

TDA7389A

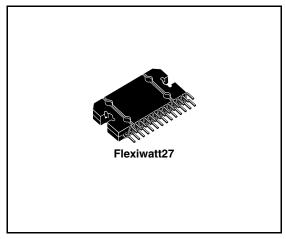
4 x 45W quad bridge car radio amplifier

Features

- High output power:
 - $-4 \times 45 \text{ W} / 4 \Omega \text{ max}.$
 - 4 x 28 W / 4 Ω @ 14.4 V, 1 kHz, 10 %
- Low distortion
- Low output noise
- Standby function
- Mute function
- Automute at min. supply voltage detection
- Low external component count:
 - Internally fixed gain (26dB)
 - No external compensation
 - No bootstrap capacitors
- Clipping detector
- Offset detector
- Diagnostic facility for:
 - Out to GND short
 - Out to V_S short
 - Thermal shutdown

Protections:

- Output short circuit to GND, to V_S, across the load
- Very inductive loads
- Overrating chip temperature with soft thermal limiter



- Load dump voltage
- Fortuitous open GND
- Output DC offset detector
- Reversed battery
- ESD

Description

The TDA7389A is an AB class audio power amplifier, packaged in Flexiwatt 27 designed for high end car radio applications.

Based on the fully complementary PNP/NPN configuration the TDA7389A allows a rail to rail output voltage swing with no need of bootstrap capacitors.

Table 1. Device summary

Order code	Package	Packing
TDA7389A	Flexiwatt27	Tube

October 2008 Rev 2 1/14

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1 Pin connection and block diagrams

Figure 1. Block diagram

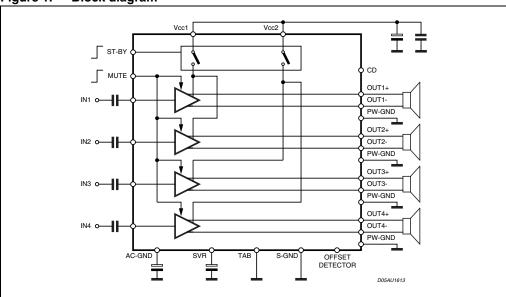
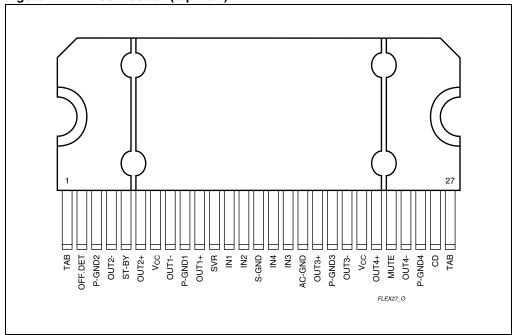


Figure 2. Pin connection (top view)



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2 Electrical specifications

2.1 Absolute maximum ratings

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V _S	Operating supply voltage	18	V
V _{S (DC)}	DC supply voltage	28	V
V _{S (pk)}	Peak supply voltage (t = 50ms)	50	V
I _O	Output peak current: Repetitive (duty cycle 10% at f = 10Hz) Non repetitive (t = 100µs)	4.5 5.5	А
P _{tot}	Power dissipation, (T _{case} = 70°C)	80	W
T _j	Junction temperature	150	°C
T _{stg}	Storage temperature	- 55 to 150	°C

2.2 Thermal data

Table 3. Thermal data

Symbol	Parameter	Value	Unit
R _{th j-case}	Thermal resistance junction to case max.	1	°C/W

2.3 Electrical characteristics

Table 4. Electrical characteristics

(V_S = 14.4V; f = 1KHz; R_g = 600Ω ; R_L = 4Ω ; T_{amb} = 25° C; Refer to the test and application diagram (*Figure 14*), unless otherwise specified.)

Symbol	Parameter	Test condition	Min.	Тур.	Max.	Unit
I _{q1}	Quiescent current	$R_L = \infty$		150	350	mA
V _{OS}	Output offset voltage	Play Mode			±80	mV
dV -	During mute on/off output offset voltagef	ITU R-ARM weighted	-10		10	- mV
dV _{OS}	During standby on/off output offset voltage	see Figure 13	-50		50	
G _v	Voltage gain		25	26	27	dB
ΔG _v	Channel gain unbalance				±1	dB
Po	Output power	THD = 10%; V _S = 14.4V	26	28		W
P _{o max}	Max.output power ⁽¹⁾	V _S = 14.4V	41	45		W

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Table 4. Electrical characteristics (continued) $(V_S = 14.4V; \ f = 1 \text{KHz}; \ R_g = 600\Omega; \ R_L = 4\Omega; \ T_{amb} = 25^{\circ}\text{C}; \ \text{Refer to the test and application diagram (\it Figure 14), unless otherwise specified.) }$

