PCI Express Jitter Attenuator

DATA SHEET

GENERAL DESCRIPTION



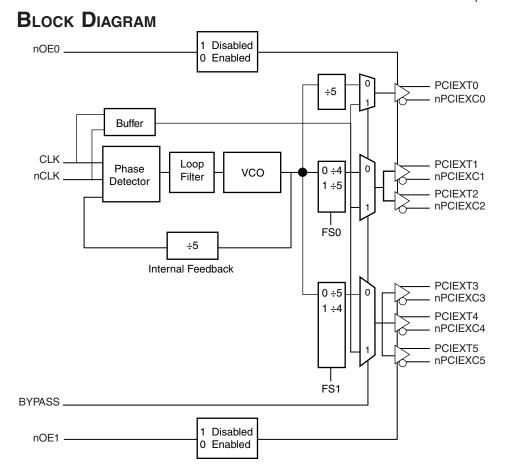
The ICS9DB306 is a high performance 1-to-6 Differential-to-LVPECL Jitter Attenuator designed for use in PCI Express™ systems. In some PCI Express systems, such as those found in desktop PCs, the PCI Express clocks are generated from a low

bandwidth, high phase noise PLL frequency synthesizer. In these systems, a zero delay buffer may be required to attenuate high frequency random and deterministic jitter components from the PLL synthesizer and from the system board. The ICS9DB306 has 2 PLL bandwidth modes. In low bandwidth mode, the PLL loop BW is about 500kHz and this setting will attenuate much of the jitter from the reference clock input while being high enough to pass a triangular input spread spectrum profile. There is also a high bandwidth mode which sets the PLL bandwidth at 1MHz which will pass more spread spectrum modulation.

For serdes which have x30 reference multipliers instead of x25 multipliers, 5 of the 6 PCI Express outputs (PCIEX1:5) can be set for 125MHz instead of 100MHz by configuring the appropriate frequency select pins (FS0:1). Output PCIEX0 will always run at the reference clock frequency (usually 100MHz) in desktop PC PCI Express Applications.

FEATURES

- Six differential LVPECL output pairs
- · One differential clock input
- CLK and nCLK supports the following input types: LVPECL, LVDS, LVHSTL, SSTL, HCSL
- Maximum output frequency: 140MHz
- Input frequency range: 90MHz 140MHz
- VCO range: 450MHz 700MHz
- Output skew: 135ps (maximum)
- Cycle-to-Cycle jitter: 30ps (maximum)
- RMS phase jitter @ 100MHz, (1.5MHz 22MHz): 3ps (typical)
- 3.3V operating supply
- 0°C to 70°C ambient operating temperature
- Available in both standard (RoHS 5) and lead-free (RoHS 6) packages
- Industrial temperature information available upon request



PIN ASSIGNMENT

VEE 🗆	1	28	□ Vcc
PCIEXT1 □	2	27	PCIEXC0
PCIEXC1 □	3	26	PCIEXT0
PCIEXT2 □	4	25	☐ FS0
PCIEXC2 □	5	24	nCLK
Vcc □	6	23	CLK
nOE0 □	7	22	☐ PLL_BW
nOE1 □	8	21	☐ VCCA
Vcc □	9	20	□ VEE
PCIEXC3 □	10	19	BYPASS
PCIEXT3 □	11	18	FS1
PCIEXC4 □	12	17	PCIEXT5
PCIEXT4 □	13	16	PCIEXC5
VEE	14	15	□ Vcc

