January 2001

1.8V to 5.0V

11µA

14nA

200pA

+/-0.3mV

670ns (20mV overdrive)

0.1V beyond rails

# National Semiconductor

# LMV7251/LMV7255 **1.8V Low Voltage Comparator with Rail-to-Rail Input General Description**

The LMV7251/LMV7255 are rail-to-rail input low voltage comparators, which can operate at supply voltage range of 1.8V to 5.0V. The LMV7251/LMV7255 are available in space saving SC-70 or SOT23-5 packages. These comparators are ideal for low voltage and space critical designs.

The LMV7251 features a push-pull output stage. This feature allows operation with minimum power consumption when driving a load.

The LMV7255 features an open drain output. This allows the connection of an external resistor at the output. The output of the comparator can be used as a level shifter.

The IC's are built with National Semiconductor's advance Submicron Silicon-Gate BICMOS process. The LMV7251/LMV7255 have bipolar inputs for improved noise performance and CMOS outputs for better rail-to-rail output performance.

### **Features**

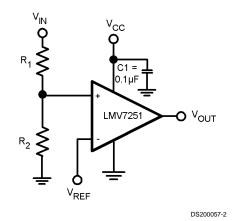
 $(V_S = 1.8V, T_A = 25^{\circ}C, Typical values unless specified).$ 

- Single or Dual Supplies
- Low supply voltage
- Ultra low supply current
- Low input bias current
- Low input offset current
- Low input offset voltage
- Response time
- Input common mode voltage

## Applications

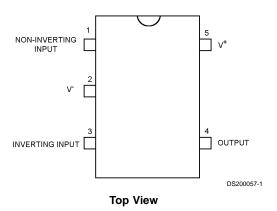
- Mobile communications
- Laptops and PDA's
- Battery powered electronics
- General purpose low voltage applications

# **Typical Circuit**



**FIGURE 1. Threshold Detector** 

# **Connection Diagram**



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# Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

1KV (Note 2)
200V (Note 6)
+/-Supply Voltage
5.5V
V <sup>+</sup> +0.1V, V <sup>-</sup> -0.1V
235°C

260°C –65°C to +150°C
+150°C
1.8V to 5.0V
–40°C to +85°C
325°C/W 265°C/W

# **1.8V Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 1.8V$ ,  $V^- = 0V$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Typ (Note 4)	Limits (Note 5)	Units
Vos	Input Offset Voltage		0.3	6	mV
				8	max
TC $V_{OS}$	Input Offset Average Drift	$V_{CM} = 0.9V$ (Note 7)	10		uV/C
I <sub>B</sub>	Input Bias Current		14		nA
l <sub>os</sub>	Input Offset Current		200		pА
I <sub>S</sub>	Supply Current		11	15 <b>17</b>	μA max
I <sub>SC</sub>	Output Short Circuit Current	utput Short Circuit Current Sourcing, V <sub>O</sub> = 0.9V 8 (LMV7251 only)	4	mA	
		Sinking, $V_O = 0.9V$	11.6	5	min
I <sub>LEAKAGE</sub>	Output Leakage Current	V <sub>O</sub> = 1.8V (LMV7255 only)	300		рА
V <sub>OH</sub>	Output Voltage High	I <sub>O</sub> = 1.5mA (LMV7251 only)	1.72	1.675	V min
V <sub>OL</sub>	Output Voltage Low	$I_{O} = -1.5 mA$	65	125	mV max
V <sub>CM</sub>	Input Common Voltage Range	CMRR > 45 dB		1.9	V max
				-0.1	V min
CMRR	Common Mode Rejection Ratio	0 < V <sub>CM</sub> < 1.8V	72	47	dB min
PSRR	Power Supply Rejection Ratio	V <sup>+</sup> = 1.8V to 5V	79	55	dB min

## **1.8V AC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 1.8V$ ,  $V^- = 0V$ ,  $V_{CM} = 0.5V$ ,  $V_O = V^+/2$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	<b>Typ</b> (Note 5)	Limits (Note 6)	Units
t <sub>PHL</sub>	Propagation Delay (High to Low)	Input Overdrive = 20mV Load = 50pF//5kΩ	720		ns
		Input Overdrive = 50mV Load = 50pF//5kΩ	380		ns

# 1.8V AC Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 1.8V$ ,  $V^- = 0V$ ,  $V_{CM} = 0.5V$ ,  $V_O = V^+/2$ . Boldface limits apply at the temperature extremes.

