

# 1:5 DIFFERENTIAL-TO-LVDS ZERO DELAY CLOCK GENERATOR

**ICS8745B** 

### **Description**



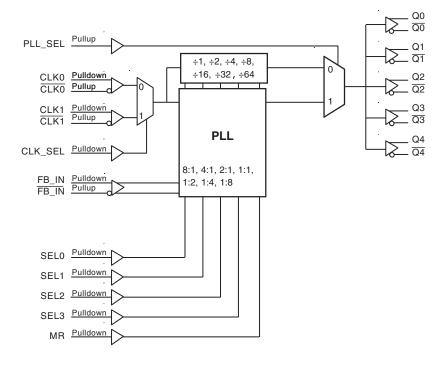
The ICS8745B is a highly versatile 1:5 LVDS Clock Generator and a member of the HiPerClockS<sup>™</sup> family of High Performance Clock Solutions from IDT. The ICS8745B has a fully integrated PLL and can be configured as zero delay buffer, multiplier or

divider, and has an output frequency range of 31.25MHz to 700MHz. The Reference Divider, Feedback Divider and Output Divider are each programmable, thereby allowing for the following output-to-input frequency ratios: 8:1, 4:1, 2:1, 1:1, 1:2, 1:4, 1:8. The external feedback allows the device to achieve "zero delay" between the input clock and the output clocks. The PLL\_SEL pin can be used to bypass the PLL for system test and debug purposes. In bypass mode, the reference clock is routed around the PLL and into the internal output dividers.

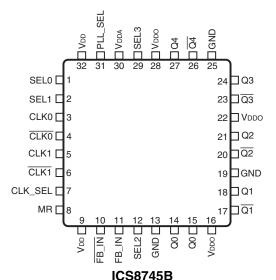
#### **Features**

- Five differential LVDS outputs designed to meet or exceed the requirements of ANSI TIA/EIA-644
- · Selectable differential clock inputs
- CLKx, CLKx pairs can accept the following differential input levels: LVPECL, LVDS, LVHSTL, HCSL, SSTL
- Output frequency range: 31.25MHz to 700MHz
- Input frequency range: 31.25MHz to 700MHz
- VCO range: 250MHz to 700MHz
- External feedback for "zero delay" clock regeneration with configurable frequencies
- Programmable dividers allow for the following output-to-input frequency ratios: 8:1, 4:1, 2:1, 1:1, 1:2, 1:4, 1:8
- Cycle-to-cycle jitter: 30ps (maximum)
- · Output skew: 35ps (maximum)
- Static phase offset: 25ps ± 125ps
- Full 3.3V supply voltage
- 0°C to 70°C ambient operating temperature
- Available in both standard (RoHS 5) and lead-free (RoHS 6) packages

