

Precision Quad Supply Monitor in 6-Lead SOT-23

FEATURES

- Ultralow Voltage Reset: V_{CC} = 0.5V Guaranteed
- Monitor Four Inputs Simultaneously 3.3V, 2.5V, 1.8V, ADJ (LTC2903-A1) 5V, 3.3V, 2.5V, 1.8V (LTC2903-B1) 5V, 3.3V, 1.8V, -5.2V (LTC2903-C1)
- Guaranteed Threshold Accuracy: ±1.5% of Monitored Voltage over Temperature
- 10% Undervoltage Monitoring
- Low Supply Current: 20µA Typical
- 200ms Reset Time Delay
- Active Low Open-Drain RST Output
- Power Supply Glitch Immunity
- Low Profile (1mm) SOT-23 (ThinSOT[™]) Package

APPLICATIONS

- Multivoltage Systems
- Optical Networking Systems
- Cell Phone Base Stations
- Network Servers

TYPICAL APPLICATION

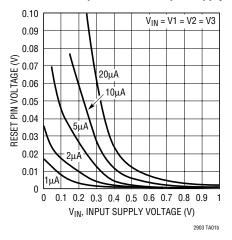
3.3V DC/DC SYSTEM 2.5V CONVERTER I OGIC 1.8V LTC2903-B1 RST V1 C1 0.1µF GND V4 C.2 0 1 u F V2 V3 2903 TAO

DESCRIPTION

The LTC[®]2903-1 monitors up to four supply voltages. The common reset output remains low until all four inputs have been in compliance for 200ms. Voltage thresholds maintain $\pm 1.5\%$ accuracy over temperature (with respect to the monitored voltage). The LTC2903-1 features an open-drain RST output with a weak internal pullup.

Internal supply voltage (V_{CC}) is generated from the greater voltage on the V1, V2 inputs. The RST output is guaranteed to sink at least 5µA ($V_{OL} = 0.15V$) for V1, V2 or V3 down to 0.5V and will typically conduct current down to 0V. Quiescent current is 20µA typical, making the LTC2903-1 ideal for power conscious systems. The LTC2903-1 is available in a 6-lead low profile (1mm) SOT-23 package.

T, LTC and LT are registered trademarks of Linear Technology Corporation. ThinSOT is a trademark of Linear Technology Corporation.



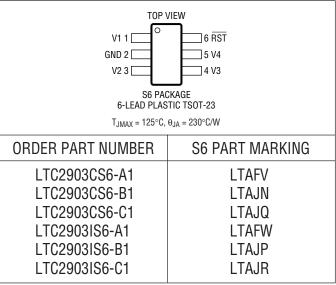
Low Voltage Reset Pull-Down Performance vs External Pull-Up Current and Input Supply Voltage

ABSOLUTE MAXIMUM RATINGS

| (Notes | 1, | 2, | 3) | |
|--------|----|----|----|--|
|--------|----|----|----|--|

| V1, V2 | 0.3V to 6.5V |
|-----------------------------------|----------------------------------|
| V3 | 2.7V or (V _{CC} + 0.3V) |
| V4 (LTC2903-A1, LTC2903-B1) | –0.3V to 6.5V |
| <u>V4 (</u> LTC2903-C1) | 6.5V to 0.3V |
| RST | 0.3V to 6.5V |
| Operating Temperature Range | |
| LTC2903C-X1 | |
| LTC2903I-X1 | |
| Storage Temperature Range | |
| Lead Temperature (Soldering, 10 s | ec) 300°C |

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS

ELECTRICAL CHARACTERISTICS (LTC2903-A1) The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T_A = 25°C. V_{CC} = 3.3V unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | | MIN | ТҮР | MAX | UNITS |
|-------------------|----------------------------|--------------------|---|-------|-------|-------|-------|
| V _{RT33} | 3.3V, 10% Reset Threshold | V1 Input Threshold | | 2.871 | 2.921 | 2.970 | V |
| V _{RT25} | 2.5V, 10% Reset Threshold | V2 Input Threshold | | 2.175 | 2.213 | 2.250 | V |
| V _{RT18} | 1.8V, 10% Reset Threshold | V3 Input Threshold | | 1.566 | 1.593 | 1.620 | V |
| V _{RTA} | Adjustable Reset Threshold | V4 Input Threshold | • | 0.492 | 0.500 | 0.508 | V |

