

LM4949 Boomer® Audio Power Amplifier Series

Stereo Class D Audio Subsystem with OCL Headphone Amplifier

General Description

The LM4949 is a fully integrated audio subsystem designed for stereo cell phone applications. The LM4949 combines a 2.5W stereo Class D amplifier plus a separate 190mW stereo headphone amplifier, volume control, and input mixer into a single device. The filterless class D amplifiers deliver 1.19W/ channel into an 8Ω load with <1% THD+N from a 5V supply. The headphone amplifier features National's Output Capacitor-less (OCL) architecture that eliminates the output coupling capacitors required by traditional headphone amplifiers. Additionally, the headphone amplifiers can be configured with capacitively coupled (CC)loads, or used to drive an external headphone amplifier. When configured for an external amplifier, the $\rm V_{DD}/2$ output (VOC) controls the external amplifier's shutdown input.

For improved noise immunity, the LM4949 features fully differential left, right and mono inputs. The three inputs can be mixed/multiplexed to either the speaker or headphone amplifiers. The left and right inputs can be used as separate single-ended inputs, mixing multiple stereo audio sources. The mixer, volume control, and device mode select are controlled through an I²C compatible interface.

Output short circuit and thermal shutdown protection prevent the device from being damaged during fault conditions. Superior click and pop suppression eliminates audible transients on power-up/down and during shutdown.

Key Specifications

■ Efficiency V _{DD} = 3.6V, 400mW	
into 8Ω	86.5%
■ Efficiency $V_{DD} = 5V$, 1W into 8Ω	87.4%
Quiescent Power Supply Current@ 3.6V	9.36mA
■ Power Output at V _{DD} = 5V	

Speaker:

$R_1 = 4\Omega$, THD+N $\leq 1\%$	
$R_1 = 8\Omega$, THD+N $\leq 1\%$	2W
$R_1 = 4\Omega$, THD+N $\leq 10\%$	1.19W 2.5W
11 - 452, 111D+N = 1070	2.5

Headphone: $B_{r} = 160 \text{ THD+N} \le 1\%$

TIL - 1032, TTID+IV = 170	153mW
$R_L = 32\Omega$, THD+N $\leq 1\%$	89mW

■ Shutdown Current 0.1µA

Features

- Output Short Circuit Protection
- Thermal Shutdown
- Stereo filterless Class D operation
- Selectable OCL/CC Headphone Drivers
- RF Suppression
- I²C Control Interface
- 32-step digital volume control
- Independent Speaker and Headphone Gain Settings
- Minimum external components
- Click and Pop suppression
- Micro-power shutdown
- Available in space-saving 25 bump µSMD package

Applications

- Mobile phones
- PDAs
- Laptops

Boomer® is a registered trademark of National Semiconductor Corporation.

Typical Application

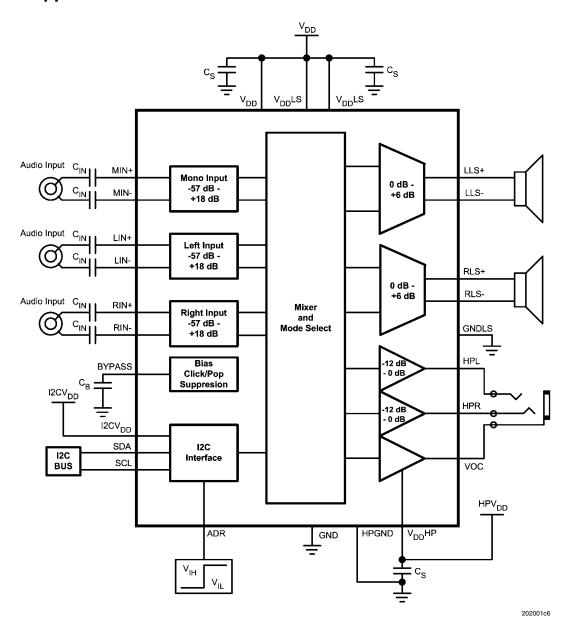
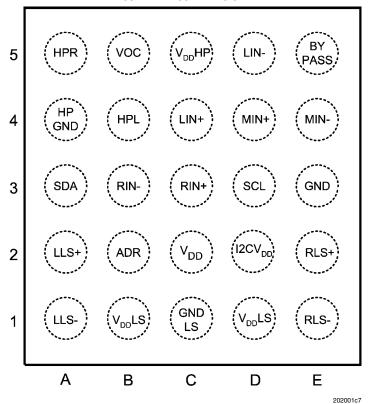


FIGURE 1. Typical Audio Amplifier Application Circuit

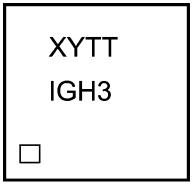
Connection Diagrams

TL Package 2.68mm x 2.68mm x 0.6mm



Top View Order Number LM4949TL See NS Package Number TLA25JJA

LM4949TL Marking



202001c0

Top View
XY — 2 digit datecode
TT — Die traceability

Absolute Maximum Ratings (Note 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ESD Susceptibility (Note 5) 200V Junction Temperature 150°C Thermal Resistance

 θ_{JA} 35.1°C/W

Operating Ratings

Temperature Range

 $\begin{aligned} &\mathsf{T_{MIN}} \leq \mathsf{T_{A}} \leq \mathsf{T_{MAX}} & -40^{\circ}\mathrm{C} \leq \mathsf{T_{A}} \leq +85^{\circ}\mathrm{C} \\ &\mathsf{Supply Voltage} \left(\mathsf{V_{DD}},\,\mathsf{V_{DD}LS}, \right. & 2.7\mathrm{V} \leq \mathsf{V_{DD}} \leq 5.5\mathrm{V} \end{aligned}$

 $V_{DD}HP$)

 I^2C Voltage (I^2CV_{DD}) 2.4V ≤ I^2CV_{DD} ≤ 5.5V

			LM	LM4949	
Symbol	Parameter	Conditions	Typical	Limit	Units (Limits)
			(Note 6)	(Notes 7, 8)	(Lillins)
		LS Mode Stereo	6	8.75	mA (max) mA
		Mono	4.5		IIIA
		OCL HP Mode			mA (max)
1	Supply Current	Stereo	5.0	6.5	mA
I _{DD}	DD Supply Current	Mono	4.3		111/4
		CC HP Mode			mA (max)
	Stereo	4.0	5.25	mA	
	Mono	3.3			
		Stereo LS + HP Mode	8.6		mA
I_{SD}	Shutdown Supply Current		0.03	2	μA (max)
V _{os}	Output Offset Voltage	Speaker (mode 1)	8.9	48.9	mV (max)
V OS	Output Offset Voltage	OCL HP (mode 1)	5.6	24.5	mV (max)
		LS Mode, f = 1 kHz			
		$R_L = 4\Omega$, THD+N = 10%	820		mW
		$R_L = 4\Omega$, THD+N = 1%	662		mW
		$R_L = 8\Omega$, THD+N = 10%	515		mW
		$R_L = 8\Omega$, THD+N = 1%	415	340	mW (min)
		OCP HP Mode, f = 1 kHz			
		$R_1 = 16\Omega$, THD+N = 10%	00.5		
P_{OUT}	Output Power	$R_1 = 16\Omega, THD + N = 1\%$	62.5 50		mW mW
001	· .	$R_1 = 32\Omega$, THD+N = 10%	37.5		mW
		$R_1 = 32\Omega$, THD+N = 1%	30.3		mW
		CC HP Mode, f = 1 kHz	1		
		$R_L = 16\Omega$, THD+N = 10%			
		$R_1 = 16\Omega$, THD+N = 1%	63		mW
		$R_1 = 32\Omega$, THD+N = 10%	50		mW
		$R_1 = 32\Omega$, THD+N = 1%	38		mW
		$\Pi_L = 3252, \Pi D + N = 1\%$	30		mW (min)

	_			4949	Units
Symbol	Parameter	Conditions	Typical Limit		(Limits)
			(Note 6)	(Notes 7, 8)	(=
		Differential Mode, f = 1kHz			
		HP Mode, $R_L = 16\Omega$, $P_{OUT} = 35$ mW			
		OCL	0.015		%
		cc	0.012		%
		HP Mode, $R_L = 32\Omega$, $P_{OUT} = 20$ mW			
ΓHD+N	Total Harmonic Distortion + Noise	OCL	0.017		%
		CC	0.018		%
		LS Mode			
		$R_L = 4\Omega$, $P_{OUT} = 300$ mW	0.023		%
		$R_L = 8\Omega, P_{OUT} = 150 \text{mW}$	0.023		%
		Single-Ended Input Mode, f = 1kHz	0.02		/0
		<u> </u>		1	
		HP Mode, $R_L = 16\Omega$, $P_{OUT} = 35$ mW	0.000		0/
		OCL CC	0.023		%
			0.017	+	%
ΓHD+N	Total Harmonic Distortion + Noise	HP Mode, $R_L = 32\Omega$, $P_{OUT} = 20$ mW	6.64-		
	OCL	0.019		%	
	CC	0.013		%	
		LS Mode			
	$R_L = 4\Omega$, $P_{OUT} = 300$ mW	0.05		%	
		$R_L = 8\Omega$, $P_{OUT} = 150$ mW	0.03		%
		Differential Input, A-weighted, Input R	eferred		
		Mono Input			
		OCL	16.4		μV
		CC	15.5		μV
		LS	43		μV
		All Inputs ON			
		OCL	29.8		μV
		CC	29.2		μV
N	Noise	LS	46.6		μV
IN		Single-Ended Input, A-weighted, Inpu	t Referred		
		Stereo Input			
		OCL	12		μV
		CC	11		μV
		LA	45		μV
		All Inputs ON			/
		OCL	23.7		μV
		CC	22.9		μV
	<u> </u>	LS	52		μV
1	Efficiency	LS Mode, $P_{OUT} = 400$ mW, $R_L = 8\Omega$	85.3		%
		LS Mode, $f = 1kHz$, $R_L = 8\Omega$, $V_{IN} = 1V_{IN}$	P-P		
/talls	Crasstalle	Differential Input Mode	84.7		dB
(talk	Crosstalk	OCL HP Mode, $f = 1kHz$, $R_L = 32\Omega$, V	$V_{INI} = 1V_{P-P}$		
		Differential Input Mode	68		dB
		CC Mode	68	+	ms
- ON	Turn on Time	OCL Mode	14		ms
UN		LS Mode	29		ms
<u></u>	Turn off Time	From any mode	683	+ +	ms
OFF	Tam on Time	<u> </u>		+	
7 -IN	Input Impedance	Maximum Gain Minimum Gain	24.8		kΩ
	1	IVIII III III Gall I	222.7	1	kΩ