## **Block Diagram**



## Pin Assignment



32-Lead LQFP 7mm x 7mm x 1.4mm package body Y Package Top View

IDT™ / ICS™ LVDS ZERO DELAY CLOCK GENERATOR

ICS8745BYREV. C OCTOBER 27, 2008

**Table 1. Pin Descriptions** 

Number	Name	Т	уре	Description
1, 2, 12, 29	SEL0, SEL1, SEL2 SEL3	Input	Pulldown	Determines output divider values in Table 3. LVCMOS / LVTTL interface levels.
3	CLK0	Input	Pulldown	Non-inverting differential clock input.
4	CLK0	Input	Pullup	Inverting differential clock input.
5	CLK1	Input	Pulldown	Non-inverting differential clock input.
6	CLK1	Input	Pullup	Inverting differential clock input.
7	CLK_SEL	Input	Pulldown	Clock select input. When HIGH, selects CLK1, CLK1. When LOW, selects CLK0, CLK0. LVCMOS / LVTTL interface levels.
8	MR	Input	Pulldown	Active HIGH Master Reset. When logic HIGH, the internal dividers are reset causing the true outputs Qx to go low and the inverted outputs $\overline{Qx}$ to go high. When logic LOW, the internal dividers and the outputs are enabled. LVCMOS / LVTTL interface levels.
9, 32	V <sub>DD</sub>	Power		Core supply pins.
10	FBIN	Input	Pullup	Inverting differential feedback input to phase detector for regenerating clocks with "Zero Delay."
11	FBIN	Input	Pulldown	Non-inverted differential feedback input to phase detector for regenerating clocks with "Zero Delay."
13, 19, 25	GND	Power		Power supply ground.
14, 15	Q0/Q0	Output		Differential output pair. LVDS interface levels.
16, 22, 28	$V_{DDO}$	Power		Output supply pins.
17, 18	Q1/Q1	Output		Differential output pair. LVDS interface levels.
20, 21	Q2/Q2	Output		Differential output pair. LVDS interface levels.
23, 24	Q3/Q3	Output		Differential output pair. LVDS interface levels.
26, 27	Q4/Q4	Output		Differential output pair. LVDS interface levels.
30	$V_{DDA}$	Power		Analog supply pin.
31	PLL_SEL	Input	Pullup	PLL select. Selects between the PLL and reference clock as the input to the dividers. When LOW, selects reference clock. LVCMOS/LVTTL interface levels.

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

### **Table 2. Pin Characteristics**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance			4		pF
R <sub>PULLUP</sub>	Input Pullup Resistor			51		kZ
R <sub>PULLDOWN</sub>	Input Pulldown Resistor			51		kZ

## **Function Tables**

**Table 3A. Control Input Function Table** 

	Inputs					
SEL3	SEL2	SEL1	SEL0	Reference Frequency Range (MHz)*	Q0:Q4/Q0:Q4	
0z	0	0	0	250 - 700	÷1	
0	0	0	1	125 - 350	÷1	
0	0	1	0	62.5 - 175	÷1	
0	0	1	1	31.25 - 87.5	÷1	
0	1	0	0	250 - 700	÷2	
0	1	0	1	125 - 350	÷2	
0	1	1	0	62.5 - 175	÷2	
0	1	1	1	250 - 700	÷4	
1	0	0	0	125 - 350	÷4	
1	0	0	1	250 - 700	÷8	
1	0	1	0	125 - 350	x2	
1	0	1	1	62.5 - 175	x2	
1	1	0	0	31.25 - 87.5	x2	
1	1	0	1	62.5 - 175	x4	
1	1	1	0	31.25 - 87.5	x4	
1	1	1	1	31.25 - 87.5	x8	

<sup>\*</sup>NOTE: VCO frequency range for all configurations above is 250MHz to 700MHz.

Table 3B. PLL Bypass Function Table

	Inp	Outputs PLL_SEL = 0 PLL Bypass Mode		
SEL3	SEL2	SEL1	SEL0	Q0:Q4/Q0:Q4
0z	0	0	0	÷4
0	0	0	1	÷4
0	0	1	0	÷4
0	0	1	1	÷8
0	1	0	0	÷8
0	1	0	1	÷8
0	1	1	0	÷16
0	1	1	1	÷16
1	0	0	0	÷32
1	0	0	1	÷64
1	0	1	0	÷2
1	0	1	1	÷2
1	1	0	0	÷4
1	1	0	1	÷1
1	1	1	0	÷2
1	1	1	1	÷1

## **Absolute Maximum Ratings**

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, V <sub>DD</sub>	4.6V
Inputs, V <sub>I</sub>	-0.5V to V <sub>DD</sub> + 0.5V
Outputs, I <sub>O</sub> Continuos Current Surge Current	10mA 15mA
Package Thermal Impedance, t <sub>JA</sub>	47.9 ·C/W (0 lfpm)
Storage Temperature, T <sub>STG</sub>	-65 ·C to 150 ·C

#### **DC Electrical Characteristics**

Table 4A. LVDS Power Supply DC Characteristics,  $V_{DD} = V_{DDA} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0$ °C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Core Supply Voltage		3.135	3.3	3.465	V
$V_{DDA}$	Analog Supply Voltage		3.135	3.3	3.465	V
$V_{DDO}$	Output Supply Voltage		3.135	3.3	3.465	V
I <sub>DD</sub>	Power Supply Current				125	mA
I <sub>DDA</sub>	Analog Supply Current				17	mA
I <sub>DDO</sub>	Output Supply Current				59	mA

