

General Description

The MAX4310-MAX4315 single-supply mux-amps combine high-speed operation, low-glitch switching, and excellent video specifications. The six products in this family are differentiated by the number of multiplexer inputs and the gain configuration. The MAX4310/ MAX4311/MAX4312 integrate 2-/4-/8-channel multiplexers, respectively, with an adjustable gain amplifier optimized for unity-gain stability. The MAX4313/MAX4314/ MAX4315 integrate 2-/4-/8-channel multiplexers, respectively, with a +2V/V fixed-gain amplifier. All devices have 40ns channel switching time and low 10mVp-p switching transients, making them ideal for video-switching applications. They operate from a single +4V to +10.5V supply, or from dual supplies of $\pm 2V$ to $\pm 5.25V$, and they feature rail-to-rail outputs and an input common-mode voltage range that extends to the negative supply rail.

The MAX4310/MAX4311/MAX4312 have a -3dB bandwidth of 280MHz/345MHz/265MHz and up to a 460V/µs slew rate. The MAX4313/MAX4314/MAX4315, with 150MHz/127MHz/97MHz -3dB bandwidths up to a 540V/µs slew rate, and a fixed gain of +2V/V, are ideally suited for driving back-terminated cables. Quiescent supply current is as low as 6.1mA, while low-power shutdown mode reduces supply current to as low as 560µA and places the outputs in a high-impedance state. The MAX4310-MAX4315's internal amplifiers maintain an open-loop output impedance of only 8Ω over the full output voltage range, minimizing the gain error and bandwidth changes under loads typical of most rail-to-rail amplifiers. With differential gain and phase errors of 0.06% and 0.08°, respectively, these devices are ideal for broadcast video applications.

Video Signal Multiplexing Video Crosspoint Switching Flash ADC Input Buffers 75Ω Video Cable Drivers **High-Speed Signal Processing**

Applications

Broadcast Video Medical Imaging Multimedia Products

Features

MAX4310-MAX4315

- Single-Supply Operation Down to +4V
- 345MHz -3dB Bandwidth (MAX4311) 150MHz -3dB Bandwidth (MAX4313)
- 540V/µs Slew Rate (MAX4313)
- Low 6.1mA Quiescent Supply Current
- 40ns Channel Switching Time
- Ultra-Low 10mVp-p Switching Transient
- ♦ 0.06%/0.08° Differential Gain/Phase Error
- Rail-to-Rail Outputs: Drives 150Ω to within 730mV of the Rails
- Input Common-Mode Range Includes **Negative Rail**
- Low-Power Shutdown Mode
- Available in Space-Saving 8-Pin µMAX[®] and **16-Pin QSOP Packages**

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX4310EUA	-40°C to +85°C	8 µMAX
MAX4310ESA	-40°C to +85°C	8 SO
MAX4311EEE	-40°C to +85°C	16 QSOP
MAX4311ESD	-40°C to +85°C	14 Narrow SO
MAX4312EEE	-40°C to +85°C	16 QSOP
MAX4312ESE	-40°C to +85°C	16 Narrow SO
MAX4313EUA	-40°C to +85°C	8 µMAX
MAX4313ESA	-40°C to +85°C	8 SO
MAX4314EEE	-40°C to +85°C	16 QSOP
MAX4314ESD	-40°C to +85°C	14 Narrow SO
MAX4315EEE	-40°C to +85°C	16 QSOP
MAX4315ESE	-40°C to +85°C	16 Narrow SO

Pin Configurations and Typical Operating Circuit appear at end of data sheet.

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Selector Guide

PART	NO. OF INPUT CHANNELS	AMPLIFIER GAIN (V/V)	PIN-PACKAGE
MAX4310	2	≥ + 1	8-Pin SO/µMAX
MAX4311	4	≥ + 1	14-Pin Narrow SO, 16-Pin QSOP
MAX4312	8	≥ + 1	16-Pin Narrow SO/QSOP
MAX4313	2	+2	8-Pin SO/µMAX
MAX4314	4	+2	14-Pin Narrow SO, 16-Pin QSOP
MAX4315	8	+2	16-Pin Narrow SO/QSOP

Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V _{CC} to V _{EE})	V
Input Voltage(VEE - 0.3V) to (VCC + 0.3V)	/)
All Other Pins(VEE - 0.3V) to (VCC + 0.3V)	√)
Output Current±120m/	A
Short-Circuit Duration (VOUT to GND, VCC or VEE)Continuou	S
Continuous Power Dissipation ($T_A = +70^{\circ}C$)	
8-Pin SO (derate 5.9mW/°C above +70°C)471mV	Ν
8-Pin µMAX (derate 4.1mW/°C above +70°C)	W

14-Pin SO (derate 8.3mW/°C above +70°C)	667mW
16-Pin SO (derate 8.7mW/°C above +70°C)	696mW
16-Pin QSOP (derate 8.3mW/°C above +70°C)	667mW
Operating Temperature Range4	40°C to +85°C
Storage Temperature Range68	5°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +5V, V_{EE} = 0V, \overline{SHDN} \ge 4V, R_L = \infty, V_{OUT} = 2.5V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$