(LTC2903-B1) The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $\dot{T}_A = 25^{\circ}C$. $V_{CC} = 5V$ unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | MIN | ТҮР | MAX | UNITS |
|-------------------|---------------------------|--------------------|-------|-------|-------|-------|
| V _{RT50} | 5V, 10% Reset Threshold | V1 Input Threshold | 4.350 | 4.425 | 4.500 | V |
| V _{RT33} | 3.3V, 10% Reset Threshold | V2 Input Threshold | 2.871 | 2.921 | 2.970 | V |
| V _{RT25} | 2.5V, 10% Reset Threshold | V3 Input Threshold | 2.175 | 2.213 | 2.250 | V |
| V _{RT18} | 1.8V, 10% Reset Threshold | V4 Input Threshold | 1.566 | 1.593 | 1.620 | V |

(LTC2903-C1) The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_{CC} = 5V$ unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | | MIN | ТҮР | MAX | UNITS |
|--------------------|----------------------------|--------------------|---|--------|--------|--------|-------|
| V _{RT50} | 5V, 10% Reset Threshold | V1 Input Threshold | • | 4.350 | 4.425 | 4.500 | V |
| V _{RT33} | 3.3V, 10% Reset Threshold | V2 Input Threshold | • | 2.871 | 2.921 | 2.970 | V |
| V _{RT18} | 1.8V, 10% Reset Threshold | V3 Input Threshold | • | 1.566 | 1.593 | 1.620 | V |
| V _{RT52N} | -5.2V, 10% Reset Threshold | V4 Input Threshold | | -4.524 | -4.602 | -4.680 | V |



ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T_A = 25°C. V_{CC} = 3.3V unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | | MIN | ТҮР | MAX | UNITS |
|------------------|---|---|-------------|--------|----------------------|-------------------------|----------------------|
| I _{V1} | V1 Input Current (Note 4) | V1 = 3.3V (LTC2903-A1) V1 = 5V (LTC2903-B1, LTC2903-C1) | • | | 20 25 | 80 80 | μA μA |
| I _{V2} | V2 Input Current (Note 4) | V2 = 2.5V (LTC2903-A1) V2 = 3.3V (LTC2903-B1, LTC2903-C1) | • | | 8 10 | 30 30 | μΑ μΑ |
| I _{V3} | V3 Input Current | V3 = 1.8V (LTC2903-A1, LTC2903-C1) V3 = 2.5V (LTC2903-B1) | • | | 6 8 | 30 30 | μA μA |
| I _{V4} | V4 Input Current | V4 = 0.55V (LTC2903-A1) V4 = 1.8V (LTC2903-B1) V4 = -5.2V (LTC2903-C1) | • | | 2 3 | ±15 4 6 | nA μA μA |
| t _{RST} | Reset Time-Out Period | | • | 140 | 200 | 260 | ms |
| t _{UV} | V_X Undervoltage Detect to \overline{RST} | V _X Less Than Threshold V _{RTX} by More Than 1% | | | 150 | | μS |
| V _{OH} | Output Voltage High RST (LTC2903-1) (Note 5) | $I_{RST(DN)} = -1\mu A$ | • | V2 – 1 | | | V |
| V _{OL} | Output Voltage Low RST (Note 6) | $V_{CC} = 0.2V, I_{RST} = 0.1\mu A \\ V_{CC} = 0.5V, I_{RST} = 5\mu A \\ V_{CC} = 1V, I_{RST} = 200\mu A \\ V_{CC} = 3V, I_{RST} = 2500\mu A$ | • • • | | 5 10 25 100 | 60 150 300 300 | mV mV mV mV |

Note 1: Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

Note 2: All currents into pins are positive, all voltages are referenced to GND unless otherwise noted.

Note 3: The internal supply voltage (V_{CC}) is generated from the greater voltage on the V1, V2 inputs.

