FAIRCHILD
SEMICONDUCTOR ™

MM74HC423A
Dual Retriggerable Monostable Multivibrator

General Description

The 74HC423A high speed monostable multivibrators (one shots) utilize advanced silicon-gate CMOS technology. They feature speeds comparable to low power Schottky TTL circuitry while retaining the low power and high noise immunity characteristic of CMOS circuits.

Each multivibrator features both a negative, A, and a positive, B, transition triggered input, either of which can be used as an inhibit input. Also included is a clear input that when taken LOW resets the one shot. The MM74HC423A cannot be triggered from clear.

The MM74HC423A is retriggerable. That is, it may be triggered repeatedly while its outputs are generating a pulse and the pulse will be extended.

Pulse width stability over a wide range of temperature and supply is achieved using linear CMOS techniques. The output pulse equation is simply: PW = (R_{EXT}) (C_{EXT}); where PW

is in seconds, R is in ohms, and C is in farads. All inputs are protected from damage due to static discharge by diodes to V_{CC} and ground.

Features

- Typical propagation delay: 40 ns
- Wide power supply range: 2V–6V
- Low quiescent current: 80 µA maximum (74HC Series)
- Low input current: 1 µA maximum
- Fanout of 10 LS-TTL loads
- Simple pulse width formula T = RC
- Wide pulse range: 400 ns to ∞ (typ)
- Part to part variation: ±5% (typ)
- Schmitt Trigger A & B inputs allow rise and fall times to be as slow as one second

Ordering Code:

| Order Number	Package Number	Package Description
MM74HC423AM (Note 1)	M16A	16-Lead Small Outline Integrated Circuit (SOIC), JEDEC MS-012, 0.150" Narrow
MM74HC423ASJ	M16D	16-Lead Small Outline Package (SOP), EIAJ TYPE II, 5.3mm Wide
MM74HC423AMTC (Note 1)	MTC16	16-Lead Thin Shrink Small Outline Package (TSSOP), JEDEC MO-153, 4.4mm Wide
MM74HC423AN	N16E	16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300" Wide

Note 1: Devices also available in Tape and Reel. Specify by appending the suffix letter "X" to the ordering code.

Connection Diagrams
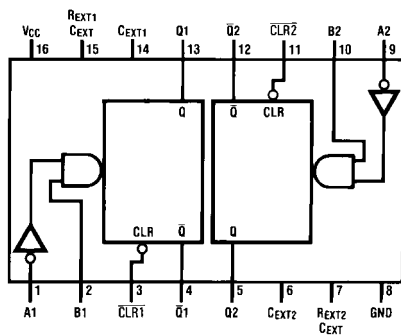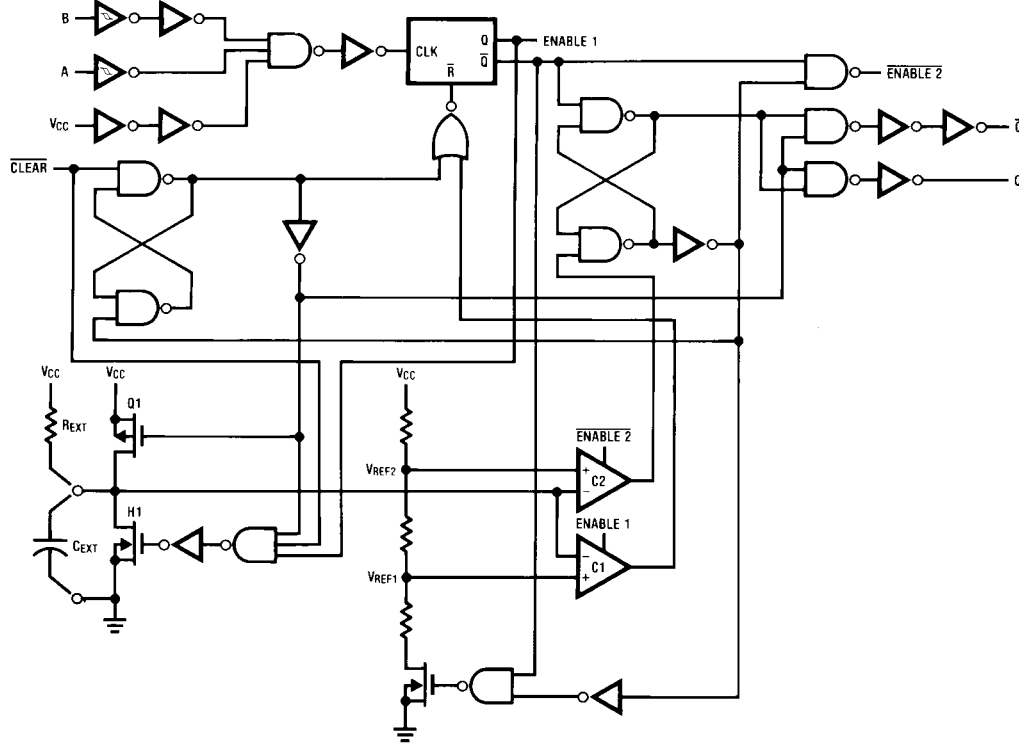