PARAMETER	SYMBOL	CON	DITIONS	MIN	ТҮР	МАХ	UNITS		
Operating Supply Voltage Range	V _{CC}	Inferred from PSRR test				10.5	V		
		MAX4310/MAX4311/M CMRR test	IAX4312, inferred from	0.035	Vc	C - 2.8	V		
Input Voltage Range		MAX4313/MAX4314/M output voltage swing	IAX4315, inferred from	0.035	Vc	CC - 2.7	V		
Common-Mode Rejection Ratio	CMRR	$0 \le V_{CM} \le 2.2V$, MAX only	4310/MAX4311/MAX4312	73	95		dB		
Input Offset Voltage	V _{OS}				±5.0	±20	mV		
Input Offset Voltage Drift	TCVOS				±7		μV/°C		
Input Offset Voltage Matching				±1		mV			
Input Bias Current	Ι _Β	lin			7	14	μΑ		
Feedback Bias Current	I _{FB}	IFB, MAX4310/MAX43	11/MAX4312 only		7	14	μA		
Input Offset Current	los	MAX4310/MAX4311/N	IAX4312 only		0.1	2	μA		
Common-Mode Input Resistance	R _{IN}	V _{IN} varied over V _{CM} , I MAX4312 only	MAX4310/MAX4311/		3		MΩ		
Differential Input Resistance	R _{IN}				70		KΩ		
		MAX4310/MAX4311/	Open loop		8				
Output Resistance	R _{OUT}	MAX4312 only	Closed loop, $A_V = +1V/V$	0.025			Ω		
		MAX4313/MAX4314/MAX4315			0.025				
Disabled Output Resistance	Rout	MAX4310/MAX4311/MAX4312, open loop		MAX4310/MAX4311/MAX4312, open loop			35		Ω
	1001	MAX4313/MAX4314/N		1		22			
Open-Loop Gain	Avol	MAX4310/MAX4311/M $R_L = 150\Omega$ to GND, 0.	50	59		dB			
Voltage Gain	Avcl	MAX4313/MAX4314/M R _L = 150Ω to GND, 0.)	1.9	2.0	2.1	V/V		

DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = +5V, V_{EE} = 0V, \overline{SHDN} \ge 4V, R_L = \infty, V_{OUT} = 2.5V, T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

PARAMETER	SYMBOL	CONI	MIN	ТҮР	МАХ	UNITS	
		D: 1500	VCC - VOH		0.73	0.9	
Output Voltage Swing	Vour	$R_L = 150\Omega$	Vol - Vee		0.03	0.06	V
Output voltage Swing	Vout	$R_{l} = 10k\Omega$	V _{CC} - V _{OH}		0.25	0.4	v
		$H_{L} = 10ksz$	V _{OL} - V _{EE}		0.04	0.07	
Output Current	IOUT	$R_L = 30\Omega$		±75	±95		mA
Power-Supply Rejection Ratio	PSRR	$V_{CC} = 4.0V$ to 10.5V	52	63		dB	
		MAX4310/MAX4313			6.1	7.8	
Quiescent Supply Current	Icc	MAX4311/MAX4314			6.9	8.8	mA
		MAX4312/MAX4315			7.4	9.4	
Shutdown Supply Current		$\overline{SHDN} \le V_{IL}$			560	750	μA
LOGIC CHARACTERISTICS	SHDN, A0, A	1, A2)					_
Logic-Low Threshold	VIL					V _{EE} + 1	V
Logic-High Threshold	VIH			V _{CC} - 1			V
Logic-Low Input Current	١ _١ ٢	$V_{IL} \le V_{EE} + 1V$	-500	-320		μA	
Logic-High Input Current	IIН	$V_{IH} \ge V_{CC} - 1V$		0.3	5	μA	

AC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +5V, V_{EE} = 0V, \overline{SHDN} \ge 4V, R_L = 150\Omega, V_{CM} = 1.5V, A_{VCL} = +1V/V (MAX4310/MAX4311/MAX4312), A_{VCL} = +2V/V (MAX4313/MAX4314/MAX4315), T_A = +25°C, unless otherwise noted.)$

PARAMETER	SYMBOL	CONDITIC	ONS	MIN	ТҮР	MAX	UNITS
			MAX4310		280		
			MAX4311		345		
-3dB Bandwidth		\/100ms\/m_m	MAX4312		265		
-30B Bandwidth	BW(-3dB)	$V_{OUT} = 100 mVp-p$	MAX4313		150		MHz
			MAX4314		127		
			MAX4315		97		
			MAX4310		60		
			MAX4311		40		1
0.1 dD Dandwidth		\/100ms\/m_m	MAX4312		35		MHz
-0.1dB Bandwidth	BW(-0.1dB)	$V_{OUT} = 100 mVp-p$	MAX4313		40		IVIHZ
			MAX4314		78]
			MAX4315		46		

MAX4310-MAX4315

AC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = +5V, V_{EE} = 0V, \overline{SHDN} \ge 4V, R_L = 150\Omega, V_{CM} = 1.5V, A_{VCL} = +1V/V (MAX4310/MAX4311/MAX4312), A_{VCL} = +2V/V (MAX4313/MAX4314/MAX4315), T_A = +25^{\circ}C, unless otherwise noted.)$