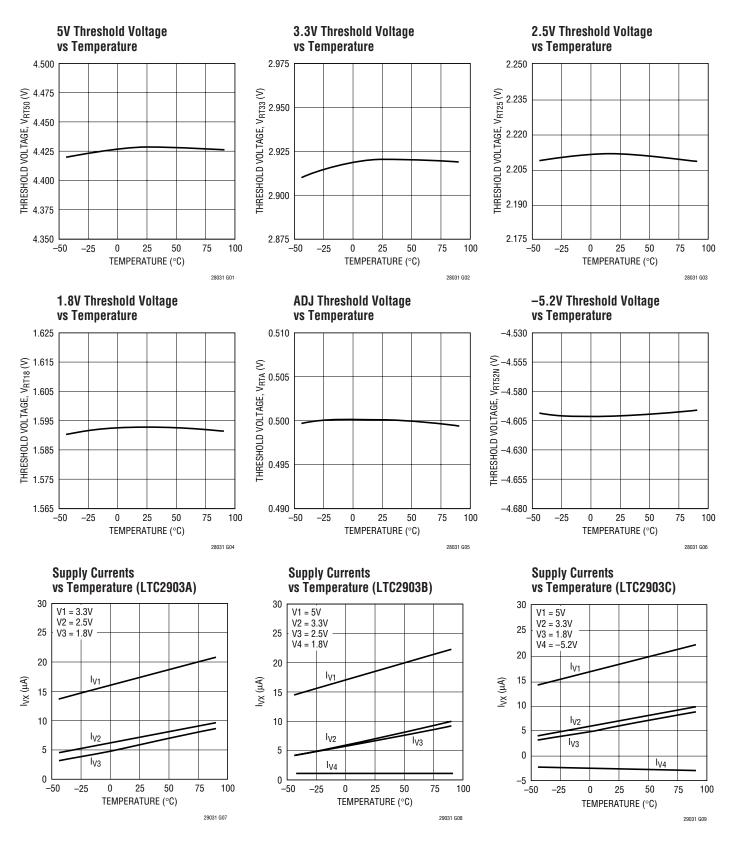
Note 4: Under typical operating conditions, quiescent current is drawn from the V1 input. When V2 exceeds V1, V2 supplies the quiescent current.

Note 5: The RST output pin on the LTC2903-1 has an internal pull-up to V2 of typically 10 μ A. However, for faster rise times or for V_{OH} voltages greater than V2, use an external pull-up resistor.

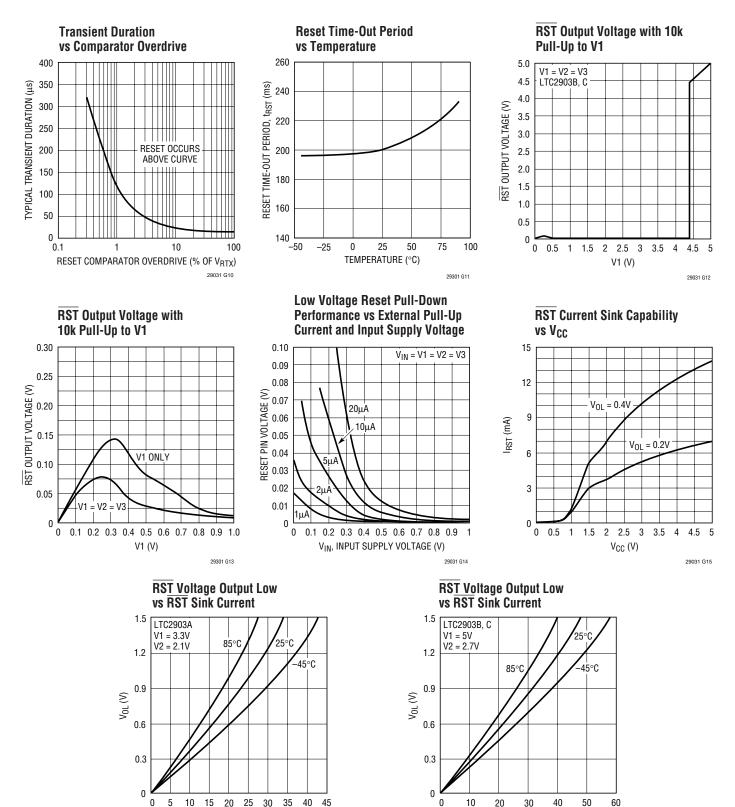
Note 6: At input voltages below 1V on V1 and V2, voltage on V3 assists pulling down the RST output.



TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL PERFORMANCE CHARACTERISTICS



29031 G16

I_{RST} (mA)

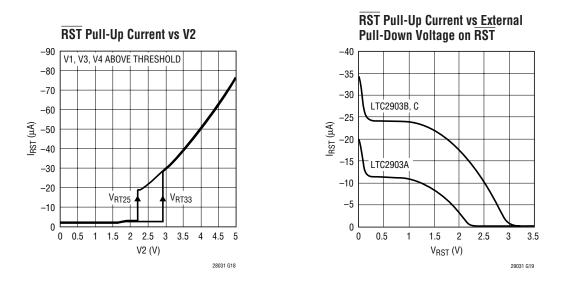
LINEAR TECHNOLOGY

Downloaded from Elcodis.com electronic components distributor

I_{RST} (mA)

29031 G16

TYPICAL PERFORMANCE CHARACTERISTICS



PIN FUNCTIONS

V1 (Pin 1): Voltage Input 1 (5V, 3.3V). Internal V_{CC} is generated from the greater voltage on the V1, V2 inputs. Bypass this pin to ground with a 0.1µF (or greater) capacitor.

GND (Pin 2): Ground.

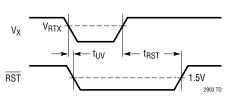
V2 (Pin 3): Voltage Input 2 (3.3V, 2.5V). Internal V_{CC} is generated from the greater voltage on the V1, V2 inputs. Bypass this pin to ground with a 0.1μ F (or greater) capacitor.

V3 (Pin 4): Voltage Input 3 (2.5V, 1.8V). This input assists the RST pull-down circuitry below 1V.

V4 (Pin 5): Voltage Input 4 (ADJ, 1.8V, -5.2V). See Table 1 for recommended ADJ resistor values.

RST (Pin 6): Reset Logic Output. Pulls low when any voltage input is below reset threshold and held low for 200ms after all voltage inputs exceed threshold. The pin contains a weak pull-up to V2. Use an external pull-up for faster rise times or output voltages greater than V2.

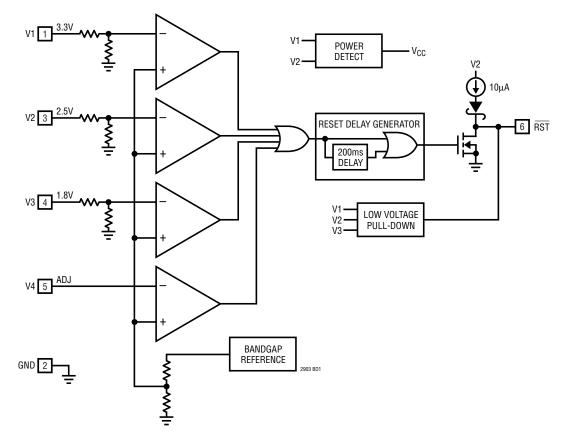
TIMING DIAGRAM





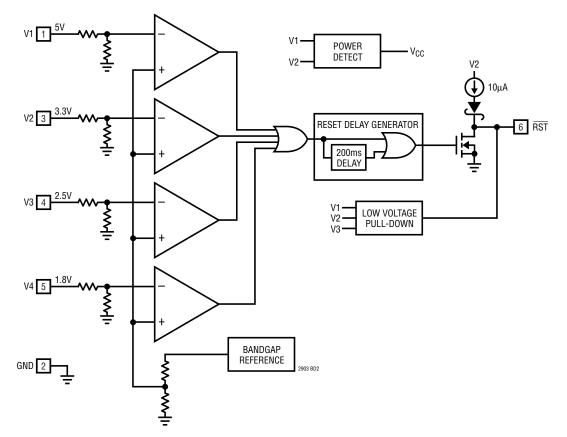
290311

BLOCK DIAGRAMS (LTC2903-A1)

