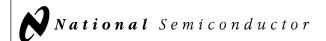
LM831 Low Voltage Audio Power Amplifier



LM831 Low Voltage Audio Power Amplifier

General Description

The LM831 is a dual audio power amplifier optimized for very low voltage operation. The LM831 has two independent amplifiers, giving stereo or higher power bridge (BTL) operation from two- or three-cell power supplies.

The LM831 uses a patented compensation technique to reduce high-frequency radiation for optimum performance in AM radio applications. This compensation also results in lower distortion and less wide-band noise.

The input is direct-coupled to the LM831, eliminating the usual coupling capacitor. Voltage gain is adjustable with a single resistor.

Features

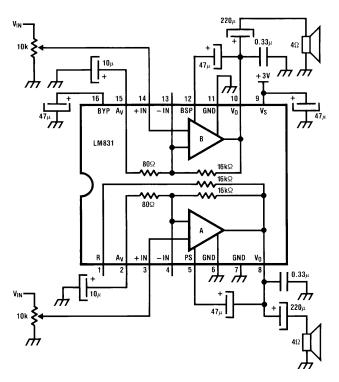
- Low voltage operation, 1.8V to 6.0V
- High power, 440 mW, 8Ω, BTL, 3V
- Low AM radiation
- Low noise
- Low THD

Applications

- Portable tape recorders
- Portable radios
- Headphone stereo
- Portable speakers

Typical Application

Dual Amplifier with Minimum Parts



 $A_V = 46 \text{ dB,BW} = 250 \text{ Hz to } 35 \text{ kHz}$ $\text{P}_{\text{OUT}} = \text{220 mW/Ch,R}_{\text{L}} = 4\Omega$

© 1995 National Semiconductor Corporation

RRD-B30M115/Printed in U. S. A.

Absolute Maximum Ratings

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

 $\begin{array}{c} {\rm 1.4W~(N~Package)} \\ {\rm Operating~Temperature~(Note~1),~T_{opr}} \\ \end{array} \\ \begin{array}{c} {\rm -40^{\circ}C~to~+85^{\circ}C} \\ \end{array}$

 $\begin{array}{lll} \mbox{Storage Temperature, $T_{\rm stg}$} & -65^{\circ}\mbox{C to } +150^{\circ}\mbox{C} \\ \mbox{Junction Temperature, $T_{\rm j}$} & +150^{\circ}\mbox{C} \\ \mbox{Lead Temp. (Soldering, 10 sec.), $T_{\rm L}$} & +260^{\circ}\mbox{C} \\ \mbox{Thermal Resistance} & & & \\ \mbox{$\theta_{\rm JC}$ (DIP)$} & 27^{\circ}\mbox{C/W} \end{array}$

 θ JC (DIP)
 27 °C/W

 θ JA (DIP)
 75 °C/W

 θ JC (SO Package)
 20 °C/W

 θ JA (SO Package)
 95 °C/W

TL/H/6754-2

Electrical Characteristics

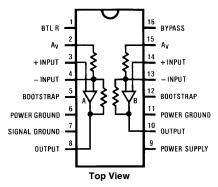
Unless otherwise specified, $T_A=25^{\circ}C$, $V_S=3V$, f=1~kHz, test circuit is dual or BTL amplifier with minimum parts.

Symbol	Parameter	Conditions	Тур	Tested Limit	Unit (Limit)
V _S	Operating Voltage		3	1.8 6	V(Min) V(Max)
IQ	Supply Current	$V_{\text{IN}} = 0$, Dual Mode $V_{\text{IN}} = 0$, BTL Mode	5 6	10 15	mA (Max) mA (Max)
Vos	Output DC Offset	V _{IN} = 0, BTL Mode	10	50	mV (Max)
R _{IN}	Input Resistance		25	15 35	k (Min) k (Max)
A _V	Voltage Gain	$V_{\rm IN}=2.25~{\rm mV_{rms}}, {\rm f}=1~{\rm kHz},$ Dual Mode	46	44 48	dB (Min) dB (Max)
PSRR	Supply Rejection	$V_S = 3V + 200 \text{ mV}_{rms} @ f = 1 \text{ kHz}$	46	30	dB (Min)
P _{OD}	Power Out	$V_{S}=3V,$ $R_{L}=4\Omega,$ 10% THD, Dual Mode	220	150	mW (Min)
P _{ODL}	Power Out Low, V _S	$V_{S}=$ 1.8V, $R_{L}=4\Omega$, 10% THD, Dual Mode	45	10	mW (Min)
P _{OB}	Power Out	$V_S=3V, R_L=8\Omega,$ 10% THD, BTL Mode	440	300	mW (Min)
P _{OBL}	Power Out Low, V _S	$V_S=1.8V, R_L=8\Omega,$ 10% THD, BTL Mode	90	20	mW (Min)
Sep	Channel Separation	Referenced to V _O = 200 mV _{rms}	52	40	dB (Min)
I _B	Input Bias Current		1	2	μΑ (Max)
E _{n0}	Output Noise	Wide Band (250 \sim 35 kHz)	250	500	μV (Max)
THD	Distortion	$V_S = 3V$, $P_O = 50$ mW, $f = 1$ kHz, Dual	0.25	1	% (Max)