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
t <sub>PLH</sub>	Propagation Delay (Low to High)	Input Overdrive = 20mV Load = 50pF//5kΩ	670		ns
		Input Overdrive = 50mV Load = 50pF//5kΩ	400		ns

# **2.7V Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 2.7V$ ,  $V^- = 0V$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Limits (Note 6)	Units
Vos	Input Offset Voltage		0.03	6	mV
				8	max
TC V <sub>OS</sub>	Input Offset Average Drift	V <sub>CM</sub> = 1.35V (Note 7)	10		μV/C
I <sub>B</sub>	Input Bias Current		15		nA
l <sub>os</sub>	Input offset Current		210		pА
I <sub>S</sub>	Supply Current		11	18	μA
				22	max
I <sub>SC</sub>	Output Short Circuit Current	Sourcing, V <sub>O</sub> = 1.35V (LMV7251 only)	28	15	mA
		Sinking, V <sub>O</sub> = 1.35V	28	15	
I <sub>LEAKAGE</sub>	Output Leakage Current	V <sub>O</sub> = 2.7V, (LMV7255 only)	320		рА
V <sub>OH</sub>	Output Voltage High	I <sub>O</sub> = 2mA (LMV7251 only)	2.63	2.575	V min
V <sub>OL</sub>	Output Voltage Low	$I_{O} = -2mA$	61	125	mV max
V <sub>CM</sub>	Input Common Voltage Range	CMRR > 45dB		2.8	V
					max
				-0.1	V
					min
CMRR	Common Mode Rejection Ratio	$0 < V_{CM} < 2.7V$	75	46	dB
					min
PSRR	Power Supply Rejection Ratio	V <sup>+</sup> = 1.8V to 5V	79	55	dB
					min

# 2.7V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 2.7V$ ,  $V^- = 0V$ .**Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
t <sub>PHL</sub>	Propagation Delay (High to Low)	Input Overdrive = $20mV$ Load = $50pF//5k\Omega$	830		ns
		Input Overdrive = 50mV Load = 50pF//5kΩ	430		ns
t <sub>PLH</sub>	Propagation Delay (Low to High)	Input Overdrive = $20mV$ Load = $50pF//5k\Omega$	730		ns
		Input Overdrive = $50mV$ Load = $50pF//5k\Omega$	410		ns

# **5V Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 5V$ ,  $V^- = 0V$ . Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Limits (Note 6)	Units
V <sub>os</sub>	Input Offset Voltage		0.03	6	mV
				8	max
TC V <sub>os</sub>	Input Offset Average Drift	V <sub>CM</sub> = 2.5V (Note 7)	10		μV/C
I <sub>B</sub>	Input Bias Current		16		nA
l <sub>os</sub>	Input Offset Current		220		pА
I <sub>S</sub>	Supply Current		12	20	μA
				25	max
I <sub>sc</sub>	Output Short Circuit Current	Sourcing, $V_0 = 2.5V$	82	50	
		(LMV7251 only)			mA
		Sinking, $V_{O} = 2.5V$	78	50	min
ILEAKAGE	Output Leakage Current	$V_{O} = 5V,$	375		pА
		(LMV7255 only)			
V <sub>OH</sub>	Output Voltage High	$I_{O} = 4mA$	4.9	4.82	V
					min
V <sub>OL</sub>	Output Voltage Low	$I_{O} = -4mA$	90	180	mV
					max

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics. Note 2: Human body model,  $1.5k\Omega$  in series with 100pF.

Note 3: The maximum power dissipation is a function of  $T_{J(max)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(max)} - T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board.

