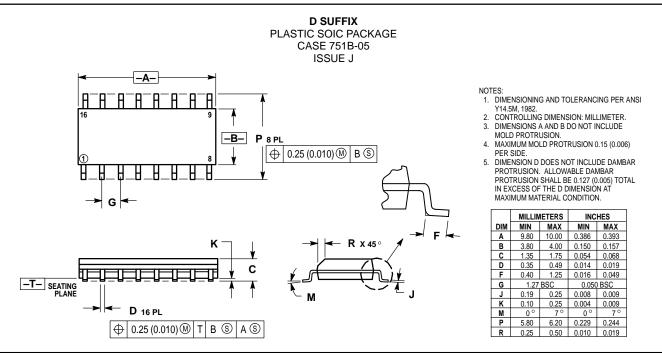


Figure 6. Optimized Dual Line Termination

SPICE level output buffer models are available for engineers who want to simulate their specific interconnect schemes. In addition IV characteristics are in the process of being generated to support the other board level simulators in general use.

OUTLINE DIMENSIONS



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