ICS9DB306

28-Lead TSSOP, 173-MIL 4.4mm x 9.7mm x 0.925mm body package L Package

Top View

ICS9DB306

28-Lead, 209-MIL SSOP 5.3mm x 10.2mm x 1.75mm body package F Package Top View

TABLE 1. PIN DESCRIPTIONS

Number	Name	Туре		Description
1, 14, 20	V _{EE}	Power		Negative supply pins.
2, 3	PCIEXT1, PCIEXC1	Output		Differential output pairs. LVPECL interface levels.
4, 5	PCIEXT2, PCIEXC2	Output		Differential output pairs. LVPECL interface levels.
6, 9, 15, 28	V _{cc}	Power		Core supply pins.
7, 8	nOE0, nOE1	Input	Pulldown	Output enable. When HIGH, forces true outputs (PCIEXTx) to go LOW and the inverted outputs (PCIEXCx) to go HIGH. When LOW, outputs are enabled. LVCMOS/LVTTL interface levels.
10, 11	PCIEXC3, PCIEXT3	Output		Differential output pairs. LVPECL interface levels.
12, 13	PCIEXC4, PCIEXT4	Output		Differential output pairs. LVPECL interface levels.
16, 17	PCIEXC5, PCIEXT5	Output		Differential output pairs. LVPECL interface levels.
18	FS1		Pulldown	Frequency select pin. LVCMOS/LVTTL interface levels.
19	BYPASS	Input	Pulldown	Bypass select pin. When HIGH, the PLL is in bypass mode, and the device can function as a 1:6 buffer. LVCMOS/LVTTL interface levels.
21	V_{CCA}	Power		Analog supply pin. Requires 24Ω series resistor.
22	PLL_BW	Input	Pullup	Selects PLL Bandwidth input. LVCMOS/LVTTL interface levels.
23	CLK	Input	Pulldown	Non-inverting differential clock input.
24	nCLK	Input	Pullup/ Pulldown	Inverting differential clock input. V _{cc} /2 default when left floating.
25	FS0	Input	Pullup	Frequency select pin. LVCMOS/LVTTL interface levels.
26, 27	PCIEXT0, PCIEXC0	Output		Differential output pairs. LVPECL interface levels.

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

Table 2. Pin Characteristics

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance			4		pF
R _{PULLUP}	Input Pullup Resistor			51		kΩ
R _{PULLDOWN}	Input Pulldown Resistor			51		kΩ

TABLE 3A. RATIO OF OUTPUT FREQUENCY TO INPUT FREQUENCY FUNCTION TABLE, FS0

Inputs	Outputs					
FS0	PCIEX0 PCIEX1 PCIEX					
0	1	5/4	5/4			
1	1	1	1			

TABLE 3C. OUTPUT ENABLE FUNCTION TABLE, nOE0

Inputs	Outputs
nOE0	PCIEX0:2
0	Enabled
1	Disabled

TABLE 3D. OUTPUT ENABLE FUNCTION TABLE, nOE1

Inputs	Outputs
nOE1	PCIEX3:5
0	Enabled
1	Disabled

TABLE 3B. RATIO OF OUTPUT FREQUENCY TO INPUT FREQUENCY FUNCTION TABLE, FS1

Inputs	Outputs					
FS1	PCIEX3 PCIEX4 PCIEX5					
0	1	1	1			
1	5/4	5/4	5/4			

TABLE 3E. PLL BANDWIDTH FUNCTION TABLE

Inputs	Bandwidth
PLL_BW	bandwidth
0	500kHz
1	1MHz

TABLE 3F. PLL MODE FUNCTION TABLE

Inputs	DI I Mada				
BYPASS	PLL Mode				
1	Disabled				
0	Enabled				

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V_{CC} 4.6V

Inputs, V_1 -0.5V to V_{cc} + 0.5V

Outputs, I_o

Continuous Current 50mA Surge Current 100mA

Package Thermal Impedance, θ_{JA} $\,$ 49.8°C/W (0 lfpm) Storage Temperature, T $_{STG}$ $\,$ -65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Table 4A. Power Supply DC Characteristics, $V_{CC} = 3.3V \pm 10\%$, $T_A = 0^{\circ}C$ to $70^{\circ}C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{cc}	Core Supply Voltage		2.97	3.3	3.63	V
V _{CCA}	Analog Supply Voltage		V _{cc} - 0.60	3.3	V _{cc}	V
I _{cc}	Power Supply Current				135	mA
I _{CCA}	Analog Supply Current				25	mA