PARAMETER	SYMBOL		CONDITIONS		MIN	ТҮР	MAX	UNITS
			MAX4310			110		
			MAX4311	MAX4311		100		
	50014		MAX4312			80		
Full-Power Bandwidth	FPBW	V _{OUT} = 2Vp-p	MAX4313			40		MHz
			MAX4314		90			
			MAX4315			70		
			MAX4310			460		
			MAX4311			430		
Olaw Data	00		MAX4312			345		
Slew Rate	SR	V _{OUT} = 2Vp-p	MAX4313			540		
			MAX4314			430		
			MAX4315			310		
Cattling Time to 0.1%			MAX4310/MAX	(4311/MAX4312		42		
Settling Time to 0.1%	ts	V _{OUT} = 2Vp-p	MAX4313/MAX	4314/MAX4315		25		ns
Gain Matching		Matching betwee bandwidth	en channels over	-3dB		0.05		dB
Differential Gain Error	DG	$A_{VCL} = +1V/V,$ $R_{L} = 150\Omega \text{ to}$ $V_{CC}/2$	MAX4310/MAX4311/ MAX4312			0.06		%
		RL = 150Ω to MAX4313/MAX4314/ V _{CC} /2 MAX4315		X4314/		0.09		
Differential Phase Error	DG	$A_{VCL} = +1V/V,$ $R_{L} = 150\Omega \text{ to}$ $V_{CC}/2$	MAX4310/MA MAX4312	X4311/		0.08		degrees
		$R_L = 150\Omega$ to $V_{CC}/2$	MAX4313/MA MAX4315	X4314/		0.03		
			MAX4310/	f = 3kHz		-89		
			MAX4311/	f = 2kHz		-80		
Spurious-Free Dynamic	SEDR		MAX4312	f = 20kHz		-47		-10-
Range	SFDR	V _{OUT} = 2Vp-p	MAX4313/	f = 3kHz		-95		dBc
			MAX4314/	f = 2kHz		-72		1
			MAX4315	f = 20kHz		-47		
0 III - D' - ''		f = 1MHz,	MAX4310/MAX4311/MAX4312			-85		15
Second Harmonic Distortion		V _{OUT} = 2Vp-p	MAX4313/MAX4314/MAX4315			-76		dBc
		f = 1MHz,	MAX4310/MAX	4311/MAX4312		-88		
Third Harmonic Distortion		$V_{OUT} = 2Vp-p$	MAX4313/MAX4314/MAX4315			-95		dBc
		f = 1MHz,	MAX4310/MAX	4311/MAX4312		-83		
Total Harmonic Distortion	THD	$V_{OUT} = 2V_{P-P}$	-	4314/MAX4315		-76		dBc

M/X/M

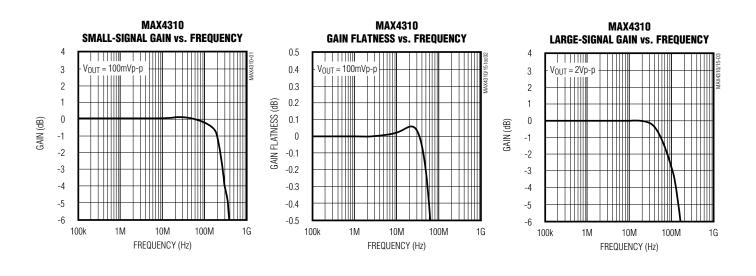
AC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = +5V, V_{EE} = 0V, \overline{SHDN} \ge 4V, R_L = 150\Omega, V_{CM} = 1.5V, A_{VCL} = +1V/V$ (MAX4310/MAX4311/MAX4312), A_{VCL} = +2V/V (MAX4313/MAX4314/MAX4315), T_A = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	(MIN	ТҮР	MAX	UNITS	
			MAX4310/MAX4313		-95		
All-Hostile Crosstalk		f = 10MHz, V _{IN} = 2Vp-p	MAX4311/MAX4314		-60		dB
		v IV – 5 v b-b	MAX4312MAX4315		-52		
Off-Isolation		$\overline{\text{SHDN}} = 0, f = 10$	MHz, V _{IN} = 2Vp-p		-82		dB
Output Impedance	Zout	f = 10MHz			3		Ω
Input Capacitance	CIN	Channel on or off			2		рF
Input Voltage-Noise Density	en	f = 10kHz			14		nV/√Hz
Input Current-Noise Density	i _n	f = 10 kHz			1.3		pA/√Hz
SWITCHING CHARACTERIS	TICS						
Channel Switching Time	tsw				40		ns
Enable Time from Shutdown	ton			50		ns	
Disable Time to Shutdown	^t OFF			120		ns	
Switching Transient				10		mVp-p	

Typical Operating Characteristics

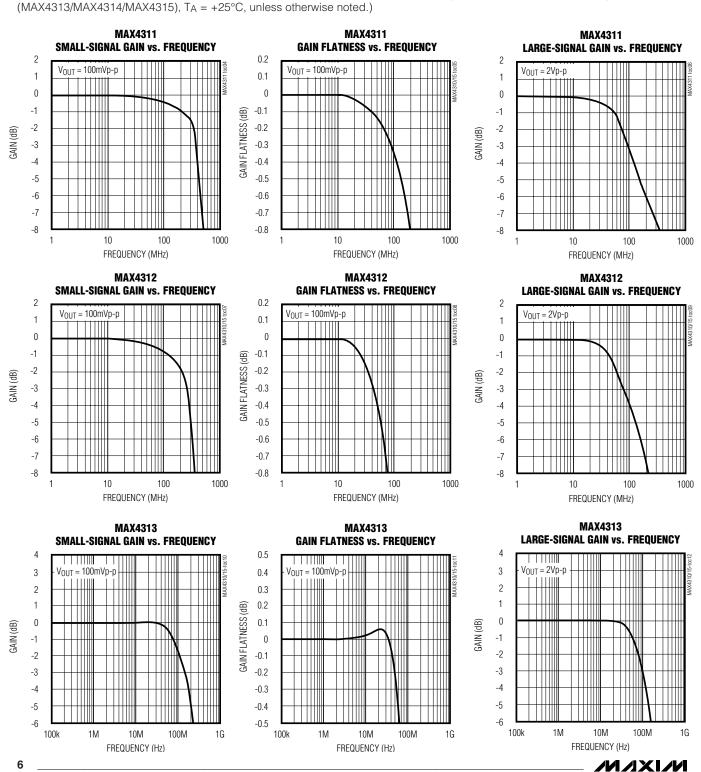
 $(V_{CC} = +5V, V_{EE} = 0V, \overline{SHDN} \ge 4V, R_L = 150\Omega$ to $V_{CC}/2, V_{CM} = 1.5V, A_{VCL} = +1V/V$ (MAX4310/MAX4311/MAX4312), $A_{VCL} = +2V/V$ (MAX4313/MAX4314/MAX4315), $T_A = +25^{\circ}C$, unless otherwise noted.)