Symbol	Parameter	Test condition	Min.	Тур.	Max.	Unit
THD	Distortion	P _o = 4W		0.04	0.10	%
		"A" Weighted		50	70	μV
e_{No}	Output noise	Bw = 20 Hz to 20 kHz		70	100	μV
SVR	Supply voltage rejection	f = 100Hz; V _r = 1V _{rms}	50	75		dB
f _{ch}	High cut-off frequency	$P_0 = 0.5W$	100	200		KHz
R _i	Input Impedance		70	100	150	ΚΩ
	Cross talk	f = 1 kHz; P _o = 4W	60	75		dB
C _T	CIOSS taik	$f = 10 \text{ kHz}; P_0 = 4W$		60		dB
1.	Standby current	$V_{St-by} = 1.2V$			50	
I _{SB}	consumption	V _{St-by} = 0V			25	μA
I _{pin4}	Standby pin current	$V_{St-by} = 1.2 \text{ to } 2.6V$			±1	μΑ
V _{SB out}	Standby Out threshold voltage	(Amp: On)	2.6			V
V _{SB IN}	Standby In threshold voltage	(Amp: Off)			1.2	٧
A _M	Mute attenuation	P _{Oref} = 4 W	80	90		dB
V _{M out}	Mute Out threshold voltage	(Amp: play)	2.6			V
$V_{M in}$	Mute In threshold voltage	(Amp: mute)			1.2	٧
$V_{AM\ in}$	V _S automute threshold	(Amp: mute); Att \geq 80 dB; $P_{Oref} = 4 \Omega$ (Amp: play); Att < 0.1 dB; $P_{O} = 0.5 \text{ W}$		7.6	6.5 8.5	V
I _{pin22}	Muting pin current	V _{MUTE} = 1.2 V (Source current)	5	11	20	μА
		V _{MUTE} = 2.6 V	-5	-	-20	
Offset detec	otor					
OD _{LK}	OD leakage current				1	μA
V _{off}	Detected diff. output offset	V _{St-by} =5 V	1.5	±2	2.5	V
V _{Pin2_Fw27}	Pin2_Fw27 voltage for Detection = True	$I_{pin2} = 1 \text{ mA}$ $V_{St-by} = 5V; V_{off} > 2.5 \text{ V}$		0.2	0.4	V
Clipping de	tector		1	ı	1	1
CD _{LK}	Clip det high leakage current	CD Off			1	μΑ
CD _{SAT}	Clip det sat voltage	DC On; I _{CD} = 1 mA		0.2	0.4	V
CD _{THD}	Clip det THD level	P _o > 16 W			2	%

^{1.} Saturated square wave output.

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2.4 Electrical characteristic curves

Figure 3. Quiescent current vs. supply voltage

Figure 4. Quiescent output voltage vs. supply voltage

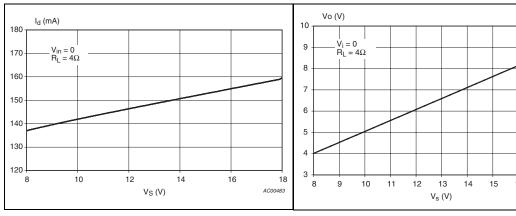


Figure 5. Output power vs. supply voltage

Figure 6. Maximum output power vs. supply voltage

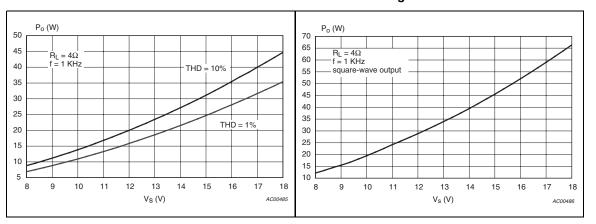


Figure 7. Distortion vs. output power

Figure 8. Distortion vs. frequency

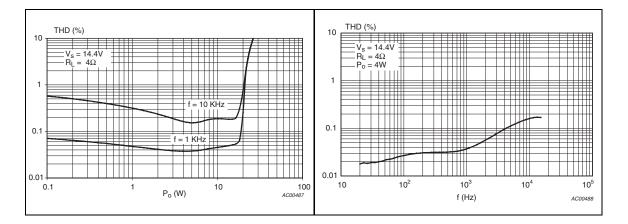


Figure 9. Supply voltage rejection vs. frequency

Figure 10. Crosstalk vs. frequency

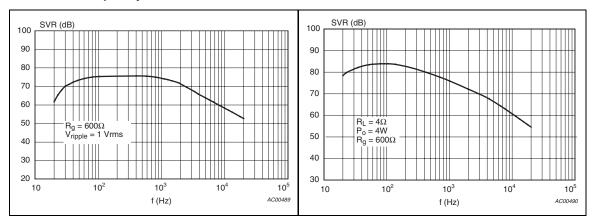


Figure 11. Output noise vs. source resistance Figure 12. Power dissipation and efficiency vs. output power