Symbol Para			LM	4949	Units
	Parameter	Conditions	Typical	Limit	(Limits
			(Note 6)	(Notes 7, 8)	(Lillins)
		Volume Control	F-7		٩D
		Minimum Gain	_57 18		dB dB
		Maximum Gain	18		uВ
		LS Second Gain Stage	•		
		Step 0			
		Differential Input	6		dB
		Single-Ended Input	12		dB
		Step 1			
		Differential Input	4		dB
		Single-Ended Input	10		dB
A _V Gain	Step 2				
	Differential Input	2		dB	
		Single-Ended Input	8		dB
	Step 3	<u> </u>	†		
		Differential Input	0		dB
		Single-Ended Input	6		dB
		HP Second Gain Stage	ļ.		
		Step 0	0		dB
		Step 1	-6		dB
		Step 2	-12		dB
	Speaker Mode	-103		dB	
Mute	Mute Attenuation	Headphone Mode	-123		dB
		Speaker Mode, f = 1kHz,			
		$V_{IN} = 200 \text{mV}_{P-P}$	66.1		dB
CMRR	Common Mode Rejection Ratio	OCL Headphone Mode, f = 1kHz,			
		$V_{IN} = 200 \text{mV}_{P-P}$	70		dB
		Differential Input Mode, V _{RIPPLE} = 20	OmV _{B B}		
		OCL HP Mode, f = 217Hz	78.1		dB
PSRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	75.4		dB
		LS Mode, f = 217Hz	74		dB
		LS Mode, f = 1kHz	72.9		dB
		Single-Ended Input Mode, V _{RIPPLE} =	200mV _{P-P}		
		OCL HP Mode, f = 217Hz	77.5		dB
PSRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	81		dB
		LS Mode, f = 217Hz	69		dB
		LS Mode, f = 70.31kHz72.8	81		dB
		All Inputs ON, Single-Ended Input M	ode, V _{BIPPI F} = 2	200mV _{P-P}	
		OCL HP Mode, f = 217Hz	66.1		dB
PSRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	70.5		dB
		LS Mode, f = 217Hz	65.4		dB
		LS Mode, f = 1kHz	72.2		dB

			LM4949		11
Symbol	Parameter	Conditions	Typical Limit		Units (Limits)
			(Note 6)	(Notes 7, 8)	(Limits)
		LS Mode			
		Stereo	6.8	7.3	mA (max)
		Mono	4.9	5.3	mA (max)
		OCL HP Mode			
ı	Supply Current	Stereo	5.8	6.5	mA (max)
DD	Supply Current	Mono	4.9	5.5	mA (max)
		CC HP Mode			
		Stereo	4.7	5.2	mA (max)
		Mono	4.1	4.6	mA (max)
		Stereo LS + HP Mode	9.36		mA
SD	Shutdown Supply Current		0.03	1	μA (max)
	Outside Official Vallages	Headphone	6.7	20	mV (max)
V _{os}	Output Offset Voltage	Speaker	8.9	49	mV (max)
		LS Mode, f = 1 kHz			
		$R_L = 4\Omega$, THD+N = 10%			147
		$R_1 = 4\Omega$, THD+N = 1%	1.24		W
		$R_1 = 8\Omega$, THD+N = 10%	1 0.765		W W
		$R_L = 8\Omega$, THD+N = 1%	0.765		W
		OCL HP Mode, f = 1 kHz	0.010		**
		$R_L = 16\Omega$, THD+N = 10%			
n	Outsid Barrer	1 -	94		mW
Pout	Output Power	$R_L = 16\Omega$, THD+N = 1%	76		mW
		$R_L = 32\Omega$, THD+N = 10%	55		mW
		$R_L = 32\Omega$, THD+N = 1%	45		mW
		CC HP Mode, f = 1 kHz			
		$R_L = 16\Omega$, THD+N = 10%	93		mW
		$R_L = 16\Omega$, THD+N = 1%	75		mW
		$R_L = 32\Omega$, THD+N = 10%	56		mW
		$R_L = 32\Omega$, THD+N = 1%	45		mW
		Differential Mode, f = 1kHz	•		
		HP Mode, $R_L = 16\Omega$, $P_{OUT} = 50$ mW			
		OCL	0.021		%
		CC	0.021		%
		HP Mode, $R_1 = 32\Omega$,			
THD+N	Total Harmonic Distortion + Noise	P _{OUT} = 30mW			
		OCL	0.01		%
		CC	0.01		%
		LS Mode	-		• • • • • • • • • • • • • • • • • • • •
		$R_L = 4\Omega$, $P_{OUT} = 400$ mW	0.000		0/
		$R_L = 8\Omega, P_{OUT} = 300 \text{mW}$	0.023 0.02		% %

Symbol Parame			LM	14949	11:4-
	Parameter	Conditions	Typical	Limit	Units (Limits)
			(Note 6)	(Notes 7, 8)	(Lillins)
		Single-Ended Input Mode, f = 1kHz			
		HP Mode, $R_L = 16\Omega$, $P_{OUT} = 50$ mW			
	OCL	0.021		%	
		cc	0.017		%
		HP Mode, $R_L = 32\Omega$, $P_{OUT} = 30$ mW			
THD+N	Total Harmonic Distortion + Noise	OCL OCL	0.02		%
		CC	0.015		%
		LS Mode			
		$R_L = 4\Omega$, $P_{OUT} = 400$ mW	0.05		%
		$R_L = 8\Omega, P_{OUT} = 300 \text{mW}$	0.03		%
		Differential Mode, A-weighted, Input			/6
			ICICIICU		
		Mono Input OCL	16.4		μV
		CC	15.5		μV
		LS	43		μV
			40	+	μν
		All Inputs ON OCL	29.8		μV
		CC	29.6		μV
		LS	46.6		μV
e _N	Noise	Single-Ended Input, A-weighted, Inpu	ļ	1	h.,
		Stereo Input			
		OCL	12		μV
		CC	11		μV
		LS	45		μV
		All Inputs ON	_		<u> </u>
		OCL	23.7		μV
		CC	22.9		μV
		LS	52		μV
η	Efficiency	LS Mode, $P_{OUT} = 400$ mW, $R_L = 8\Omega$	86.5		%
•		LS Mode, $f = 1kHz$, $R_L = 8\Omega$, $V_{IN} = 1V_{IN}$			
		Differential Input Mode	86	T	dB
Xtalk	Crosstalk	OCL HP Mode, $f = 1kHz$, $R_L = 32\Omega$, V		1	
		Differential Input Mode	68	<u> </u>	dB
		CC Mode	75	+	45
T _{ON}	Turn on Time	OCL Mode	14		ms
· ON		LS Mode	31		ms
T _{OFF}	Turn off Time	From any mode	692	+	ms
OFF	Tani on Time	-		+	
Z _{IN}	Input Impedance	Maximum Gain	24.8		kΩ
	· · ·	Minimum Gain	222.7		kΩ

	Parameter	Conditions	LM4949		Units
Symbol			Typical Limit		(Limits)
			(Note 6)	(Notes 7, 8)	(Lillins)
		Volume Control	F-7		٩D
		Minimum Gain	-57		dB
		Maximum Gain	18		dB
	LS Second Gain Stage		•		
		Step 0			
		Differential Input	6		dB
		Single-Ended Input	12		dB
		Step 2			
		Differential Input	4		dB
	, Gain	Single-Ended Input	10		dB
\ _V		Step 2			
		Differential Input	2		dB
	Single-Ended Input	8		dB	
		Step 3			
		Differential Input	0		dB
		Single-Ended Input	6		dB
		HP Second Gain Stage	•	, ,	
		Step 0	0		
		Step 1	-6		dB
		Step 2	-12		dB
ute Mute Attenuation	Speaker Mode	-84		dB	
nute	Mute Attenuation	Headphone Mode	-95		dB
		Speaker Mode, f = 1kHz,	66		dB
NADD	Common Made Daisetian Datie	$V_{IN} = 200 \text{mV}_{P-P}$	66		ив
CMRR	Common Mode Rejection Ratio	OCL Headphone Mode, f = 1kHz,	60.6		-ID
		$V_{IN} = 200 \text{mV}_{P-P}$	68.6		dB
		Differential Input Mode, V _{RIPPLE} = 20	0mV _{P-P}	•	
		OCL HP Mode, f = 217Hz	75		dB
SRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	75		dB
		LS Mode, f = 217Hz	73		dB
		LS Mode, f = 1kHz	73		dB
		Single-Ended Input Mode, V _{RIPPLE} =	200mV _{P-P}	'	
		OCL HP Mode, f = 217Hz	75		dB
SRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	75		dB
		LS Mode, f = 217Hz	67		dB
		LS Mode, f = 1kHz	71		dB
		All Inputs ON, Single-Ended Input M		200mV _{P-P}	
		OCL HP Mode, f = 217Hz	72	<u> </u>	dB
SRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	70		dB
		LS Mode, f = 217Hz	60		dB
		LS Mode, f = 1kHz	65		dB

(01)			LM4949		
Symbol	Parameter	Conditions	Typical Limit		Units
•			(Note 6)	(Notes 7, 8)	(Limits)
		LS Mode	,		
		Stereo	9.9	10.9	mA (max)
		Mono	6.6	7.2	mA (max)
		OCL HP Mode			
ı	Cumply Cumpnt	Stereo	6.6	7.3	mA (max)
DD	Supply Current	Mono	5.5	6.2	mA (max)
		CC HP Mode			
		Stereo	5.4	5.9	mA (max)
		Mono	4.3	4.8	mA (max)
		Stereo LS + HP Mode	13		mA
SD	Shutdown Supply Current		0.1	1	μA (max)
· · · · · · · · · · · · · · · · · · ·	Output Offset Voltage	Headphone	10	52	mV (max)
V _{os}	Output Offset Voltage	Speaker	9.6	50	mV (max)
		LS Mode, f = 1 kHz			
		$R_L = 4\Omega$, THD+N = 10%	2.5		W
		$R_L = 4\Omega$, THD+N = 1%	2.01		W
		$R_L = 8\Omega$, THD+N = 10%	1.48		W
		$R_1 = 8\Omega$, THD+N = 1%	1.19		W
		OCL HP Mode, f = 1 kHz			
		$R_L = 16\Omega$, THD+N = 10%			
POUT	Output Power	$R_1 = 16\Omega$, THD+N = 1%	190		mW
. 001		$R_L = 32\Omega$, THD+N = 10%	154		mW mW
		$R_L = 32\Omega$, THD+N = 1%	109 89		mW mW
		CC HP Mode, f = 1 kHz	09		11100
		$R_L = 16\Omega$, THD+N = 10%			
		_	188		mW
		$R_L = 16\Omega$, THD+N = 1%	153		mW
		$R_L = 32\Omega$, THD+N = 10%	105		mW
		$R_L = 32\Omega$, THD+N = 1%	88		mW
		Differential Input Mode, f = 1kHz			
		HP Mode, $R_L = 16\Omega$, $P_{OUT} = 100$ mW			
		OCL	0.02		%
		cc	0.027		%
ΓHD + N	Total Harmonic Distortion + Noise	HP Mode, $R_L = 32\Omega$, $P_{OUT} = 50$ mW			
	Total Harmonia Biotortion 1 140100	OCL	0.02		%
		cc	0.022		%
		LS Mode			
		$R_L = 4\Omega, P_{OUT} = 1W$	0.022		%
		$R_L = 8\Omega$, $P_{OUT} = 600$ mW	0.02		%