Note 1: For operation in ambient temperatures above 25°C, the device must be derated based on a 150°C maximum junction temperature and a thermal resistance of 98°C/W junction to ambient for the M package or 90°C/W junction to ambient for the N package.

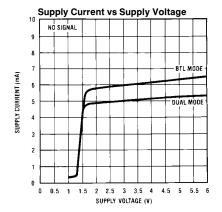
Connection Diagram

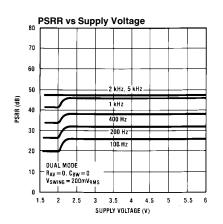
Dual-In-Line Package



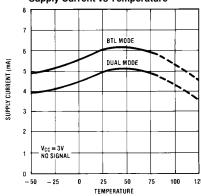
Order Number LM831M or N See NS Package Number M16B or N16E

Typical Performance Characteristics

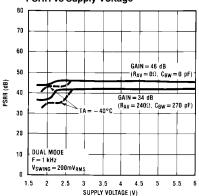




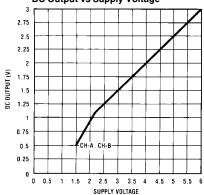
Supply Current vs Temperature



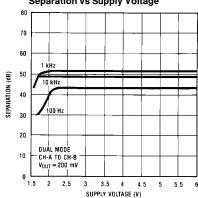
PSRR vs Supply Voltage



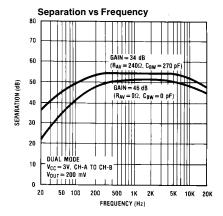
DC Output vs Supply Voltage

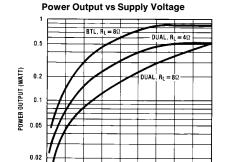


Separation vs Supply Voltage

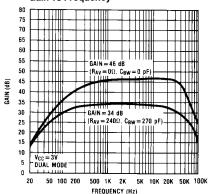








Gain vs Frequency



Power Output vs Temperature

3.5

SUPPLY VOLTAGE (V)

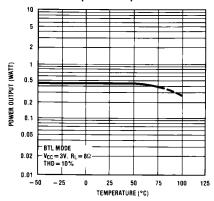
4.5

5.5

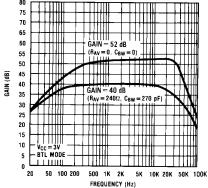
0.01

1.5 2

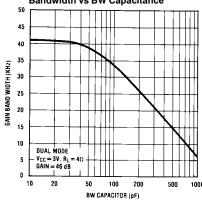
2.5



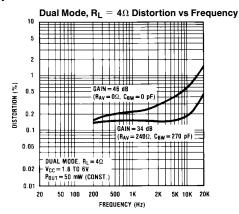
Gain vs Frequency

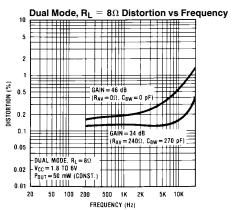


Bandwidth vs BW Capacitance

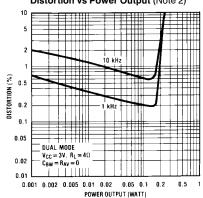


Typical Performance Characteristics (Continued)

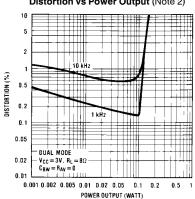




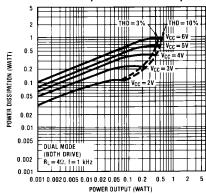
Distortion vs Power Output (Note 2)