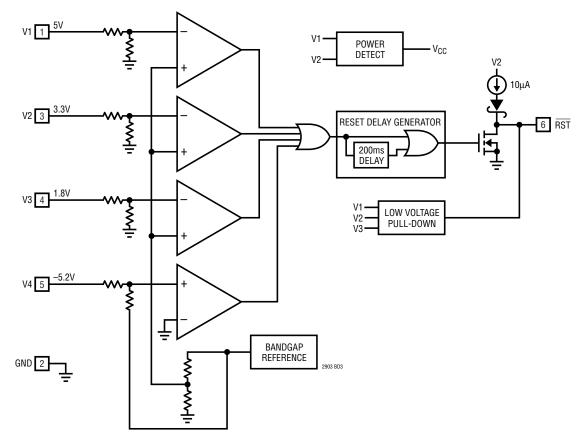




BLOCK DIAGRAMS (LTC2903-B1)



BLOCK DIAGRAMS (LTC2903-C1)





9

APPLICATIONS INFORMATION

Power-Up

The LTC2903-1 issues a logic low on the RST output when an input supply voltage resides below the prescribed threshold voltage. Ideally, the RST logic output would remain low with the input supply voltage down to zero volts. Most supervisors lack pull-down capability below 1V. The LTC2903-1 power supply supervisors incorporate a new low voltage pull-down circuit that can hold the RST line low with as little as 200mV of input supply voltage on V1, V2 or V3. The pull-down circuit helps maintain a low impedance path to ground, reducing the risk of floating the RST node to undetermined voltages. Such voltages may trigger external logic causing erroneous reset operation(s). Furthermore, a mid-scale voltage could cause external circuits to operate in the middle of their voltage transfer characteristic, consuming more guiescent current than normal. These conditions could cause serious system reliability problems.

When V1, V2 and V3 are ramped simultaneously, the reset pull-down current increases up to three times the current that may be pulled with a single input. Figure 1 demonstrates the reset pin current sinking ability for single supply and triple supply-tracking applications. Figure 2 shows a detailed view of the reset pin voltage with a 10k pull-up resistor to V1.

The LTC2903-1 supervisors derive their internal supply voltage (V_{CC}) automatically from the greater voltage on the V1 and V2 inputs. With all supply inputs above threshold, the quiescent current drawn from V_{CC} is 20µA (typ).

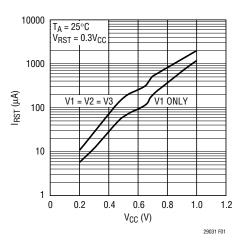


Figure 1. RST Pull-Down Current vs V_{CC}

Supply Monitoring

The LTC2903-1 accurately monitors four inputs in a small 6-lead SOT-23 package. The low voltage reset output includes an integrated 200ms reset delay timer. The reset line pulls high 200ms after all voltage inputs exceed their respective thresholds. The reset output remains low during power-up, power-down and brownout conditions on any of the voltage inputs.

For applications requiring an adjustable trip threshold, use the V4 input on the LTC2903-A1. Connect the tap point on an external resistive divider (R1, R2) placed between the positive voltage being sensed and ground, to the high impedance input on V4. The LTC2903-A1 compares the voltage on the V4 pin to the internal 0.5V reference. Figure 3 shows a generic setup for the positive adjustable application.

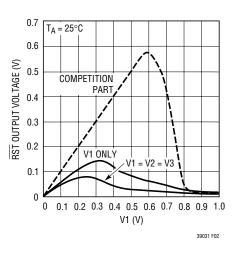
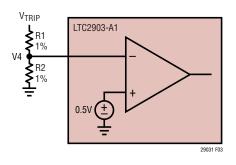


Figure 2. $\overrightarrow{\text{RST}}$ Output Voltage with a 10k Pull-Up to V1 (Enlarged Area of Detail)