			LM	4949	11!
Symbol Parameter	Parameter	Conditions	Typical	Limit	Units (Limits)
			(Note 6)	(Notes 7, 8)	(Lillins)
		Single-Ended Input Mode, f = 1kHz			
		HP Mode, $R_L = 16\Omega$, $P_{OUT} = 100$ mW			
		OCL	0.021		%
		CC	0.02		%
TUD N	Table 11 Property Name	HP Mode, $R_L = 32\Omega$, $P_{OUT} = 50$ mW			
THD + N	Total Harmonic Distortion + Noise	OCL	0.02		%
		CC	0.017		%
		LS Mode			
		$R_L = 4\Omega$, $P_{OUT} = 1W$	0.05		%
		$R_L = 8\Omega$, $P_{OUT} = 600$ mW	0.033		%
		Differential Input, A-weighted, Input F		<u> </u>	
		Mono Input			
		OCL	16.4		μV
		cc	15.5		μV
		LS	43		μV
		All Inputs ON			
		OCL	29.8		μV
		CC	29.2		μV
۵	Noise	LS	46.6		μV
GN	Noise	Single-Ended Input, A-weighted, Inpu	ıt Rrferred		
		Stereo Input			
		OCL	12		μV
		CC	11		μV
		LS	45		μV
		All Inputs ON			.,
		OCL	23.7		μV
		CC	22.9		μV
		LS	52		μV
η	Efficiency	LS Mode, $P_{OUT} = 1W$, $R_L = 8\Omega$	87.4		%
		LS Mode, $f = 1kHz$, $R_L = 8\Omega$, $V_{IN} = 1$	/ _{P-P}		
Xtalk	Crosstalk	Differential Input Mode	105.8		dB
Aldin	Ciossiaik	OCL HP Mode, $f = 1kHz$, $R_L = 32\Omega$, $V_{IN} = 1V_{P-P}$			
		Differential Input Mode	69.6		dB
T _{ON}		CC Mode	89		ms
5.1	Turn on Time	OCL Mode	14		ms
		LS Mode	35		ms
T _{OFF}	Turn off Time	From any mode	716		ms
		Maximum Gain	24.8		kΩ
Z_{IN}	Input Impedance	Minimum Gain	222.7		kΩ

			LM	4949	11	
Symbol	Parameter	Conditions	Typical Limit		Units	
			(Note 6)	(Notes 7, 8)	(Limits)	
		Volume Control			JD.	
		Minimum Gain	-57		dB	
		Maximum Gain	18		dB	
		LS Second Gain Stage	•	•	•	
		Step 0				
		Differential Input	6		dB	
		Single-Ended Input	12		dB	
Λ_V Gain	Step 1					
	Differential Input	4		dB		
	Single-Ended Input	10		dB		
, .V	v Sam	Step 2				
		Differential Input	8		dB	
	Single-Ended Input	2		dB		
	Step 3					
		Differential Input	0		dB	
	Single-Ended Input	6		dB		
		HP Second Gain Stage			1	
		Step 0	0		dB	
	Step 1	-6		dB		
	Step 2	-12		dB		
Mute	Mute Attenuation	Speaker Mode	-102.7		dB	
		Headphone Mode	-123		dB	
		Speaker Mode, f = 1kHz,	64.4		dB	
CMRR	Common Mode Rejection Ratio	$V_{IN} = 200 \text{mV}_{P-P}$				
	_	OCL Headphone Mode, f = 1kHz,	74.3		dB	
		$V_{IN} = 200 \text{mV}_{P-P}$				
		Differential Input Mode, V _{RIPPLE} = 200mV _{P-P}				
		OCL HP Mode, f = 217Hz	68.3		dB	
PSRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	67.9		dB	
		LS Mode, f = 217Hz	73.8		dB	
		LS Mode, f = 1kHz	72		dB	
		Single-Ended Input Mode, V _{RIPPLE} =	200mV _{P-P}			
		OCL HP Mode, f = 217Hz	70.55		dB	
PSRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	63.05		dB	
		LS Mode, f = 217Hz	64.6		dB	
		LS Mode, f = 1kHz	70.3		dB	
		All Inputs ON, Single-Ended Input M		200mV _{P-P}	L	
		OCL HP Mode, f = 217Hz	63.1	F *F	dB	
PSRR	Power Supply Rejection Ratio	OCL HP Mode, f = 1kHz	66.4		dB	
. 5	- Shor Supply Hojeotion Hallo	LS Mode, f = 217Hz	59.1		dB	
			+		•	
		LS Mode, f = 1kHz	69.3		dB	

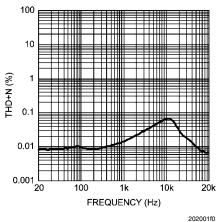
- Note 1: All voltages are measured with respect to the ground pin, unless otherwise specified.
- **Note 2:** Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which guarantee specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not guaranteed for parameters where no limit is given, however, the typical value is a good indication of device performance.
- Note 3: The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX} , θ_{JA} , and the ambient temperature T_A . The maximum allowable power dissipation is $P_{DMAX} = (T_{JMAX} T_A) / \theta_{JA}$ or the number given in Absolute Maximum Ratings, whichever is lower. For the LM4949, see power derating currents for additional information.
- Note 4: Human body model, 100pF discharged through a 1.5k Ω resistor.
- Note 5: Machine Model, 220pF 240pF discharged through all pins.
- Note 6: Typicals are measured at 25°C and represent the parametric norm.
- Note 7: Limits are guaranteed to National's AOQL (Average Outgoing Quality Level).
- Note 8: Datasheet min/max specification limits are guaranteed by design, test or statistical analysis.

TABLE 1. Bump Description

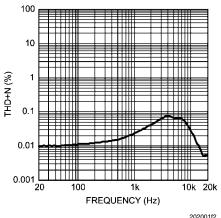
BUMP	NAME	DESCRIPTION
A1	LLS-	Left Channel Loudspeaker Inverting Output
A2	LLS+	Left Channel Loudspeaker Non-inverting Output
A3	SDA	Serial Data Input
A4	HPGND	Headphone Ground
A5	HPR	Right Channel Headphone Output
B1	VDDLS	Speaker Power Supply
B2	ADR	Address Select Bit
B3	RIN-	Right Channel Inverting Input
B4	HPL	Left Channel Headphone Output
B5	VOC	Headphone Return Bias Output
C1	GNDLS	Speaker Ground
C2	VDD	Power Supply
C3	RIN+	Right Channel Non-Inverting Input
C4	LIN+	Left Channel Non-inverting Input
C5	VDDHP	Headphone Power Supply
D1	VDDLS	Speaker Power Supply
D2	I2CVDD	I2C Power Supply
D3	SCL	Serial Clock Input
D4	MIN+	Mono Channel Non-inverting Input
D5	LIN-	Left Channel Inverting Input
E1	RLS-	Right Channel Loudspeaker Inverting Output
E2	RLS+	Right Channel Loudspeaker Non-inverting Output
E3	GND	Ground
E4	MIN-	Mono Channel Inverting Input
E5	BYPASS	Mid-rail Bias Bypass

Typical Performance Characteristics

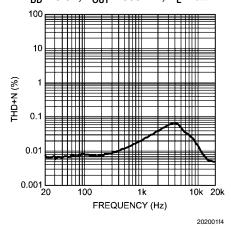
THD+N vs Frequency Speaker Mode, Differential Input $V_{DD}=3.0V,\,P_{OUT}=300mW,\,R_{L}=4\Omega$



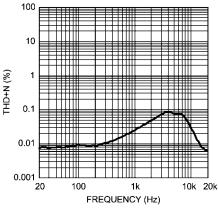
THD+N vs Frequency Speaker Mode, Differential Input V_{DD} = 5.0V, P_{OUT} = 1W, R_L = 4Ω



THD+N vs Frequency Speaker Mode, Differential Input $V_{DD} = 3.6V$, $P_{OUT} = 300$ mW, $R_L = 8\Omega$

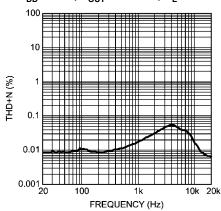


THD+N vs Frequency Speaker Mode, Differential Input $V_{DD}=3.6V,\,P_{OUT}=400mW,\,R_{L}=4\Omega$



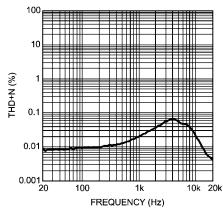
202001f1

THD+N vs Frequency Speaker Mode, Differential Input $V_{DD} = 3.0V$, $P_{OUT} = 150$ mW, $R_L = 8\Omega$



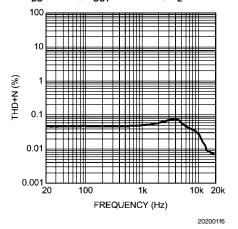
202001f3

THD+N vs Frequency Speaker Mode, Differential Input $V_{DD} = 5.0V$, $P_{OUT} = 600$ mW, $R_L = 8\Omega$

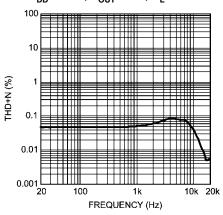


202001f5

THD+N vs Frequency Speaker Mode, Single-Ended Input $V_{DD}=3.0V,\,P_{OUT}=300mW,\,R_L=4\Omega$

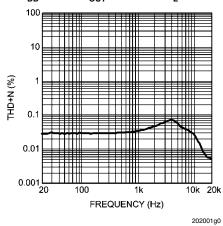


THD+N vs Frequency Speaker Mode, Single-Ended Input V_{DD} = 5.0V, P_{OUT} = 1W, R_L = 4Ω

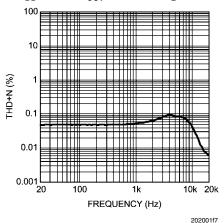


THD+N vs Frequency Speaker Mode, Single-Ended Input $V_{DD}=3.6V,\,P_{OUT}=300 mW,\,R_{L}=8\Omega$

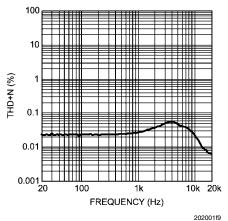
202001f8



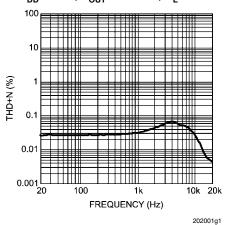
THD+N vs Frequency Speaker Mode, Single-Ended Input $V_{DD}=3.6V,\,P_{OUT}=400mW,\,R_L=4\Omega$



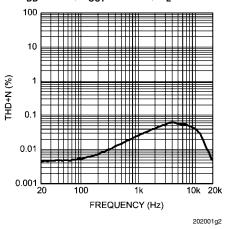
THD+N vs Frequency Speaker Mode, Single-Ended Input $V_{DD}=3.0V,\,P_{OUT}=150mW,\,R_{L}=8\Omega$



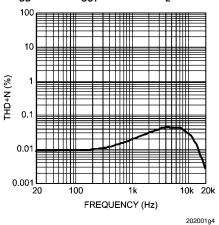
THD+N vs Frequency Speaker Mode, Single-Ended Input $V_{DD} = 5.0V$, $P_{OUT} = 600$ mW, $R_L = 8\Omega$



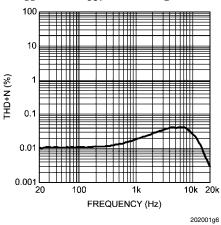
THD+N vs Frequency OCL Headphone Mode, Differential Input $V_{DD}=3.0V,\,P_{OUT}=35mW,\,R_L=16\Omega$