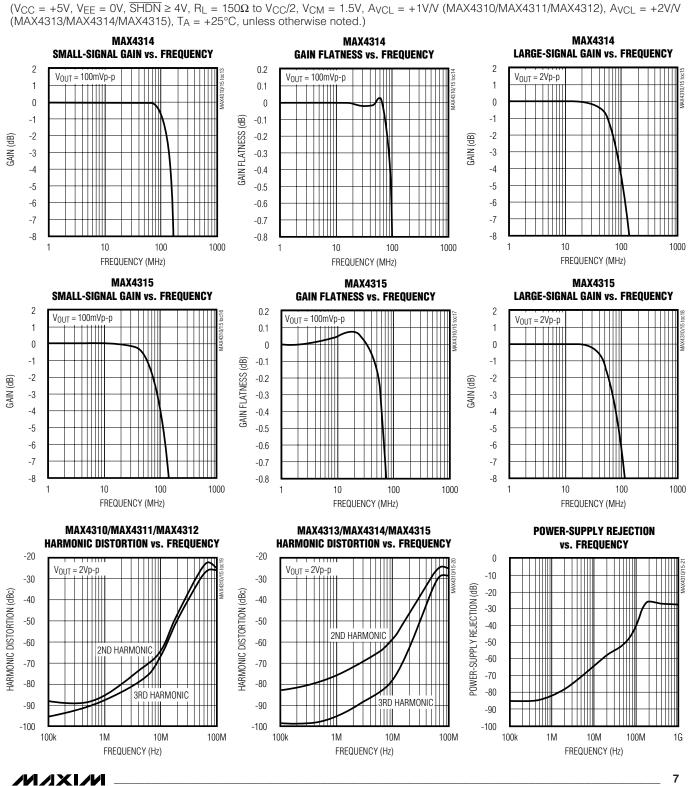
(V_{CC} = +5V, V_{EE} = 0V, SHDN ≥ 4V, R_L = 150Ω to V_{CC}/2, V_{CM} = 1.5V, A_{VCL} = +1V/V (MAX4310/MAX4311/MAX4312), A_{VCL} = +2V/V

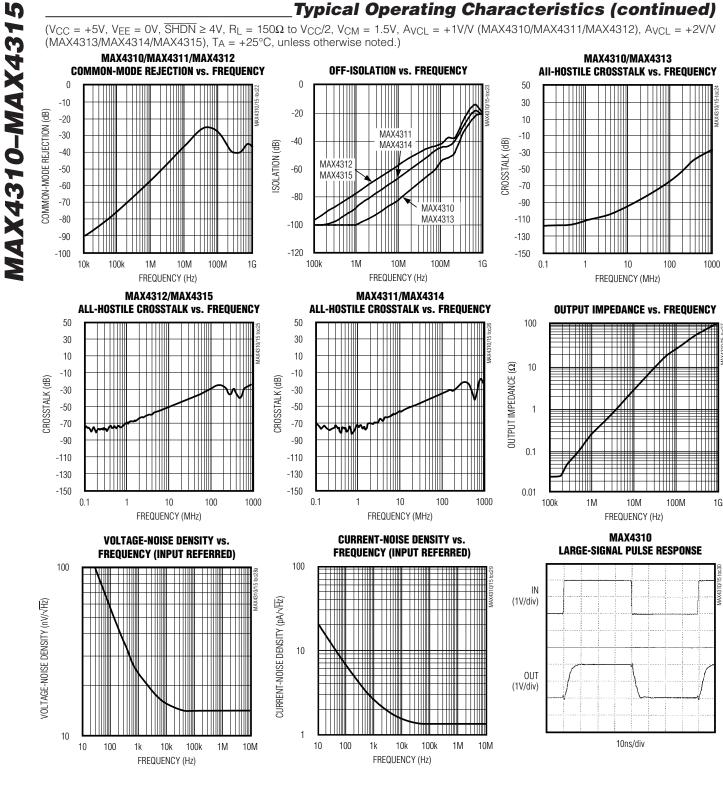
Typical Operating Characteristics (continued)