Top View

Timing Component
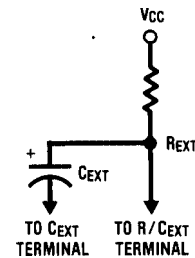
Note: Pin 6 and Pin 14 must be hard-wired to GND.

Truth Table

Inputs			Outputs	
$\overline{\text{Clear}}$	A	B	Q	$\overline{Q}$
L	X	X	L	H
X	H	X	L	H
X	X	L	L	H
H	L	↑	⎍	⎍
H	↓	H	⎍	⎍

H = HIGH Level
L = LOW Level
↑ = Transition from LOW-to-HIGH
↓ = Transition from HIGH-to-LOW
⎍ = One HIGH Level Pulse
⎍ = One LOW Level Pulse
X = Irrelevant

Logic Diagram

Theory of Operation

① POSITIVE EDGE TRIGGER   ④ POSITIVE EDGE RE-TRIGGER (PULSE LENGTHENING)

② NEGATIVE EDGE TRIGGER   ⑤ RESET PULSE SHORTENING
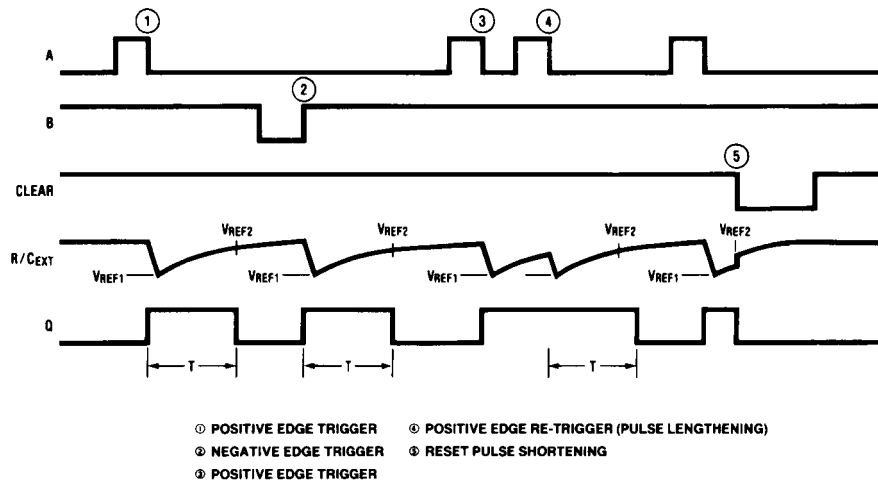
③ POSITIVE EDGE TRIGGER

FIGURE 1.