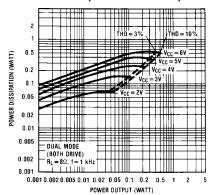
Distortion vs Power Output (Note 2)



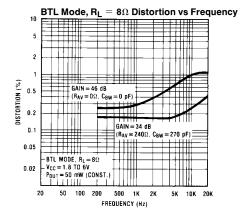
Power Dissipation vs Power Output

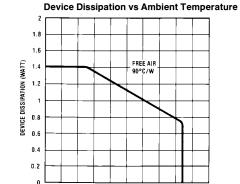


Power Dissipation vs Power Output

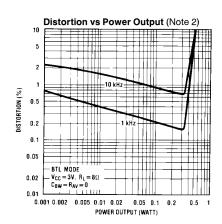


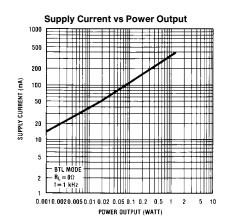
Typical Performance Characteristics (Continued)



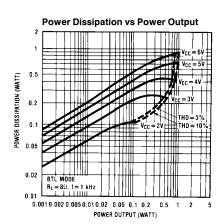


90 100





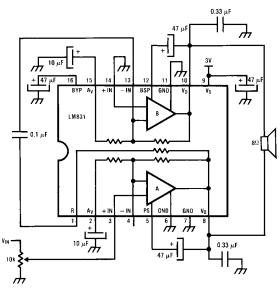
10 20 30 40 50 60 70



Note 2: 1 kHz curve is measured with 400 Hz-30 kHz Filter.

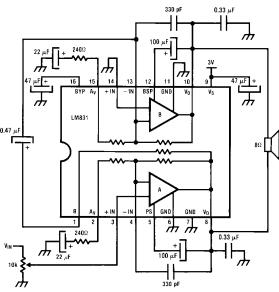
Typical Applications

BTL Amplifier with Minimum Parts



 $\rm A_V = 52~dB,\,BW = 250~Hz$ to 25 kHz $\rm P_{OUT} = 440~mW,\,R_L = 8\Omega$

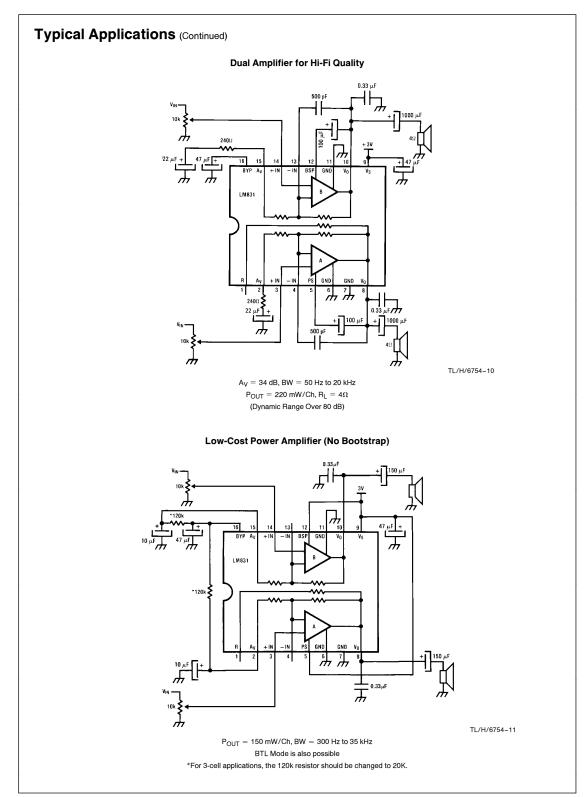
BTL Amplifier for Hi-Fi Quality



TL/H/6754-9

TL/H/6754-8

 $\rm A_V=40~dB,\,BW=20~Hz$ to 20 kHz $\rm P_{OUT}=440~mW,\,R_L=8\Omega$ (Dynamic Range Over 80 dB)



LM831 Circuit Description Refer to the external component diagram and equivalent schematic.

The power supply is applied to Pin 9 and is filtered by resistor $\rm R_1$ and capacitor $\rm C_{BY}$ on Pin 16. This filtered voltage at Pin 16 is used to bias all of the LM831 circuits except the power output stage. Resistor $\rm R_0$ generates a biasing current that sets the output DC voltage for optimum output power for any given supply voltage.