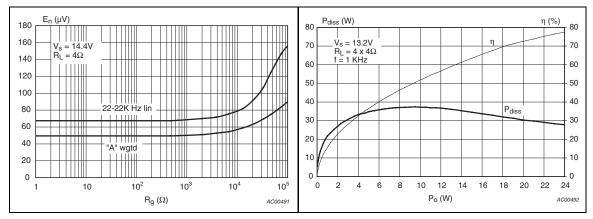
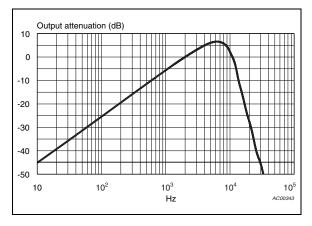
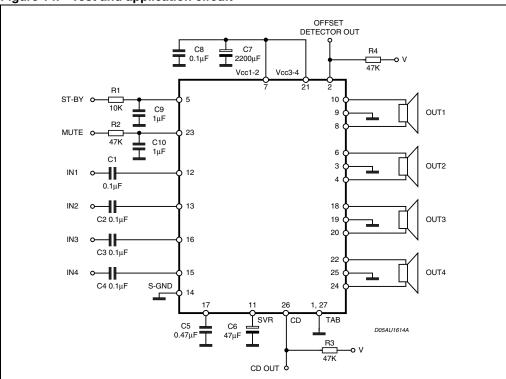


Figure 13. ITU R-ARM frequency response, weighting filter for transient pop



2.5 Test and application circuit

Figure 14. Test and application circuit



TDA7389A Application hints

3 Application hints

Ref. to the circuit of Figure 14.

3.1 SVR

Besides its contribution to the ripple rejection, the SVR capacitor governs the turn ON/OFF time sequence and, consequently, plays an essential role in the pop optimization during ON/OFF transients.

To conveniently serve both needs, its minimum recommended value is $22\mu F$, $47\mu F$ can be used to optimize pop performances.

3.2 Input stage

The TDA7389A's inputs are ground-compatible and can stand very high input signals (± 8Vpk) without any performances degradation.

If the standard value for the input capacitors $(0.1\mu F)$ is adopted, the low frequency cut-off will amount to 16 Hz.

The input capacitors should be 1/4 of the capacitor connected to AC-GND pin for optimum pop performances.

3.3 Standby and muting

Standby and muting facilities are both 3.3 V CMOS-compatible. If unused, a straight connection to $V_{\rm S}$ of their respective pins would be admissible.

Conventional/low-power transistors can be employed to drive muting and stand-by pins in absence of true CMOS ports or microprocessors. R-C cells have always to be used in order to smooth down the transitions for preventing any audible transient noises.

Since a DC current of about $10\mu A$ normally flows out of pin 23, the maximum allowable muting-series resistance (R₂) is $70 \text{ k}\Omega$, which is sufficiently high to permit a muting capacitor reasonably small (about $1\mu F$).

If R_2 is higher than recommended, the involved risk will be that the voltage at pin 23 may rise to above the 1.2 V threshold voltage and the device will consequently fail to turn OFF when the mute line is brought down.

About the stand-by, the time constant to be assigned in order to obtain a virtually pop-free transition has to be slower than 2.5V/ms.

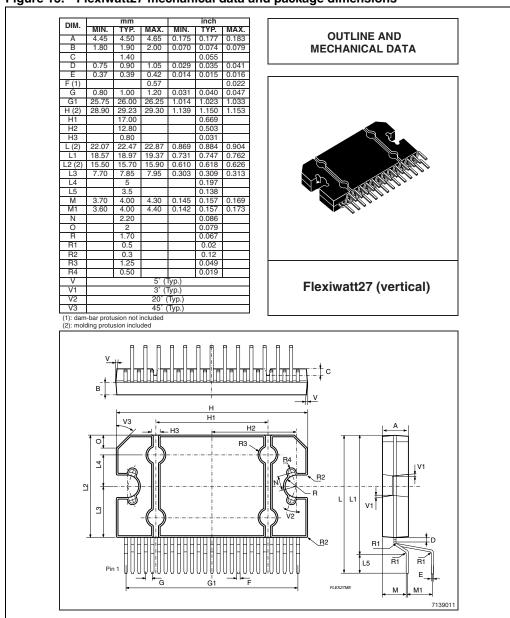
Package information TDA7389A

4 Package information

In order to meet environmental requirements, ST (also) offers these devices in ECOPACK[®] packages. ECOPACK[®] packages are lead-free. The category of second Level Interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label.

ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com.

Figure 15. Flexiwatt27 mechanical data and package dimensions



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TDA7389A Revision history

5 Revision history

Table 5. Document revision history

Date	Revision	Changes	
14-Nov-2007	1	Initial release.	
20-Oct-2008	2	Updated the Table 3: Thermal data on page 6.	

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