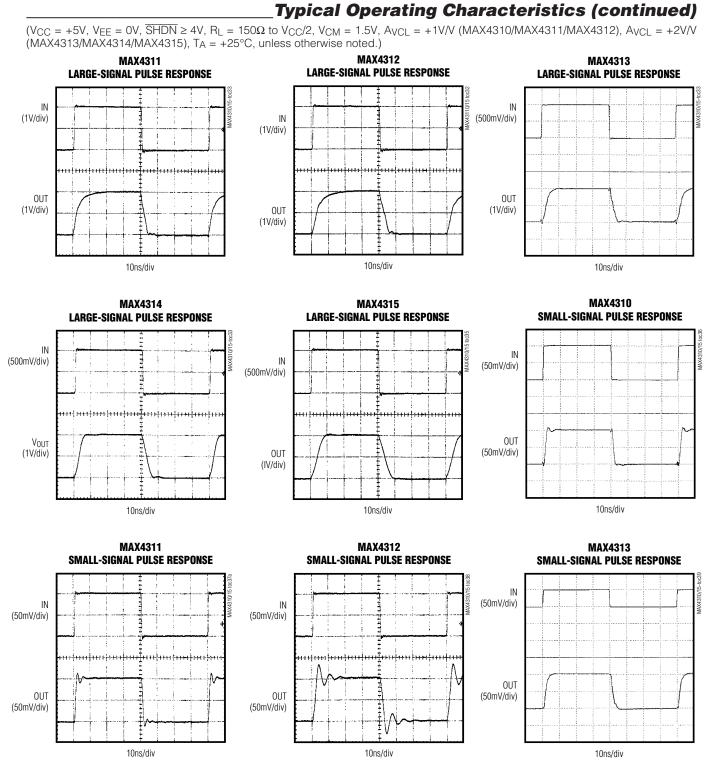
MAX4310-MAX4315



Typical Operating Characteristics (continued)



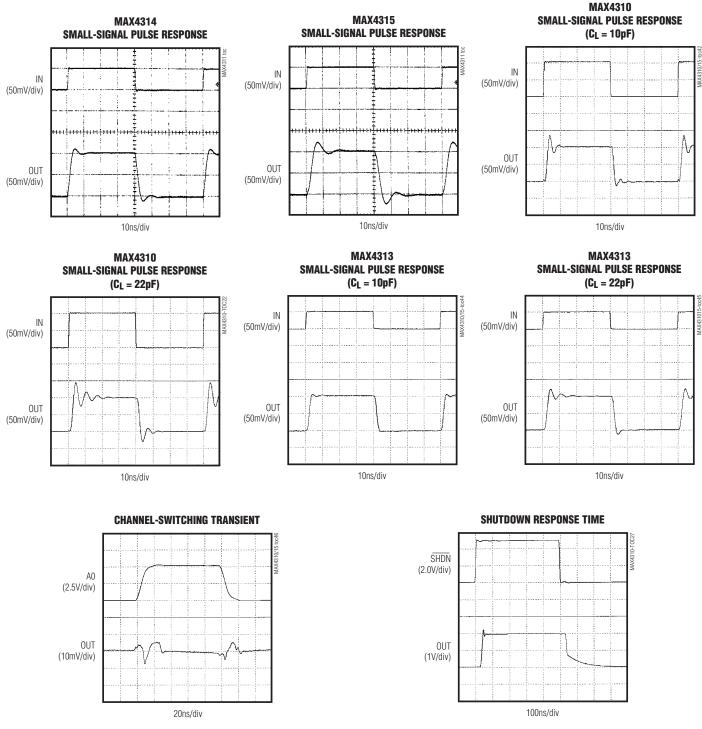




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 $(V_{CC} = +5V, V_{EE} = 0V, \overline{SHDN} \ge 4V, R_L = 150\Omega$ to $V_{CC}/2$, $V_{CM} = 1.5V$, $A_{VCL} = +1V/V$ (MAX4310/MAX4311/MAX4312), $A_{VCL} = +2V/V$ (MAX4313/MAX4314/MAX4315), $T_A = +25^{\circ}C$, unless otherwise noted.)



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Pin Description

	PIN								
MAX4310	MA	X4311	MAX4312	MAX4313	MA	X4314	MAX4315	NAME	FUNCTION
SO/µMAX	SO	QSOP	SO/QSOP	SO/µMAX	SO	QSOP	SO/QSOP		
1	2	2	3	1	2	2	3	AO	Channel Address Logic Input 0
_	1	1	2	_	1	1	2	A1	Channel Address Logic Input 1
_		_	1	_		_	1	A2	Channel Address Logic Input 2
2	12	14	14	2	12	14	14	SHDN	Shutdown Input
3	4	4	4	3	4	4	4	Vcc	Positive Power Supply
4	5	5	5	4	5	5	5	IN0	Amplifier Input 0
5	7	7	6	5	7	7	6	IN1	Amplifier Input 1
_	8	10	7		8	10	7	IN2	Amplifier Input 2
_	10	12	8	_	10	12	8	IN3	Amplifier Input 3
_		_	9	_		_	9	IN4	Amplifier Input 4
_		_	10	_	_	_	10	IN5	Amplifier Input 5
_	—		11	_	—	_	11	IN6	Amplifier Input 6
_	—		12	_	—	_	12	IN7	Amplifier Input 7
6	11	13	13	6	11	13	13	V _{EE}	Negative Power Supply. Ground for single-supply operation.
7	13	15	15	_		_	_	FB	Amplifier Feedback Input
_	_	_	_	7	13	15	15	GND	Ground
8	14	16	16	8	14	16	16	OUT	Amplifier Output
	3, 6, 9	3, 6, 8, 9, 11		_	3, 6, 9	3, 6, 8, 9, 11	_	N.C.	Not connected. Tie to ground plane for optimal performance.