Feedback is provided to the input transistor \textbf{Q}_1 emitter by \textbf{R}_6 and $\textbf{R}_7.$

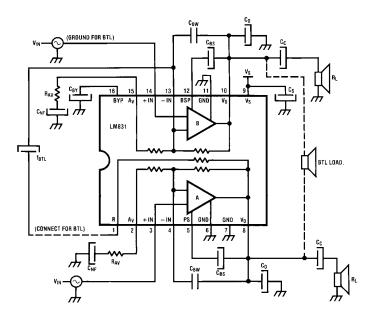
The capacitor $C_{\mbox{\scriptsize NF}}$ on Pin 2 provides unity DC gain for maximum DC accuracy.

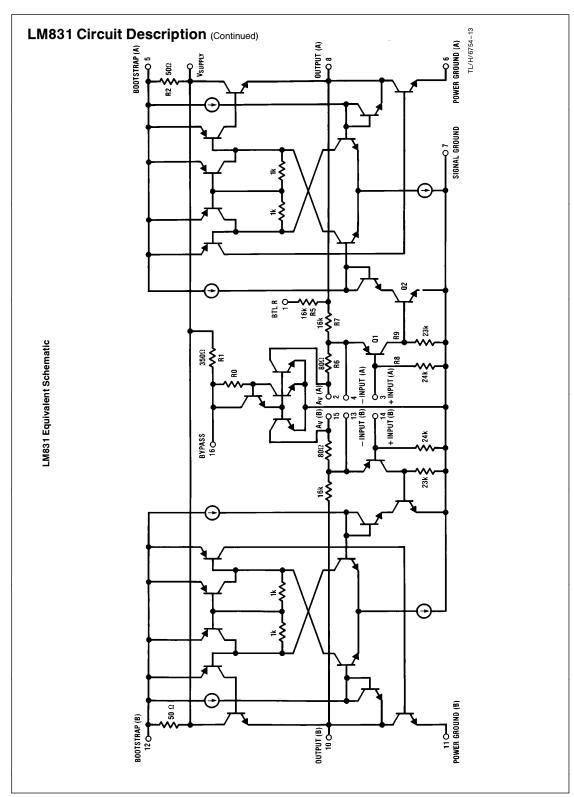
 ${\rm Q}_2$ provides voltage gain and the rest of the devices buffer the output load from ${\rm Q}_2$'s collector.

Bootstrapping of Pin 5 by $C_{\mbox{\footnotesize{BS}}}$ allows maximum output swing and improved supply rejection.

 $\ensuremath{\mathsf{R}}_5$ is provided for bridge (BTL) operation.

External Component Diagram

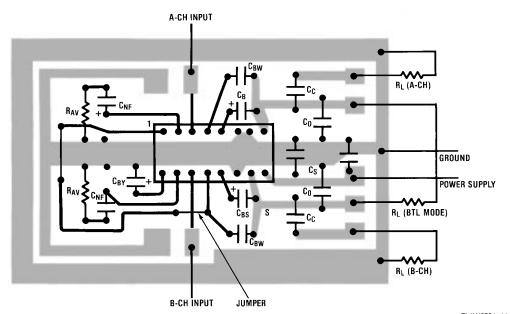




Component	Comments	Min	Max
CO	Required to stabilize output stage.	0.33 μF	1 μF
C _c	Output coupling capacitors for Dual Mode. Sets a low-frequency pole in the frequency response. $f_L = \frac{1}{2\pi C_c R_L}$	100 μF	10,000 μF
C _{BS}	Bootstrap capacitors. Sets a low-frequency pole in the power BW. Recommended value is $C_{BS} = \frac{1}{10^{\bullet}2\pi^{\bullet}f_{L}^{\bullet}R_{L}}$	22 μF or (short Pins 4 & 12 to 9)	470 μF
C _S	Supply bypass. Larger values improve low-battery performance by reducing supply ripple.	47 μF	10,000 μF
C _{BY}	Filters the supply for improved low-voltage operation. Also sets turn-on delay.	47 μF	470 μF
C _{NF}	Sets a low-frequency response. Also affects turn-on delay. $f_L = \frac{1}{2\pi^\bullet C_{NF}^\bullet (R_{AV} + 80)}$ In BTL Mode, C_{NF} on Pin 15 can be reduced without affecting the frequency response. However, the turn-on "POP" will be worsened.	10 μF	100 μF
C _{BTL}	Used only in the Bridge Mode. Connects the output of the first amplifier to the inverting input of the other through an internal resistor. Sets a low-frequency pole in one-half the frequency response. $f_L = \frac{1}{2\pi \bullet C_{BTL} \bullet 16k}$	0.1 μF	1 μF
C _{BW}	Improves clipping waveform and sets the high-frequency bandwidth. Works with an internal 16k resistor. (This equation applies for $R_{AV} \neq 0$. For 46 dB application, see BW–CBW curve.) $f_H = \frac{1}{2\pi^{\bullet}C_{BW}^{\bullet}16k}$	See table below	
R _{AV}	Used to reduce the gain and improve the distortion and signal to noise. If this is desired, C _{BW} must also be used.	See table below	