TRIGGER OPERATION

As shown in Figure 1 and the Logic Diagram before an input trigger occurs, the one-shot is in the quiescent state with the Q output LOW, and the timing capacitor C_{EXT} completely charged to V_{CC}. When the trigger input A goes from V_{CC} to GND (while inputs B and clear are held to V_{CC}) a valid trigger is recognized, which turns on comparator C1 and N-Channel transistor N11. At the same time the output latch is set. With transistor N1 on, the capacitor C_{EXT} rapidly discharges toward GND until V_{REF1} is reached. At this point the output of comparator C1 changes state and transistor N1 turns OFF. Comparator C1 then turns OFF while at the same time comparator C2 turns on. With transistor N1 OFF, the capacitor C_{EXT} begins to charge through the timing resistor, R_{EXT}, toward V_{CC}. When the voltage across C_{EXT} equals V_{REF2}, comparator C2 changes state causing the output latch to reset (Q goes LOW) while at the same time disabling comparator C2. This ends the timing cycle with the one-shot in the quiescent state, waiting for the next trigger.

A valid trigger is also recognized when trigger input B goes from GND to V_{CC} (while input A is at GND and input clear is at V_{CC}2.)

It should be noted that in the quiescent state C_{EXT} is fully charged to V_{CC} causing the current through resistor R_{EXT} to be zero. Both comparators are "OFF" with the total device current due only to reverse junction leakages. An added feature of the MM74HC423A is that the output latch is set via the input trigger without regard to the capacitor voltage. Thus, propagation delay from trigger to Q is independent of the value of C_{EXT}, R_{EXT}, or the duty cycle of the input waveform.

RETRIGGER OPERATION

The MM74HC423A is retriggered if a valid trigger occurs 3 followed by another trigger 4 before the Q output has returned to the quiescent (zero) state. Any retrigger, after the timing node voltage at pin or has begun to rise from V_{REF1}, but has not yet reached V_{REF2}, will cause an increase in output pulse width T. When a valid retrigger is initiated 4, the voltage at the R/C_{EXT} pin will again drop to V_{REF1} before progressing along the RC charging curve toward V_{CC}. The Q output will remain high until time T, after the last valid retrigger.

Because the trigger-control circuit flip-flop resets shortly after C_X has discharged to the reference voltage of the lower reference circuit, the minimum retrigger time, t_{rr} is a function of internal propagation delays and the discharge time of C_X:

$$t_{rr} = 20 + \frac{187}{V_{CC} - 0.7} + \frac{565 + (0.256\ V_{CC})\ C_X}{(V_{CC} - 0.7)^2}\ ns$$

Another removal/retrigger time occurs when a short clear pulse is used. Upon receipt of a clear, the one shot must charge the capacitor up to the upper trip point before the one shot is ready to receive the next trigger. This time is dependent on the capacitor used and is approximately:

$$t_{rr} = 20 + \frac{187}{V_{CC} - 0.7} + \frac{565 + (0.256\ V_{CC})\ C_X}{(V_{CC} - 0.7)^2}\ ns$$

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Theory of Operation (Continued)

RESET OPERATION

These one shots may be reset during the generation of the output pulse. In the reset mode of operation, an input pulse on clear sets the reset latch and causes the capacitor to be fast charged to V_{CC} by turning on transistor Q1 5. When the voltage on the capacitor reaches V_{REF2}, the reset latch will clear and then be ready to accept another pulse. If the clear input is held LOW, any trigger inputs that occur will be inhibited and the Q and $\overline{Q}$ outputs of the output latch will not change. Since the Q output is reset when an input low level is detected on the Clear input, the output pulse T can be made significantly shorter than the minimum pulse width specification.
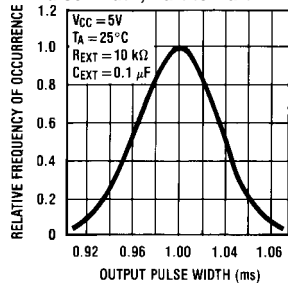
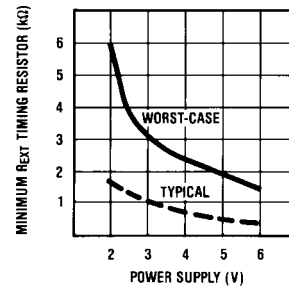**Typical Output Pulse Width vs. Timing Components**
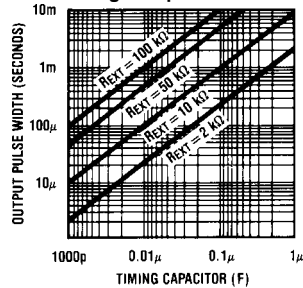
Typical 1ms Pulse Width Variation vs. Supply
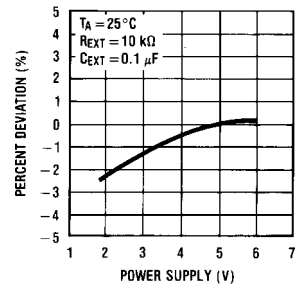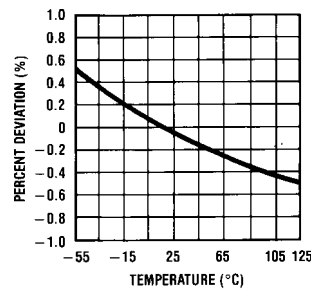