Table 4B. LVCMOS/LVTTL DC Characteristics,  $V_{DD} = V_{DDA} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0$ °C to 70°°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V <sub>IH</sub>	Input High Voltage			2		V <sub>DD</sub> + 0.3	V
V <sub>IL</sub>	Input Low Voltage			-0.3		0.8	٧
I <sub>IH</sub>	Input High Current	CLK_SEL, SEL[0:3], MR	V <sub>DD</sub> = V <sub>IN</sub> = 3.465V			150	μA
		PLL_SEL	$V_{DD} = V_{IN} = 3.465V$			5	μΑ
I <sub>IL</sub>	Input Low Current	CLK_SEL, SEL[0:3], MR	V <sub>DD</sub> = 3.465V, V <sub>IN</sub> = 0V	-5			μA
		PLL_SEL	V <sub>DD</sub> = 3.465V, V <sub>IN</sub> = 0V	-150			μΑ

Table 4C. Differential DC Characteristics,  $V_{DD} = V_{DDA} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0$ °C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
	Input High Current	CLK0, CLK1, FB_IN	V <sub>DD</sub> = V <sub>IN</sub> = 3.465V			150	μΑ
l IH	Input High Current	CLKO, CLK1, FB_IN	$V_{DD} = V_{IN} = 3.465V$			5	μΑ
	Innut Lou Current	CLK0, CLK1, FB_IN	$V_{DD} = 3.465V, V_{IN} = 0V$	-5			μA
IIL	Input Low Current	CLKO, CLK1, FB_IN	$V_{DD} = 3.465V,$ $V_{IN} = 0V$	-150			μΑ
V <sub>PP</sub>	Peak-to-Peak Voltage			0.15		1.3	V
V <sub>CMR</sub>	Common Mode Input	Voltage; NOTE 1, 2		GND + 0.5		V <sub>DD</sub> – 0.85	V

NOTE 1: Common mode input voltage is defined as V<sub>IH</sub>.

NOTE 2: For single-ended applications, the maximum input voltage for CLKx,  $\overline{\text{CLKx}}$  is  $V_{DD}$  + 0.3V.

Table 4D. LVDS DC Characteristics,  $V_{DD} = V_{DDA} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0^{\circ}C$  to  $70^{\circ}C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OD</sub>	Differential Output Voltage		320	440	550	mV
$\Delta V_{OD}$	V <sub>OD</sub> Magnitude Change			0	50	mV
V <sub>OS</sub>	Offset Voltage		1.05	1.2	1.35	V
$\Delta V_{OS}$	V <sub>OS</sub> Magnitude Change				25	mV

## Table 5. Input Frequency Characteristics, $V_{DD} = V_{DDA} = V_{DDO} = 3.3V \pm 5\%$ , $T_A = 0$ °C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
_	Input Frequency	CLK0/CLK0,	PLL_SEL = 1	31.25		700	μΑ
FIN	input Frequency	CLK1/CLK1	PLL_SEL = 0			700	V

#### **AC Electrical Characteristics**

Table 6. AC Characteristics,  $V_{DD} = V_{DDA} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0$ °C to 70°C

Parameter	Symbol	Test Conditions	Minimum	Typical	Maximum	Units
f <sub>MAX</sub>	Output Frequency				700	MHz
t <sub>PD</sub>	Propagation Delay; NOTE 1	PLL_SEL = 0V, $f \le 700MHz$	3.1	3.4	4.0	ns
tsk(Ø)	Static Phase Offset; NOTE 2, 5	PLL_SEL = 3.3V	-100	25	150	ps
tsk(o)	Output Skew; NOTE 3, 5				35	ps
tjit(cc)	Cycle-to-Cycle Jitter; NOTE 5, 6				30	ps
<i>t</i> jit(θ)	Phase Jitter; NOTE 4, 5, 6				±52	ps
tL	PLL Lock Time				1	ms
t <sub>R</sub> / t <sub>F</sub>	Output Rise/Fall Time; NOTE 7	20% to 80%	200		700	ps
odc	Output Duty Cycle		46	50	54	%

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as the time difference between the input reference clock and the averaged feedback input signal across all conditions, when the PLL is locked and the input reference frequency is stable.