_Detailed Description

The MAX4310/MAX4311/MAX4312 combine 2-channel, 4-channel, or 8-channel multiplexers, respectively, with an adjustable-gain output amplifier optimized for closed-loop gains of +1V/V (0dB) or greater. The MAX4313/MAX4314/MAX4315 combine 2-channel, 4channel, or 8-channel multiplexers, respectively, with a +2V/V (6dB) fixed-gain amplifier, optimized for driving back-terminated cables. These devices operate from a single supply voltage of +4V to +10.5V, or from dual supplies of $\pm 2V$ to $\pm 5.25V$. The outputs may be placed in a high-impedance state and the supply current minimized by forcing the SHDN pin low. The input multiplexers feature short 40ns channel-switching times and small 10mVp-p switching transients. The input capacitance remains constant at 1pF whether the channel is on or off, providing a predictable input impedance to the signal source. These devices feature single-supply. rail-to-rail, voltage-feedback output amplifiers that achieve up to 540V/µs slew rates and up to 345MHz -3dB bandwidths. These devices also feature excellent harmonic distortion and differential gain/phase performance.

Applications Information

Rail-to-Rail Outputs, Ground-Sensing Input

The input common-mode range extends from the negative supply rail to V_{CC} - 2.7V with excellent commonmode rejection. Beyond this range, multiplexer switching times may increase and the amplifier output is a nonlinear function of the input, but does not undergo phase reversal or latchup.

The output swings to within 250mV of V_{CC} and 40mV of V_{EE} with a 10k Ω load. With a 150 Ω load to ground, the output swings from 30mV above V_{EE} to within 730mV of

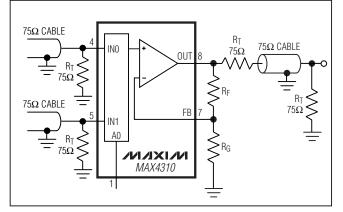


Figure 1. MAX4310 Noninverting Gain Configuration

the supply rail. Local feedback around the output stage ensures low open-loop output impedance to reduce gain sensitivity to load variations. This feedback also produces demand-driven bias current to the output transistors for ± 95 mA drive capability while constraining total supply current to only 6.1mA.

Feedback and Gain Resistor Selection (MAX4310/MAX4311/MAX4312)

Select the MAX4310/MAX4311/MAX4312 gain-setting feedback (RF) and input (RG) resistors to fit your application. Large resistor values increase voltage noise and interact with the amplifier's input and PC board capacitance. This can generate undesirable poles and zeros, and can decrease bandwidth or cause oscillations. For example, a noninverting gain of +2V/V configuration (RF = RG) using 1k Ω resistors, combined with 2pF of input capacitance and 1pF of PC board capacitance, causes a pole at 159MHz. Since this pole is within the amplifier bandwidth, it jeopardizes stability. Reducing the 1k Ω resistors to 100 Ω extends the pole frequency to 1.59GHz, but could limit output swing by adding 200 Ω in parallel with the amplifier's load resistor.

Table 1 shows suggested RF and RG values for the MAX4310/MAX4311/MAX4312 when operating in the noninverting configuration (shown in Figure 1). These values provide optimal AC response using surface-mount resistors and good layout techniques, as discussed in the *Layout and Power-Supply Bypassing* section.

Stray capacitance at the FB pin causes feedback resistor decoupling and produces peaking in the frequencyresponse curve. Keep the capacitance at FB as low as possible by using surface-mount resistors and by avoiding the use of a ground plane beneath or beside these resistors and the FB pin. Some capacitance is unavoidable; if necessary, its effects can be neutralized by adjusting RF. Use 1% resistors to maintain consistency over a wide range of production lots.

Table 1. Bandwidth and Gain withSuggested Gain-Setting resistors(MAX4310/MAX4311/MAX4312)

GAIN (V/V)	GAIN (d B)	R F (Ω)	R G (Ω)	-3dB BW (MHz)	0.1dB BW (MHz)
1	0	0	~	280	60
2	6	500	500	80	30
5	14	500	120	20	4
10	20	500	56	10	2

Resistor

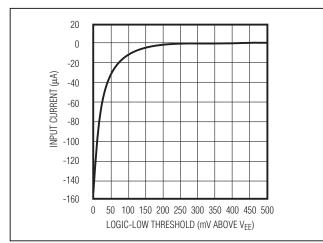


Figure 2. Logic-Low Input Current vs. VIL (SHDN, A0, A1, A2)

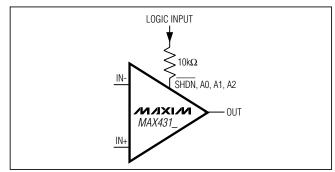


Figure 3. Circuit to Reduce Logic-Low Input Current

Low-Power Shutdown Mode

All parts feature a low-power shutdown mode that is activated by driving the SHDN input low. Placing the amplifier in shutdown mode reduces the quiescent supply current to 560µA and places the output into a highimpedance state, typically $35k\Omega$. By tying the outputs of several devices together and disabling all but one of the paralleled amplifiers' outputs, multiple devices may be paralleled to construct larger switch matrices.

For MAX4310/MAX4311/MAX4312 application circuits operating with a closed-loop gain of +2V/V or greater, consider the external-feedback network impedance of all devices used in the mux application when calculating the total load on the output amplifier of the active device. The MAX4313/MAX4314/MAX4315 have a fixed gain of +2V/V that is internally set with two 500 Ω thinfilm resistors. The impedance of the internal feedback resistors must be taken into account when operating multiple MAX4313/MAX4314/MAX4315s in large multiplexer applications. For normal operation, drive SHDN high. If the shutdown function is not used, connect SHDN to Vcc.