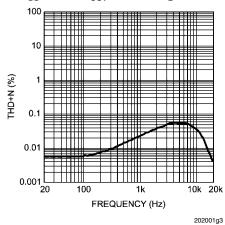
THD+N vs Frequency OCL Headphone Mode, Differential Input $V_{DD}=5.0V,\,P_{OUT}=100mW,\,R_L=16\Omega$



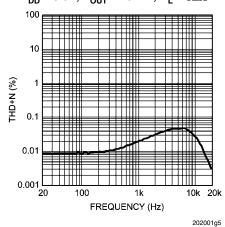
THD+N vs Frequency OCL Headphone Mode, Differential Input $V_{DD}=3.6V,\,P_{OUT}=30mW,\,R_L=32\Omega$



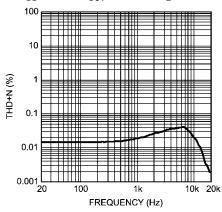
THD+N vs Frequency OCL Headphone Mode, Differential Input $V_{DD}=3.6V,\,P_{OUT}=50mW,\,R_L=16\Omega$



THD+N vs Frequency OCL Headphone Mode, Differential Input $V_{DD}=3.0V,\,P_{OUT}=20mW,\,R_L=32\Omega$

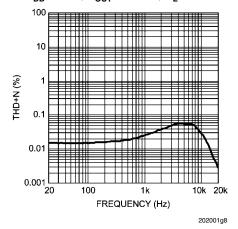


THD+N vs Frequency OCL Headphone Mode, Differential Input V_{DD} = 5.0V, P_{OUT} = 50mW, R_{L} = 32 Ω

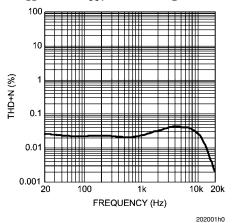


202001g7

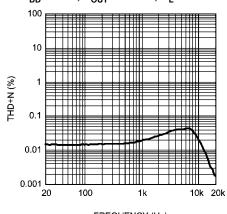
THD+N vs Frequency OCL Headphone Mode, Single-Ended Input V_{DD} = 3.0V, P_{OUT} = 35mW, R_L = 16 Ω



THD+N vs Frequency OCL Headphone Mode, Single-Ended Input V_{DD} = 5.0V, P_{OUT} = 100mW, R_L = 16 Ω

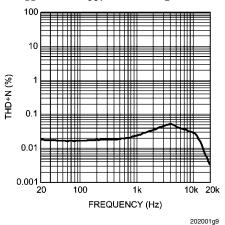


THD+N vs Frequency OCL Headphone Mode, Single-Ended Input V_{DD} = 3.6V, P_{OUT} = 30mW, R_L = 32 Ω

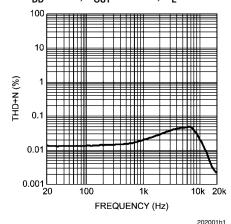


FREQUENCY (Hz) 202001h2

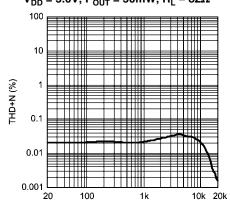
THD+N vs Frequency OCL Headphone Mode, Single-Ended Input V_{DD} = 3.6V, P_{OUT} = 50mW, R_L = 16 Ω



THD+N vs Frequency OCL Headphone Mode, Single-Ended Input V_{DD} = 3.0V, P_{OUT} = 20mW, R_L = 32 Ω



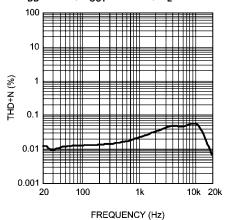
THD+N vs Frequency OCL Headphone Mode, Single-Ended Input V_{DD} = 5.0V, P_{OUT} = 50mW, R_L = 32 Ω



FREQUENCY (Hz)

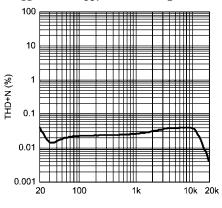
202001h3

THD+N vs Frequency CC Headphone Mode, Differential Input $V_{DD}=3.0V,\,P_{OUT}=35mW,\,R_L=16\Omega$



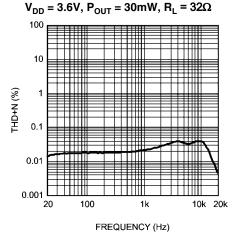
202001h4

THD+N vs Frequency CC Headphone Mode, Differential Input $V_{DD}=5.0V,\,P_{OUT}=100mW,\,R_{I}=16\Omega$



FREQUENCY (Hz)

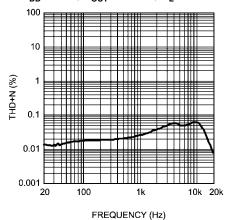
THD+N vs Frequency CC Headphone Mode, Differential Input



202001h8

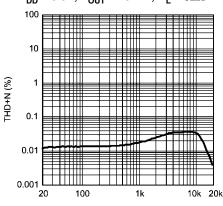
202001h6

THD+N vs Frequency CC Headphone Mode, Differential Input $V_{DD}=3.6V,\,P_{OUT}=50mW,\,R_L=16\Omega$



202001h5

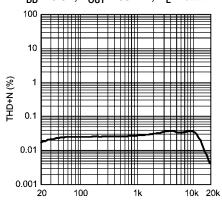
THD+N vs Frequency CC Headphone Mode, Differential Input $V_{DD}=3.0V,\,P_{OUT}=20mW,\,R_L=32\Omega$



FREQUENCY (Hz)

202001h7

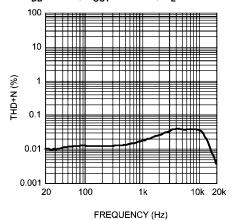
THD+N vs Frequency CC Headphone Mode, Differential Input $V_{DD} = 5.0V,\, P_{OUT} = 50 mW,\, R_L = 32 \Omega$



FREQUENCY (Hz)

202001h9

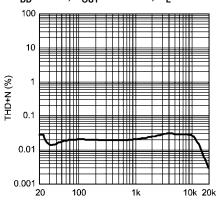
THD+N vs Frequency CC Headphone Mode, Single-Ended Input V_{DD} = 3.0V, P_{OUT} = 35mW, R_L = 16 Ω



14021101 (112)

202001i3

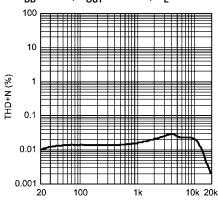
THD+N vs Frequency CC Headphone Mode, Single-Ended Input V_{DD} = 5.0V, P_{OUT} = 100mW, R_L = 16 Ω



FREQUENCY (Hz)

202001i5

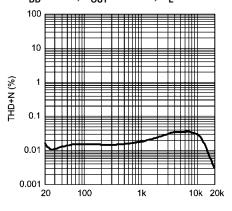
THD+N vs Frequency CC Headphone Mode, Single-Ended Input V_{DD} = 3.6V, P_{OUT} = 30mW, R_{L} = 32 Ω



FREQUENCY (Hz)

202001i7

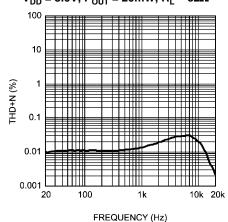
THD+N vs Frequency CC Headphone Mode, Single-Ended Input V_{DD} = 3.6V, P_{OUT} = 50mW, R_L = 16 Ω



FREQUENCY (Hz)

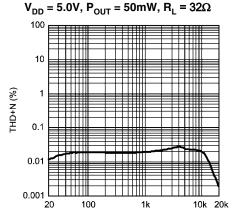
202001i4

THD+N vs Frequency CC Headphone Mode, Single-Ended Input V_{DD} = 3.0V, P_{OUT} = 20mW, R_L = 32 Ω



202001i6

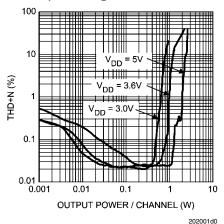
THD+N vs Frequency CC Headphone Mode, Single-Ended Input



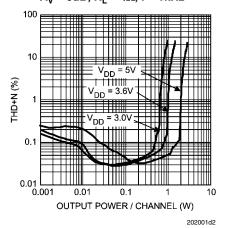
FREQUENCY (Hz)

202001i8

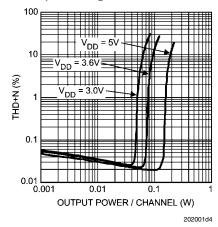
THD+N vs Output Power Speaker Mode, Differential Input $A_V = 6$ dB, $R_L = 4\Omega$, f = 1kHz



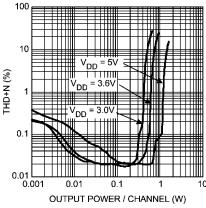
THD+N vs Output Power Speaker Mode, Single-Ended Input $A_V = 6dB$, $R_L = 4\Omega$, f = 1kHz



THD+N vs Output Power OCL Headphone Mode, Differential Input $A_V = 0 dB, R_L = 16\Omega, f = 1 kHz$

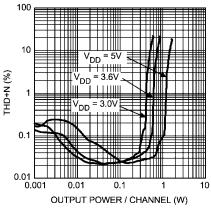


THD+N vs Output Power Speaker Mode, Differential Input $A_V = 6dB$, $R_L = 8\Omega$, f = 1kHz



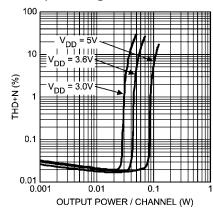
202001

THD+N vs Output Power Speaker Mode, Single-Ended Input $A_V = 6dB$, $R_L = 8\Omega$, f = 1kHz



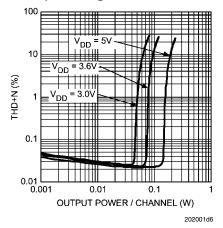
202001d3

THD+N vs Output Power OCL Headphone Mode, Differential Input $A_V = 0$ dB, $R_L = 32\Omega$, f = 1kHz

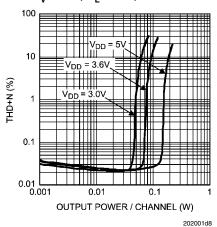


202001d5

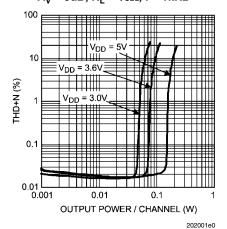
THD+N vs Output Power OCL Headphone Mode, Single-Ended Input A_V = 0dB, R_L = 16 Ω , f = 1kHz



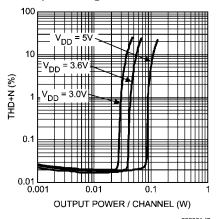
THD+N vs Output Power CC Headphone Mode, Differential Input $A_V = 0$ dB, $R_L = 16\Omega$, f = 1kHz



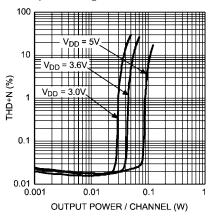
THD+N vs Output Power CC Headphone Mode, Single-Ended Input $A_V = 0$ dB, $R_L = 16\Omega$, f = 1kHz



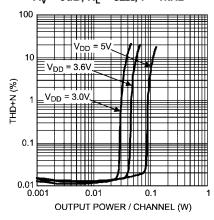
THD+N vs Output Power OCL Headphone Mode, Single-Ended Input $A_V = 0$ dB, $R_L = 32\Omega$, f = 1kHz