Note 4: Typical values represent the most likely parametric norm.

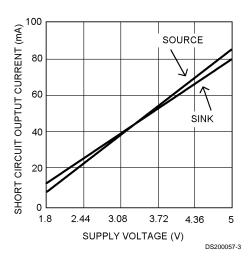
Note 5: All limits are guaranteed by testing or statistical analysis.

Note 6: Machine Model,  $0\Omega$  in series with 200pF.

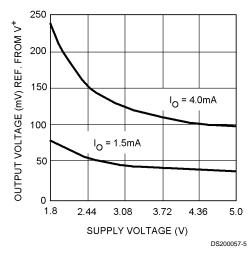
Note 7: Offset Voltage average drift determined by dividing the change in VOS at temperature extremes into the total temperature change.

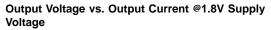
# **Typical Performance Characteristics** (T<sub>A</sub> = 25°C, Unless otherwise specified).

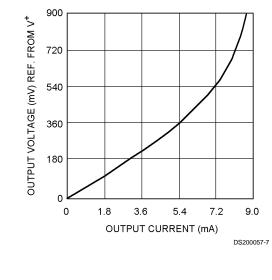
### Short Circuit Current vs. Supply Voltage



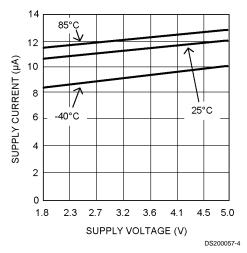
### Output Voltage vs. Supply Voltage

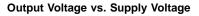


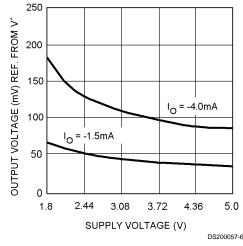


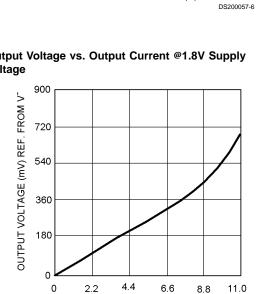


### Supply Current vs. Supply Voltage









OUTPUT CURRENT (mA)

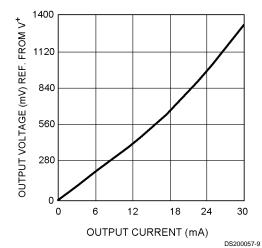
Output Voltage vs. Output Current @1.8V Supply Voltage

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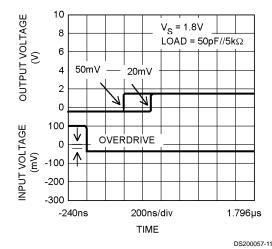
# LMV7251/LMV7255

# **Typical Performance Characteristics** ( $T_A = 25^{\circ}C$ , Unless otherwise specified). (Continued)

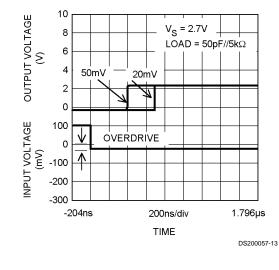
### Output Voltage vs. Output Current @2.7V Supply Voltage

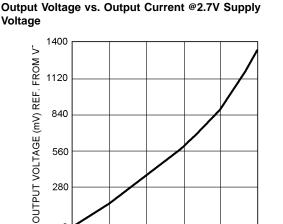






Response Time for Various Input Overdrives - t<sub>PLH</sub>





Response Time for Various Input Overdrives - t<sub>PHL</sub>

12

OUTPUT CURRENT (mA)

18

24

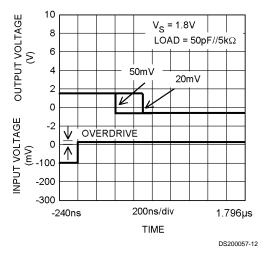
30

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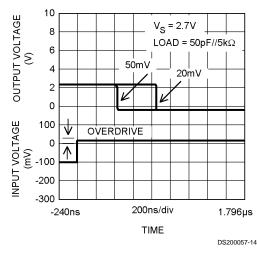
6