Table 4B. LVCMOS / LVTTL DC Characteristics, $V_{CC} = 3.3V \pm 10\%$, TA = 0°C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V _{IH}	Input High Voltage			2		V _{cc} + 0.3	mV
V _{IL}	Input Low Voltage			-0.3		0.8	mV
I _{IH}	Input High Current	nOE0, nOE1, FS1, BYPASS	$V_{CC} = V_{IN} = 3.63V$			150	μΑ
"		FS0, PLL_BW				5	μΑ
I _{IL}	Input Low Current	nOE0, nOE1, FS1, BYPASS	$V_{CC} = 3.63V, V_{IN} = 0V$	-5			μΑ
IL		FS0, PLL_BW		-150			μΑ

Table 4C. Differential DC Characteristics, $V_{CC} = 3.3V \pm 10\%$, $TA = 0^{\circ}C$ to $70^{\circ}C$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
I _{IH}	Input High Current	CLK, nCLK	$V_{CC} = V_{IN} = 3.63V$			150	μΑ
I	Input Low Current	CLK, nCLK	$V_{CC} = 3.63V, V_{IN} = 0V$			150	μΑ
V _{PP}	Peak-to-Peak Input Voltage; NOTE 1			0.15		1.3	V
V _{CMR}	Common Mode Inpu	it Voltage; NOTE 1, 2		V _{EE} + 0.5		V _{cc} - 0.85	V

NOTE 1: V_{IL} should not be less than -0.3V.

NOTE 2: Common mode voltage is defined as $V_{\rm in}$.

Table 4D. LVPECL DC Characteristics, $V_{CC} = 3.3V \pm 10\%$, Ta = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{OH}	Output High Voltage; NOTE 1		V _{cc} - 1.4		V _{cc} - 0.9	V
V _{OL}	Output Low Voltage; NOTE 1		V _{cc} - 2.0		V _{cc} - 1.7	V
V _{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with 50 Ω to V_{cc} - 2V.

Table 5A. AC Characteristics, $V_{CC} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f _{MAX}	Output Frequency				140	MHz
tsk(o)	Output Skew; NOTE 1, 2			55	135	ps
tjit(cc)	Cycle-to-Cycle Jitter, NOTE 2				25	ps
tjit(Ø)	RMS Phase Jitter (Random); NOTE 3	Integration Range: 1.5MHz - 22MHz		3		ps
t _R / t _F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle		48		52	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential cross points.

NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: Please refer to the Phase Noise Plot following this section.

Table 5B. AC Characteristics, $V_{cc} = 3.3V \pm 10\%$, Ta = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f _{MAX}	Output Frequency				140	MHz
tsk(o)	Output Skew; NOTE 1, 2			25	100	ps
tjit(cc)	Cycle-to-Cycle Jitter, NOTE 2				30	ps
<i>t</i> jit(Ø)	RMS Phase Jitter (Random); NOTE 3	Integration Range: 1.5MHz - 22MHz		3		ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle		47		53	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

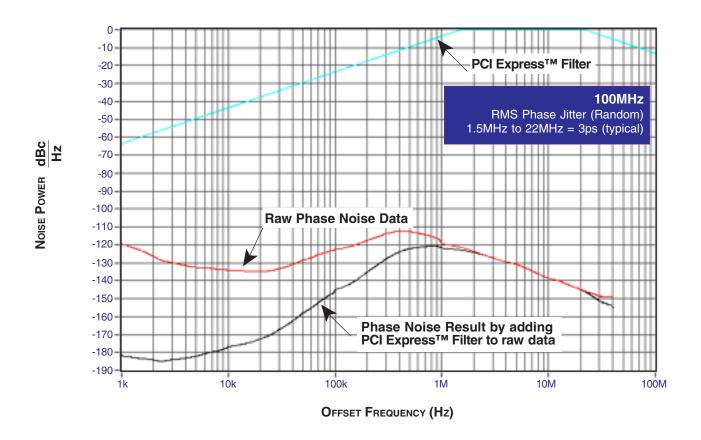
NOTE 1: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential cross points.

NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: Please refer to the Phase Noise Plot following this section.

Typical Phase Noise at 100MHz

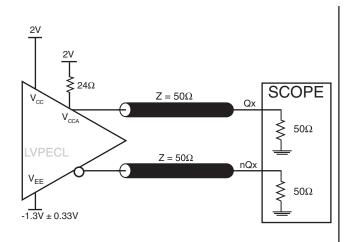


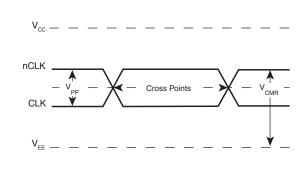
The illustrated phase noise plot was taken using a low phase noise signal generator, the noise floor of the signal generator is less than that of the device under test.