NOTE 3: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the output differential cross points.

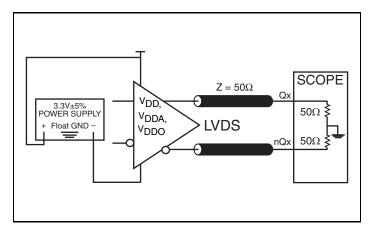
NOTE 4: Phase jitter is dependent on the input source used.

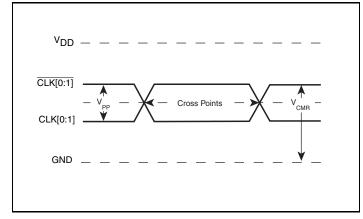
NOTE 5: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 6: Characterized at VCO frequency of 622MHz.

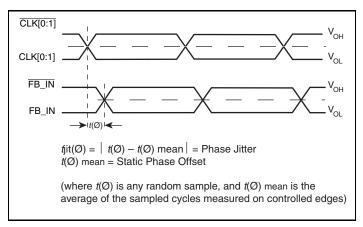
NOTE 7: Measured from the 20% to 80% points. Guaranteed by characterization. Not production tested.

#### **Parameter Measurement Information**

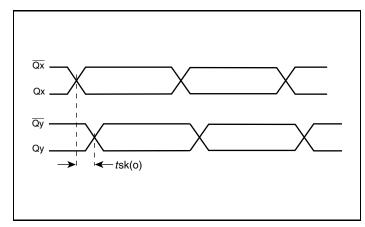




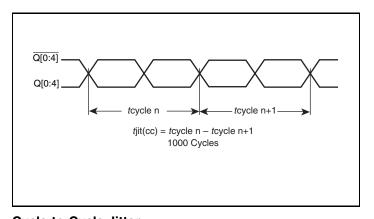
#### 3.3V LVDS Output Load AC Test Circuit



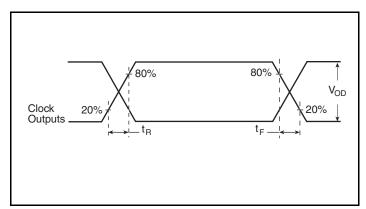
**Differential Input Level** 



#### **Phase Jitter and Static Phase Offset**



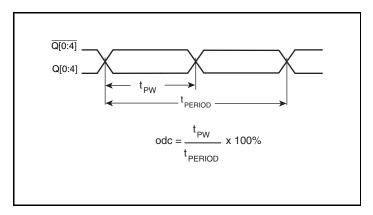
**Output Skew** 

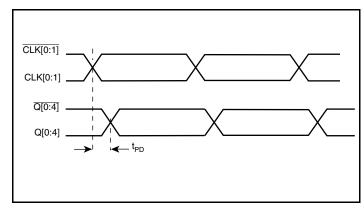


Cycle-to-Cycle Jitter

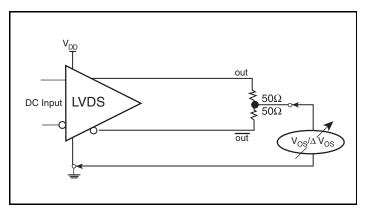
**Output Rise/Fall Time** 

## **Parameter Measurement Information, continued**



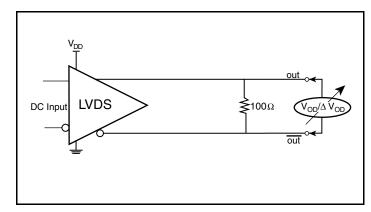


#### **Output Duty Cycle**



**Offset Voltage Setup** 

#### **Propagation Delay**



**Differential Output Voltage Setup** 

## **Application Information**

#### **Power Supply Filtering Technique**

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The ICS8745B provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL.  $V_{DD_i}$ ,  $V_{DDA}$  and  $V_{DDO}$  should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. Figure 1 illustrates how a  $10\Omega$  resistor along with a  $10\mu F$  and a  $.01\mu F$  bypass capacitor should be connected to each  $V_{DDA}$  pin. The  $10\Omega$  resistor can also be replaced by a ferrite bead.