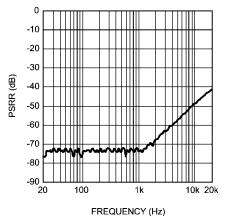
THD+N vs Output Power CC Headphone Mode, Differential Input $A_V = 0$ dB, $R_L = 32\Omega$, f = 1kHz



THD+N vs Output Power CC Headphone Mode, Single-Ended Input $A_V = 0$ dB, $R_L = 32\Omega$, f = 1kHz



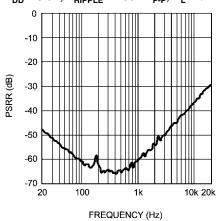
PSRR vs Frequency Speaker Mode, Differential Input $V_{DD} = 3.6V$, $V_{RIPPLE} = 200 mV_{P-P}$, $R_L = 8\Omega$



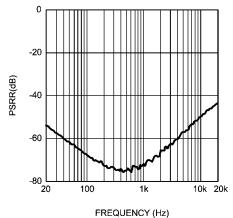
202001i9

202001j1

PSRR vs Frequency Speaker Mode, Single-Ended Input Stereo and Mono Inputs Active $V_{DD}=3.6V,\,V_{RIPPLE}=200mV_{P-P},\,R_L=8\Omega$

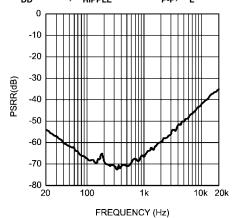


PSRR vs Frequency OCL Headphone Mode, Single-Ended Input V_{DD} = 3.6V, V_{RIPPLE} = 200m V_{P-P} , R_L = 32 Ω



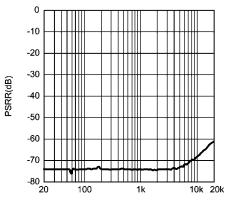
202001j3

PSRR vs Frequency Speaker Mode, Differential Input $V_{DD}=3.6V,\,V_{RIPPLE}=200mV_{P-P},\,R_L=8\Omega$



202001j0

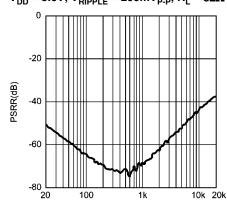
PSRR vs Frequency OCL Headphone Mode, Differential Input $V_{DD} = 3.6V$, $V_{RIPPLE} = 200 mV_{P-P}$, $R_L = 32 \Omega$



FREQUENCY (Hz)

202001j2

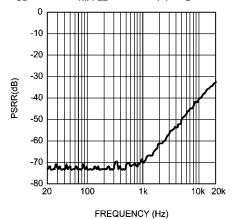
PSRR vs Frequency OCL Headphone Mode, Single-Ended Input Stereo and Mono Inputs Active $V_{DD}=3.6V,\,V_{RIPPLE}=200mV_{P.P},\,R_L=32\Omega$



FREQUENCY (Hz)

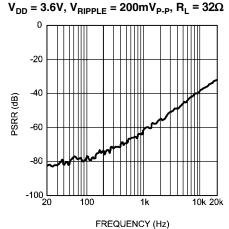
202001j4

PSRR vs Frequency CC Headphone Mode, Differential Input $V_{DD}=3.6V,\,V_{RIPPLE}=200mV_{P-P},\,R_L=32\Omega$



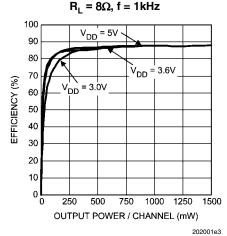
202001j5

PSRR vs Frequency
CC Headphone Mode, Single-Ended Input
Stereo and Mono Inputs Active

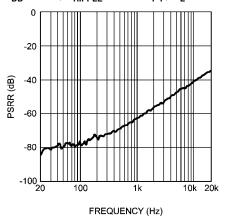


202001j7

Efficiency vs Output Power Speaker Mode

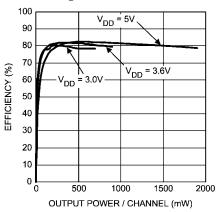


PSRR vs Frequency CC Headphone Mode, Single-Ended Input V_{DD} = 3.6V, V_{RIPPLE} = 200m V_{P-P} , R_L = 32 Ω



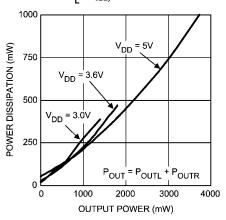
202001j6

Efficiency vs Output Power Speaker Mode $R_1 = 32\Omega$, f = 1kHz



202001e2

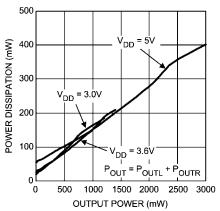
Power Dissipation vs Output Power Speaker Mode $R_L = 4\Omega, \, f = 1 \text{kHz}$



20200139

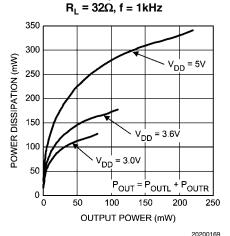
Power Dissipation vs Output Power Speaker Mode



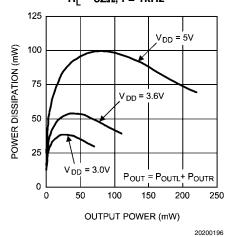


20200140

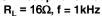
Power Dissipation vs Output Power OCL Headphone Mode

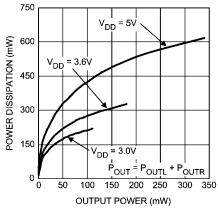


Power Dissipation vs Output Power CC Headphone Mode $R_1 = 32\Omega$, f = 1kHz



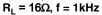
Power Dissipation vs Output Power OCL Headphone Mode

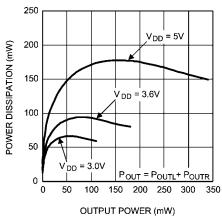




20200168

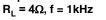
Power Dissipation vs Output Power CC Headphone Mode

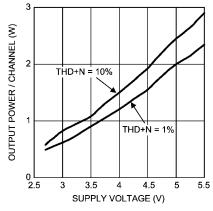




20200195

Output Power vs Supply Voltage Speaker Mode

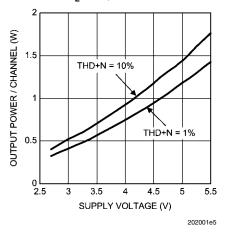




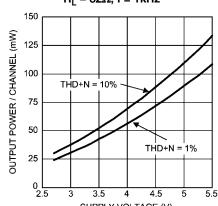
202001e4

Output Power vs Supply Voltage Speaker Mode

 $R_L = 8\Omega$, f = 1kHz



Output Power vs Supply Voltage OCL Headphone Mode $R_L = 32\Omega$, f = 1kHz



202001e7

5.5

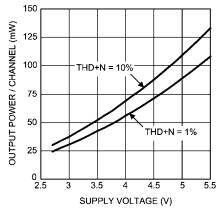
5

4.5

Output Power vs Supply Voltage CC Headphone Mode $R_L = 32\Omega$, f = 1kHz

SUPPLY VOLTAGE (V)

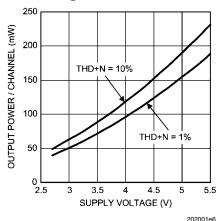
3.5



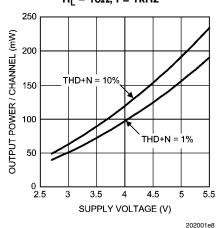
202001e9

Output Power vs Supply Voltage OCL Headphone Mode

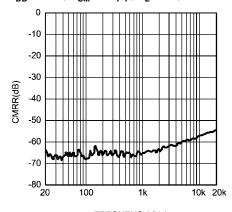
 $R_L = 16\Omega$, f = 1kHz



Output Power vs Supply Voltage CC Headphone Mode $R_L = 16\Omega$, f = 1kHz



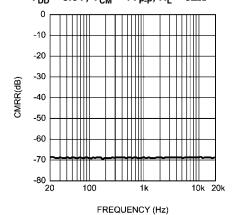
CMRR vs Frequency Speaker Mode, Differential Input V_{DD} = 3.6V, V_{CM} = 1 V_{P-P} , R_L = 8 Ω , f = 1kHz



FREQUENCY (Hz)

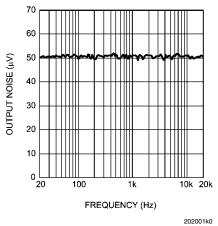
202001j8

CMRR vs Frequency OCL Headphone Mode V_{DD} = 3.6V, V_{CM} = 1 V_{P-P} , R_L = 32 Ω

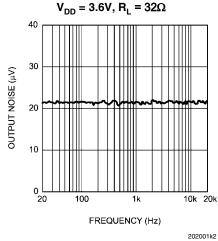


Output Noise vs Frequency Speaker Mode, Single-Ended Input Stereo and Mono Inputs Active $V_{DD} = 3.6V, R_L = 8\Omega$

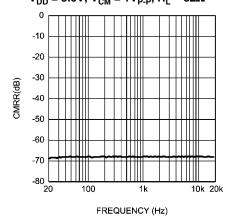
202001j9



Output Noise vs Frequency CC Headphone Mode, Single-Ended Input Stereo and Mono Inputs Active

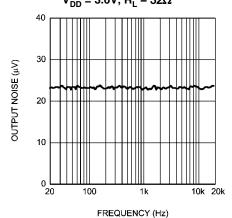


CMRR vs Frequency CC Headphone Mode V_{DD} = 3.6V, V_{CM} = 1 V_{P-P} , R_L = 32 Ω



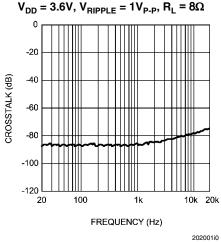
202001k3

Output Noise vs Frequency OCL Headphone Mode, Single-Ended Input Stereo and Mono Inputs Active $V_{DD} = 3.6V, R_{L} = 32\Omega$



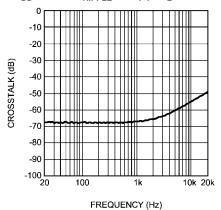
202001k1

Crosstalk vs Frequency Speaker Mode



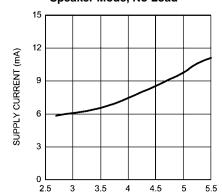
Crosstalk vs Frequency OCL Headphone Mode

 $V_{DD} = 3.6V$, $V_{RIPPLE} = 1V_{P-P}$, $R_L = 32\Omega$



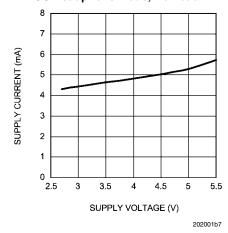
Supply Current vs Supply Voltage Speaker Mode, No Load

202001i1



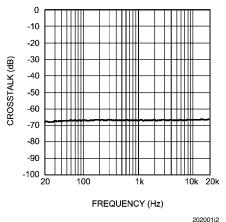
Supply Current vs Supply Voltage CC Headphone Mode, No Load

SUPPLY VOLTAGE (V)