Using this configuration allows one to see the true spectral purity or phase noise performance of the PLL in the device under test.

Due to the tracking ability of a PLL, it will track the input signal up to its loop bandwidth. Therefore, if the input phase noise is greater than that of the PLL, it will increase the output phase noise performance of the device. It is recommended that the phase noise performance of the input is verified in order to achieve the above phase noise performance.

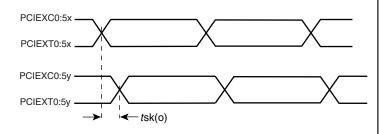
PARAMETER MEASUREMENT INFORMATION

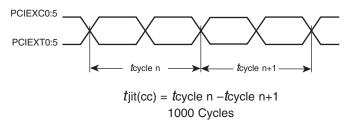




3.3V LVPECL OUTPUT LOAD AC TEST CIRCUIT

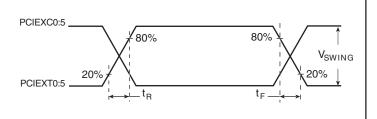
DIFFERENTIAL INPUT LEVEL

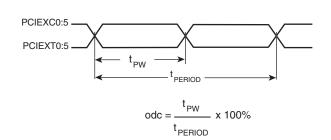




OUTPUT SKEW

CYCLE-TO-CYCLE JITTER





OUTPUT RISE/FALL TIME

OUTPUT DUTY CYCLE/PULSE WIDTH/PERIOD

APPLICATION INFORMATION

Power Supply Filtering Techniques

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The ICS9DB306 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. $V_{\rm CC}$ and $V_{\rm CCA}$ should be individually connected to the power supply plane through vias, and 0.01µF bypass capacitors should be used for each pin. Figure 1 illustrates this for a generic $V_{\rm CC}$ pin and also shows that $V_{\rm CCA}$ requires that an additional 24Ω resistor along with a $10\mu{\rm F}$ bypass capacitor be connected to the $V_{\rm CCA}$ pin.

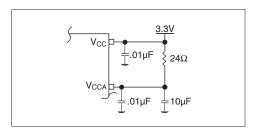


FIGURE 1. POWER SUPPLY FILTERING

WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LEVELS

Figure 2 shows how the differential input can be wired to accept single ended levels. The reference voltage V_REF = $V_{_{DD}}/2$ is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio

of R1 and R2 might need to be adjusted to position the V_REF in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and $V_{\rm cc}$ = 3.3V, V_REF should be 1.25V and R2/R1 = 0.609.

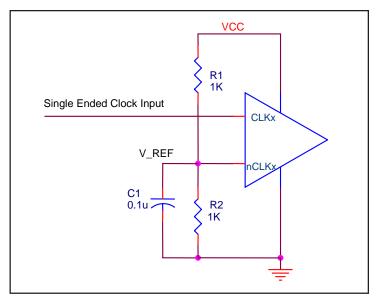


FIGURE 2. SINGLE ENDED SIGNAL DRIVING DIFFERENTIAL INPUT

DIFFERENTIAL CLOCK INPUT INTERFACE

The CLK /nCLK accepts LVDS, LVPECL, LVHSTL, SSTL, HCSL and other differential signals. Both V $_{\mbox{\tiny SWINIG}}$ and V $_{\mbox{\tiny CMR}}$ input requirements. Figures 3A to 3E show interface examples for the HiPerClockS CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are

examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example in *Figure 3A*, the input termination applies for IDT HiPerClockS LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

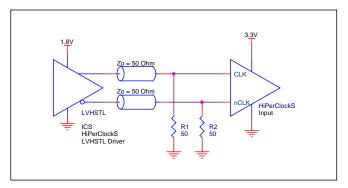


FIGURE 3A. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY IDT HIPERCLOCKS LVHSTL DRIVER