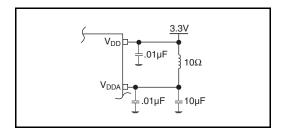


Figure 1. Power Supply Filtering

#### Wiring the Differential Input to Accept Single Ended Levels

Figure 2 shows how the differential input can be wired to accept single ended levels. The reference voltage V\_REF =  $V_{DD}/2$  is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio of R1 and R2 might need to be adjusted to position the V\_REF in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and  $V_{DD} = 3.3V$ , V\_REF should be 1.25V and R2/R1 = 0.609.

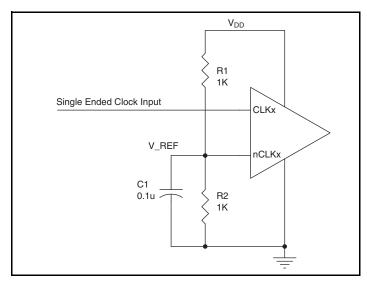


Figure 2. Single-Ended Signal Driving Differential Input

#### **Differential Clock Input Interface**

The CLK/CLK accepts LVDS, LVPECL, LVHSTL, SSTL, HCSL and other differential signals. Both signals must meet the  $V_{PP}$  and  $V_{CMR}$  input requirements. Figures 3A to 3D show interface examples for the HiPerClockS CLK/CLK input driven by the most common driver types. The input interfaces suggested here are

examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example, in Figure 3A, the input termination applies for IDT HiPerClockS LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

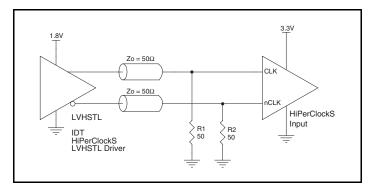


Figure 3A. HiPerClockS CLK/CLK Input Driven by an IDT HiPerClockS LVHSTLDriver

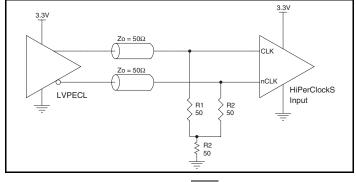


Figure 3B. HiPerClockS CLK/CLK Input
Driven by a 3.3V LVPECL Driver

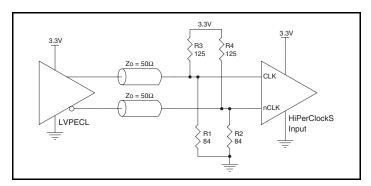


Figure 3C. HiPerClockS CLK/CLK Input
Driven by a 3.3V LVPECL Driver

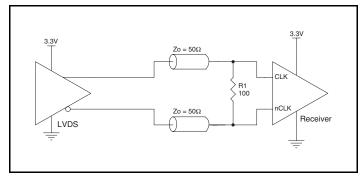


Figure 3D. HiPerClockS CLK/CLK Input Driven by a 3.3V LVDS Driver

#### **Recommendations for Unused Input and Output Pins**

#### Inputs:

#### **LVCMOS Control Pins:**

All control pins have internal pull-ups or pull-downs; additional resistance is not required but can be added for additional protection. A  $1k\Omega$  resistor can be used.

#### **CLK/CLK** INPUT:

For applications not requiring the use of the differential input, both CLK and  $\overline{\text{CLK}}$  can be left floating. Though not required, but for additional protection, a  $1 k\Omega$  resistor can be tied from CLK to ground.