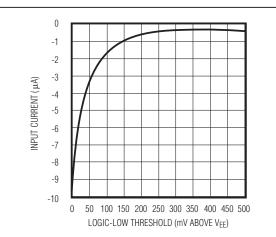


Figure 4. Logic-Low Input Current vs. V_{IL} with 10k Ω Series

Layout and Power-Supply Bypassing

The MAX4310–MAX4315 have very high bandwidths and consequently require careful board layout, including the possible use of constant-impedance microstrip or stripline techniques.

To realize the full AC performance of these high-speed amplifiers, pay careful attention to power-supply bypassing and board layout. The PC board should have at least two layers: a signal and power layer on one side, and a large, low-impedance ground plane on the other side. The ground plane should be as free of voids as possible, with one exception: the feedback (FB) should have as low a capacitance to ground as possible. Therefore, layers that do not incorporate a signal or power trace should not have a ground plane.

Whether or not a constant-impedance board is used, it is best to observe the following guidelines when designing the board:

- 1) Do not use wire-wrapped boards (they are too inductive) or breadboards (they are too capacitive).
- 2) Do not use IC sockets; they increase parasitic capacitance and inductance.
- 3) Keep signal lines as short and straight as possible. Do not make 90° turns; round all corners.
- 4) Observe high-frequency bypassing techniques to maintain the amplifier's accuracy and stability.
- 5) Use surface-mount components. They generally have shorter bodies and lower parasitic reactance, yielding better high-frequency performance than through-hole components.



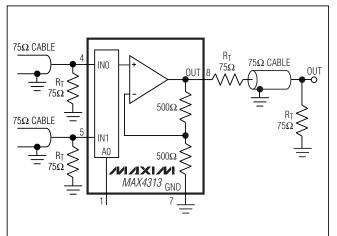


Figure 5. Video Line Driver

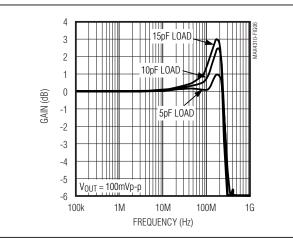


Figure 6. Small-Signal Gain vs. Frequency with a Capacitive Load and No-Isolation Resistor

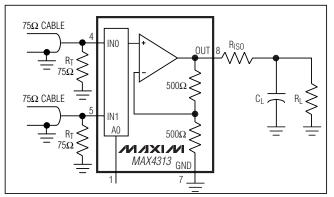


Figure 7. Using an Isolation Resistor (R_{ISO}) for High-Capacitive Loads

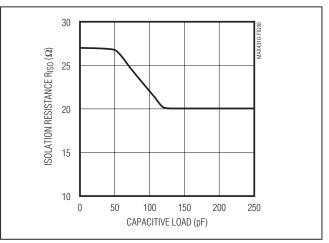


Figure 8. Optimal Isolation Resistance vs. Capacitive Load

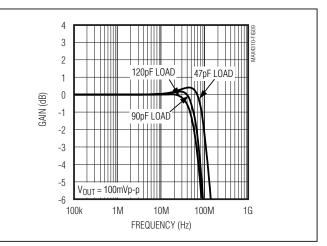


Figure 9. Small-Signal Gain vs. Frequency with a Capacitive Load and 27 Ω No-Isolation Resistor

The bypass capacitors should include a 100nF, ceramic surface-mount capacitor between each supply pin and the ground plane, located as close to the package as possible. Optionally, place a 10 μ F tantalum capacitor at the power-supply pin's point of entry to the PC board to ensure the integrity of incoming supplies. The power-supply trace should lead directly from the tantalum capacitor to the V_{CC} and V_{EE} pins. To minimize parasitic inductance, keep PC traces short and use surface-mount components. If input termination resistors and output back-termination resistors are used, they should be surface-mount types, and should be placed as close to the IC pins as possible.



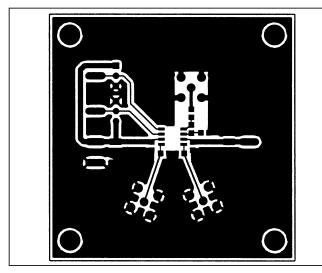


Figure 10. High-Speed EV Board Layout—Component Side

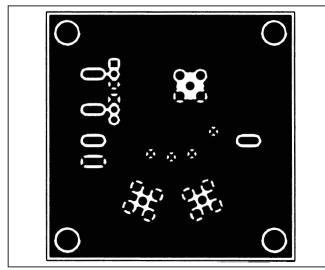


Figure 11. High-Speed EV Board Layout—Solder Side

Video Line Driver

The MAX4310–MAX4315 are well-suited to drive coaxial transmission lines when the cable is terminated at both ends, as shown in Figure 5. Cable frequency response can cause variations in the signal's flatness.

Driving Capacitive Loads

A correctly terminated transmission line is purely resistive and presents no capacitive load to the amplifier. Reactive loads decrease phase margin and may produce excessive ringing and oscillation (see *Typical Operating Characteristics*).



Table 2. Input Control Logic

MAX4310/MAX4313 SHDN A2 A1 A0 CHANNEL SELECTED 0 X None, High-Z Output 1 0 0 1 1 1 MAX4311/MAX4314 MAX4311/MAX4314 SHDN A2 A1 A0 CHANNEL SELECTED 0 X X None, High-Z Output 1 0 0 0 0 X X None, High-Z Output 1 0 0 0 1 0 1 1 1 0 1 1 1 1 0 2 1 1 1 3 MAX4312/MAX4315 SHDN A2 A1 A0 CHANNEL SELECTED 0 X X X None, High-Z Output 1 1 0 <th></th> <th></th> <th></th> <th></th> <th>5</th>					5								
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1 1 0 0 4 1 1 0 1 5	1	0	1	0	2								
1 1 0 1 5	1	0	1	1	3								
	1	1	0	0	4								
	1	1	0	1	5								
	1	1	1	0	6								
1 1 1 1 7	1	1	1	1	7								

Another concern when driving capacitive loads originates from the amplifier's output impedance, which appears inductive at high frequencies. This inductance forms an L-C resonant circuit with the capacitive load, which causes peaking in the frequency response and degrades the amplifier's phase margin.