Typical Distribution of Output Pulse Width, Part to Part

Minimum R_{EXT} vs. Supply Voltage

Typical 1ms Pulse Width Variation vs. Temperature

Note: R and C are not subjected to temperature. The C is polypropylene.

Absolute Maximum Ratings(Note 2)

(Note 3)

Supply Voltage (V_{CC})	−0.5V to +7.0V
DC Input Voltage (V_{IN})	−1.5V to V_{CC} +1.5V
DC Output Voltage (V_{OUT})	−0.5V to V_{CC} +0.5V
Clamp Diode Current (I_{IK}, I_{OK})	±20 mA
DC Output Current, per pin (I_{OUT})	±25 mA
DC V_{CC} or GND Current,	
per pin (I_{CC})	±50 mA
Storage Temperature Range (T_{STG})	−65°C to +150°C
Power Dissipation (P_D)	
(Note 4)	600 mW
S.O. Package only	500 mW
Lead Temperature (T_L)	
(Soldering 10 seconds)	260°C

Recommended Operating Conditions

	Min	Max	Units
Supply Voltage (V_{CC})	2	6	V
DC Input or Output Voltage	0	V_{CC}	V
(V_{IN}, V_{OUT})			
Operating Temperature Range (T_A)	−40	+85	°C
Maximum Input Rise and Fall Time			
(Clear Input)			
V_{CC} = 2.0V		1000	ns
V_{CC} = 4.5V		500	ns
V_{CC} = 6.0V		400	ns

Note 2: Maximum Ratings are those values beyond which damage to the device may occur.

Note 3: Unless otherwise specified all voltages are referenced to ground.

Note 4: Power Dissipation Temperature Derating: Plastic "N" Package: − 12mW/°C from 65°C to 85°C.

DC Electrical Characteristics (Note 5)

Symbol	Parameter	Conditions	V_{CC}	T_A = 25°C		T_A = −40 to 85°C	T_A = −55 to 125°C	Units
				Typ	Guaranteed Limits			
V_{IH}	Minimum HIGH Level		2.0V		1.5	1.5	1.5	
	Input Voltage		4.5V		3.15	3.15	3.15	V
			6.0V		4.2	4.2	4.2	
V_{IL}	Maximum LOW Level		2.0V		0.3	0.3	0.3	
	Input Voltage		4.5V		0.9	0.9	0.9	V
			6.0V		1.2	1.2	1.2	
V_{OH}	Minimum HIGH Level	V_{IN} = V_{IH} or V_{IL}						
	Output Voltage	$|I_{OUT}| \leq$ 20 μA	2.0V	2.0	1.9	1.9	1.9	
			4.5V	4.5	4.4	4.4	4.4	
			6.0V	6.0	5.9	5.9	5.9	V
		V_{IN} = V_{IH} or V_{IL}						
		$|I_{OUT}| \leq$ 4.0 mA	4.5V		3.96	3.84	3.7	
		$|I_{OUT}| \leq$ 5.2 mA	6.0V		5.46	5.34	5.2	
V_{OL}	Maximum LOW Level	V_{IN} = V_{IH} or V_{IL}						
	Output Voltage	$|I_{OUT}| \leq$ 20 μA	2.0V	0	0.1	0.1	0.1	
			4.5V	0	0.1	0.1	0.1	
			6.0V	0	0.1	0.1	0.1	V
		V_{IN} = V_{IH} or V_{IL}						
		$|I_{OUT}| \leq$ 4 mA	4.5V		0.26	0.33	0.4	
		$|I_{OUT}| \leq$ 5.2 mA	6.0V		0.26	0.33	0.4	
I_{IN}	Maximum Input Current (Pins 7, 15)	V_{IN} = V_{CC} or GND	5.0V		0.5	5.0	5.0	μA
I_{IN}	Maximum Input Current (all other pins)	V_{IN} = V_{CC} or GND	6.0V		±0.1	±1.0	±1.0	μA
I_{CC}	Maximum Quiescent Supply Current (standby)	V_{IN} = V_{CC} or GND I_{OUT} = 0 μA	6.0V		8.0	80	160	μA
I_{CC}	Maximum Active Supply Current (per monostable)	V_{IN} = V_{CC} or GND R/C_{EXT} = 0.5V_{CC}	2.0V	36	80	110	130	μA
			4.5V	0.33	1.0	1.3	1.6	mA
			6.0V	0.7	2.0	2.6	3.2	mA