#### **Outputs:**

#### **LVDS Output**

All unused LVDS output pairs can be either left floating or terminated with  $100\Omega$  across. If they are left floating, we recommend that there is no trace attached.

#### 3.3V LVDS Driver Termination

A general LVDS interface is shown in *Figure 4*. In a  $100\Omega$  differential transmission line environment, LVDS drivers require a matched load termination of  $100\Omega$  across near the receiver input.

For a multiple LVDS outputs buffer, if only partial outputs are used, it is recommended to terminate the unused outputs.

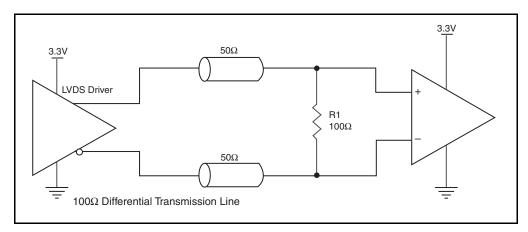


Figure 4. Typical LVDS Driver Termination

#### **Schematic Example**

The schematic of the ICS8745B layout example is shown in *Figure 5A*. The ICS8745B recommended PCB board layout for this example is shown in *Figure 5B*. This layout example is used as a

general guideline. The layout in the actual system will depend on the selected component types, the density of the components, the density of the traces, and the stack up of the P.C. board.

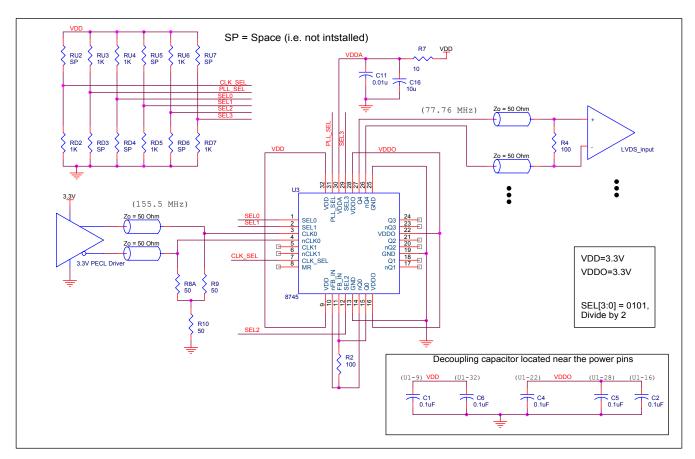


Figure 5A. ICS8745B LVDS Zero Delay Buffer Schematic Example

#### 1:5 DIFFERENTIAL-TO-LVDS ZERO DELAY CLOCK GENERATOR

The following component footprints are used in this layout example.

All the resistors and capacitors are size 0603.

#### **Power and Grounding**

Place the decoupling capacitors as close as possible to the power pins. If space allows, placement of the decoupling capacitor on the component side is preferred. This can reduce unwanted inductance between the decoupling capacitor and the power pin caused by the via.

Maximize the power and ground pad sizes and number of vias capacitors. This can reduce the inductance between the power and ground planes and the component power and ground pins.

The RC filter consisting of R7, C11, and C16 should be placed as close to the  $V_{DDA}$  pin as possible.

#### **Clock Traces and Termination**

Poor signal integrity can degrade the system performance or cause system failure. In synchronous high-speed digital systems, the clock signal is less tolerant to poor signal integrity than other signals. Any ringing on the rising or falling edge or excessive ring back can cause system failure. The shape of the trace and the trace delay might be restricted by the available space on the board

and the component location. While routing the traces, the clock signal traces should be routed first and should be locked prior to routing other signal traces.

- The differential  $50\Omega$  output traces should have the same length.
- Avoid sharp angles on the clock trace. Sharp angle turns cause the characteristic impedance to change on the transmission lines.
- Keep the clock traces on the same layer. Whenever possible, avoid placing vias on the clock traces.
   Placement of vias on the traces can affect the trace characteristic impedance and hence degrade signal integrity.
- To prevent cross talk, avoid routing other signal traces in parallel with the clock traces. If running parallel traces is unavoidable, allow a separation of at least three trace widths between the differential clock trace and the other signal trace.
- Make sure no other signal traces are routed between the clock trace pair.
- The matching termination resistors should be located as close to the receiver input pins as possible.