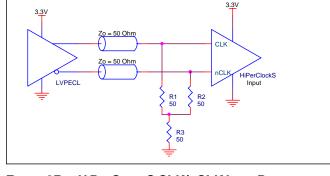


FIGURE 3B. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER

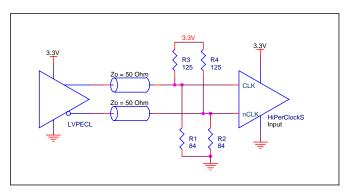


FIGURE3C. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER

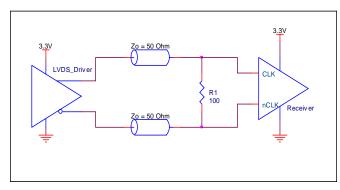


FIGURE 3D. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY 3.3V LVDS DRIVER

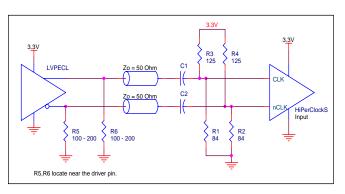


FIGURE 3E. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER WITH AC COUPLE

TERMINATION FOR LVPECL OUTPUTS

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are de-

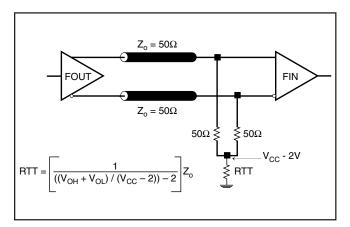


FIGURE 4A. LVPECL OUTPUT TERMINATION

signed to drive 50Ω transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. Figures 4A and 4B show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

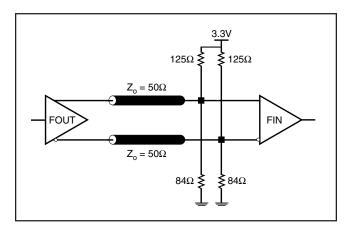


FIGURE 4B. LVPECL OUTPUT TERMINATION

RECOMMENDATIONS FOR UNUSED INPUT AND OUTPUT PINS INPUTS:

LVCMOS CONTROL PINS

All control pins have internal pullups or pulldowns; additional resistance is not required but can be added for additional protection. A 1kW resistor can be used.

OUTPUTS:

LVPECL OUTPUTS

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

SCHEMATIC EXAMPLE

Figure 5 shows an example of ICS9DB306 application schematic. In this example, the device is operated at $V_{cc} = 3.3V$. The decoupling capacitor should be located as close as possible to the power pin. The input is driven by a HCSL driver. For LVPECL

output drivers, one of terminations approaches is shown in this schematic. For additional termination approaches, please refer to the LVPECL Termination Application Note.

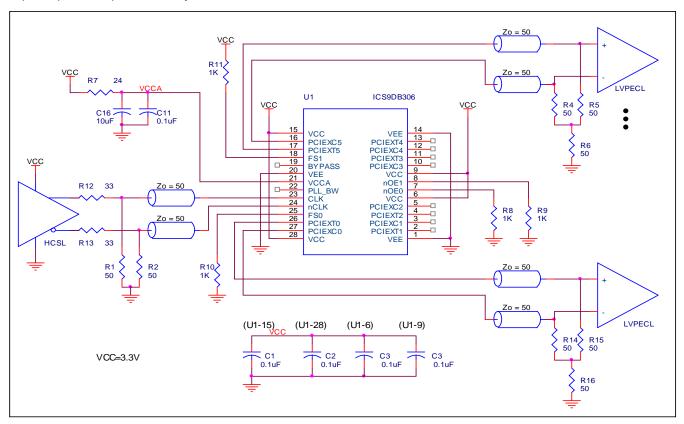


FIGURE 5. EXAMPLE OF ICS9DB306 SCHEMATIC

POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS9DB306. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS9DB306 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{cc} = 3.3V + 10\% = 3.63V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{CC MAX} * I_{EE MAX} = 3.63V * 135mA = 490.1mW
- Power (outputs)_{MAX} = 30mW/Loaded Output pair
 If all outputs are loaded, the total power is 6 * 30mW = 180mW

Total Power $_{\text{MAX}}$ (3.63V, with all outputs switching) = 490.1mW + 180mW = 670.1mW