Although the MAX4310–MAX4315 are optimized for AC performance and are not designed to drive highly capacitive loads, they are capable of driving up to 20pF without oscillations. However, some peaking may occur in the frequency domain (Figure 6). To drive larger capacitive loads or to reduce ringing, add an isolation resistor between the amplifier's output and the load (Figure 7).

The value of RISO depends on the circuit's gain and the capacitive load (Figure 8). Figure 9 shows the MAX4310–MAX4315 frequency response with the isolation resistor and a capacitive load. With higher capacitive values, bandwidth is dominated by the RC network formed by RISO and CL; the bandwidth of the amplifier itself is much higher. Also note that the isolation resistor forms a divider that decreases the voltage delivered to the load.

Digital Interface

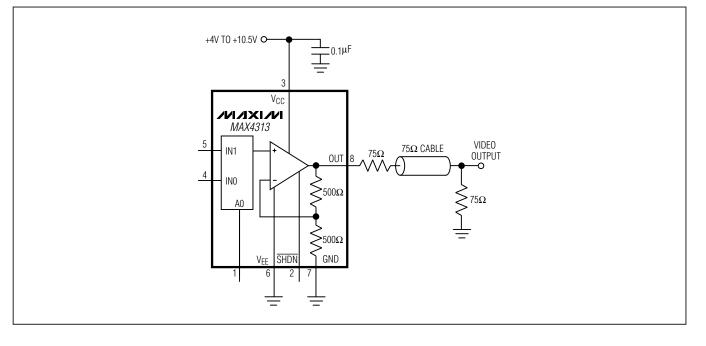
The multiplexer architecture of the MAX4310–MAX4315 ensures that no two input channels are ever connected together. Channel selection is accomplished by applying a binary code to channel address inputs. The address decoder selects input channels, as shown in Table 2. All digital inputs are CMOS compatible.

High-Speed Evaluation Board

Figures 10 and 11 show the evaluation board and present a suggested layout for the circuits. This board was developed using the techniques described in the *Layout and Power-Supply Bypassing* section. The smallest available surface-mount resistors were used for feedback and back-termination to minimize their distance from the part, reducing the capacitance associated with longer lead lengths.

SMA connectors were used for best high-frequency performance. Inputs and outputs do not match a 75 Ω line, but this does not affect performance since distances are extremely short. However, in applications that require lead lengths greater than one-quarter of the wavelength of the highest frequency of interest, use constant-impedance traces. Fully assembled evaluation boards are available for the MAX4313 in an SO package.

Typical Operating Circuit



Chip Information

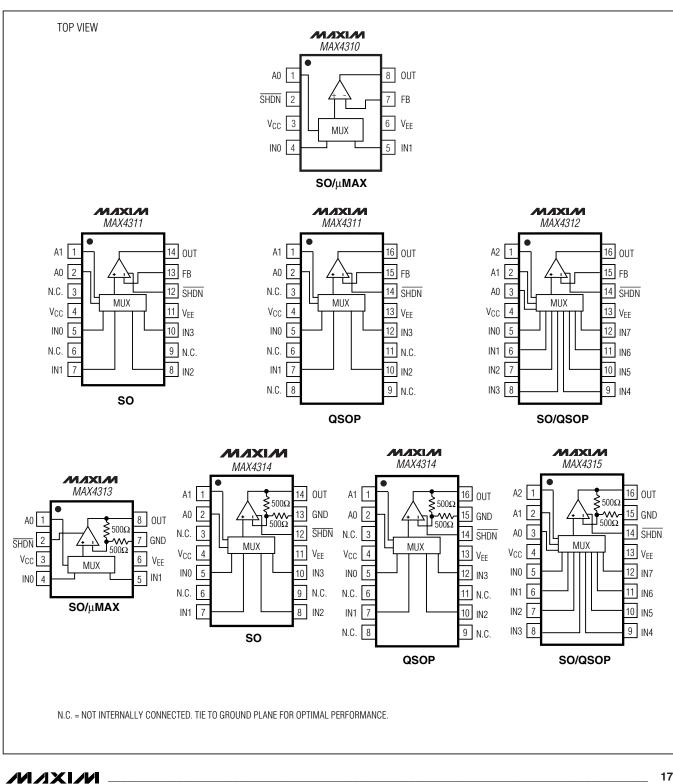
TRANSISTOR COUNT: 156

Package Information

For the latest package outline information, go to **www.maxim-ic.com/packages**.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.	
8 SO	S8-4	<u>21-0041</u>	
8 µMAX	U8-1	<u>21-0036</u>	
14 Narrow SO	S14-1	<u>21-0041</u>	
16 Narrow SO	S16-1	<u>21-0041</u>	
16 QSOP	E16-1	<u>21-0055</u>	

Pin Configurations



Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	7/98	Initial release	—
1	4/99	Added new parts to data sheet.	1–20
2	12/02	Corrected MAX4314 Pin Configuration.	17
3	3/08	Updated Typical Operating Characteristics.	8

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