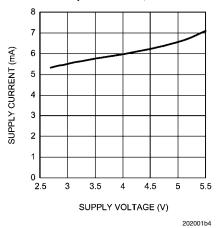


Crosstalk vs Frequency CC Headphone Mode

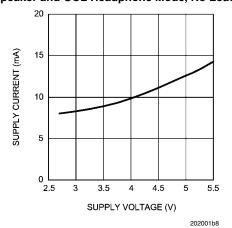
 V_{DD} = 3.6V, V_{RIPPLE} = 1 V_{P-P} , R_L = 32 Ω



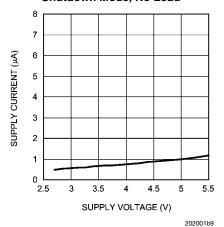
Supply Current vs Supply Voltage OCL Headphone Mode, No Load



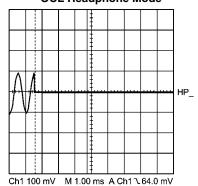
Supply Current vs Supply Voltage Speaker and OCL Headphone Mode, No Load



Supply Current vs Supply Voltage Shutdown Mode, No Load

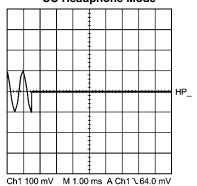


Turn-Off OCL Headphone Mode



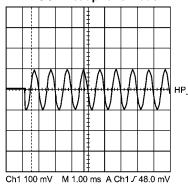
20200114

Turn-Off CC Headphone Mode



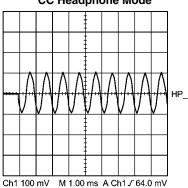
20200116

Turn-On OCL Headphone Mode



20200113

Turn-On CC Headphone Mode



20200115

Application Information

12C COMPATIBLE INTERFACE

The LM4949 is controlled through an I²C compatible serial interface that consists of two wires; clock (SCL) and data (SDA). The clock line is uni-directional. The data line is bi-directional (open-collector) although the LM4949 does not

write to the I²C bus. The maximum clock frequency specified by the I²C standard is 400kHz.

To avoid an address conflict with another device on the I^2C bus, the LM4949 address is determined by the ADR pin, the state of ADR determines address bit A1 (Table 2). When ADR = 0, the address is 1111 1000. When ADR = 1 the device address is 1111 1010.

TABLE 2. Device Address

ADR	A 7	A6	A 5	A 4	A3	A2	A 1	A0
X	1	1	1	1	1	0	Х	0
0	1	1	1	1	1	0	0	0
1	1	1	1	1	1	0	1	0

BUS FORMAT

The I²C bus format is shown in Figure 2. The "start" signal is generated by lowering the data signal while the clock is high. The start signal alerts all devices on the bus that a device address is being written to the bus.

The 8-bit device address is written to the bus next, most significant bit first. The data is latched in on the rising edge of the clock. Each address bit must be stable while the clock is high. After the last address bit is sent, the master device releases the data line, during which time, an acknowledge clock pulse

is generated. If the LM4949 receives the address correctly, then the LM4949 pulls the data line low, generating an acknowledge bit (ACK).

Once the master device has registered the ACK bit, the 8-bit register address/data word is sent. Each data bit should be stable while the clock level is high. After the 8-bit word is sent, the LM4949 sends another ACK bit. Following the acknowledgement of the data word, the master device issues a "stop" bit, allowing SDA to go high while the clock signal is high.

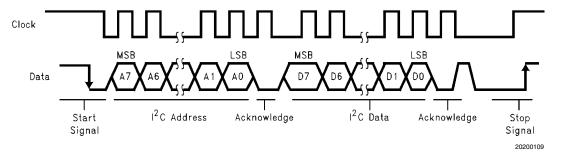


FIGURE 2. I²C Bus Format

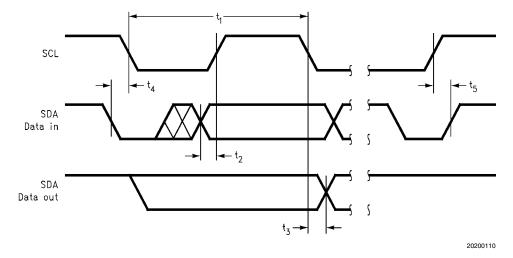


FIGURE 3. I²C Timing Diagram

TABLE 3. I²C Control Registers

REGISTE R	REGISTE R NAME	D7	D6	D5	D4	D3	D2	D1	D0
0.0	Shutdown Control	0	0	0	0	0	OCL_LGC *	OCL*	PWR_ON
0.1	Stereo Input Mode Control	0	0	0	1	L1_INSEL	L2_INSEL	SDB_HPSEL	SDB_MUXSE L
1	Speaker Output Mux Control	0	0	1	LS_XSEL	LSR_MSEL	LSR_SSEL	LSL_MSEL	LSL_SSEL
2	Headphon e Output Mux Control	0	1	0	HP_XSEL	HPR_MSEL	HPR_SSEL	HPL_MSEL	HPL_SSEL
3.0	Output On/ Off Control	0	1	1	0	HPR_ON	HPL_ON	LSR_ON	LSL_ON
3.1	Reserved	0	1	1	1	RESERVED	RESERVED	RESERVED	RESERVED
4.0	Headphon e Output Stage Gain Control	1	0	0	0	HPG1	HPG0	RESERVED	RESERVED
4.1	Speaker Output Stage Gain Control	1	0	0	1	LSRG1	LSRG0	LSLG1	LSLG0
5	Mono Input Gain Control	1	0	1	MG4	MG3	MG2	MG1	MG0
6	Left Input Gain Control	1	1	0	LG4	LG3	LG2	LG1	LG0
7	Right Input Gain Control	1	1	1	RG4	RG3	RG2	RG1	RG0

^{*} Note: OCL_LGC = 1 and OCL = 1 at the same time is not allowed.

GENERAL AMPLIFIER FUNCTION

Class D Amplifier

The LM4949 features a high-efficiency, filterless, Class D stereo amplifier. The LM4949 Class D amplifiers feature a filterless modulation scheme, the differential outputs of each channel switch at 300khz, from $\rm V_{DD}$ to GND. When there is no input signal applied, the two outputs (_LS+ and _LS-) switch with a 50% duty cycle, with both outputs in phase. Because the outputs of the LM4949 are differential, the two signals cancel each other. This results in no net voltage across the speaker, thus no load current during the idle state, conserving power.

When an input signal is applied, the duty cycle (pulse width) changes. For increasing output voltages, the duty cycle of _LS+ increases, while the duty cycle of _LS- decreases. For decreasing output voltages, the converse occurs, the duty cycle of _LS- increases while the duty cycle of _LS+ decreases. The difference between the two pulse widths yields the differential output voltage.

Headphone Amplifier

The LM4949 headphone amplifier features three different operating modes, output capacitorless (OCL), capacitor-coupled (CC), and external amplifier mode.

The OCL architecture eliminates the bulky, expensive output coupling capacitors required by traditional headphone amplifiers. The LM4949 headphone section uses three amplifiers. Two amplifiers drive the headphones while the third (VOC) is set to the internally generated bias voltage (typically $V_{\rm DD}/2$). The third amplifier is connected to the return terminal of the headphone jack. In this configuration, the signal side of the headphones are biased to $V_{\rm DD}/2$, the return is biased to $V_{\rm DD}/2$, thus there is no net DC voltage across the headphone, eliminating the need for an output coupling capacitor. Re-

moving the output coupling capacitors from the headphone signal path reduces component count, reducing system cost and board space consumption, as well as improving low frequency performance.

In OCL mode, the headphone return sleeve is biased to $V_{DD}/2$. When driving headphones, the voltage on the return sleeve is not an issue. However, if the headphone output is used as a line out, the $V_{DD}/2$ can conflict with the GND potential that a line-in would expect on the return sleeve. When the return of the headphone jack is connected to GND, the VOC amplifier of the LM4949 detects an output short circuit condition and is disabled, preventing damage to the LM4949, and allowing the headphone return to be biased at GND.

Capacitor Coupled Headphone Mode

In capacitor coupled (CC) mode, the VOC pin is disabled, and the headphone outputs are coupled to the jack through series capacitors, allowing the headphone return to be connected to GND (Figure 4). In CC mode, the LM4949 requires output coupling capacitors to block the DC component of the amplifier output, preventing DC current from flowing to the load. The output capacitor and speaker impedance form a high pass filter with a -3dB roll-off determined by:

$$f_{-3dB} = 1 / 2\pi R_1 C_{OUT}$$

Where R_L is the headphone impedance, and C_{OUT} is the output coupling capacitor. Choose C_{OUT} such that f_{-3dB} is well below the lowest frequency of interest. Setting f_{-3dB} too high results in poor low frequency performance. Select capacitor dielectric types with low ESR to minimize signal loss due to capacitor series resistance and maximize power transfer to the load.

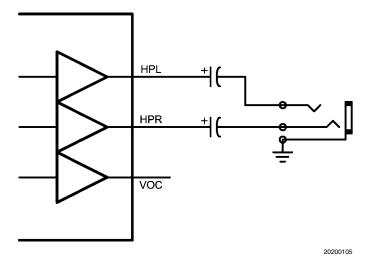


FIGURE 4. Capacitor Coupled Headphone Mode

External Headphone Amplifier

The LM4949 features the ability to drive and control a separate headphone amplifier for applications that require a True Ground headphone output (Figure 5). Configure the LM4949 into external headphone amplifier mode by setting bit D2 (OCL_LGC) in register 0.0 to 1 and bit D1 (OCL) to 0. In this mode the VOC output becomes a logic output used to drive the shutdown input of the external amplifier. The output level of VOC is controlled by bits D1 (SDB_HPSEL) and D2

(SDB_MUXSEL) in register 0.1. SDB_MUXSEL determines the source of the VOC control signal. With SDB_MUXSEL = 0, the VOC signal comes from the internal start-up circuitry of the LM4949. This allows the external headphone amplifier to be turned on and off simultaneously with the LM4949. When SDB_MUXSEL = 1, the VOC signal comes from the I²C bus, bit D1. With SDB_HPSEL = 0, VOC is a logic low, with SDB_HPSEL = 1, VOC is a logic high.

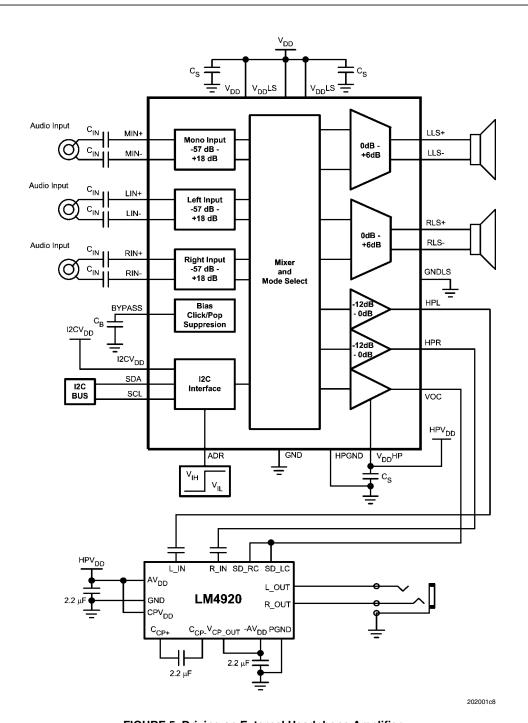


FIGURE 5. Driving an External Headphone Amplifier

Single-Ended Input

The left and right stereo inputs of the LM4949 can be configured for single-ended sources (Figure 6). In single-ended input mode, the LM4949 can accept up to 4 different single-ended audio sources. Set bits L1_INSEL = 1 and L2_INSEL

= 0 to use the RIN+ and LIN+ inputs. Set L1 _INSEL = 0 and L2_INSEL = 1 to use the RIN- and LIN- inputs. Set L1_INSEL = L2_INSEL = 1 to use both input pairs. Table 4 shows the single ended input combinations.