2. Junction Temperature.

Junction temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS™ devices is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{La} * Pd_total + T_a

Tj = Junction Temperature

 $\theta_{\text{\tiny IA}}$ = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_a = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance $\theta_{\text{\tiny M}}$ must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 43.9°C/W per Table 6 below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is: $70^{\circ}\text{C} + 0.670\text{W} * 43.9^{\circ}\text{C/W} = 99.4^{\circ}\text{C}$. This is well below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

Table 6. Thermal Resistance $\theta_{_{JA}}$ for 28-pin TSSOP, Forced Convection

0 200 500 Single-Layer PCB, JEDEC Standard Test Boards 82.9°C/W 68.7°C/W 60.5°C/W Multi-Layer PCB, JEDEC Standard Test Boards 49.8°C/W 43.9°C/W 41.2°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

 θ_{M} by Velocity (Linear Feet per Minute)

3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in Figure 6.

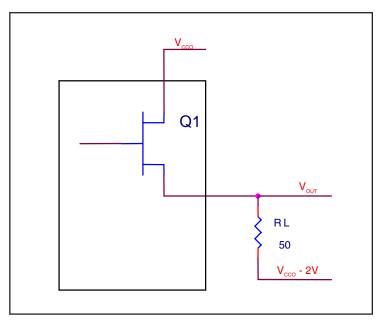


FIGURE 6. LVPECL DRIVER CIRCUIT AND TERMINATION

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{cco} - 2V.

• For logic high,
$$V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} - 0.9V$$

$$(V_{CCO,MAX} - V_{OH,MAX}) = 0.9V$$

• For logic low, $V_{OUT} = V_{OL_MAX} = V_{CCO_MAX} - 1.7V$

$$(V_{CCO,MAX} - V_{OL,MAX}) = 1.7V$$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$$Pd_H = [(V_{\text{OH_MAX}} - (V_{\text{CCO_MAX}} - 2V))/R_{\text{L}}] * (V_{\text{CCO_MAX}} - V_{\text{OH_MAX}}) = [(2V - (V_{\text{CCO_MAX}} - V_{\text{OH_MAX}}))/R_{\text{L}}] * (V_{\text{CCO_MAX}} - V_{\text{OH_MAX}}) = [(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$$

$$Pd_L = [(V_{OL_MAX} - (V_{CCO_MAX} - 2V))/R_{L}] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - (V_{CCO_MAX} - V_{OL_MAX}))/R_{L}] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$$

Total Power Dissipation per output pair = Pd_H + Pd_L = 30mW

RELIABILITY INFORMATION

Table 7A. $\theta_{\text{\tiny JA}}$ vs. Air Flow Table For 28 Lead TSSOP Package

θ_{JA} by Velocity (Linear Feet per Minute)

O200500Single-Layer PCB, JEDEC Standard Test Boards82.9°C/W68.7°C/W60.5°C/WMulti-Layer PCB, JEDEC Standard Test Boards49.8°C/W43.9°C/W41.2°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

Table 7B. $\theta_{_{,|A}}$ vs. Air Flow Table For 28 Lead SSOP Package

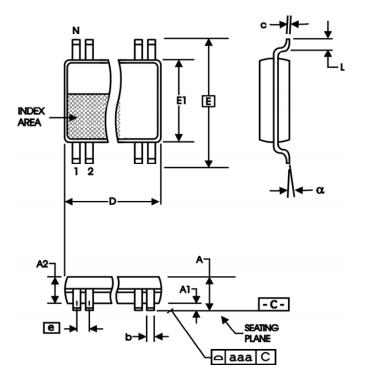
θ_{A} by Velocity (Linear Feet per Minute)