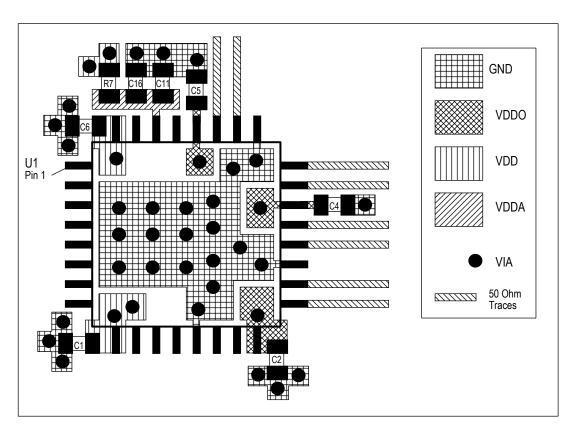


Figure 5B. PCB Board Layout for ICS8745B

#### **Power Considerations**

This section provides information on power dissipation and junction temperature for the ICS8745B. Equations and example calculations are also provided.

#### 1. Power Dissipation.

The total power dissipation for the ICS8745B is the sum of the core power plus the analog power plus the power dissipated in the load(s). The following is the power dissipation for  $V_{DD} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> = V<sub>DD MAX</sub> \* (I<sub>DD MAX</sub> + I<sub>DDA MAX</sub>) = 3.465V \* (125mA + 17mA) = 492mW
- Power (outputs)<sub>MAX</sub> = V<sub>DDO MAX</sub> \* I<sub>DDO MAX</sub> = 3.465V \* 59mA = 204mW

Total Power\_MAX = 492mW + 204mW = 696mW

#### 2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS devices is 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA}$  \* Pd\_total + T<sub>A</sub>

Tj = Junction Temperature

 $\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

T<sub>A</sub> = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 42.1°C/W per Table 7below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is:

 $70^{\circ}\text{C} + 0.696\text{W} * 42.1^{\circ}\text{C/W} = 99.3^{\circ}\text{C}$ . This is well below the limit of  $125^{\circ}\text{C}$ .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (single layer or multi-layer).

Table 7. Thermal Resitance  $\theta_{JA}$  for 32 Lead LQFP, Forced Convection

$\theta_{JA}$ vs. Air Flow								
Linear Feet per Minute	0	200	500					
Single-Layer PCB, JEDEC Standard Test Boards	67.8°C/W	55.9°C/W	50.1°C/W					
Multi-Layer PCB, JEDEC Standard Test Boards	47.9°C/W	42.1°C/W	39.4°C/W					
NOTE: Most modern PCB designs use multi-layered	NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.							

## **Reliability Information**

### Table 8. $\theta_{\text{JA}}$ vs. Air Flow Table for a 32 Lead LQFP

$\theta_{JA}$ vs. Air Flow					
Linear Feet per Minute	0	200	500		
Single-Layer PCB, JEDEC Standard Test Boards	67.8°C/W	55.9°C/W	50.1°C/W		
Multi-Layer PCB, JEDEC Standard Test Boards	47.9°C/W	42.1°C/W	39.4°C/W		
NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.					