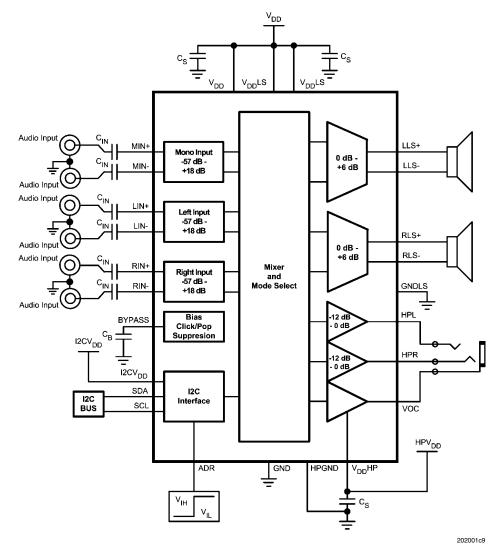


FIGURE 6. Single-Ended Input Configuration

TABLE 4. Single-Ended Stereo Input Modes

INPUT MODE	L1_INSEL	L2_INSEL	INPUT DESCRIPTION
0	0	0	Fully Differential Input Mode
1	0	1	Single-ended input. RIN- and LIN- selected
2	1	0	Single-ended input. RIN+ and LIN+ selected
3	1	1	Single-ended input. RIN+ mixed with RIN- and LIN+ mixed with LIN-

Input Mixer / Multiplexer

The LM4949 includes a comprehensive mixer/multiplexer controlled through the I2C interface. The mixer/multiplexer allows any input combination to appear on any output of the LM4949. Control bits LSR_SSEL and LSL_SSEL (loudspeakers), and HPR_SSEL and HPL_SSEL (headphones) select the individual stereo input channels; for example, LSR_SSEL = 1 outputs the right channel stereo input on the right channel loudspeaker, while LSL_SSEL = 1 outputs the left channel stereo input on the left channel loudspeaker. Control bits LSR_MSEL and LSL_MSEL (loudspeaker), and HPR_MSEL and HPR_LSEL (headphones) direct the mono input to the selected output. Setting HPR_MSEL = 1 outputs the mono input on the right channel headphone. Control bits LS_XSEL

(loudspeaker) and HP_XSEL (headphone) selects both stereo input channels and directs the signals to the opposite outputs, for example, LS_XSEL = 1 outputs the right channel stereo input on the left channel loudspeaker, while the left channel stereo input is output on the right channel loudspeaker. Setting __XSEL = selects both stereo inputs simultaneously, unlike the __SSEL controls which select the stereo input channels individually.

Multiple input paths can be selected simultaneously. Under these conditions, the selected inputs are mixed together and output on the selected channel. Tables 5 and 6 show how the input signals are mixed together for each possible input selection combination.

TABLE 5. Loudspeaker Multiplexer Control

LS MODE	LS_XSEL	LSR_MSEL/ LSL_MSEL	LSR_SSEL/ LSL_SSEL	LEFT CHANNEL OUTPUT	RIGHT CHANNEL OUTPUT
0		0	0	MUTE	MUTE
1	0	1	0	MONO	MONO
2	0	0	1	LEFT (DIFF)/ /LIN+/LIN-/ (LIN+ - LIN-)	RIGHT (DIFF)//RIN+/RIN-/ (RIN+ - RIN-)
3	0	1	1	MONO + LEFT (DIFF)/ /LIN+/ LIN-/ (LIN+ - LIN-)	MONO + RIGHT (DIFF)/ /RIN+/ RIN-/ (RIN+ - RIN-)
4	1	0	1	LEFT (DIFF)/ /LIN+/LIN-/ (LIN+ - LIN-) + RIGHT (DIFF)/ /RIN+/ RIN-/ (RIN+ - RIN-)	LEFT (DIFF)/ /LIN+/LIN-/ (LIN+ - LIN-) + RIGHT (DIFF)/ /RIN+/ RIN-/ (RIN+ - RIN-)
5	1	1	1	MONO + LEFT (DIFF)/ /LIN+/ LIN-/ (LIN+ - LIN-) + RIGHT (DIFF)/ /RIN+/RIN-/ (RIN+ - RIN-)	MONO + LEFT (DIFF)/ /LIN+/ LIN-/ (LIN+ - LIN-) + RIGHT (DIFF)/ /RIN+/RIN-/ (RIN+ - RIN-)

TABLE 6. Headphone Multiplexer Control

HP MODE	HP_XSEL	HPR_MSEL/	HPR_SSEL/	LEFT CHANNEL OUTPUT	RIGHT CHANNEL OUTPUT
		HPL_MSEL	LSL_SSEL		
0		0	0	MUTE	MUTE
1	0	1	0	MONO	MONO
2	0	0	1	LEFT (DIFF)//LIN+/LIN-/ (LIN	RIGHT (DIFF)//RIN+/RIN-/
				+ - LIN-)	(RIN+ - RIN-)
3	0	1	1	MONO + LEFT (DIFF)//LIN+/	MONO + RIGHT (DIFF)//RIN
				LIN-/ (LIN+ - LIN-)	+/RIN-/ (RIN+ - RIN-)
4	1	0	1	LEFT (DIFF)//LIN+/LIN-/ (LIN	LEFT (DIFF)//LIN+/LIN-/ (LIN
				+ - LIN-) + RIGHT (DIFF)//	+ - LIN-) + RIGHT (DIFF)//RIN
				RIN+/RIN-/ (RIN+ - RIN-)	+/RIN-/ (RIN+ - RIN-)
5	1	1	1	MONO + LEFT (DIFF)//LIN+/	MONO + LEFT (DIFF)/ /LIN+/
				LIN-/ (LIN+ - LIN-) + RIGHT	LIN-/ (LIN+ - LIN-) + RIGHT
				(DIFF)/ /RIN+/RIN-/ (RIN+ -	(DIFF)//RIN+/RIN-/(RIN+ -
				RIN-)	RIN-)

Power Supplies

The LM4949 uses different supplies for each portion of the device, allowing for the optimum combination of headroom, power dissipation and noise immunity. The speaker amplifier gain stage is powered from VDD, while the output stage is powered from VDDLS. The headphone amplifiers, input amplifiers and volume control stages are powered from VDDHP. The separate power supplies allow the speakers to operate from a higher voltage for maximum headroom, while the headphones operate from a lower voltage, improving power

dissipation. VDDHP may be driven by a linear regulator to further improve performance in noisy environments. The I²C portion if powered from I²CVDD, allowing the I²C portion of the LM4949 to interface with lower voltage digital controllers.

Shutdown Function

The LM4949 features five shutdown modes, configured through the l²C interface. Bit D0 (PWR_ON) in the Shutdown Control register shuts down/turns on the entire device. Set PWR_ON = 1 to enable the LM4949, set PWR_ON 0 to disable the device. Bits D0 - D3 in the Output On/Off Control

shutdown/turn on the individual channels. HPR_ON (D3) controls the right channel headphone output, HPL_ON (D2) controls the left channel headphone output, LSR_ON (D1) controls the right channel loudspeaker output, and LRL_ON (D0) controls the left channel loudspeaker output. The PWR_ON bit takes precedence over the individual channel controls.

Audio Amplifier Gain Setting

The each channel of the LM4949 has two separate gain stages. Each input stage features a 32 step volume control with a range of -57dB to +18dB (Table 7). Each speaker output stage has 4 gain settings (Table 8); 0dB, 2dB, 4dB, and 6dB when either a fully differential signal or two single ended signals are applied on the _IN+ and _IN- pins; and 6dB, 8dB,

10dB and 12dB in single-ended input mode with only one signal applied. The headphone gain is not affected by the input mode. Each headphone output stage has 3 gain settings (Table 9), 0dB, -6dB, and -12dB. This allows for a maximum separation of 24dB between the speaker and headphone outputs when both are active.

Calculate the total gain of a given signal path as follows:

$$A_{VOL} + A_{OS} = A_{TOTAL}$$

Where A_{VOL} is the volume control level, A_{OS} is the gain setting of the output stage, and A_{TOTAL} is the total gain for the signal path.

TABLE 7. 32 Step Volume Control

Volume Step	MG4/LG4/RG4	MG3/LG3/RG3	MG2/LG2/RG2	MG1/LG1/RG1	MG0/LG0/RG0	Gain (dB)
1	0	0	0	0	0	– 57
2	0	0	0	0	1	-49
3	0	0	0	1	0	-42
4	0	0	0	1	1	-34.5
5	0	0	1	0	0	-30.5
6	0	0	1	0	1	-27
7	0	0	1	1	0	-24
8	0	0	1	1	1	-21
9	0	1	0	0	0	-18
10	0	1	0	0	1	-15
11	0	1	0	1	0	-13.5
12	0	1	0	1	1	-12
13	0	1	1	0	0	-10.5
14	0	1	1	0	1	-9
15	0	1	1	1	0	-7.5
16	0	1	1	1	1	-6
17	1	0	0	0	0	-4.5
18	1	0	0	0	1	-3
19	1	0	0	1	0	-1.5
20	1	0	0	1	1	0
21	1	0	1	0	0	1.5
22	1	0	1	0	1	3
23	1	0	1	1	0	4.5
24	1	0	1	1	1	6
25	1	1	0	0	0	7.5
26	1	1	0	0	1	9
27	1	1	0	1	0	10.5
28	1	1	0	1	1	12
29	1	1	1	0	0	13.5
30	1	1	1	0	1	15
31	1	1	1	1	0	16.5
32	1	1	1	1	1	18

TABLE 8. Loudspeaker Gain Setting

LSRG1/LSLG1	LSRG0/LSLG0	Gain (dB)		
LSRG I/LSLG I	LSRG0/LSLG0	_IN+ _IN-	_IN+ =_IN-	
0	0	12	6	
0	1	10	4	
1	0	8	2	
1	1	6	0	

TABLE 9. Headphone Gain Setting

HPG1	HPG0	Gain (dB)
0	0	0
0	1	-6
1	0	-12

Differential Audio Amplifier Configuration

As logic supply voltages continue to shrink, system designers increasingly turn to differential signal handling to preserve signal to noise ratio with decreasing voltage swing. The LM4949 can be configured as a fully differential amplifier, amplifying the difference between the two inputs. The advantage of the differential architecture is any signal component that is common to both inputs is rejected, improving commonmode rejection (CMRR) and increasing the SNR of the amplifier by 6dB over a single-ended architecture. The improved CMRR and SNR of a differential amplifier reduce sensitivity to ground offset related noise injection, especially important in noisy applications such as cellular phones. Driving the LM4949 differentially also allows the inputs to be DC coupled. eliminating two external capacitors per channel. Set bits L1_INSEL and L2_INSEL = 0 for differential input mode. The left and right stereo inputs have selectable differential or single-ended input modes, while the mono input is always differential.