Note 5: For a power supply of 5V ±10% the worst-case output voltages (V_{OH}, V_{OL}) occur for HC at 4.5V. Thus the 4.5V values should be used when designing with this supply. Worst-case V_{IH} and V_{IL} occur at V_{CC} = 5.5V and 4.5V respectively. (The V_{IH} value at 5.5V is 3.85V.) The worst-case leakage current (I_{IN}, I_{CC}, and I_{OZ}) occur for CMOS at the higher voltage and so the 6.0V values should be used.

MM74HC423A

AC Electrical Characteristics

$V_{CC} = 5V$, $T_A = 25°C$, $C_L = 15$ pF, $t_r = t_f = 6$ ns

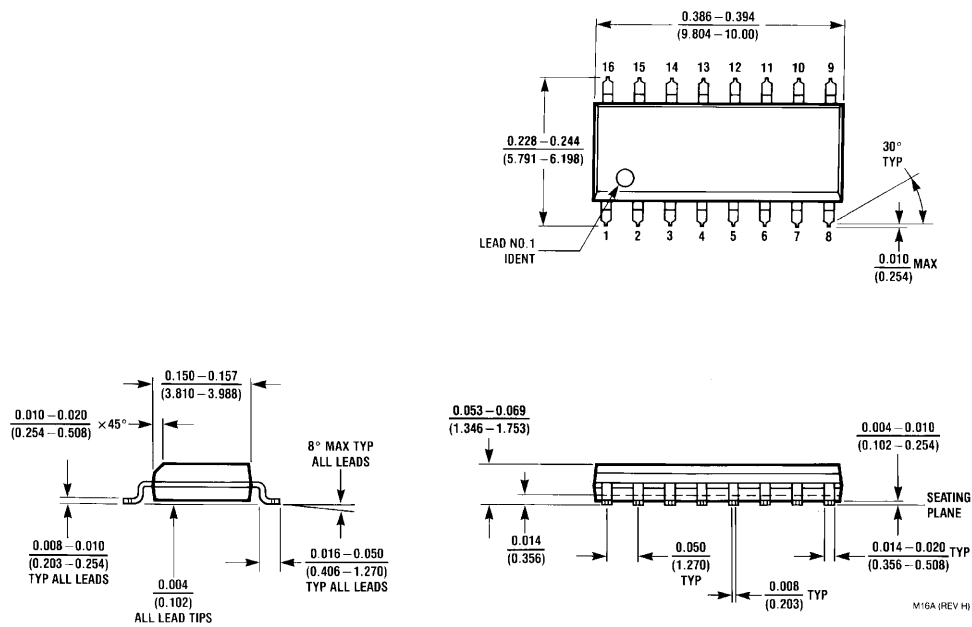| Symbol	Parameter	Conditions	Typ	Guatanteed Limit	Units
t_{PLH}	Maximum Trigger Propagation Delay, A, B to Q		22	33	ns
t_{PHL}	Maximum Trigger Propagation Delay, A, B to $\overline{Q}$		25	42	ns
t_{PHL}	Maximum Propagation Delay, Clear to Q		20	27	ns
t_{PLH}	Maximum Propagation Delay, Clear to $\overline{Q}$		22	33	ns
t_W	Minimum Pulse Width, A, B or Clear		14	26	ns
t_{REM}	Minimum Clear Removal Time			0	ns
$t_{WQ(MIN)}$	Minimum Output Pulse Width	$C_{EXT} = 28$ pF $R_{EXT} = 2$ kΩ	400		ns
t_{WQ}	Output Pulse Width	$C_{EXT} = 1000$ pF $R_{EXT} = 10$ kΩ	10		μs