#### **Transistor Count**

The transistor count for ICS8745B is: 2772

## **Package Outline and Package Dimension**

Package Outline - Y Suffix for 32 Lead LQFP

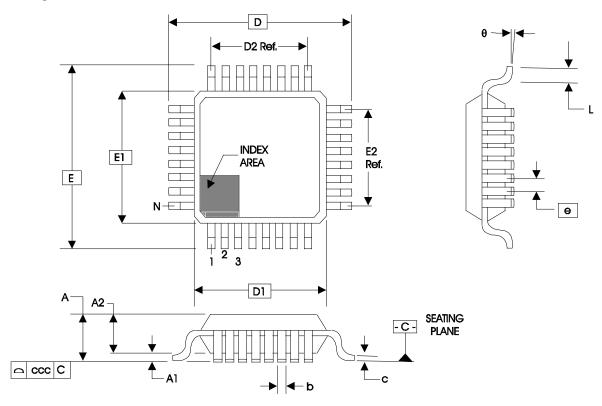


Table 9. Package Dimensions for 32 Lead LQFP

JEDEC Variation: BBC - HD All Dimensions in Millimeters					
Symbol	Minimum	Nominal	Maximum		
N	32				
Α			1.60		
A1	0.05	0.10	0.15		
A2	1.35	1.40	1.45		
b	0.30	0.37	0.45		
С	0.09		0.20		
D&E		9.00 Basic			
D1 & E1	7.00 Basic				
D2 & E2	5.60 Ref.				
е	0.80 Basic				
L	0.45	0.60	0.75		
θ	0°		7°		
ccc			0.10		

Reference Document: JEDEC Publication 95, MS-026

## **Ordering Information**

#### **Table 10. Ordering Information**

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8745BY	ICS8745BY	32 Lead LQFP	Tray	0°C to 70°C
8745BYT	ICS8745BY	32 Lead LQFP	1000 Tape & Reel	0°C to 70°C
8745BYLF	ICS8745BYLF	"Lead-Free" 32 Lead LQFP	Tray	0°C to 70°C
8745BYLFT	ICS8745BYLF	"Lead-Free" 32 Lead LQFP	1000 Tape & Reel	0°C to 70°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

While the information presented herein has been checked for both accuracy and reliability, Integrated Device Technology (IDT) assumes no responsibility for either its use or for the infringement of any patents or other rights of third parties, which would result from its use. No other circuits, patents, or licenses are implied. This product is intended for use in normal commercial applications. Any other applications, such as those requiring extended temperature ranges, high reliability or other extraordinary environmental requirements are not recommended without additional processing by IDT. IDT reserves the right to change any circuitry or specifications without notice. IDT does not authorize or warrant any IDT product for use in life support devices or critical medical instruments.

## **Revision History Sheet**

Rev	Table	Page	Description of Change	Date
В	T4D	5	LVDS DC Characteristics Table - modified VOS 0.90V min. to 1.05V min, 1.15V typical to 1.2V typical, and 1.4V max. to 1.35V max.	3/17/04
В	Т9	1 14	Added Lead-Free bullet. Ordering Information Table - added Lead-Free part.	12/2/04
В	Т9	1 14	Features Section - delete bullet, ""Industrial temperature available upon request."" Ordering Information Table - added Lead-Free note.	3/18/05
С	T6	7 15	AC Characteristics Table - changed t <sub>PD</sub> max limit from 3.7ns to 4.0ns. Added Power Considerations section. Updated format throughout the datasheet.	4/16/07

## Innovate with IDT and accelerate your future networks. Contact:

www.IDT.com

#### For Sales

800-345-7015 408-284-8200 Fax: 408-284-2775

#### For Tech Support

netcom@idt.com 480-763-2056

#### **Corporate Headquarters**

Integrated Device Technology, Inc. 6024 Silver Creek Valley Road San Jose, CA 95138 United States 800 345 7015 +408 284 8200 (outside U.S.)

#### **Asia Pacific and Japan**

Integrated Device Technology Singapore (1997) Pte. Ltd. Reg. No. 199707558G 435 Orchard Road #20-03 Wisma Atria Singapore 238877 +65 6 887 5505

#### **Europe**

IDT Europe, Limited 321 Kingston Road Leatherhead, Surrey KT22 7TU England +44 (0) 1372 363 339 Fax: +44 (0) 1372 378851



© 2006 Integrated Device Technology, Inc. All rights reserved. Product specifications subject to change without notice. IDT and the IDT logo are trademarks of Integrated Device Technology, Inc. Accelerated Thinking is a service mark of Integrated Device Technology, Inc. All other brands, product names and marks are or may be trademarks or registered trademarks used to identify products or services of their respective owners.