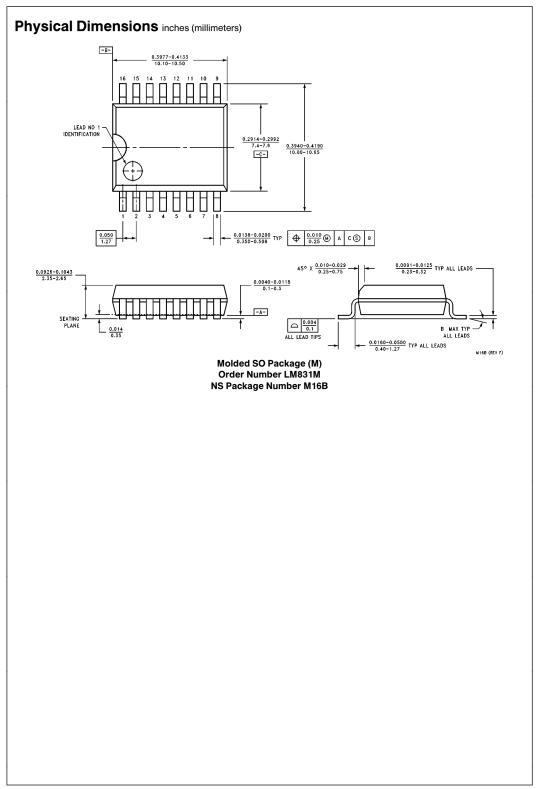
Typical A _V	R _{AV}	C _{BW}		
	··AV	Min	Max	
46 dB	Short	Open	4700 pF	
40 dB	82	100 pF	4700 pF	
34 dB	240	270 pF	4700 pF	
28 dB	560	500 pF	4700 pF	

Printed Circuit Layout for LM831N (Foil Side View) Refer to External Component Diagram

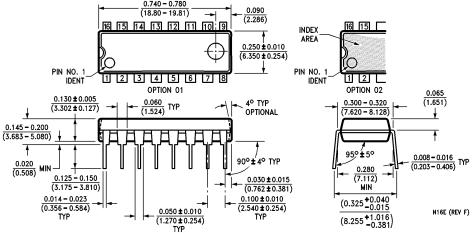


TL/H/6754-14

Note: Power ground pattern should be as wide as possible. Supply bypass capacitor should be as close to the IC as possible. Output compensation capacitors should also be close to the IC.



Physical Dimensions inches (millimeters) (Continued)



Molded Dual-In-Line Package (N) Order Number LM831N NS Package Number N16E

LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

- 1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform, when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
- A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.



National Semiconductor Corporation 1111 West Bardin Road Arlington, TX 76017 Tel: 1(800) 272-9959 Fax: 1(800) 737-7018 National Semiconductor Europe

Fax: (+49) 0-180-530 85 86 Email: cnjwge@tevm2.nsc.com Deutsch Tel: (+49) 0-180-530 85 85 English Tel: (+49) 0-180-532 78 32 Français Tel: (+49) 0-180-532 93 58 Italiano Tel: (+49) 0-180-534 16 80 National Semiconductor Hong Kong Ltd. 13th Floor, Straight Block, Ocean Centre, 5 Canton Rd. Tsimshatsui, Kowloon Hong Kong Tel: (852) 2737-1600 Fax: (852) 2736-9960 National Semiconductor Japan Ltd. Tel: 81-043-299-2309 Fax: 81-043-299-2408

National does not assume any responsibility for use of any circuitry described, no circuit patent licenses are implied and National reserves the right at any time without notice to change said circuitry and specifications