AC Electrical Characteristics

$C_L = 50$ pF $t_r = t_f = 6$ ns (Unless otherwise specified)

Symbol	Parameter	Conditions	V_{CC}	$T_A = 25°C$ Typ	$T_A = 25°C$ Guaranteed Limits	$T_A = -40$ to $85°C$	$T_A = -55$ to $125°C$	Units
t_{PLH}	Maximum Trigger Propagation Delay, A or B to Q		2.0V	77	169	194	210	
			4.5V	26	42	51	57	ns
			6.0V	21	32	39	44	
t_{PHL}	Maximum Trigger Propagation Delay, A or B to $\overline{Q}$		2.0V	88	197	229	250	
			4.5V	29	48	60	67	ns
			6.0V	24	38	46	51	
t_{PHL}	Maximum Propagation Delay, Clear to Q		2.0V	54	114	132	143	
			4.5V	23	34	41	45	ns
			6.0V	19	28	33	36	
t_{PLH}	Maximum Propagation Delay, Clear to $\overline{Q}$		2.0V	56	116	135	147	
			4.5V	25	36	42	46	ns
			6.0V	20	29	34	37	
t_W	Minimum Pulse Width A, B, Clear		2.0V	57	123	144	157	
			4.5V	17	30	37	42	ns
			6.0V	12	21	27	30	
t_{REM}	Minimum Clear Removal Time		2.0V	0	0	0	0	
			4.5V	0	0	0	0	ns
			6.0V	0	0	0	0	
t_{WQ}	Output Pulse Width	$C_{EXT} = 0.1$ μF $R_{EXT} = 10$ kΩ Min	5.0V	1	0.9	0.86	0.85	ms
		Max	5.0V	1	1.1	1.14	1.15	ms
t_{TLH}, t_{THL}	Maximum Output Rise and Fall Time		2.0V	30	75	95	110	
			4.5V	8	15	19	22	ns
			6.0V	7	13	16	19	
C_{PD}	Power Dissipation Capacitance (Note 6)			83				pF
C_{IN}	Maximum Input Capacitance (Pins 7 & 15)			12	20	20	20	pF
C_{IN}	Maximum Input Capacitance (other inputs)			6	10	10	10	pF

Note 6: C_{PD} determines the no load dynamic power consumption, $P_D = C_{PD} V_{CC}^2 f + I_{CC} V_{CC}$, and the no load dynamic current consumption, $I_S = C_{PD} V_{CC} f + I_{CC}$.

Physical Dimensions inches (millimeters) unless otherwise noted

0.386 − 0.394
(9.804 − 10.00)

16 15 14 13 12 11 10 9

0.228 − 0.244
(5.791 − 6.198)

30°
TYP

LEAD NO.1
IDENT

1 2 3 4 5 6 7 8

0.010 MAX
(0.254)

0.150 − 0.157
(3.810 − 3.988)

0.010 − 0.020
(0.254 − 0.508) × 45°

8° MAX TYP
ALL LEADS

0.008 − 0.010
(0.203 − 0.254)
TYP ALL LEADS

0.004
(0.102)
ALL LEAD TIPS

0.016 − 0.050
(0.406 − 1.270)
TYP ALL LEADS

0.053 − 0.069
(1.346 − 1.753)

0.004 − 0.010
(0.102 − 0.254)

SEATING
PLANE

0.014
(0.356)

0.050
(1.270)
TYP

0.008
(0.203) TYP

0.014 − 0.020
(0.356 − 0.508) TYP

M16A (REV H)