Single-Ended Audio Amplifier Configuration

In single-ended input mode, the audio sources must be capacitively coupled to the LM4949. With LIN+ LIN- and RIN+ RIN-, the loud speaker gain is 6dB more than in differential input mode, or when LIN+ = LIN- and RIN+ = RIN-. The headphone gain does not change. The mono input channel is not affected by L1_INSEL and L2_INSEL, and is always configured as a differential input.

Power Dissipation and Efficiency

The major benefit of Class D amplifiers is increased efficiency versus Class AB. The efficiency of the LM4949 speaker amplifiers is attributed to the output transistors' region of operation. The Class D output stage acts as current steering switches, consuming negligible amounts of power compared to their Class AB counterparts. Most of the power loss associated with the output stage is due to the IR loss of the MOSFET on-resistance, along with the switching losses due to gate charge.

The maximum power dissipation per headphone channel in Capacitor-Coupled mode is given by:

$$P_{DMAX} = V_{DD}^2 / 2\pi^2 R_L$$

In OCL mode, the maximum power dissipation per headphone channel increases due to the use of a third amplifier as a buffer. The power dissipation is given by:

$$P_{DMAX} = V_{DD}^2 / \pi^2 R_L$$

PROPER SELECTION OF EXTERNAL COMPONENTS

Audio Amplifier Power Supply Bypassing / Filtering

Proper power supply bypassing is critical for low noise performance and high PSRR. Place the supply bypass capacitors as close to the device as possible. Typical applications employ a voltage regulator with $10\mu F$ and $0.1\mu F$ bypass capacitors that increase supply stability. These capacitors do not eliminate the need for bypassing of the LM4949 supply pins. A $1\mu F$ ceramic capacitor placed close to each supply pin is recommended.

Bypass Capacitor Selection

The LM4949 generates a $V_{DD}/2$ common-mode bias voltage internally. The BYPASS capacitor, C_B , improves PSRR and THD+N by reducing noise at the BYPASS node. Use a $1\mu F$ capacitor, placed as close to the device as possible for C_B .

Audio Amplifier Input Capacitor Selection

Input capacitors, $C_{\rm IN}$, in conjunction with the input impedance of the LM4949 forms a high pass filter that removes the DC bias from an incoming signal. The AC-coupling capacitor allows the amplifier to bias the signal to an optimal DC level. Assuming zero source impedance, the -3dB point of the high pass filter is given by:

$$f_{-3dB} = 1 / 2\pi R_{IN}C_{IN}$$

Choose $C_{\rm IN}$ such that $f_{-3{\rm dB}}$ is well below the lowest frequency of interest. Setting $f_{-3{\rm dB}}$ too high affects the low-frequency response of the amplifier. Use capacitors with low voltage coefficient dielectrics, such as tantalum or aluminum electrolytic. Capacitors with high-voltage coefficients, such as ceramics, may result in increased distortion at low frequencies. Other factors to consider when designing the input filter include the constraints of the overall system. Although high fidelity audio requires a flat frequency response between 20Hz and 20kHz, portable devices such as cell phones may only concentrate on the frequency range of the spoken human voice (typically 300Hz to 4kHz). In addition, the physical size of the speakers used in such portable devices limits the low frequency response; in this case, frequencies below 150Hz may be filtered out.

PCB LAYOUT GUIDELINES

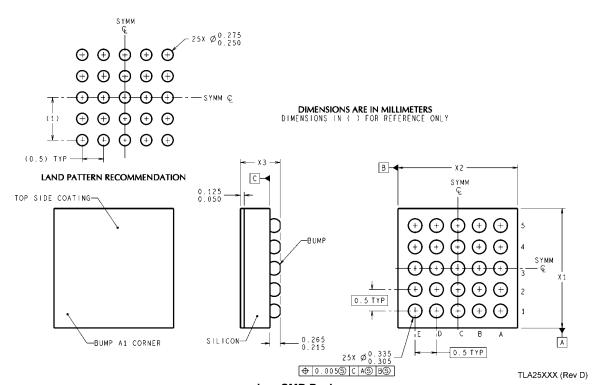
Minimize trace impedance of the power, ground and all output traces for optimum performance. Voltage loss due to trace resistance between the LM4949 and the load results in decreased output power and efficiency. Trace resistance between the power supply and GND of the LM4949 has the same effect as a poorly regulated supply, increased ripple and reduced peak output power. Use wide traces for power-supply inputs and amplifier outputs to minimize losses due to trace resistance, as well as route heat away from the device. Proper grounding improves audio performance, minimizes crosstalk between channels and prevents switching noise from interfering with the audio signal. Use of power and ground planes is recommended.

Place all digital components and digital signal traces as far as possible from analog components and traces. Do not run digital and analog traces in parallel on the same PCB layer.

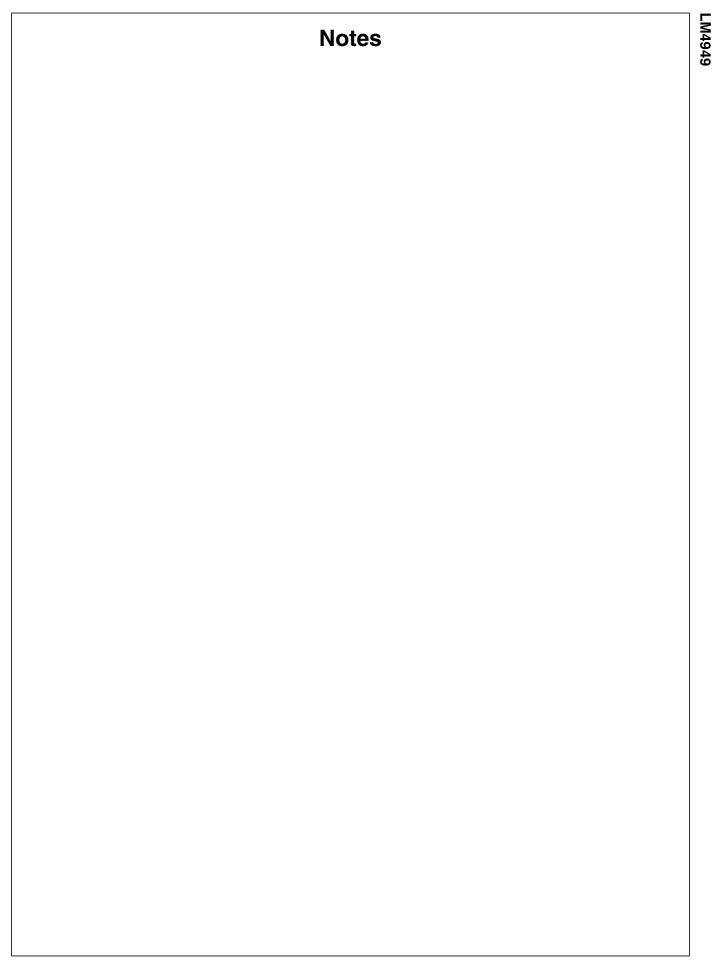
Revision History

Rev	Date	Description
1.0	09/06/06	Initial release.
1.1	09/27/06	Fixed some of the Typical Performance Curves.
1.2	01/17/07	Added the X1, X2, and X3 numerical values of the TLA25JJA mktg outline (back
		page).

Physical Dimensions inches (millimeters) unless otherwise noted



micro SMD Package Order Number LM4949TL NS Package Number TLA25JJA X1 = 2.722, X2 = 2.722, X3 = 0.600



Notes

THE CONTENTS OF THIS DOCUMENT ARE PROVIDED IN CONNECTION WITH NATIONAL SEMICONDUCTOR CORPORATION ("NATIONAL") PRODUCTS. NATIONAL MAKES NO REPRESENTATIONS OR WARRANTIES WITH RESPECT TO THE ACCURACY OR COMPLETENESS OF THE CONTENTS OF THIS PUBLICATION AND RESERVES THE RIGHT TO MAKE CHANGES TO SPECIFICATIONS AND PRODUCT DESCRIPTIONS AT ANY TIME WITHOUT NOTICE. NO LICENSE, WHETHER EXPRESS, IMPLIED, ARISING BY ESTOPPEL OR OTHERWISE, TO ANY INTELLECTUAL PROPERTY RIGHTS IS GRANTED BY THIS DOCUMENT.

TESTING AND OTHER QUALITY CONTROLS ARE USED TO THE EXTENT NATIONAL DEEMS NECESSARY TO SUPPORT NATIONAL'S PRODUCT WARRANTY. EXCEPT WHERE MANDATED BY GOVERNMENT REQUIREMENTS, TESTING OF ALL PARAMETERS OF EACH PRODUCT IS NOT NECESSARILY PERFORMED. NATIONAL ASSUMES NO LIABILITY FOR APPLICATIONS ASSISTANCE OR BUYER PRODUCT DESIGN. BUYERS ARE RESPONSIBLE FOR THEIR PRODUCTS AND APPLICATIONS USING NATIONAL COMPONENTS. PRIOR TO USING OR DISTRIBUTING ANY PRODUCTS THAT INCLUDE NATIONAL COMPONENTS, BUYERS SHOULD PROVIDE ADEQUATE DESIGN, TESTING AND OPERATING SAFEGUARDS.

EXCEPT AS PROVIDED IN NATIONAL'S TERMS AND CONDITIONS OF SALE FOR SUCH PRODUCTS, NATIONAL ASSUMES NO LIABILITY WHATSOEVER, AND NATIONAL DISCLAIMS ANY EXPRESS OR IMPLIED WARRANTY RELATING TO THE SALE AND/OR USE OF NATIONAL PRODUCTS INCLUDING LIABILITY OR WARRANTIES RELATING TO FITNESS FOR A PARTICULAR PURPOSE, MERCHANTABILITY, OR INFRINGEMENT OF ANY PATENT, COPYRIGHT OR OTHER INTELLECTUAL PROPERTY RIGHT.

LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS PRIOR WRITTEN APPROVAL OF THE CHIEF EXECUTIVE OFFICER AND GENERAL COUNSEL OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

Life support devices or systems are devices which (a) are intended for surgical implant into the body, or (b) support or sustain life and whose failure to perform when properly used in accordance with instructions for use provided in the labeling can be reasonably expected to result in a significant injury to the user. A critical component is any component in a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system or to affect its safety or effectiveness.

National Semiconductor and the National Semiconductor logo are registered trademarks of National Semiconductor Corporation. All other brand or product names may be trademarks or registered trademarks of their respective holders.

Copyright© 2007 National Semiconductor Corporation

For the most current product information visit us at www.national.com



National Semiconductor Americas Customer Support Center Email:

Email: new.feedback@nsc.com Tel: 1-800-272-9959 National Semiconductor Europe Customer Support Center Fax: +49 (0) 180-530-85-86 Email: europe.support@nsc.com Deutsch Tel: +49 (0) 69 9508 6208

English Tel: +49 (0) 870 24 0 2171

Français Tel: +33 (0) 1 41 91 8790

National Semiconductor Asia Pacific Customer Support Center Email: ap.support@nsc.com National Semiconductor Japan Customer Support Center Fax: 81-3-5639-7507 Email: jpn.feedback@nsc.com Tel: 81-3-5639-7560