0200500Multi-Layer PCB, JEDEC Standard Test Boards49°C/W36°C/W30°C/W

TRANSISTOR COUNT

The transistor count for ICS9DB306 is: 2190

PACKAGE OUTLINE - G SUFFIX FOR 20 LEAD TSSOP



PACKAGE OUTLINE - F SUFFIX FOR 20 LEAD SSOP

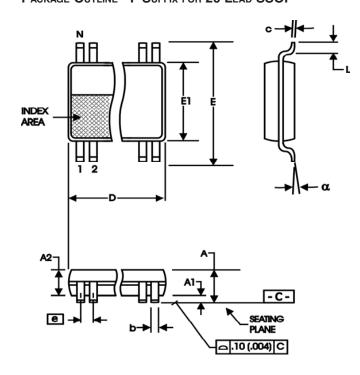


TABLE 8A. PACKAGE DIMENSIONS

CVMPOL	Millimeters			
SYMBOL	Minimum	Maximum		
N	2	0		
А		1.20		
A1	0.05	0.15		
A2	0.80	1.05		
b	0.19	0.30		
С	0.09	0.20		
D	6.40	6.60		
Е	6.40 E	BASIC		
E1	4.30	4.50		
е	0.65 BASIC			
L	0.45	0.75		
α	0°	8°		
aaa		0.10		

Reference Document: JEDEC Publication 95, MO-153

TABLE 8B. PACKAGE DIMENSIONS

CVMPOL	Millimeters			
SYMBOL	Minimum	Maximum		
N	20			
Α		2.0		
A1	0.05			
A2	1.65	1.85		
b	0.22	0.38		
С	0.09	0.25		
D	6.90	7.50		
E	7.40	8.20		
E1	5.0	5.60		
е	0.65 BASIC			
L	0.55	0.95		
α	0°	8°		

Reference Document: JEDEC Publication 95, MO-150

TABLE 9. ORDERING INFORMATION

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
9DB306BL	ICS9DB306BL	28 Lead TSSOP	Tube	0°C to 70°C
9DB306BLT	ICS9DB306BL	28 Lead TSSOP	1000 tape & reel	0°C to 70°C
9DB306BLLF	ICS9DB306BLLF	28 Lead "Lead-Free" TSSOP	Tube	0°C to 70°C
9DB306BLLFT	ICS9DB306BLLF	28 Lead "Lead-Free" TSSOP	1000 tape & reel	0°C to 70°C
9DB306BF	ICS9DB306BF	28 Lead SSOP	Tube	0°C to 70°C
9DB306BFT	ICS9DB306BF	28 Lead SSOP	1000 tape & reel	0°C to 70°C
9DB306BFLF	ICS9DB306BFLF	28 Lead "Lead-Free" SSOP	Tube	0°C to 70°C
9DB306BFLFT	ICS9DB306BFLF	28 Lead "Lead-Free" SSOP	1000 tape & reel	0°C to 70°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

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	REVISION HISTORY SHEET						
Rev	Table	Page	Description of Change	Date			
Α	T3F	2	Added PLL Mode Function Table.	4/7/05			
	T4A	3	Power Supply Table - minimum $V_{\rm CCA}$ changed from 3.135V to $V_{\rm CC}$ - 0.60V, and maximum set to $V_{\rm CC}$.				
В		6	Corrected 3.3V Output Load AC Test Circuit diagram to correspond with Power Supply table.	6/16/06			
		8	Added Recommendations for Unused Input and Output Pins.				
	T9	15	Ordering Information Table - added lead-free SSOP part number.				
В		1	Features Section - added Input Frequency Range and VCO Range bullets.	7/14/06			
			Changed power supply from 3.3V±5% to 3.3V±10% throughout the datasheet.				
	T5	4	AC Characteristics Table - changed Output Skew from 55ps typ./135ps max. to				
С			25ps typ./100ps max. Changed Cycle-toCycle Jitter from 25ps max. to 30ps max. Changed Output Duty Cycle from 48% min./52% max. to 47% min./53% max.	9/22/06			
		11	Power Considerations - correct Power Dissipation to coincide with the power supply change.				
	T4C	3	Differential DC Characteristics Table - updated notes.				
	T5A - T5B	4	AC Characteristics Tables - added thermal note.				
		7	Power Supply Filtering Techniques - deleted last line " The 10ohm resistor can				
С		8	also be replaced by a ferrite bead." Updated Differential Clock Input Interface section.	8/13/09			
		9	Updated Figures 4A and 4B.				
	Т9	15	Ordering Information Table - deleted ICS prefix in Part/Order Number column. Added 28 Lead SSOP Lead-free marking.				



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