APPLICATIONS INFORMATION

Calculate the trip voltage from:

$$V_{\text{TRIP}} = 0.5 V \left(1 + \frac{\text{R1}}{\text{R2}} \right)$$

Table 1 contains suggested 1% resistor values for the ADJ input to obtain nominal -11.5% thresholds.

| able 1. Suggested 176 nesistor values for the ADJ input | | | | | | |
|---|-----------------------|-----------------|----------------|--|--|--|
| V _{SUPPLY} (V) | V _{TRIP} (V) | R1 (k Ω) | R2(k Ω) | | | |
| 12 | 10.75 | 2050 | 100 | | | |
| 10 | 8.95 | 1690 | 100 | | | |
| 8 | 7.15 | 1330 | 100 | | | |
| 7.5 | 6.7 | 1240 | 100 | | | |
| 6 | 5.38 | 976 | 100 | | | |
| 5 | 4.435 | 787 | 100 | | | |
| 3.3 | 2.935 | 487 | 100 | | | |
| 3 | 2.66 | 432 | 100 | | | |
| 2.5 | 2.2 | 340 | 100 | | | |
| 1.8 | 1.605 | 221 | 100 | | | |
| 1.5 | 1.325 | 165 | 100 | | | |
| 1.2 | 1.065 | 113 | 100 | | | |
| 1 | 0.884 | 76.8 | 100 | | | |
| 0.9 | 0.795 | 59 | 100 | | | |

Table 1. Suggested 1% Resistor Values for the ADJ Input

Connect unused supervisor inputs to the highest supply voltage available (typically V1). On the LTC2903-C1, the negative V4 input must always be applied.

Implications of Threshold Accuracy

Specifying system voltage margin for worst-case operation requires consideration of three factors: power supply tolerance, IC supply voltage tolerance and supervisor reset threshold accuracy. Highly accurate supervisors ease the design challenge by decreasing the overall voltage margin required for reliable system operation. Consider a 5V system with a $\pm 10\%$ power supply tolerance band. System ICs powered by this supply must operate reliably within this band (and a little more, as explained below). The bottom of the supply tolerance band, at 4.5V (5V – 10%), is the exact voltage at which a *perfectly accurate* supervisor generates a reset. Such a perfectly accurate supervisor does not exist—the actual reset threshold may vary over a specified band ($\pm 1.5\%$ for the LTC2903-1 supervisors). Figure 4 shows the typical relative threshold accuracy for all four inputs, over temperature.

With this variation of reset threshold in mind, the nominal reset threshold of the supervisor resides *below* the minimum supply voltage; just enough so that the reset threshold band and the power supply tolerance bands do not overlap. If the two bands overlap, the supervisor could generate a false or nuisance reset when the power supply remains within its specified tolerance band (say, at 4.6V).

Adding half of the reset threshold accuracy spread (1.5%) to the ideal 10% thresholds puts the LTC2903-1 thresholds at 11.5% (typ) below the nominal input voltage. For example, the 5V typical threshold is 4.425V, or 75mV below the ideal threshold of 4.500V. The guaranteed threshold lies in the band between 4.500V and 4.350V over temperature.

The powered system must work reliably down to the lowest voltage in the threshold band or risk malfunction before the reset line falls. In the 5V example, using the 1.5% accurate supervisor, the system ICs must work down to 4.35V. System ICs working with $a\pm 2.5\%$ accurate supervisor must operate down to 4.25V, increasing the required system voltage margin and the probability of system malfunction.

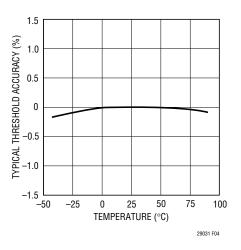


Figure 4. LTC2903 Typical Threshold Accuracy vs Temperature



APPLICATIONS INFORMATION

In any supervisory application, supply noise riding on the monitored DC voltage can cause spurious resets, particularly when the monitored voltage approaches the reset threshold. A less than desirable but commonly used technique used to mitigate this problem adds hysteresis to the input comparator. The amount of added hysteresis, usually specified as a percentage of the trip threshold, effectively degrades the advertised accuracy of the part. To maintain high accuracy, the LTC2903-1 does not use hysteresis.