16-Lead Small Outline Integrated Circuit (SOIC), JEDEC MS-012, 0.150" Narrow
Package Number M16A

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MM74HC423A

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)

10.2±0.1
-A-
0.47 TYP
16   9
7.8
3.9
5.3±0.1
-B-
PIN #1 IDENT.   1   8
0.2 C B A
ALL LEAD TIPS

16 15   10 9
5.01 TYP
9.27 TYP
(2.13 TYP)
1 2   7 8
1.27 TYP   0.6 TYP
LAND PATTERN RECOMMENDATION

2.1 MAX.
ALL LEAD TIPS
0.1 C
1.8±0.1
-C-
1.27 TYP   0.35-0.51
0.15±0.05
0.12 M C A
DIMENSIONS ARE IN MILLIMETERS

SEE DETAIL A
0.15-0.25

7° TYP
GAGE PLANE
0°-8° TYP
0.25
0.60±0.15
1.25
SEATING PLANE
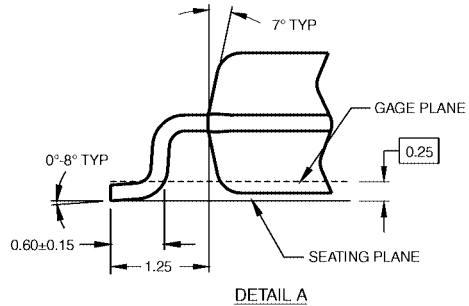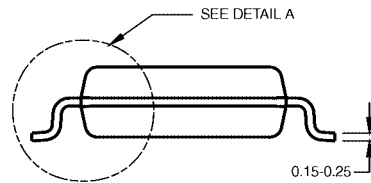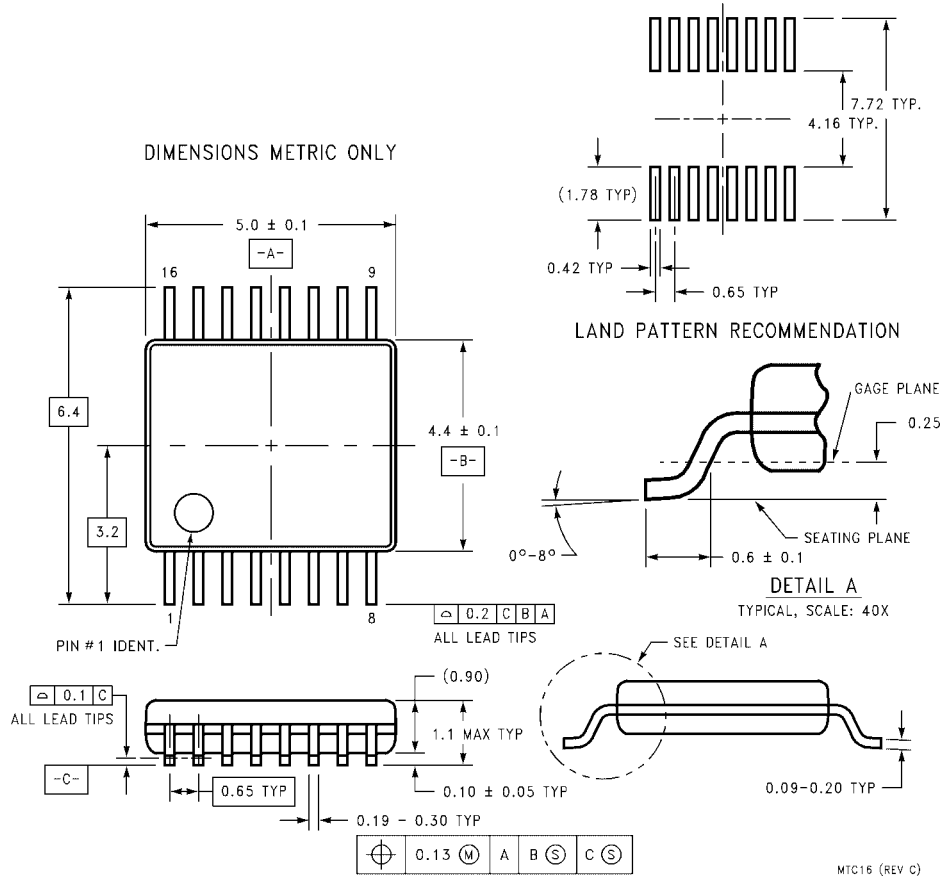DETAIL A