280

0 0



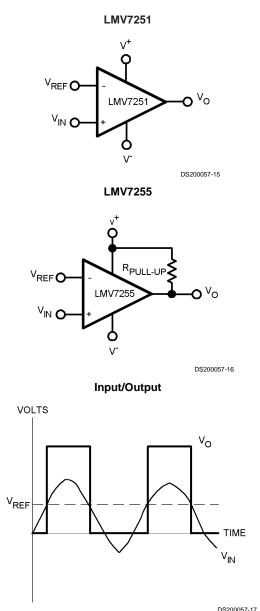




# **Application Info**

### **Basic Comparators**

A comparator is quite often used to convert an analog signal to a digital signal. The comparator compares an input voltage (V<sub>IN</sub>) at the non-inverting pin to the reference voltage (V<sub>REF</sub>) at the inverting pin. If V<sub>IN</sub> is less than V<sub>REF</sub> the output (V<sub>O</sub>) is low (V<sub>OL</sub>). However, if V<sub>IN</sub> is greater than V<sub>REF</sub>, the output voltage (V<sub>O</sub>) is high (V<sub>OH</sub>).



### FIGURE 2. Basic Comparator

### Hysteresis

The basic comparator configuration may oscillate or produce a noisy output if the applied differential input is near the comparator's input offset voltage. This tends to occur when the voltage on the input is equal or very close to the other input voltage. Adding hysteresis can prevent this problem. Hysteresis creates two switching thresholds (one for the rising input voltage and the other for the falling input voltage). Hysteresis is the voltage difference between the two switching thresholds. When both inputs are nearly equal, hysteresis causes one input to effectively move quickly pass the other. Thus, effectively moving the input out of region that oscillation may occur.

Hysteresis can easily be added to a comparator in a non-inverting configuration with two resistors and positive feedback *Figure 3*. The output will switch from low to high when  $V_{IN}$  rises up to  $V_{IN1}$ , where  $V_{IN1}$  is calculated by

$$V_{IN1} = (V_{REF} (R1 + R2))/R2$$

The output will switch from high to low when  $V_{\rm IN}$  falls to  $V_{\rm IN2},$  where  $V_{\rm IN2}$  is calculated by

$$V_{IN2} = (V_{REF} (R1 + R2) - V_{CC} R1) / R2$$

The Hysteresis is the difference between V<sub>IN1</sub> and V<sub>IN2</sub>.  $\Delta V_{IN} = V_{IN1} - V_{IN2} = ((V_{REF} (R1 + R2)) / R2) - ((V_{REF} (R1 + R2)) - V_{CC} R1) / R2) = V_{CC} R1 / R2.$ 

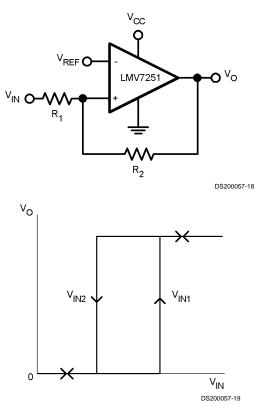


FIGURE 3. Non-Inverting Comparator Configuration — LMV7251

For an inverting configured comparator, hysteresis can be added with a three resistor network and positive feedback. When input voltage  $(V_{IN})$  at the inverting node is less than non-inverting node  $(V_T)$ , the output is high. The equivalent circuit for the three resistor network is R1 in parallel with R3 and in series with R2. The lower threshold voltage  $V_{T1}$  is calculated by:

$$V_{T1} = ((V_{CC} R2) / ((R1 R3) / (R1+R3)) + R2$$

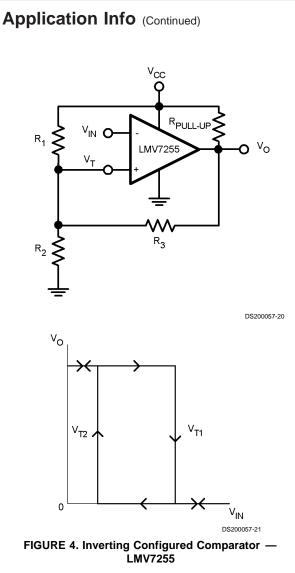
When  $V_{IN}$  is greater than  $V_T$ , the output voltage is low. The equivalent circuit for the three resistor network is R2 in parallel with R3 and in series with R1. The upper threshold voltage  $V_{T2}$  is calculated by:

 $V_{T2} = V_{CC} ((R2 R3) / (R2 + R3)) / ((R1 + ((R2 R3) / (R2 + R3))))$ 

### The hysteresis is defined as

 $\begin{array}{l} \Delta V_{\mathsf{IN}} = V_{\mathsf{T1}} - V_{\mathsf{T2}} = ((V_{\mathsf{CC}} \; \mathsf{R2}) \, / \, ((\mathsf{R1} \; \mathsf{R3}) \, / \, (\mathsf{R1} + \mathsf{R3})) + \mathsf{R2}) - \\ (V_{\mathsf{CC}} \; ((\mathsf{R2} \; \mathsf{R3}) \, / \, (\mathsf{R2} + \mathsf{R3})) \, / \; ((\mathsf{R1} + ((\mathsf{R2} \; \mathsf{R3}) \, / \, (\mathsf{R2} + \mathsf{R3}))) \end{array}$ 

# LMV7251/LMV7255



Power supply decoupling is critical and improves stability. Place decoupling capacitors  $0.1\mu$ F as close as possible to the V<sup>+</sup> pin. For split supply applications, place decoupling capacitors  $0.1\mu$ F on both the V<sup>+</sup> and V<sup>-</sup> pins. The decoupling capacitors will help keep the comparator from oscillating under various load conditions.

### Input Stage

The LMV7251 and LMV7255 have rail-to-rail input stages. The input common mode voltage range is from –100mV to (V\_{CC} + 100mV).

### Output Stage

The LMV7251 has a push-pull CMOS output stage. Large push-pull output drivers allows rail-to-rail output swings with load currents in the miliampere range.

The LMV7255 has a open drain CMOS output stage. This requires an external pull-up resistor connected between the positive supply voltage and the output. The external pull-up resistor should be high enough resistance so to avoid excessive power dissipation. In addition, the pull-up resistor should be low enough resistance to enable the comparator to switch with the load circuitry connected.