To minimize spurious resets while maintaining threshold accuracy, the LTC2903-1 employs two forms of noise filtering. The first line of defense incorporates proprietary tailoring of the comparator transient response. Transient events receive electronic integration in the comparator and must exceed a certain magnitude and duration to cause the comparator to switch. Figure 5 illustrates the typical transient duration versus comparator overdrive (as a percentage of the trip threshold V_{RT}) required to trip the comparators. Once any comparator is switched, the reset line pulls low. The reset time-out counter starts once all inputs return above threshold. The nominal reset delay time is 200ms. The counter clears whenever any input

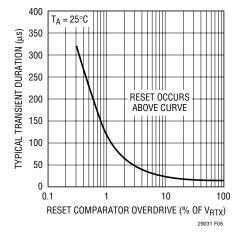


Figure 5. Typical Transient Duration vs Overdrive Required to Trip Comparator

drops back below threshold. This reset delay time effectively provides further filtering of the voltage inputs. A noisy input with frequency components of sufficient magnitude above $f = 1/t_{RST} = 5Hz$ holds the reset line low, preventing oscillatory behavior on the reset line.

Although all four comparators have built-in glitch filtering, use bypass capacitors on the V1 and V2 inputs because the greater of V1 or V2 supplies the V_{CC} for the part (a 0.1µF ceramic capacitor satisfies most applications). Apply filter capacitors on the V3 and V4 inputs in extremely noisy situations.

Reset Output Rise and Fall Time Estimation

The reset output line contains a weak pull-up current source to the V2 supply. Use an external pull-up resistor when the output needs to pull to another voltage and/or when the reset output needs a faster rise time. The opendrain output allows for wired-OR connections when more than one signal needs to pull down on the reset line. Estimate output rise time for the open-drain output without an external pull-up using:

$$t_{RISE} \approx 2.2 \bullet R_{PU} \bullet C_{LOAD}$$

where R_{PU} is the on-resistance of the pull-up transistor and C_{LOAD} is the external load capacitance on the pin. At room temperature, the average R_{PU} is approximately 50k Ω . When externally pulling up to voltages higher than V2, an internal network automatically protects the weak pull-up circuitry from reverse currents.

The reset output has very strong pull-down capability. Estimate the output fall time using:

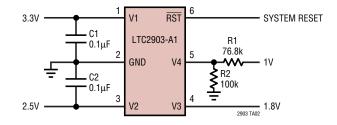
$$t_{FALL} \approx 2.2 \bullet R_{PD} \bullet C_{LOAD}$$

where R_{PD} is the on-resistance of the pull-down transistor and C_{LOAD} is the external load capacitance on the pin. At room temperature, the average R_{PD} is approximately 40Ω . With a 150pF load capacitance the reset line can pull down in about 13ns.

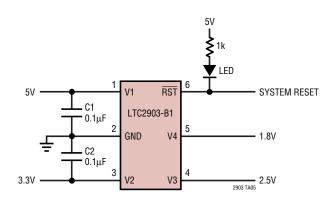


TYPICAL APPLICATIONS

Quad Supply Monitor with Adjustable Input

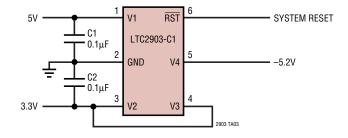


Fixed Quad Supply Monitor with LED Indication on $\overline{\text{RST}}$



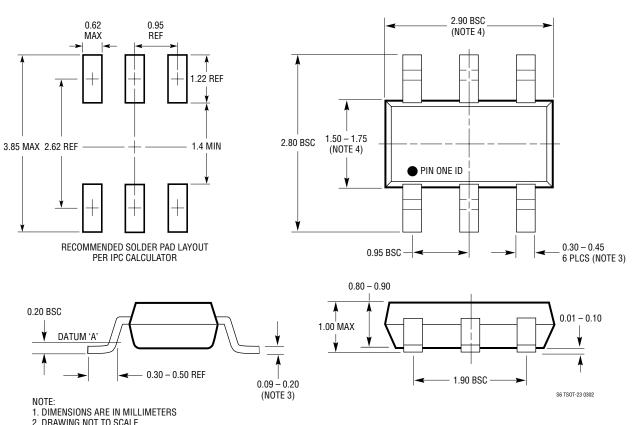
TYPICAL APPLICATIONS

Quad Supply Monitor with Unused Input Pulled Above Threshold





PACKAGE DESCRIPTION



S6 Package 6-Lead Plastic TSOT-23 (Reference LTC DWG # 05-08-1636)