NOTES:
A. CONFORMS TO EIAJ EDR-7320 REGISTRATION, ESTABLISHED IN DECEMBER, 1998.
B. DIMENSIONS ARE IN MILLIMETERS.
C. DIMENSIONS ARE EXCLUSIVE OF BURRS, MOLD FLASH, AND TIE BAR EXTRUSIONS.

M16DRevB1

16-Lead Small Outline Package (SOP), EIAJ TYPE II, 5.3mm Wide
Package Number M16D

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)

DIMENSIONS METRIC ONLY

5.0 ± 0.1
−A−
16   9

6.4

4.4 ± 0.1
−B−

3.2

1   8

PIN #1 IDENT.

⌐ 0.2 C B A
ALL LEAD TIPS

⌐ 0.1 C
ALL LEAD TIPS

−C−

0.65 TYP

0.19 − 0.30 TYP

(0.90)

1.1 MAX TYP

0.10 ± 0.05 TYP

⊕ 0.13 Ⓜ A B Ⓢ C Ⓢ

7.72 TYP.
4.16 TYP.

(1.78 TYP)

0.42 TYP   0.65 TYP

LAND PATTERN RECOMMENDATION

GAGE PLANE

0.25

SEATING PLANE

0°−8°   0.6 ± 0.1

DETAIL A
TYPICAL, SCALE: 40X

SEE DETAIL A

0.09−0.20 TYP

MTC16 (REV C)

16-Lead Thin Shrink Small Outline Package (TSSOP), JEDEC MO-153, 4.4mm Wide
Package Number MTC16

MM74HC423A Dual Retriggerable Monostable Multivibrator

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)

16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300" Wide
Package Number N16E
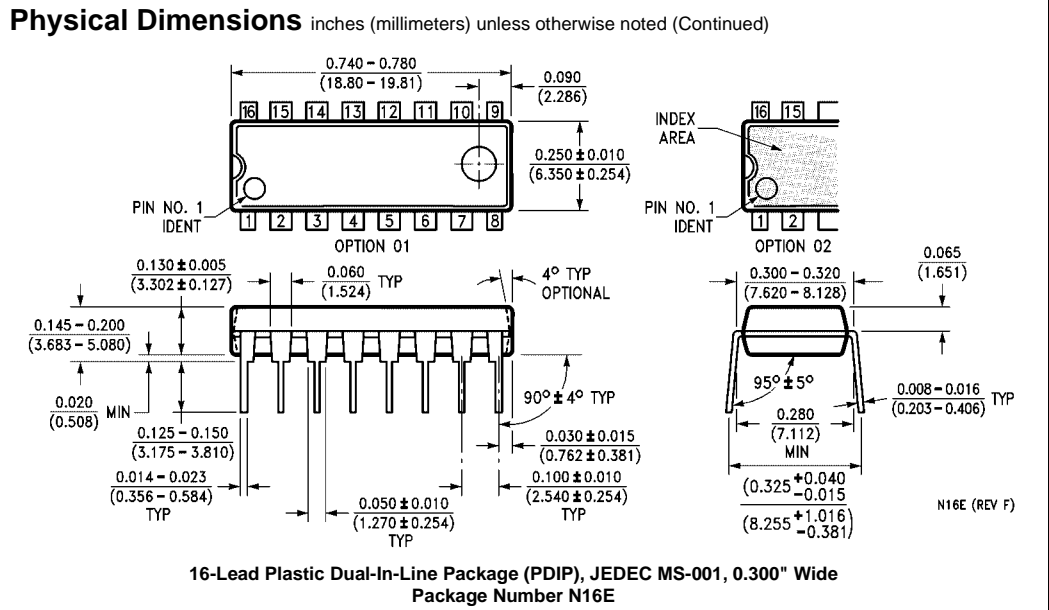
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2. A critical component in any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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