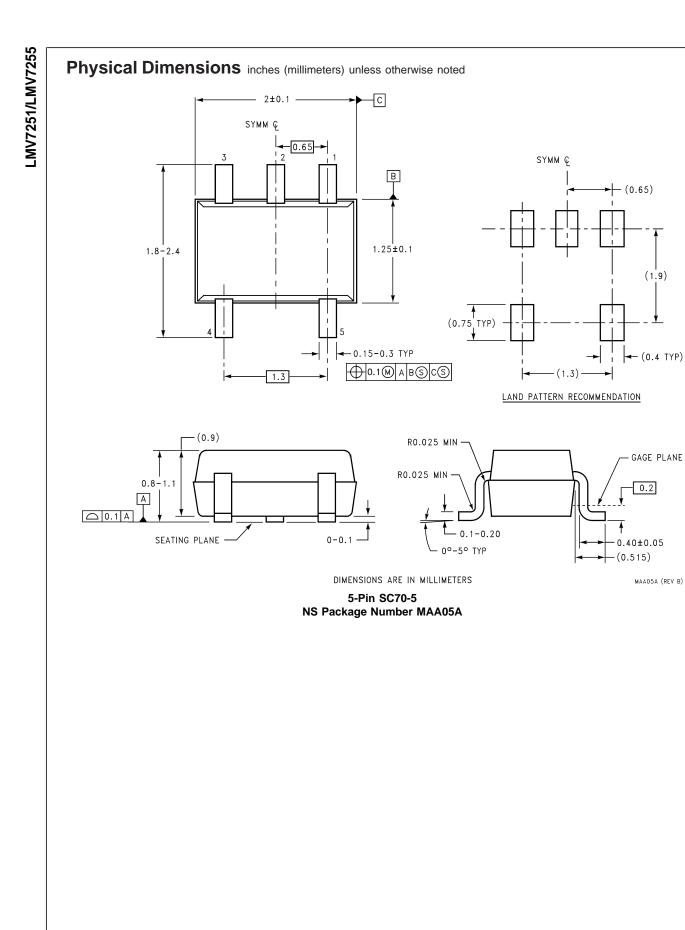
### **Power Supply Considerations**

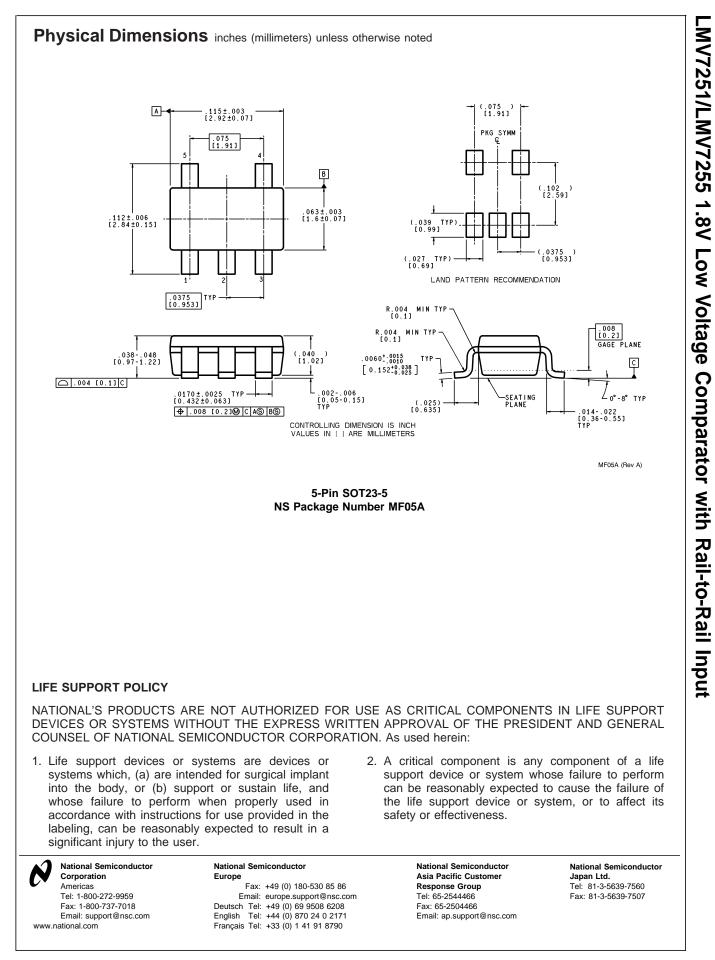
The LMV7251/LMV7255 are well suited for many batterypowered applications. The LMV7251/LMV7255 can operate from single power supply of +1.8V to +5V. The device typically consumes only 11µA with a 2.7V supply. With a high power supply rejection ratio (PSRR) of 79 dB (typical), the comparator is well suited for operating under conditions of a decaying battery voltage.

Ordering Information						
Package	Ordering Info	Pkg Marking	Supplied As	NSC Drawing		
5-Pin SOT23-5	LMV7251M5	C16A	1k Units Tape and Reel	MF05A		
	LMV7251M5X	C16A	3k units Tape and Reel			
	LMV7255M5	C18A	1k Units Tape and Reel			
	LMV7255M5X	C18A	3k units Tape and Reel			
5-Pin SC-70	LMV7251M7	C17	1k Units Tape and Reel	MAA05A		
	LMV7251M7X	C17	3k units Tape and Reel			
	LMV7255M7	C19	1k Units Tape and Reel			
	LMV7255M7X	C19	3k units Tape and Reel			

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LMV7251/LMV7255





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