2. DRAWING NOT TO SCALE

3. DIMENSIONS ARE INCLUSIVE OF PLATING 4. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR

5. MOLD FLASH SHALL NOT EXCEED 0.254mm

6. JEDEC PACKAGE REFERENCE IS MO-193

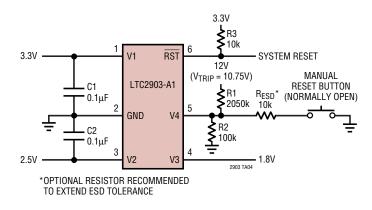


Downloaded from Elcodis.com electronic components distributor

Information furnished by Linear Technology Corporation is believed to be accurate and reliable. However, no responsibility is assumed for its use. Linear Technology Corporation makes no representation that the interconnection of its circuits as described herein will not infringe on existing patent rights.

TYPICAL APPLICATION





RELATED PARTS

| PART NUMBER | DESCRIPTION | COMMENTS |
|-------------------------|---|--|
| LTC690 | 5V Supply Monitor, Watchdog Timer and Battery Backup | 4.65 Threshold |
| LTC694-3.3 | 3.3V Supply Monitor, Watchdog Timer and Battery Backup | 2.9V Threshold |
| LTC699 | 5V Supply Monitor and Watchdog Timer | 4.65 Threshold |
| LTC1232 | 5V Supply Monitor, Watchdog Timer and Pushbutton Reset | 4.37V/4.62V Threshold |
| LTC1326 | Micropower Precision Triple Supply Monitor for 5V, 3.3V and ADJ | 4.725V, 3.118V, 1V Thresholds (±0.75%) |
| LTC1326-2.5 | Micropower Precision Triple Supply Monitor for 2.5V, 3.3V and ADJ | 2.363V, 3.118V, 1V Thresholds (±0.75%) |
| LTC1536 | Precision triple Supply Monitor for PCI Applications | Meets PCI t _{FAIL} Timing Specifications |
| LTC1726-2.5 | Micropower Triple Supply Monitor for 2.5V, 3.3V and ADJ | Adjustable RESET and Watchdog Time Outs |
| LTC1726-5 | Micropower Triple Supply Monitor for 5V, 3.3V and ADJ | Adjustable RESET and Watchdog Time Outs |
| LTC1727-2.5/LTC1727-5 | Micropower Triple Supply Monitors with Open-Drain Reset | Individual Monitor Outputs in MSOP |
| LTC1728-1.8/LTC1728-3.3 | Micropower Triple Supply Monitors with Open-Drain Reset | 5-Lead SOT-23 Package |
| LTC1728-2.5/LTC1728-5 | Micropower Triple Supply Monitors with Open-Drain Reset | 5-Lead SOT-23 Package |
| LTC1985-1.8 | Micropower Triple Supply Monitor with Push-Pull Reset Output | 5-Lead SOT-23 Package |
| LTC2900 | Quad Voltage Monitor in MSOP | 16 User Selectable Combinations, ±1.5% Threshold Accuracy |
| LTC2901 | Quad Voltage Monitor with Watchdog | 16 User Selectable Combinations, Adjustable Timers |
| LTC2902 | Quad Voltage Monitor with RST Disable | 16 User Selectable Combinations, Adjustable Tolerance |
| LTC2920-1/LTC2920-2 | Single/Dual Power Supply Margining Controller | <0.4% Margin Voltage Precision |
| LTC2921/LTC2922 | Power Supply Trackers with Input Monitors | 3 (LTC2921) and 5 (LTC2922) Remote Sense Switches |
| LTC2923 | Power Supply Tracking Controller | Tracks Up and Down, Supply Sequencing |